



# Low-Cost, High-Slew-Rate, Rail-to-Rail I/O Op Amps in SC70

MAX4490/MAX4491/MAX4492

## General Description

The MAX4490/MAX4491/MAX4492 single/dual/quad, low-cost CMOS op amps feature Rail-to-Rail® input and output capability from either a single +2.7V to +5.5V supply or dual ±1.35V to ±2.75V supplies. These amplifiers exhibit a high slew rate of 10V/μs and a gain-bandwidth product of 10MHz. They can drive 2kΩ resistive loads to within 55mV of either supply rail and remain unity-gain stable with capacitive loads up to 300pF.

The MAX4490 is offered in the ultra-small, 5-pin SC70 package, which is 50% smaller than the standard 5-pin SOT23 package. Specifications for all parts are guaranteed over the automotive (-40°C to +125°C) temperature range.

## Features

- ◆ +2.7V to +5.5V Single-Supply Operation
- ◆ 10V/μs Slew Rate
- ◆ Rail-to-Rail Input Common-Mode Voltage Range
- ◆ Rail-to-Rail Output Voltage Swing
- ◆ 10MHz Gain-Bandwidth Product
- ◆ Unity-Gain Stable with Capacitive Loads Up to 300pF
- ◆ 50pA Input Bias Current
- ◆ Ultra-Small, 5-Pin SC70 Package (MAX4490)

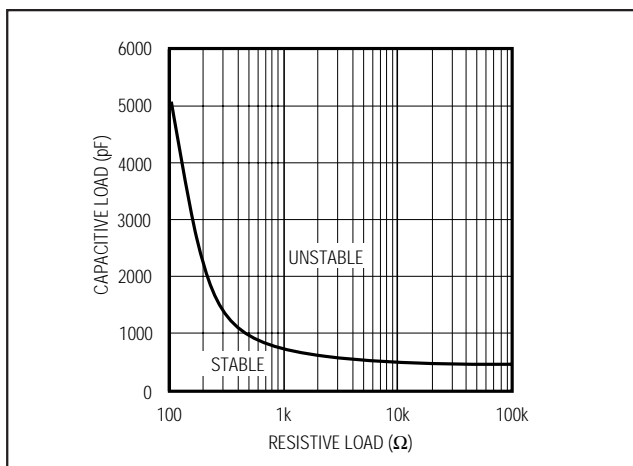
## Applications

- Battery-Powered Instruments
- Portable Equipment
- Audio Signal Conditioning
- Low-Power/Low-Voltage Applications
- Sensor Amplifiers
- RF Power Amplifier Control
- High-Side/Low-Side Current Sensors

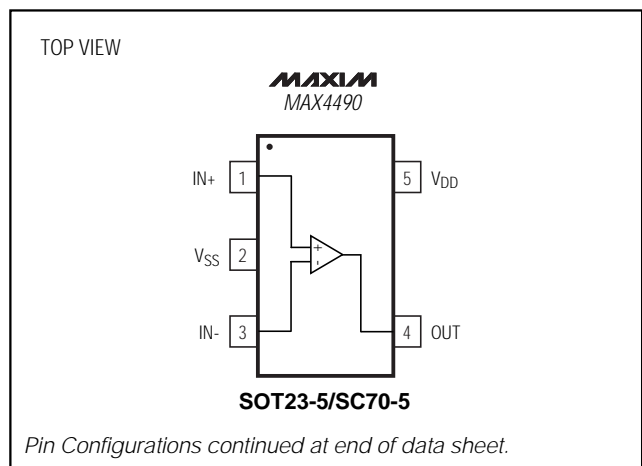
## Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE	TOP MARK
MAX4490AXK-T	-40°C to +125°C	5 SC70-5	AAB
MAX4490AUK-T	-40°C to +125°C	5 SOT23-5	ADKQ
MAX4491AKA-T	-40°C to +125°C	8 SOT23-8	AADB
MAX4492AUD	-40°C to +125°C	14 TSSOP	—
MAX4492ASD	-40°C to +125°C	14 SO	—

## Capacitive Load Stability



## Pin Configurations/Functional Diagrams



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## ABSOLUTE MAXIMUM RATINGS

Supply Voltage ( $V_{DD}$ to $V_{SS}$ ).....	+6V	14-Pin TSSOP (derate 8.3mW/°C above +70°C) .....	667mW
All Other Pins .....	( $V_{SS} - 0.3V$ ) to ( $V_{DD} + 0.3V$ )	14-Pin SO (derate 8.3mW/°C above +70°C).....	667mW
Output Short-Circuit Duration .....	10s	Operating Temperature Range .....	-40°C to +125°C
Continuous Power Dissipation ( $T_A = +70^\circ\text{C}$ )		Junction Temperature .....	+150°C
5-Pin SC70 (derate 2.5mW/°C above +70°C).....	200mW	Storage Temperature Range .....	-65°C to +150°C
5-Pin SOT23 (derate 7.1mW/°C above +70°C).....	571mW	Lead Temperature (soldering, 10s) .....	+300°C
8-Pin SOT23 (derate 5.26mW/°C above +70°C).....	421 mW		

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

( $V_{DD} = +5V$ ,  $V_{SS} = 0$ ,  $V_{CM} = 0$ ,  $V_{OUT} = V_{DD}/2$ ,  $R_L = 100k\Omega$  connected to  $V_{DD}/2$ ,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^\circ\text{C}$ .) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Supply Voltage Range	$V_{DD}$	(Note 2)	2.7		5.5	V	
Supply Current (per amplifier)	$I_S$			0.8	2	mA	
Input Offset Voltage	$V_{OS}$	(Note 3)		$T_A = +25^\circ\text{C}$	$\pm 1.5$	$\pm 10$	mV
		$T_A = T_{MIN}$ to $T_{MAX}$			16		
Input Bias Current	$I_B$	(Note 3)		$\pm 0.05$	$\pm 2.5$	nA	
Input Offset Current	$I_{OS}$	(Note 3)		$\pm 0.05$	$\pm 2.5$	nA	
Input Resistance	$R_{IN}$			1000		M $\Omega$	
Input Common-Mode Range	$V_{CM}$	Inferred from CMRR test	$V_{SS}$		$V_{DD}$	V	
Common-Mode Rejection Ratio	CMRR	$V_{SS} \leq V_{CM} \leq V_{DD}$	54	75		dB	
Power-Supply Rejection Ratio	PSRR	$2.7V \leq V_{DD} \leq 5.5V$	65	100		dB	
Large-Signal Voltage Gain	$A_V$	$(V_{SS} + 0.25V) \leq V_{OUT} \leq (V_{DD} - 0.25V)$	$R_L = 100k\Omega$	110		dB	
			$R_L = 2k\Omega$	65	85		
Output Voltage Swing High	$V_{OH}$	Specified as $V_{DD} - V_{OH}$	$R_L = 100k\Omega$	1.5		mV	
			$R_L = 2k\Omega$	55	200		
Output Voltage Swing Low	$V_{OL}$	Specified as $V_{OL} - V_{SS}$	$R_L = 100k\Omega$	1.5		mV	
			$R_L = 2k\Omega$	35	150		
Output Short-Circuit Current	$I_{OUT(SC)}$	Sourcing or sinking		$\pm 50$		mA	
Gain-Bandwidth Product	GBWP	$C_L = 10pF$		10		MHz	
Input Capacitance	$C_{IN}$			5		pF	
Phase Margin		$C_L = 10pF$		60		deg	
Gain Margin		$C_L = 10pF$		10		dB	
Slew Rate	SR	Measured from 10% to 90% of 4Vp-p step		10		V/ $\mu\text{s}$	
Voltage Noise Density	$e_n$	$f = 10kHz$		12		nV/ $\sqrt{\text{Hz}}$	
Current Noise Density	$i_n$	$f = 10kHz$		1		fA/ $\sqrt{\text{Hz}}$	
Capacitive Load Drive		$A_V(C_L) = 1$ , no sustained oscillations		300		pF	

**Note 1:** All units production tested at  $T_A = +25^\circ\text{C}$ . Limits over temperature guaranteed by design.

**Note 2:** Guaranteed by the Power-Supply Rejection Ratio (PSRR) test.

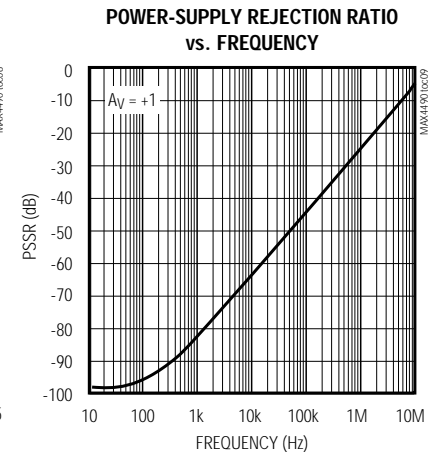
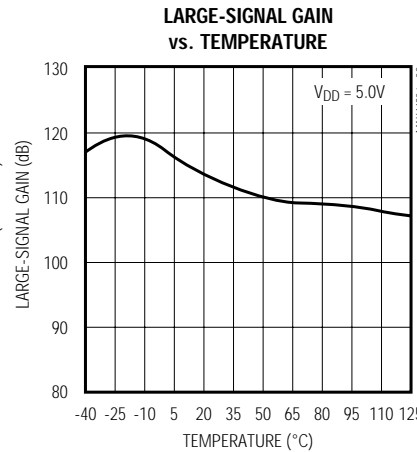
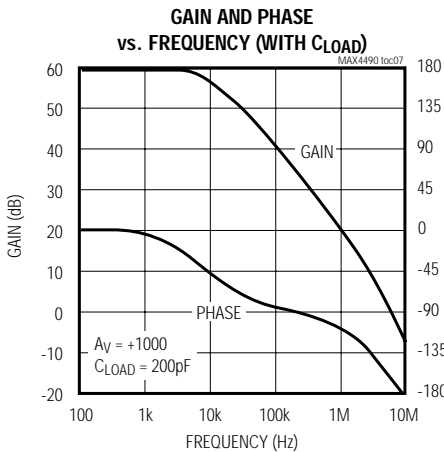
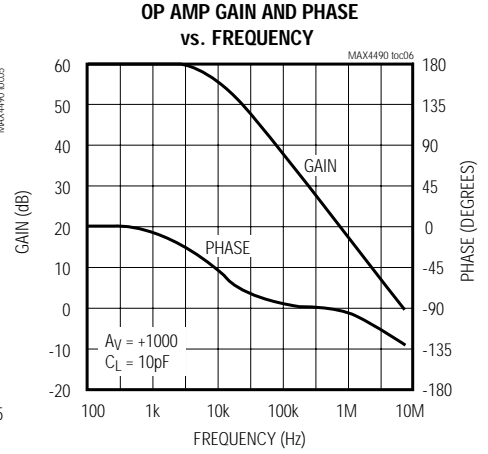
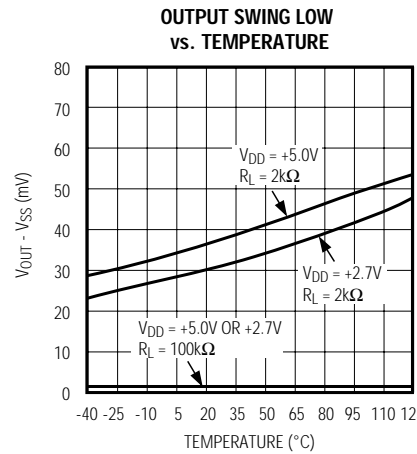
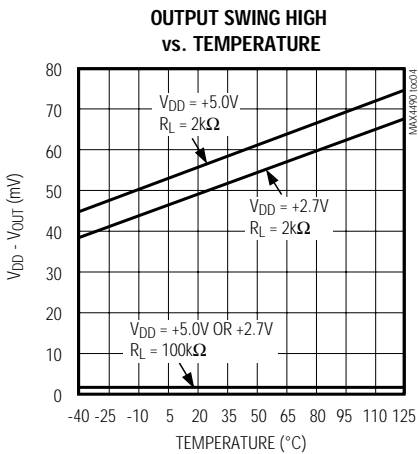
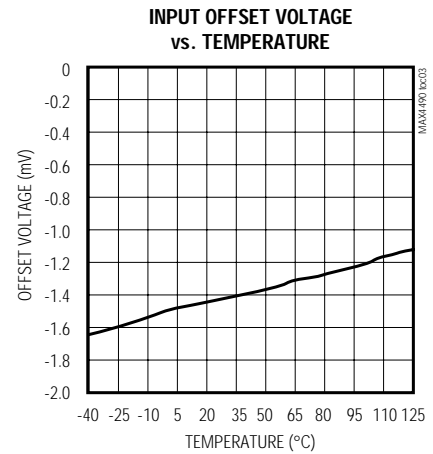
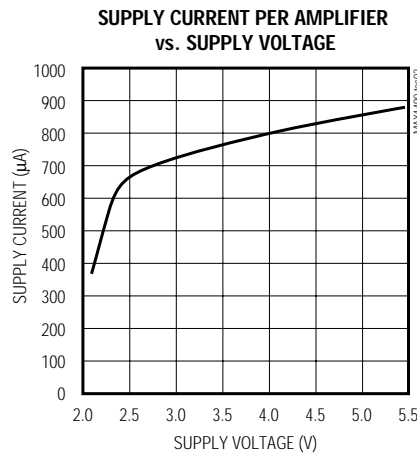
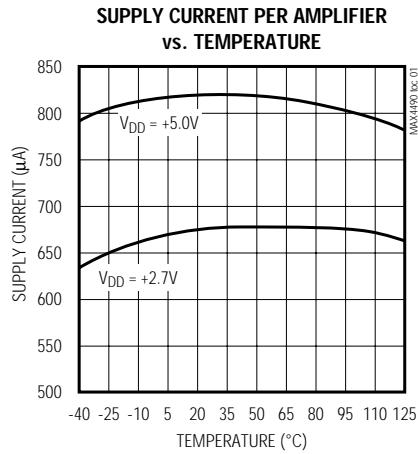
**Note 3:** Input Offset Voltage, Input Bias Current, and Input Offset Current are all tested and guaranteed at both ends of the common-mode range.

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## Typical Operating Characteristics

( $V_{DD} = +5V$ ,  $V_{SS} = 0$ ,  $V_{CM} = V_{DD}/2$ ,  $R_L = 100k\Omega$  to  $V_{DD}/2$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)

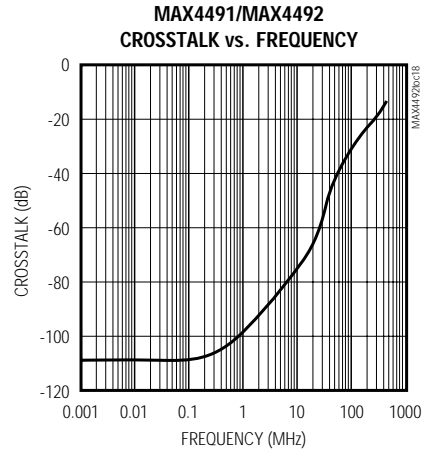
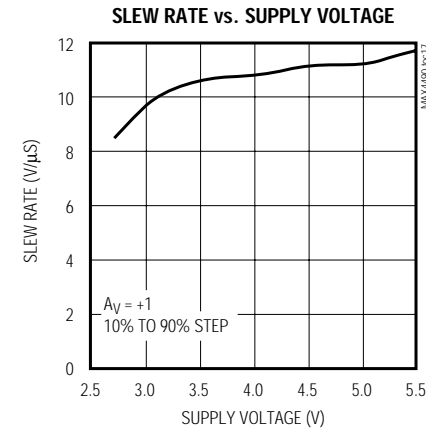
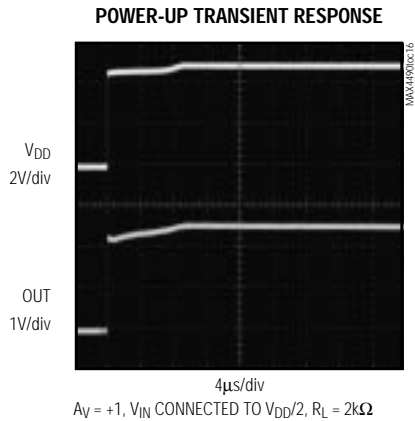
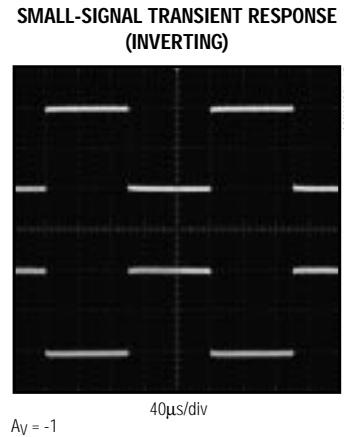
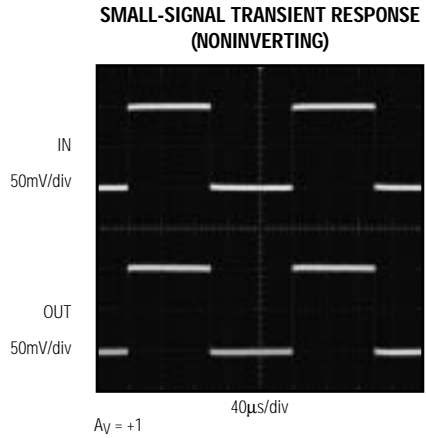
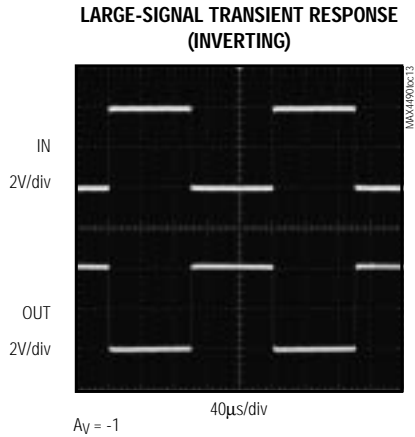
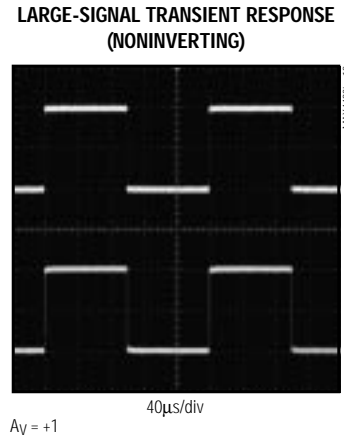
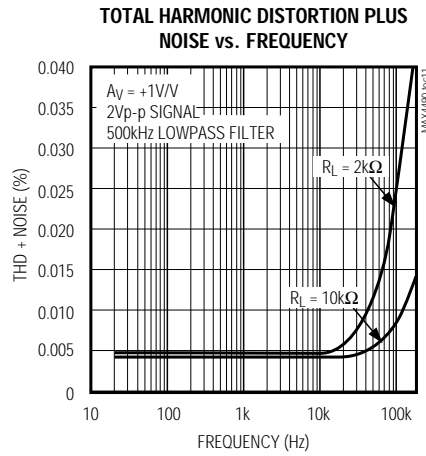
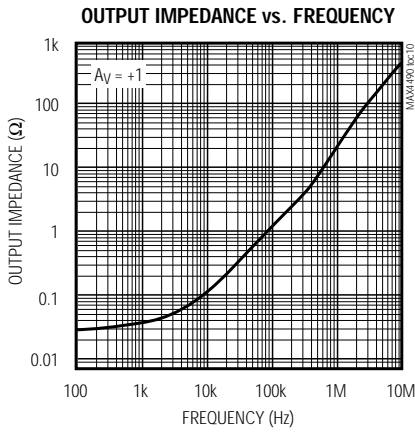
MAX4490/MAX4491/MAX4492



# Low-Cost, High-Slew-Rate, Rail-to-Rail I/O Op Amps in SC70

## Typical Operating Characteristics (continued)

( $V_{DD} = +5V$ ,  $V_{SS} = 0$ ,  $V_{CM} = V_{DD}/2$ ,  $R_L = 100k\Omega$  to  $V_{DD}/2$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)



# Low-Cost, High-Slew-Rate, Rail-to-Rail I/O Op Amps in SC70

## Pin Description

MAX4490/MAX4491/MAX4492

PIN			NAME	FUNCTION
MAX4490	MAX4491	MAX4492		
1	–	–	IN+	Noninverting Input
2	4	11	V <sub>SS</sub>	Negative Supply Input. Connect to ground for single-supply operation.
3	–	–	IN-	Inverting Input
4	–	–	OUT	Amplifier Output
5	8	4	V <sub>DD</sub>	Positive Supply Input
–	3	3	INA+	Noninverting Input to Amplifier A
–	2	2	INA-	Inverting Input to Amplifier A
–	1	1	OUTA	Amplifier A Output
–	5	5	INB+	Noninverting Input to Amplifier B
–	6	6	INB-	Inverting Input to Amplifier B
–	7	7	OUTB	Amplifier B Output
–	–	10, 12	INC+, IND+	Noninverting Inputs to Amplifiers C and D
–	–	9, 13	INC-, IND-	Inverting Inputs to Amplifiers C and D
–	–	8, 14	OUTC, OUTD	Amplifiers C and D Outputs

## Detailed Description

### Rail-to-Rail Input Stage

The MAX4490/MAX4491/MAX4492 CMOS operational amplifiers have parallel-connected N- and P-channel differential input stages that combine to accept a common-mode range extending to both supply rails. The N-channel stage is active for common-mode input voltages typically greater than (V<sub>SS</sub> + 1.2V), and the P-channel stage is active for common-mode input voltages typically less than (V<sub>DD</sub> - 1.2V).

### Rail-to-Rail Output Stage

The MAX4490/MAX4491/MAX4492 CMOS operational amplifiers feature class-AB push-pull output stages that can drive a 100k $\Omega$  load to within 1.5mV of either supply rail. Short-circuit output current is typically  $\pm$ 50mA.

Figures 1a and 1b show the typical temperature dependence of output source and sink currents, respectively, for three fixed values of (V<sub>DD</sub> - V<sub>OH</sub>) and (V<sub>OL</sub> - V<sub>SS</sub>). For example, at V<sub>DD</sub> = +5.0V, the load currents that maintain (V<sub>DD</sub> - V<sub>OH</sub>) = 100mV and (V<sub>OL</sub> - V<sub>SS</sub>) = 100mV at T<sub>A</sub> = +25°C are 2.2mA and 3.3mA, respectively,

when the load is connected to V<sub>DD</sub>/2. Consistent resistive drive capability is (2.5 - 0.1) / 2.2 = 1.1k $\Omega$ . For the same application, resistive drive capability is 2.2k $\Omega$  when the load is connected to V<sub>DD</sub> or V<sub>SS</sub>.

## Applications Information

### Power-Supply Considerations

The MAX4490/MAX4491/MAX4492 operate from a single +2.7V to +5.5V supply or from dual  $\pm$ 1.35V to  $\pm$ 2.75V supplies with typically 800 $\mu$ A supply current per amplifier. A high power-supply rejection ratio of 100dB allows for extended operation from a decaying battery voltage, thereby simplifying designs for portable applications. For single-supply operation, bypass the power supply with a 0.1 $\mu$ F ceramic capacitor placed close to the V<sub>DD</sub> pin. For dual-supply operation, bypass each supply to ground.

### Input Capacitance

One consequence of the parallel-connected differential input stages for rail-to-rail operation is a relatively large input capacitance C<sub>IN</sub> (typically 5pF). This introduces a

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pole at frequency  $(2\pi R' C_{IN})^{-1}$ , where  $R'$  is the parallel combination of the gain-setting resistors for the inverting or noninverting amplifier configuration (Figure 2). If the pole frequency is less than or comparable to the unity-gain bandwidth (10MHz), the phase margin will be reduced, and the amplifier will exhibit degraded AC performance through either ringing in the step response or sustained oscillations. The pole frequency is 10MHz when  $R' = 3.2k\Omega$ . To maximize stability,  $R' < 3k\Omega$  is recommended.

Applications that require rail-to-rail operation with minimal loading (for small  $V_{DD} - V_{OH}$  and  $V_{OL} - V_{SS}$ ) will typically require  $R'$  values  $> 3k\Omega$ . To improve step response under these conditions, connect a small

capacitor  $C_f$  between the inverting input and output. Choose  $C_f$  as follows:

$$C_f = 5(R / R_f) \text{ [pF]}$$

where  $R_f$  is the feedback resistor and  $R$  is the gain-setting resistor (Figure 2).

Figure 3 shows the step response for a noninverting amplifier subject to  $R' = 4k\Omega$  with and without the  $C_f$  feedback capacitor.

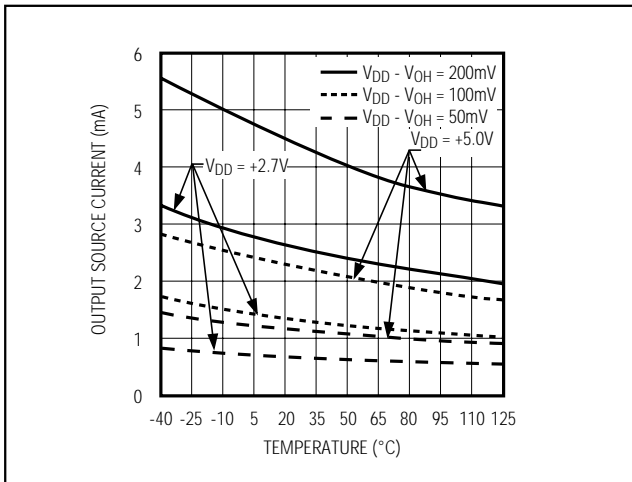


Figure 1a. Output Source Current vs. Temperature

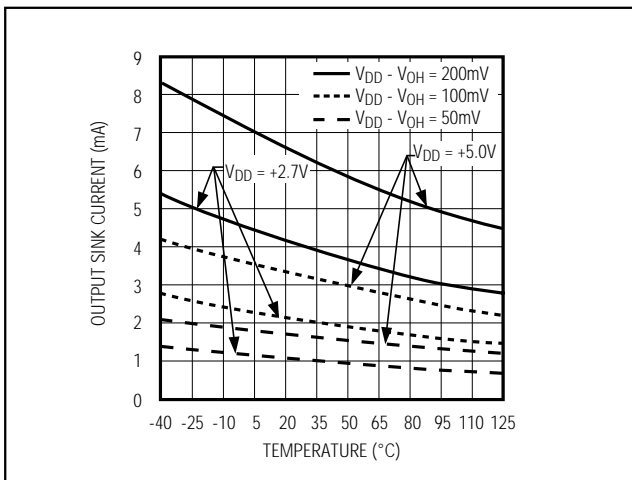


Figure 1b. Output Sink Current vs. Temperature

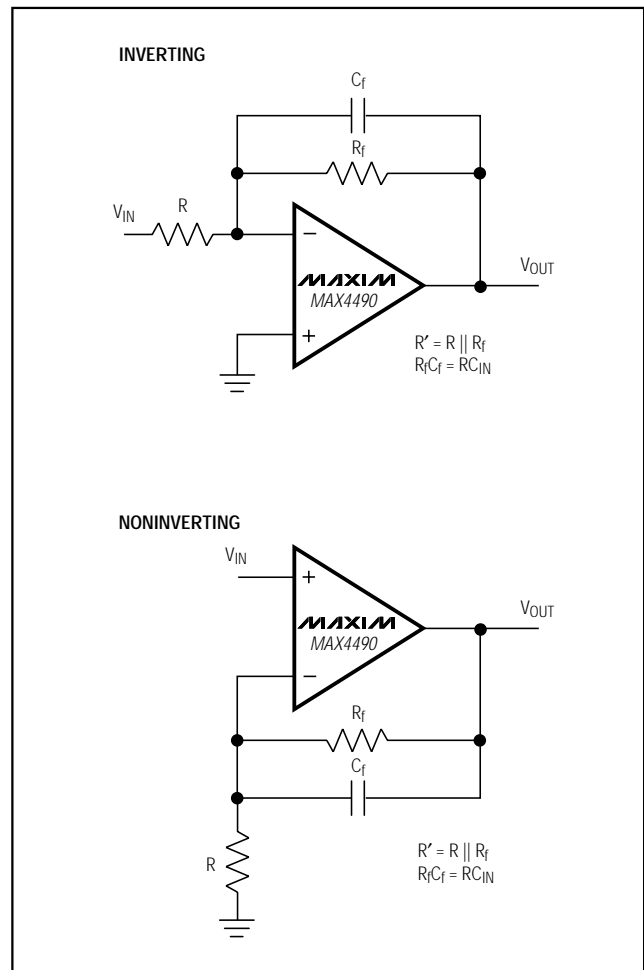


Figure 2. Inverting and Noninverting Amplifier with Feedback Compensation

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## Driving Capacitive Loads

In conjunction with op amp output resistance, capacitive loads introduce a pole frequency that can reduce phase margin and lead to unstable operation. The MAX4490/MAX4491/MAX4492 drive capacitive loads up to 300pF without significant degradation of step response and slew rate (Figure 4). *Capacitive Load Stability* (page 1) shows regions of stable and marginally stable (step overshoot <10%) operation for different combinations of capacitive and resistive loads.

Improve stability for large capacitive loads by adding an isolation resistor (typically 10 $\Omega$ ) in series with the output (Figure 5). Note that the isolation resistor forms a voltage divider with potential for gain error.

## Chip Information

MAX4490 TRANSISTOR COUNT: 60  
 MAX4491 TRANSISTOR COUNT: 120  
 MAX4492 TRANSISTOR COUNT: 240  
 SUBSTRATE CONNECTED TO V<sub>SS</sub>

MAX4490/MAX4491/MAX4492

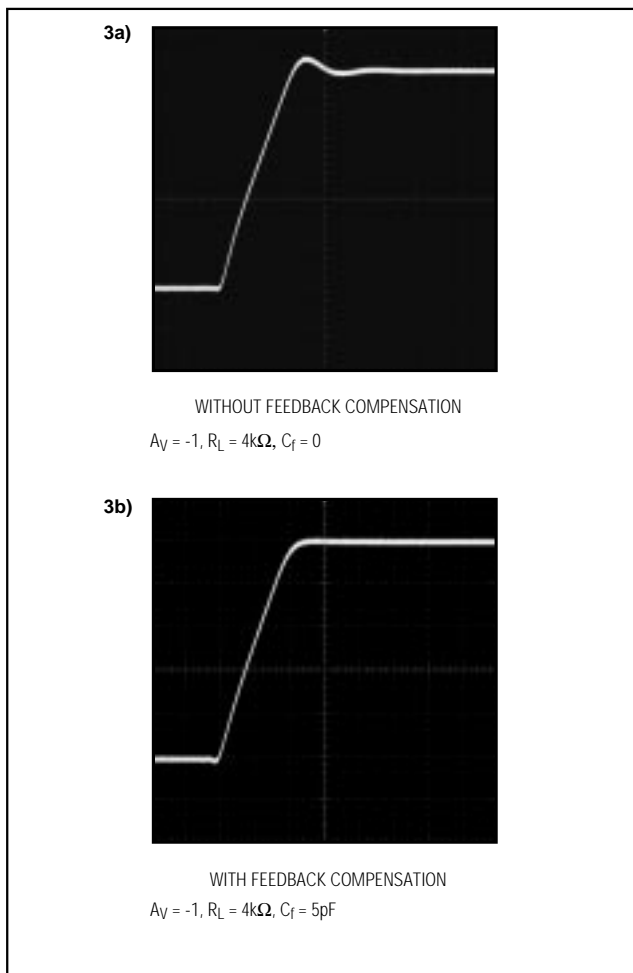


Figure 3. Step Response With and Without Feedback Compensation

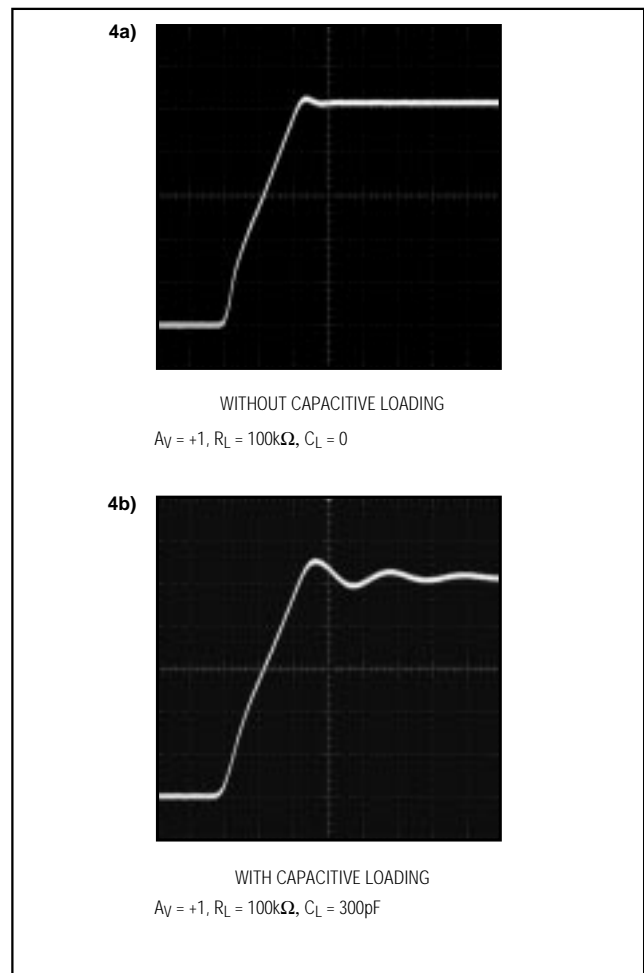


Figure 4. Step Response With and Without Capacitive Loading

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## Pin Configurations/Functional Diagrams (continued)

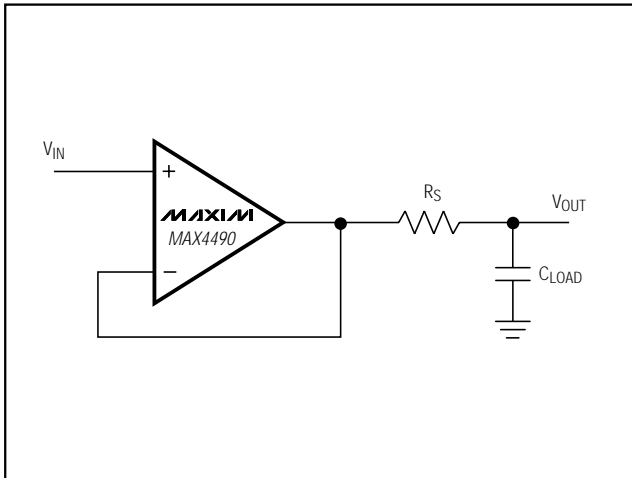
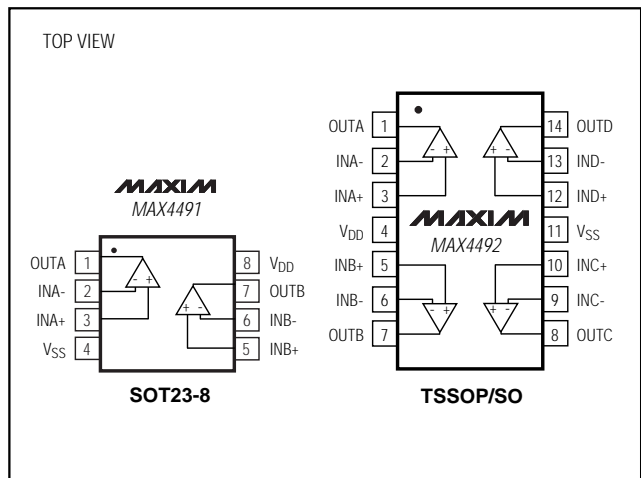
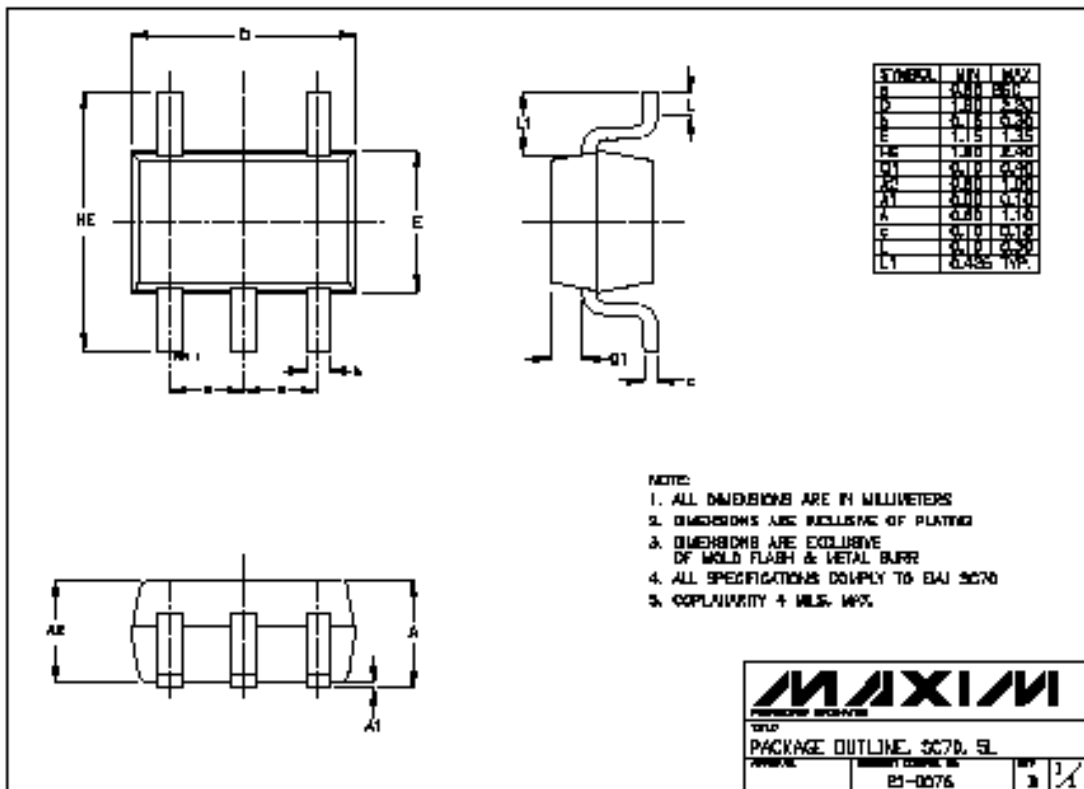


Figure 5. Isolation Resistor for Large Capacitive Loads



## Package Information



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