DATA SHEET



BIPOLAR ANALOG INTEGRATED CIRCUIT $\mu PC1857A$

SOUND CONTROL IC WITH SURROUND AND I'C BUS

DESCRIPTION

The μ PC1857A is a sound control IC with I²C bus.

It has functions to control volume, balance, and tone, and a phase shift matrix surround function.

The surround function achieves wide sound expansion using only two front speakers. Three modes can be selected: movie mode that increases the presence of sound with stereo sound input, music mode emphasizing vocal music, and simulated mode that gives expansion and left and right sound depth with monaural sound input.

The μ PC1857A can perform all control (mode switching, volume control and so on) using I²C.

FEATURES

Volume control function
 Attenuation adjustable from 0 to -80 dB in 64 steps

Balance control function
 The difference in attenuation adjustable from 0 to -80 dB in 64 steps

 \bullet Tone (bass, treble) control function : Adjustable in 32 steps from +10 to -10 dB

Surround function (gain adjustable): Three modes (movie, music, and simulated)

· Mute function

- · Mixing function
- Output selection function (for two mono channels input)
- All parameters can be controlled via I²C bus.

APPLICATIONS

· TV, PC monitor

ORDERING INFORMATION

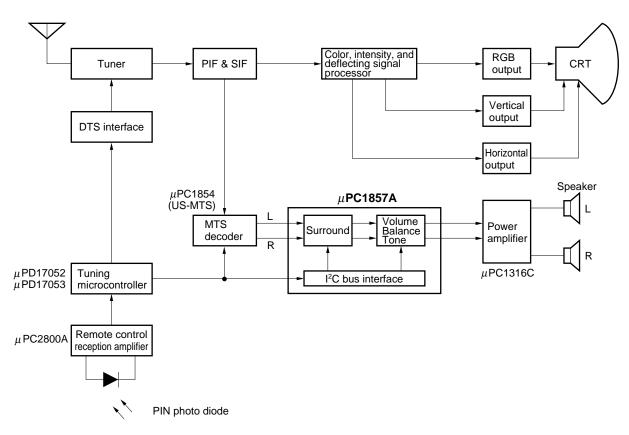
Part Number	Package
μ PC1857ACT	30-pin plastic shrink DIP (400 mil)

The information in this document is subject to change without notice.



SYSTEM BLOCK DIAGRAM

TV

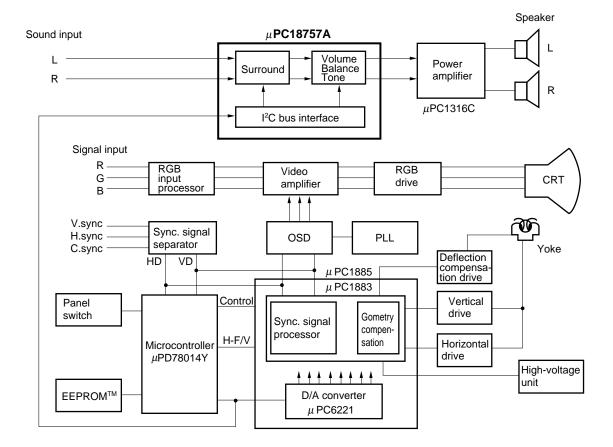


Remark DTS: Digital Tuning System

MTS: Multichannel Television Sound

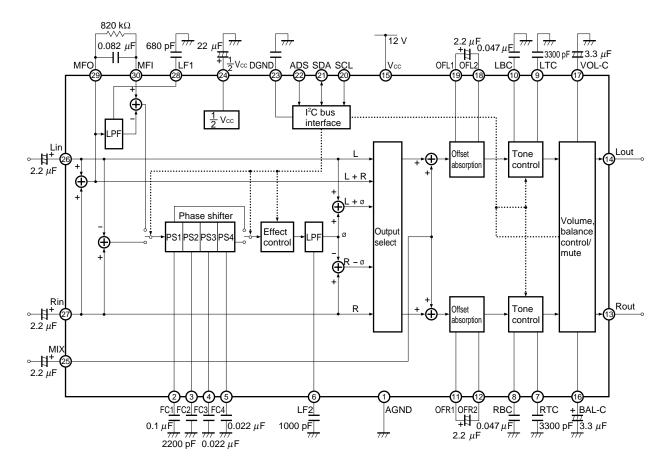


• PC monitor





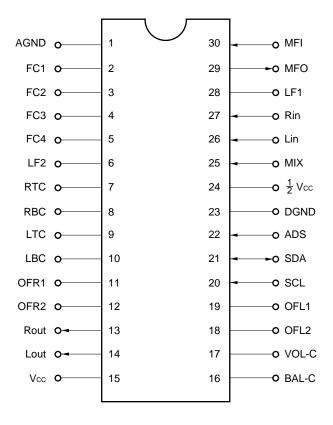
BLOCK DIAGRAM





PIN CONFIGURATION (Top View)

30-pin plastic shrink DIP (400 mil)



$\frac{1}{2}$ Vcc	: Reference Voltage Filter	MFO	: Monaural Filter Output
ĀDS	: Slave Address Select	MIX	: Mixer Input
AGND	: Analog Ground	OFL1, OFL2	: L-channel Offset Absorption
BAL-C	: Balance Control Offset Absorption	OFR1, OFR2	: R-channel Offset Absorption
DGND	: Ground for I ² C Bus	RBC	: R-channel Bass Capacitor
FC1-FC4	: Phase Shift Filter	Rin	: R-channel Signal Input
LBC	: L-channel Bass Capacitor	Rout	: R-channel Signal Output
LF1, LF2	: Low-pass Filter	RTC	: R-channel Treble Capacitor
Lin	: L-channel Signal Input	SCL	: Serial Clock for I ² C Bus
Lout	: L-channel Signal Output	SDA	: Serial Data for I ² C Bus

LTC : L-channel Treble Capacitor Vcc : Power Supply

MFI : Monaural Filter Input VOL-C : Volume Control Offset Absorption

CONTENTS

1.	I. PIN FUNCTIONS	7
2.	2. ATTENTIONS	14
3.	B. I'C BUS INTERFACE	15
	3.1 Data Transfer	15
	3.1.1 Start condition	15
	3.1.2 Stop condition	16
	3.1.3 Data transfer	16
	3.2 Data Transfer Format	17
	3.2.1 1-byte data transfer	18
	3.2.2 Successive data transfer	18
	3.2.3 Acknowledge	18
4.	I. EXPLANATION OF EACH COMMAND	19
	4.1 Subaddress List	19
	4.2 Initialization	20
	4.3 Surround Function	20
	4.4 Explanation of Each Command	21
	4.4.1 Mute	21
	4.4.2 Output selection	21
	4.4.3 Surround mode	22
	4.4.4 Surround effect	23
	4.4.5 Mix	23
	4.4.6 Automatic increment	24
	4.4.7 Volume level	25
	4.4.8 Balance	25
	4.4.9 Bass level	26
	4.4.10 Treble level	26
5.	5. ELECTRICAL CHARACTERISTICS	27
6.	CHARACTERISTIC CURVES	33
	6.1 Frequency Characteristic in Each Mode	33
	6.2 Control Characteristic	36
	6.3 I/O Characteristic	38
7.	7. PACKAGE DRAWING	39
8.	B. RECOMMENDED SOLDERING CONDITIONS	40



1. PIN FUNCTIONS

Table 1-1. Pin Function List (1/7)

Pin Number	Pin Name	Equivalent Circuit	Description
1	AGND	15 (3)	Ground for analog signal. Pin voltage: approx. 0.0 V
2	FC1	$36 \text{ k}\Omega$	Connection pin for capacitor which determines time constant of phase shifter. Pin voltage: approx. 6.0 V
3	FC2	36 kΩ 36 kΩ 3 kΩ 36 kΩ 36 kΩ 37 y 2200 pF	
4	FC3	$36 \text{ k}\Omega$	

Remark Pin voltage is the reference value when Vcc = 12 V.



Table 1-1. Pin Function List (2/7)

Pin Number	Pin Name	Equivalent Circuit	Description
5	FC4	$36 \text{ k}\Omega$	Connection pin for capacitor which determines time constant of phase shifter. Pin voltage: approx. 6.0 V
6	LF2	7/7 kΩ 5 kΩ 5 kΩ 17.7 kΩ 1000 pF	Low-pass filter. Pin voltage: approx. 6.0 V
7	RTC	3 kΩ 3 kΩ 12 kΩ 12 kΩ 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7	Connection pin for capacitor for treble boost/cut frequency characteristic of R-channel signal. Pin voltage: approx. 6.0 V
8	RBC	$3 \text{ k}\Omega$	Connection pin for capacitor for bass boost/cut frequency characteristic of R-channel signal. Pin voltage: approx. 6.0 V



Table 1-1. Pin Function List (3/7)

Pin Number	Pin Name	Equivalent Circuit	Description
9	LTC	3 kΩ 3 kΩ 12 kΩ Vcc 12 kΩ 3 800 pF	Connection pin for capacitor for treble boost/cut frequency characteristic of L-channel signal. Pin voltage: approx. 6.0 V
10	LBC	$3 \text{ k}\Omega$	Connection pin for capacitor for bass boost/cut frequency characteristic of L-channel signal. Pin voltage: approx. 6.0 V
11	OFR1	$\begin{array}{c c} & & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & &$	Pin that absorbs offset voltage of R channel. Pin voltage: approx. 6.0 V
12	OFR2	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	



Table 1-1. Pin Function List (4/7)

Pin Number	Pin Name	Equivalent Circuit	Description
13	Rout	V_{CC} V	R-channel signal output pin. Pin voltage: approx. 6.0 V
14	Lout	V_{CC} V	L-channel signal output pin. Pin voltage: approx. 6.0 V
15	Vcc	1 23	Supply voltage. Pin voltage: approx. 12.0 V
16	BAL-C	3.3 μF 2// 16 50 kΩ	Pin for D/A converter capacitor for balance control. Pin voltage: approx. 4.8 V



Table 1-1. Pin Function List (5/7)

Pin Number	Pin Name	Equivalent Circuit	Description
17	VOL-C	3.3 μF ## 17	Pin for D/A converter capacitor for volume control. Pin voltage: approx. 6.0 V
18	OFL2	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Pin that absorbs offset voltage of L channel. Pin voltage: approx. 6.0 V
19	OFL1	$\begin{array}{c c} & 6 & k\Omega \\ & 6 & k\Omega \\ & 5 & k\Omega \end{array}$	
20	SCL	20 5 kΩ	Serial clock line (I ² C bus clock Input) pin. Pin voltage: approx. 0.0 V
21	SDA	21 	Serial data line (l²C bus data l/O) pin Pin voltage: approx. 0.2 V
22	ADS	22 5 κΩ	Slave address select pin. Pin voltage: approx. 0.0 V
23	DGND	1) (23)	GND for I ² C bus signal. Pin voltage: approx. 0.0 V



Table 1-1. Pin Function List (6/7)

Pin Number	Pin Name	Equivalent Circuit	Description
24	1/2 Vcc	$\begin{array}{c c} & & & & & \\ & & & & \\ & & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & \\ & & & \\ & &$	Filter pin for middle point of supply voltage. Pin voltage: approx. 6.0 V
25	MIX	1/2Vcc 3 kΩ 60 kΩ 5 kΩ 2.2 μF 277 ΜΙΧ	Mixing signal input pin. Input impedance: 60 kΩ Pin voltage: approx. 6.0 V
26	Lin	$ \begin{array}{c c} 1/2Vcc \\ Vcc \\ \hline & 3 k\Omega \end{array} $ $ \begin{array}{c} 3 k\Omega \end{array} $ $ \begin{array}{c} 3 k\Omega \end{array} $ $ \begin{array}{c} 5 k\Omega \end{array} $ $ \begin{array}{c} 2.2 \mu \text{F} \\ \end{array} $	L-channel signal input pin. Input impedance: 60 kΩ Pin voltage: approx. 6.0 V
27	Rin	1/2Vcc Vcc 3 kΩ 60 kΩ 5 kΩ 2.2 μF ΖΖΖ Rin	R-channel signal input pin. Input impedance: 60 kΩ Pin voltage: approx. 6.0 V



Table 1-1. Pin Function List (7/7)

Pin Number	Pin Name	Equivalent Circuit	Description
28	LF1	$ \begin{array}{c} 18 \text{ k}\Omega \\ \hline V_{\text{CC}} \\ \hline 5 \text{ k}\Omega \end{array} $ $ \begin{array}{c} 5 \text{ k}\Omega \end{array} $ $ \begin{array}{c} 680 \text{ pF} \\ \hline \end{array} $	Low-pass filter. Pin voltage: approx. 6.0 V
29	MFO	29 18 kΩ 18 kΩ	Filter output pin for surround function (simulated mode) (see 4.3 Surround Function). Pin voltage: approx. 6.0 V
30	MFI	820 k Ω $\stackrel{\bigcirc}{\downarrow}$ 0.082 μ F $\stackrel{\bigvee_{CC}}{\downarrow}$ 15 k Ω ${\downarrow}$ 47 k Ω	Filter input pin for surround function (simulated mode) (see 4.3 Surround Function). Pin voltage: approx. 6.0 V



2. ATTENTIONS

(1) Attention on Pop Noise Reduction

When changing the surround mode, use the mute function (approx. 200 ms) for pop noise reduction (see **4.4.1 Mute**).

When turning ON/OFF power to the μ PC1857A, use the external mute function for pop noise reduction.

(2) Attention on Supply Voltage

Drive data on the I²C bus after supply voltage of total application system becomes stable.



3. I'C BUS INTERFACE

The μ PC1857A has serial bus function.

This serial bus (I²C bus) is a double-wired bus developed by Philips. It is composed of 2 wires: serial clock line (SCL) and serial data line (SDA).

The μ PC1857A has built-in I²C bus interface circuit, and five rewritable registers (8 bits).

SCL (Serial Clock Line)

The host CPU outputs a serial clock to synchronize with the data. The μ PC1857A takes in the serial data based on this clock.

Input level is compatible with CMOS.

Clock frequency is 0 to 100 kHz.

SDA (Serial Data Line)

The host CPU outputs the data which is synchronized with the serial clock. The μ PC1857A takes in this data based on the clock.

Input level is compatible with CMOS.

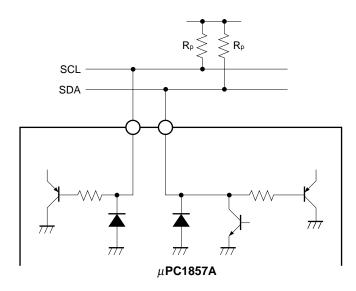


Figure 3-1. Internal Equivalent Circuit of Interface Pin

3.1 Data Transfer

3.1.1 Start condition

Start condition is made by SDA falling from "High" to "Low" while SCL is "High" as shown in Figure 3-2.

When this start condition is received, the μ PC1857A takes in the data synchronized with the serial clock after that.



3.1.2 Stop condition

Stop condition is made by SDA rising from "Low" to "High" while SCL is "High" as shown in Figure 3-2. When this stop condition is received, the μ PC1857A stops taking in or outputting data.

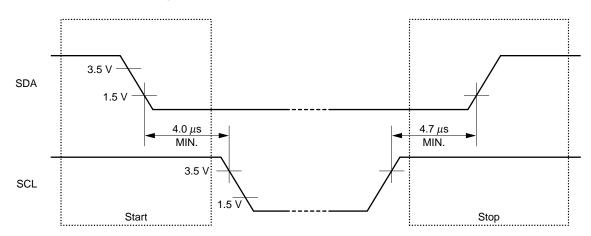


Figure 3-2. Start/Stop Condition of Data Transfer

3.1.3 Data transfer

When transferring data, the data must be changed while SCL is "Low" as shown in Figure 3-3. Never change the data while SCL is "High".

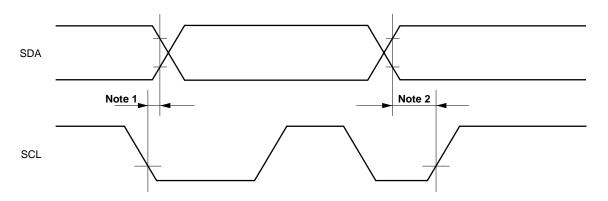


Figure 3-3. Data Transfer

Notes 1. Data hold time for I^2C device: 300 ns MIN., Data hold time for CPU: 5 μ s MIN.

2. Data setup time: 250 ns MIN.

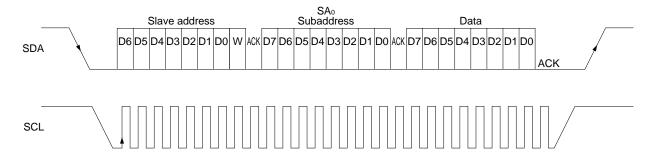
Remark Clock frequency: 0 to 100 kHz



3.2 Data Transfer Format

Figure 3-4 shows an example of data transfer in write mode.

Figure 3-4. Example of Data Transfer in Write Mode



Remark W: Write mode, ACK: Acknowledge bit

Data is composed of 8 bits. One acknowledge bit always follows these 8 bits of data. Data must be transferred starting from the MSB.

The 1 byte immediately following the start condition specifies a slave address (chip address). This slave address is composed of 7 bits.

Table 3-1 shows the slave address of the μ PC1857A. This slave address is registered by Philips.

Table 3-1. Slave Address of μ PC1857A

Bias Voltage of ADS (Pin 22)		Slave Address						
	D6	D5	D4	D3	D2	D1 ^{Note}	D0	
5 V	1	0	0	0	1	1	0	
GND	1	0	0	0	1	0	0	

Note The user can set bit D1 freely.

0: Bias voltage of ADS (pin 22) is 0 V.

1: Bias voltage of ADS (pin 22) is 5 V.

The 1 bit following the slave address is a read/write bit which specifies the direction of the data to be subsequently transferred. Write "0" to this read/write bit because the μ PC1857A is write mode only.

The byte following the slave address is the subaddress byte of the μ PC1857A.

The μ PC1857A has five subaddresses, from SA₀ to SA₄, and each of these addresses is composed of 8 bits. The data to be set to a subaddress follows this subaddress byte.



3.2.1 1-byte data transfer

The format in which 1-byte data is to be transferred is as follows:

Start	Slave address	Write mode	ACK	Subaddress	ACK	Data	ACK	Stop
-------	---------------	---------------	-----	------------	-----	------	-----	------

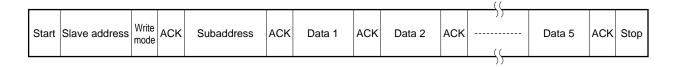
3.2.2 Successive data transfer

The μ PC1857A has an automatic increment function which can be used to transfer successive data (refer to **4.4.6 Automatic increment**).

By using this function, the internal subaddress is automatically incremented if a slave address and a subaddress have been set, so that the data from subsequent subaddresses can be transferred in succession.

Incrementing the subaddress of the μ PC1857A is stopped when the subaddress reaches "04H".

The format in which 5 bytes of data are to be transferred in succession by using the automatic increment function is as follows:



The host CPU transfers "00H" as subaddress SA₀ after start and slave addresses, as shown above. Data SA₀ is transferred after this subaddress SA₀, and without transferring the stop condition the data SA₁, SA₂, SA₃, and SA₄ are transferred successively, and then the stop condition is transferred.

To successively change data at a fixed subaddress, for example to turn up/down the volume, turn off the automatic increment function.

3.2.3 Acknowledge

On the I²C bus, an acknowledge bit is appended to the 9th bit following the data. This acknowledge bit is used to judge whether data transfer has been successful. The host CPU judges whether data transfer has been successful or not, depending on whether the status of the acknowledge bit is "H" or "L".

When the acknowledge bit is "L", it indicates success. When the acknowledge bit is "H", it indicates failure of transfer or forced release of bus (NAK status). The NAK status occurs when a wrong slave address is transferred to a slave IC or data transfer from slave side is finished in the read status.



4. EXPLANATION OF EACH COMMAND

4.1 Subaddress List

Bit Subaddress	MSB D7	D6		D5		D4	D	3	D2			D1		LSB D0
00H	Output mute	0		Outp	ut sele	ct	,	Surround	mode			Surro	ound effe	ect
	0: OFF		D5	D4	Lout	Rout	D3	D2	Мо	de	D'	1 I	D0	Gain
	1: ON		0	0	L	R	0	0	Simu	lated	0		0	0 dB
			0	1	L	L	0	1	Mu	sic	0		1	–3 dB
			1	0	R	R	1	0	Mo	vie	1		0	–6 dB
			1	1	L+R	L+R	1	1	OF	F	1		1 -	-12 dB
01H	Mix	Automatic				Volume	level							
	0: OFF	increment				Volume	e :	MAX	to	MI	N			
	1: ON	0: OFF				Data	:	111111	to	0000	000			
		1: ON												
02H	0	Automatic				Balanc	е							
		increment				L volun	ne:	MIN	to	MA	λX	to	MAX	
		0: OFF				R volur	ne:	MAX	to	MA	λX	to	MIN	
		1: ON				Data	:	111111	to	1000	000	to	000000	
03H	0	Automatic		0		Bass lev	el							
		increment				Gain	:	Boost	to	0		to	Cut	
		0: OFF				Data	:	11111	to	1000	0	to	00000	
		1: ON												
04H	0	Automatic		0		Treble le	evel							
		increment				Gain	:	Boost	to	0		to	Cut	
		0: OFF				Data	:	11111	to	1000	0	to	00000	
		1: ON												

- Cautions 1. Be sure to write "0" to bit D7 of subaddresses 02H through 04H, bit D6 of subaddress 00H, and bit D5 of subaddresses 03H and 04H.
 - 2. The surround mode is OFF: 00H (D3, D2 = 11) in any mode other than stereo mode is selected for output: 00H (D5, D4 = 00).



4.2 Initialization

After power application, be sure to initialize the subaddresses as shown below.

Table 4-1. Initialization of μ PC1857A (recommendation value)

Bit Subaddress	MSB D7	D6	D5	D4	D3	D2	D1	LSB D0
00H	0	0	0	0	1	1	0	0
01H	0	-	ı	_	-	-	-	-
02H	0	-	1	0	0	0	0	0
03H	0	_	0	1	0	0	0	0
04H	0	_	0	1	0	0	0	0

Caution Until initialization is completed, mute using an external unit.

Remark -: Don't care.

4.3 Surround Function

For how to set the surround mode, refer to the table below.

Table 4-2. Setting Surround Mode

Setting	Subaddr	ess: 00H	De	escription
Surround Mode	D3	D2	Units of Phase Shifter	Effect
Simulated	0	0	4 units	Monaural to pseudo-stereo
Music	0	1	1 unit	Stereo sound to surround
Movie	1	0	4 units	
OFF	1	1	-	Through

Caution When changing the surround mode, use the mute function (approx. 200 ms) for pop noise reduction.

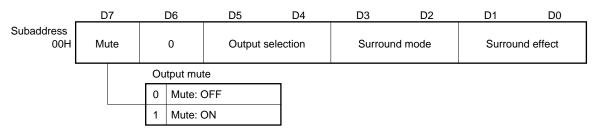


4.4 Explanation of Each Command

4.4.1 Mute

The mute function can be turned ON/OFF by using data of bit D7 of subaddress 00H.

Figure 4-1. Mute



Caution When changing the surround mode, and when turning ON/OFF power, use the mute function (approx. 200 ms) for pop noise reduction.

4.4.2 Output selection

Output can be selected by using data of bits D5 and D4 of subaddress 00H.

Figure 4-2. Output Selection

_	D7	D6	D5	D4	ļ		D3	D2	D1	D0
Subaddress 00H	Mute	0	Output :	out selection			Surrour	nd mode	Surrour	nd effect
					Ou	tput	selection			
				Da		ata	Output			
					D5	D4	Lout	Rout		
					0	0	L	R		
					0	1	L	L		
					1	0	R	R		
					1	1	L+R	L+R		

Caution The surround mode is OFF (D3, D2 = 11) in modes other than the stereo mode (D5, D4 = 00).



Music

4.4.3 Surround mode

The following surround modes can be selected by using data of bits D3 and D2 of subaddress 00H.

Simulated: Simulated stereo sound for monaural source. The difference between the signal that has gone

through HPF and the signal that has gone through LPF is calculated, and the phase of the difference is shifted and added to the original signal. The simulated stereo effect is created if the

output frequency characteristics of the L-channel and R-channel signals is comb-shaped.

: Surround sound for stereo source. The phase of the differential signal between L and R channels

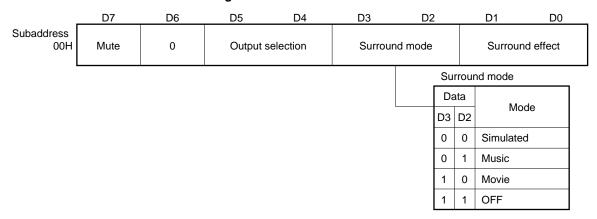
(L-R signal) is rotated by a phase shifter (1-unit), and is added to the original signal.

Movie : Surround sound for stereo source. The phase of the differential signal between L and R channels

(L-R signal) is rotated by a phase shifter (4-unit), and is added to the original signal.

OFF : Original signal as is.

Figure 4-3. Surround Mode



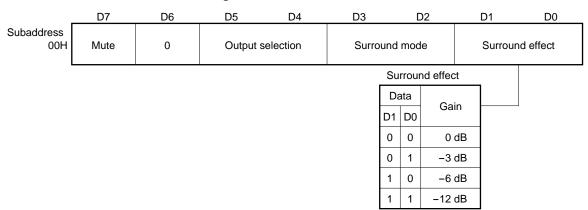
Caution The surround mode is OFF (D3, D2 = 11) if the stereo mode is not selected by the output selection bits (D5, D4 = 00).



4.4.4 Surround effect

The surround effect can be changed in four steps by using the data of bits D1 and D0 of subaddress 00H.

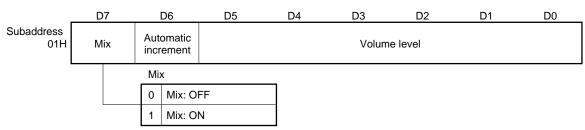
Figure 4-4. Surround Effect



4.4.5 Mix

Mixing of the signal input to the MIX pin can be turned ON/OFF by using the data of bit D7 of subaddress 01H.

Figure 4-5. Mix



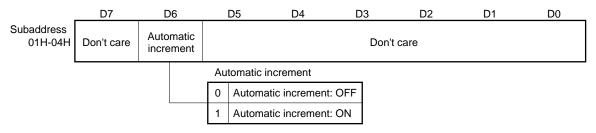


4.4.6 Automatic increment

The automatic increment function can be turned ON/OFF by using the data of bit D6 of subaddresses 01H through 04H.

This is effective when transmitting data successively (refer to 3.2.2 Successive data transfer).

Figure 4-6. Automatic Increment



Caution Subaddress 00H does not have an automatic increment function. It is always set to ON.

The automatic increment function automatically increments the subaddress when data is transferred successively.

Automatic increment ON $\,:\,$ The subaddress is automatically incremented immediately after byte data with D6 = 1

has been transferred.

This setting is useful if the data at every subaddress is to be set at once for initialization. The subaddress is always incremented immediately after the data of

subaddress: 00H has been transferred.

Automatic increment OFF: The subaddress is fixed immediately after byte data with D6 = 0 has been

transferred. This setting is useful when the data at the same subaddress is to be

successively changed, for example to turn up/down the volume.

There is an automatic increment function ON/OFF bit in subaddresses 01H through 04H. Incrementing subaddresses is individually controlled by the automatic increment function ON/OFF bit of each subaddress.

For example, if the automatic increment function of subaddress 01H is turned ON, and that of subaddress 02H is turned OFF, the subaddress is automatically incremented from 01H to 02H, and is fixed to 02H.

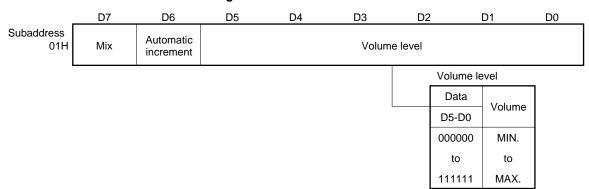
Even if the automatic increment function ON/OFF bit of subaddress 04H is set to ON, the subaddress is not incremented. If the next data is transferred after the data of 04H has been set (acknowledge bit: L), acknowledge enters the NAK status (acknowledge bit: H), and data transfer from the host CPU is stopped.



4.4.7 Volume level

The volume of output can be controlled in 64 steps by using the data of bits D5 through D0 of subaddress 01H.

Figure 4-7. Volume Level



4.4.8 Balance

The balance of output of the Lout and Rout pins can be controlled in 64 steps by using the data of bits D5 through D0 of subaddress 02H.

Figure 4-8. Balance

-	D7	D6	D5	D4	D3	D2		D1	D0	
Subaddress 02H	0	Automatic increment			Bala	ince				
							Balance			
				Data Volur						
							D5-D0	Lout	Rout	
							000000	MAX.	MIN.	
							to	to	to	

100000

to

111111

MAX.

to

MIN.

MAX.

to

MAX.

to

11111

to

Boost



4.4.9 Bass level

The bass level of output can be controlled in 32 steps by using the data of bits D4 through D0 of subaddress 03H.

Figure 4-9. Bass Level

	D7	D6	D5	D4	D3	D2	D1	D0				
Subaddress 03H	0	Automatic increment	0	Bass level								
							Bass leve	I				
							Data	Gain				
							D4-D0	Gaiii				
							00000	Cut				
							to	to				
							10000	0				

4.4.10 Treble level

The treble level of output can be controlled in 32 steps by using the data of bits D4 through D0 of subaddress 04H.

Figure 4-10. Treble Level

			•					
-	D7	D6	D5	D4	D3	D2	D1	D0
Subaddress 04H	0	Automatic increment	0			Treble level		
							Treble lev	rel
							Data	Onin
							D4-D0	Gain
							00000	Cut
							to	to
							10000	0
							to	to
							11111	Boost



5. ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings (Unless otherwise specified, T_A = 25 °C)

Parameter	Symbol	Condition	Rating	Unit
Supply voltage	Vcc	Without signal	14.0	V
Input signal voltage	Vin	Pins Lin, Rin, MIX	Vcc	V
I ² C bus input signal voltage	Vcnt	Pins SDA, SCL	Vcc + 0.2	V
Permissible package dissipation	Po	T _A = 75 °C	500	mW
Operating temperature	TA	Vcc = 12 V	−20 to +75	°C
Storage temperature	T _{stg}		-40 to +125	°C

Caution If any of the parameters exceeds the absolute maximum ratings, even momentarily, the quality of the product may be impaired. The absolute maximum ratings are values that may physically damage the product(s). Be sure to use the product(s) within the ratings.

Recommended Operating Conditions (Unless otherwise specified, T_A = 25 °C)

Damanatan	O. mah al	O and Million			1.1-26	
Parameter Symbol		Condition	MIN.	TYP.	MAX.	Unit
Supply voltage	Vcc	Gain between input and output: 0 dB	8.1	12.0	13.2	V
Input signal voltage	Vin	Vcc = 12 V, gain between input and output: 0 dB	0.0	1.4	7.9	V _{p-p}
I ² C bus input voltage (H)	VcntH	Pins SDA, SCL	3.5	5.0	6.0	٧
I ² C bus input voltage (L)	VcntL		-0.1	0	+1.5	٧



Electrical Characteristics (1/5)

(Unless otherwise specified, Vcc = 12 V, T_A = 25 °C, $RH \le 70\%$, f = 1 kHz, V_{IN} = 0.5 $V_{r.m.s.}$, no load)

D	O. male al	Total Constitions		Suba	ddress	Data			Rating		11-2
Parameter	Symbol	Test Condition	00	01	02	03	04	MIN.	TYP.	MAX.	Unit
Circuit current	Icc	No signal	0D	3F	20	10	10	12	18	25	mA
$\begin{array}{c} \text{Maximum input voltage} \\ \text{Lin} \rightarrow \text{Lout} \end{array}$	VOM-L	Lin = variable (tested) Lout = THD 1%	0D	3F	20	10	10	2.8	3.1	-	Vr.m.s.
$\begin{array}{c} \text{Maximum input voltage} \\ \text{Rin} \rightarrow \text{Rout} \end{array}$	VOM-R	Rin = variable (tested) Rout = THD 1%						2.8	3.1	-	Vr.m.s.
Distortion rate Lin → Lout	THDL	Lin = 2.0 V _{r.m.s.} Rin = GND	0D	2B	20	10	10	_	0.1	0.5	%
Distortion rate $Rin \rightarrow Rout$	THDR	Lin = GND Rin = 2.0 V _{r.m.s.}						_	0.1	0.5	%
Voltage gain Lin → Lout	GV-LL	Lin = 0.5 V _{r.m.s.} Rin = GND	0D	3F	20	10	10	-1.0	0	+1.0	dB
Voltage gain Rin → Rout	GV-RR	Lin = GND Rin = 0.5 V _{r.m.s.}						-1.0	0	+1.0	dB
Voltage gain MIX → Lout	GV-ML	MIX = 0.5 V _{r.m.s.} Lin, Rin = GND	0D	BF	20	10	10	5.0	6.0	7.0	dB
Voltage gain MIX → Rout	GV-MR							5.0	6.0	7.0	dB
Ripple rejection ratio $V_{\text{CC}} \rightarrow \text{Lout}$	SVRR-L	$V_{cc} = 100 \text{ mV}_{r.m.s.}$ f = 100 Hz	0D	14	20	10	10	_	_	-50	dB
Ripple rejection ratio $V_{\text{CC}} \rightarrow \text{Rout}$	SVRR-R	Lin, Rin = GND						_	_	-50	dB
Output noise voltage (surround OFF) Lout	Vn-L (OFF)	Rg = 0 Ω , JIS-A	0D	3F	20	10	10	-	-	50	μVr.m.s.
Output noise voltage (surround OFF) Rout	Vn-R (OFF)							-	_	50	μVr.m.s.
Cross talk Lin → Rout	CT-L	Lin = 0.5 V _{r.m.s.} Rin = GND	0D	3F	20	10	10	-	-80	-70	dB
Cross talk Rin → Lout	CT-R	Lin = GND Rin = 0.5 V _{r.m.s.}						_	-80	-70	dB



Electrical Characteristics (2/5)

(Unless otherwise specified, Vcc = 12 V, T_A = 25 °C, $RH \le 70\%$, f = 1 kHz, V_{IN} = 0.5 $V_{r.m.s.}$, no load)

5 .		T . O . IV		Suba	ddress	Data			Rating		11.7
Parameter	Symbol	Test Condition	00	01	02	03	04	MIN.	TYP.	MAX.	Unit
Volume attenuation 1 Lin → Lout1	VOL-L1	Lin = 0.5 V _{r.m.s.} Rin = GND	0D	3F	20	10	10	-1.5	0	+1.5	dB
Volume attenuation 2 Lin → Lout2	VOL-L2			20				-20.0	-14.0	-7.0	dB
Volume attenuation 3 Lin → Lout3	VOL-L3			00				-	-	-80.0	dB
Volume attenuation deviation Rin \rightarrow R/Lout1	VOL-RL1	Lin = GND Rin = 0.5 V _{r.m.s.} Difference from VOL-L1	0D	3F	20	10	10	-1.0	0	+1.0	dB
Volume attenuation deviation Rin \rightarrow R/ Lout2	VOL-RL2	Lin = GND Rin = 0.5 V _{r.m.s.} Difference from VOL-L2		20				-1.0	0	+1.0	dB
Volume attenuation deviation Rin → R/ Lout3	VOL-RL3	Lin = GND Rin = 0.5 V _{r.m.s.} Difference from VOL-L3		00				-3.0	0	+3.0	dB
Mute attenuation Lin → Lout	MUTE-L	Lin = 2.0 V _{r.m.s.} Rin = GND	8D	3F	20	10	10	-	-	-80.0	dB
Mute attenuation Rin → Rout	MUTE-R	Lin = GND Rin = 2.0 V _{r.m.s.}						_	-	-80.0	dB
Balance attenuation L1 Lin \rightarrow Lout1	BAL-L1	Lin = 0.5 V _{r.m.s.} Rin = GND	0D	3F	01	10	10	-1.5	0	+1.5	dB
Balance attenuation L2 Lin \rightarrow Lout2	BAL-L2				28			-2.5	-0.5	+1.0	dB
Balance attenuation L3 Lin → Lout3	BAL-L3				30			-15.0	-10.0	-5.0	dB
Balance attenuation L4 $Lin \rightarrow Lout4$	BAL-L4				3F			_	-	-80.0	dB
Balance attenuation R1 Rin → Rout1	BAL-R1	Lin = GND Rin = 0.5 V _{r.m.s.}	0D	3F	3F	10	10	-1.5	0	+1.5	dB
Balance attenuation R2 Rin → Rout2	BAL-R2				18			-2.5	-0.5	+1.0	dB
Balance attenuation R3 Rin → Rout3	BAL-R3				10			-15.0	-10.0	-5.0	dB
Balance attenuation R4 Rin → Rout4	BAL-R4				01			_	_	-80.0	dB



Electrical Characteristics (3/5)

(Unless otherwise specified, Vcc = 12 V, T_A = 25 °C, $RH \le 70\%$, f = 1 kHz, V_{IN} = 0.5 $V_{r.m.s.}$, no load)

				Suba	ddress	Data			Rating		
Parameter	Symbol	Test Condition	00	01	02	03	04	MIN.	TYP.	MAX.	Unit
Tone control, bass characteristic Lin → Lout1	BASS-L1	$f = 100 \text{ Hz}$ $Lin = 0.5 \text{ V}_{r.m.s.}$ $Rin = GND$	0D	3F	20	1F	10	7.0	10.0	13.0	dB
Tone control, bass characteristic Lin → Lout2	BASS-L2					10		-2.0	0	+2.0	dB
Tone control, bass characteristic Lin → Lout3	BASS-L3					01		-13.0	-10.0	-7.0	dB
Tone control, bass characteristic deviation Rin → Rout1/Lout1	BASS-RL1	f = 100 Hz, Lin = GND Rin = 0.5 V _{r.m.s.} Difference from BASS-L1	0D	3F	20	1F	10	-1.0	0	+1.0	dB
Tone control, bass characteristic deviation Rin → Rout2/Lout2	BASS-RL2	$f = 100 \text{ Hz}, \text{ Lin} = \text{GND}$ $Rin = 0.5 \text{ V}_{r.m.s.}$ $Difference \text{ from BASS-L2}$				10		-1.0	0	+1.0	dB
Tone control, bass characteristic deviation Rin → Rout3/Lout3	BASS-RL3	$f = 100 \text{ Hz}, \text{ Lin} = \text{GND}$ $Rin = 0.5 \text{ V}_{r.m.s.}$ $Difference \text{ from BASS-L3}$				01		-1.0	0	+1.0	dB
Tone control, treble characteristic Lin → Lout1	TREB-L1	$f = 10 \text{ kHz}$ $Lin = 0.5 \text{ V}_{r.m.s.}$ $Rin = GND$	0D	3F	20	10	1F	7.0	10.0	13.0	dB
Tone control, treble characteristic Lin → Lout2	TREB-L2						10	-2.0	0	+2.0	dB
Tone control, treble characteristic Lin → Lout3	TREB-L3						01	-13.0	-10.0	-7.0	dB
Tone control, treble characteristic deviation Rin → Rout1/Lout1	TREB-RL1	f = 10 kHz, Lin = GND Rin = 0.5 V _{r.m.s.} Difference from TREB-L1	0D	3F	20	10	1F	-1.0	0	+1.0	dB
Tone control, treble characteristic deviation Rin → Rout2/Lout2	TREB-RL2	f = 10 kHz, Lin = GND Rin = 0.5 V _{r.m.s.} Difference from TREB-L2					10	-1.0	0	+1.0	dB
Tone control, treble characteristic deviation Rin → Rout3/Lout3	TREB-RL3	$f = 10 \text{ kHz, Lin} = \text{GND}$ $Rin = 0.5 \text{ V}_{\text{r.m.s.}}$ $Difference \text{ from TREB-L3}$					01	-1.0	0	+1.0	dB



Electrical Characteristics (4/5)

(Unless otherwise specified, Vcc = 12 V, T_A = 25 °C, $RH \le 70\%$, f = 1 kHz, V_{IN} = 0.5 $V_{r.m.s.}$, no load)

Damanatan	Overhal Tool Over	Test Ossalition	Subaddress Data				Rating				
Parameter	Symbol	Test Condition	00	01	02	03	04	MIN.	TYP.	MAX.	Unit
Surround voltage gain, music mode Lin → Lout	MUS-L	f = 1 kHz $Lin = 0.5 Vr.m.s.$ $Rin = GND$	05	3F	20	10	10	3.5	5.5	7.5	dB
Surround voltage gain, music mode $\label{eq:surround} \mbox{Lin} \rightarrow \mbox{Rout}$	MUS-R							-2.5	-0.5	+1.5	dB
Surround voltage gain, movie mode Lin → Lout	MOV-L	$f = 1 \text{ kHz}$ $Lin = 0.5 \text{ V}_{r.m.s.}$ $Rin = GND$	09	3F	20	10	10	3.0	7.0	11.0	dB
Surround voltage gain, movie mode $\label{eq:surround} \mbox{Lin} \rightarrow \mbox{Rout}$	MOV-R							0	4.0	8.0	dB
Surround voltage gain, simulated mode LRin → Lout1	SIM-L1	f = 250 Hz Lin = 0.5 V _{r.m.s.} Rin = 0.5 V _{r.m.s.}	01	3F	20	10	10	-0.5	+3.5	+6.5	dB
Surround voltage gain, simulated mode LRin \rightarrow Lout2	SIM-L2	f = 1 kHz $Lin = 0.5 Vr.m.s.$ $Rin = 0.5 Vr.m.s.$						ı	-3.0	+4.5	dB
Surround voltage gain, simulated mode LRin \rightarrow Lout3	SIM-L3	$f = 4 \text{ kHz}$ $Lin = 0.5 \text{ V}_{r.m.s.}$ $Rin = 0.5 \text{ V}_{r.m.s.}$						2.0	6.0	10.0	dB
Surround voltage gain, simulated mode LRin \rightarrow Rout1	SIM-R1	f = 250 Hz Lin = 0.5 V _{r.m.s.} Rin = 0.5 V _{r.m.s.}	01	3F	20	10	10	1	-5.5	-1.0	dB
Surround voltage gain, simulated mode LRin \rightarrow Rout2	SIM-R2	f = 1 kHz $Lin = 0.5 Vr.m.s.$ $Rin = 0.5 Vr.m.s.$						0	3.0	6.0	dB
Surround voltage gain, simulated mode LRin \rightarrow Rout3	SIM-R3	f = 4 kHz $Lin = 0.5 Vr.m.s.$ $Rin = 0.5 Vr.m.s.$						-	-7.0	+5.0	dB

Remark For the surround mode, refer to **4.3 Surround Function**.



Electrical Characteristics (5/5)

(Unless otherwise specified, Vcc = 12 V, T_A = 25 °C, $RH \le 70\%$, f = 1 kHz, V_{IN} = 0.5 $V_{r.m.s.}$, no load)

Parameter Symb	Cymah al	ol Test Condition	Subaddress Data				Rating			Unit	
Parameter	Symbol		00	01	02	03	04	MIN.	TYP.	MAX.	Unit
Output selector, DC offset Lin → Lout	OFST LRL	No signal Voltage conversion of Lout Lout: L output \rightarrow R output	1D ↓ 2D	3F	20	10	10	-100	0	+100	mV
Output selector, DC offset Lin → Lout	OFST LL + RL	No signal Voltage conversion of Lout Lout: L output → L+R output	1D ↓ 3D					-100	0	+100	mV
Output selector, DC offset Rin → Lout	OFST RL + RL	No signal Voltage conversion of Lout Lout: R output → L+R output	2D ↓ 3D					-100	0	+100	mV
Output selector, DC offset Rin → Rout	OFST RLR	No signal Voltage conversion of Rout Rout: R output → L output	2D ↓ 1D	3F	20	10	10	-100	0	+100	mV
Output selector, DC offset $Rin \rightarrow Rout$	OFST RL + RR	No signal Voltage conversion of Rout Rout: R output → L+R output	2D ↓ 3D					-100	0	+100	mV
Output selector, DC offset Lin → Rout	OFST LL + RR	No signal Voltage conversion of Rout Rout: L output → L+R output	1D ↓ 3D					-100	0	+100	mV



6. CHARACTERISTIC CURVES

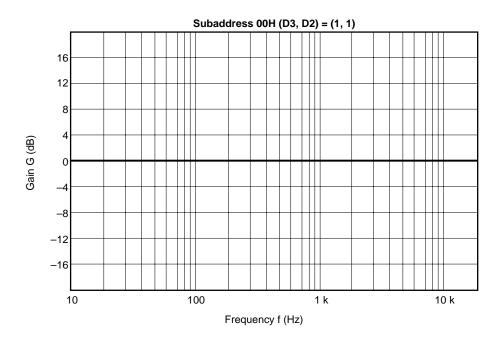
6.1 Frequency Characteristic in Each Mode

Vcc = 12 V, Vin = 0.5 Vr.m.s.

Stereo mode: subaddress 00H (D5, D4) = (0,0)

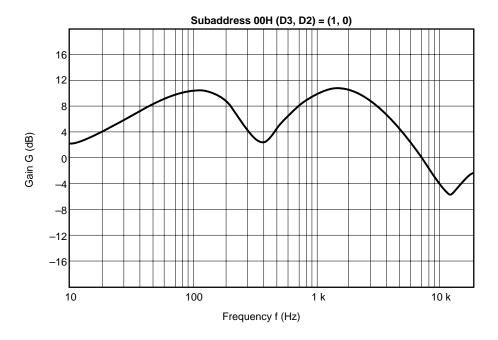
Surround effect (0 dB attenuation): subaddress 00H (D1, D0) = (0,0)

(1) OFF mode Lch/Rch

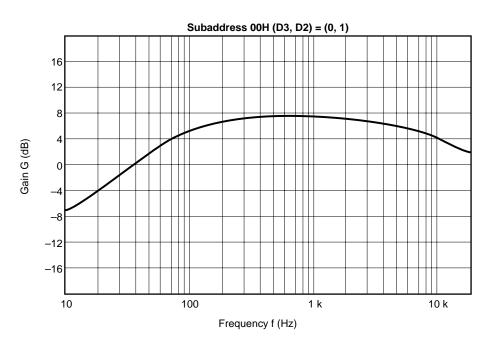




(2) Movie mode Lch/Rch

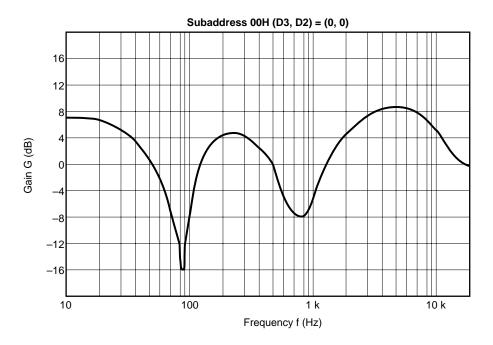


(3) Music mode Lch/Rch

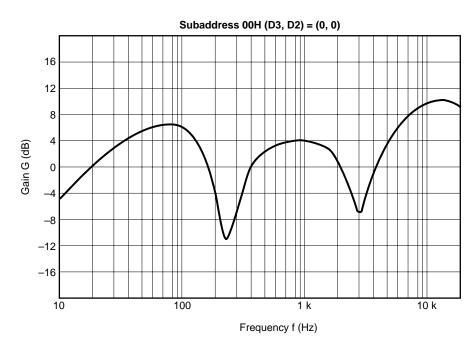




(4) Simulated mode Lch



(5) Simulated mode Rch

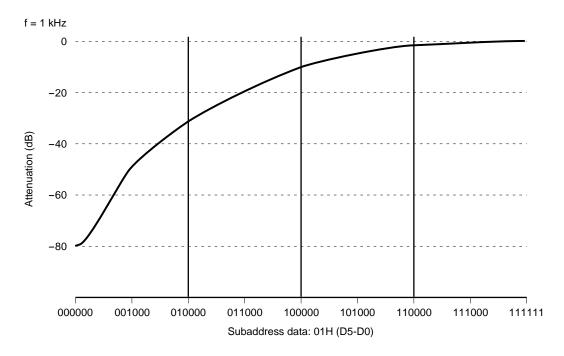




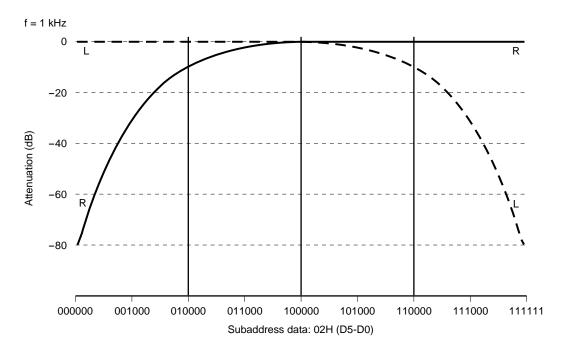
6.2 Control Characteristic

Vcc = 12 V, VIN = 0.5 Vr.m.s Surround mode (OFF): subaddress 00H (D3, D2) = (1, 1)

(1) Volume control characteristic



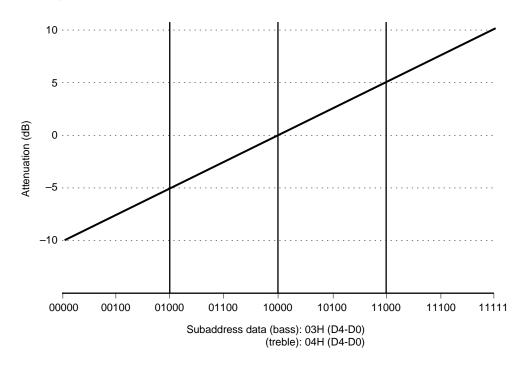
(2) Balance control characteristic



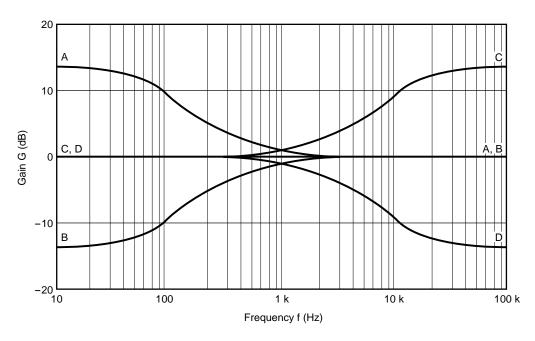


(3) Tone control characteristic (bass/treble)

Bass: f = 100 Hz, treble: f = 10 kHz



(4) Tone frequency characteristic

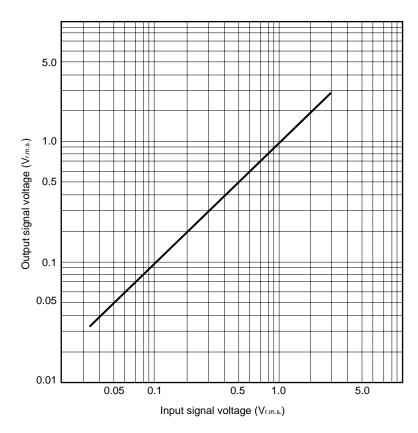


Curve	Subaddress	Data (D4-D0)
Α	0011	11111
В	03H	00001
С	0.411	11111
D	04H	00001



6.3 I/O Characteristic

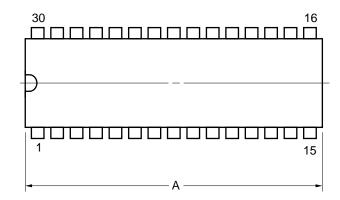
Vcc = 12 V

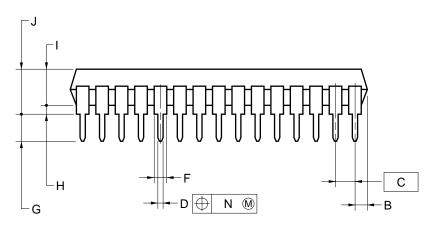


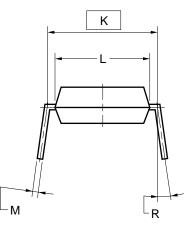


7. PACKAGE DRAWING

30 PIN PLASTIC SHRINK DIP (400 mil)







NOTES

- 1. Controlling dimension millimeter.
- 2. Each lead centerline is located within 0.17 mm (0.007 inch) of its true position (T.P.) at maximum material condition.
- 3. Item "K" to center of leads when formed parallel.

ITEM	MILLIMETERS	INCHES
Α	27.3±0.2	$1.075^{+0.008}_{-0.009}$
В	1.78 MAX.	0.070 MAX.
С	1.778 (T.P.)	0.070 (T.P.)
D	0.50±0.10	$0.020^{+0.004}_{-0.005}$
F	1.0±0.15	0.039+0.007
G	3.2±0.3	0.126±0.012
Н	0.51 MIN.	0.020 MIN.
I	3.45±0.2	$0.136^{+0.008}_{-0.009}$
J	5.08 MAX.	0.200 MAX.
K	10.16 (T.P.)	0.400 (T.P.)
L	8.6±0.2	$0.339^{+0.008}_{-0.009}$
М	0.25 ^{+0.10} -0.05	0.010+0.004
N	0.17	0.007
R	0~15°	0~15°

S30C-70-400B-2



8. RECOMMENDED SOLDERING CONDITIONS

It is recommended to solder this product under the conditions described below.

For details of the recommended soldering conditions, refer to the **Semiconductor Device Mounting Technology Manual (C10535E).**

For soldering methods and conditions other than those recommended, consult NEC.

Soldering condition of through-hole type

 μ PC1857ACT: 30-pin plastic shrink DIP (400 mil)

Soldering Method	Soldering Condition				
Wave soldering (only pins)	Soldering bath temperature: 260 °C MAX., Time: 10 seconds				
Partial heating	Pin temperature: 300 °C MAX., Time: 3 seconds MAX. (per pin)				

Caution Apply wave soldering only to the pins, and exercise care that solder does not directly contact the package.

[MEMO]

[MEMO]

[MEMO]



Purchase of NEC I²C components conveys a license under the Philips I²C Patent Rights to use these components in an I²C system, provided that the system conforms to the I²C Standard Specification as defined by Philips.

EEPROM is a trademark of NEC Corp.

The application circuits and their parameters are for reference only and are not intended for use in actual design-ins.

No part of this document may be copied or reproduced in any form or by any means without the prior written consent of NEC Corporation. NEC Corporation assumes no responsibility for any errors which may appear in this document.

NEC Corporation does not assume any liability for infringement of patents, copyrights or other intellectual property rights of third parties by or arising from use of a device described herein or any other liability arising from use of such device. No license, either express, implied or otherwise, is granted under any patents, copyrights or other intellectual property rights of NEC Corporation or others.

While NEC Corporation has been making continuous effort to enhance the reliability of its semiconductor devices, the possibility of defects cannot be eliminated entirely. To minimize risks of damage or injury to persons or property arising from a defect in an NEC semiconductor device, customers must incorporate sufficient safety measures in its design, such as redundancy, fire-containment, and anti-failure features.

NEC devices are classified into the following three quality grades:

"Standard", "Special", and "Specific". The Specific quality grade applies only to devices developed based on a customer designated "quality assurance program" for a specific application. The recommended applications of a device depend on its quality grade, as indicated below. Customers must check the quality grade of each device before using it in a particular application.

Standard: Computers, office equipment, communications equipment, test and measurement equipment, audio and visual equipment, home electronic appliances, machine tools, personal electronic equipment and industrial robots

Special: Transportation equipment (automobiles, trains, ships, etc.), traffic control systems, anti-disaster systems, anti-crime systems, safety equipment and medical equipment (not specifically designed for life support)

Specific: Aircrafts, aerospace equipment, submersible repeaters, nuclear reactor control systems, life support systems or medical equipment for life support, etc.

The quality grade of NEC devices is "Standard" unless otherwise specified in NEC's Data Sheets or Data Books. If customers intend to use NEC devices for applications other than those specified for Standard quality grade, they should contact an NEC sales representative in advance.

Anti-radioactive design is not implemented in this product.

M4 96.5