

## Wideband Two Quadrant Analog Multiplier (Voltage Output)

July 1994

### Features

- This Circuit is Processed in Accordance to MIL-STD-883 and is Fully Conformant Under the Provisions of Paragraph 1.2.1.
- High Speed Voltage Output. . . . . 300V/μs (Min)
- Low Multiplication Error . . . . . 3.0% (Max)  
1.6% (Typ)
- Input Bias Currents . . . . . 5μA (Max)  
1.2μA (Typ)
- Signal Input Feedthrough . . . . . -52dB (Typ)
- Wide Signal Bandwidth . . . . . 30MHz (Typ)
- Wide Control Bandwidth . . . . . 17MHz (Typ)
- Gain Flatness to 5MHz. . . . . 0.10dB (Typ)

### Applications

- Military Avionics
- Missile Guidance Systems
- Medical Imaging Displays
- Video Mixers
- Sonar AGC Processors
- Radar Signal Conditioning
- Voltage Controlled Amplifier
- Vector Generator

### Description

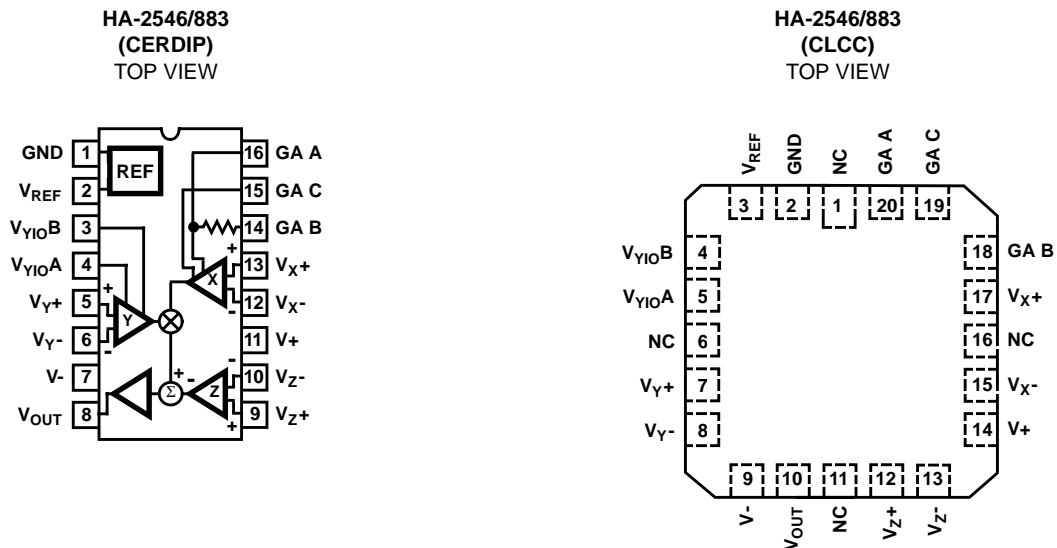
The HA-2546/883 is a monolithic, high speed, two quadrant, analog multiplier constructed in the Intersil Dielectrically Isolated High Frequency Process. The HA-2546/883 has a voltage output with a 30MHz signal bandwidth, 300V/μs slew rate and a 17MHz control input bandwidth. High bandwidth and slew rate make this part an ideal component for use in video systems. The suitability for precision video applications is demonstrated further by the 0.1dB gain flatness at 5MHz, 1.6% multiplication error, -52dB feedthrough and differential inputs with 1.2μA bias currents. The HA-2546/883 also has low differential gain (0.1% typ.) and phase (0.1° typ.) errors.

The HA-2546/883 is well suited for AGC circuits as well as mixer applications for sonar, radar, and medical imaging equipment. The voltage output of the HA-2546/883 simplifies many designs by eliminating the current-to-voltage conversion stage required for current output multipliers.

### Ordering Information

PART NUMBER	TEMPERATURE RANGE	PACKAGE
HA1-2546/883	-55°C to +125°C	16 Lead CerDIP
HA4-2546/883	-55°C to +125°C	20 Lead Ceramic LCC

### Pinouts



# Specifications HA2546/883

## Absolute Maximum Ratings

Voltage Between V+ and V-	35V
Differential Input Voltage	6V
Output Current	.60mA
Junction Temperature	+175°C
Storage Temperature Range	-65°C ≤ T <sub>A</sub> ≤ +150°C
ESD Rating	<2000V
Lead Temperature (Soldering 10s)	+300°C

## Thermal Information

Thermal Resistance	θ <sub>JA</sub>	θ <sub>JC</sub>
CerDIP Package	80°C/W	25°C/W
Ceramic LCC	61°C/W	12°C/W
Maximum Package Power Dissipation		
CerDIP Package at +75°C	1.25W	
Ceramic LCC Package at +75°C	1.64W	
Package Power Dissipation Derating Factor above +75°C		
CerDIP Package	12mW/°C	
Ceramic LCC Package	16mW/°C	

*CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.*

## Operating Conditions

Operating Temperature Range . . . . . -55°C to +125°C    Operating Supply Voltage . . . . . ±8V to ±15V

**TABLE 1. DC ELECTRICAL PERFORMANCE CHARACTERISTICS**

Device Tested at: V<sub>SUPPLY</sub> = ±15V, R<sub>LOAD</sub> = 1kΩ, C<sub>LOAD</sub> = 50pF, Unless Otherwise Specified.

PARAMETERS	SYMBOL	CONDITIONS	GROUP A SUBGROUP PS	TEMPERATURE	LIMITS		UNITS
					MIN	MAX	
Multiplication Error	ME	V <sub>Y</sub> = ±5V	1	+25°C	-3	3	%FS
			2, 3	+125°C, -55°C	-5	5	%FS
Scale Factor Error	SF		1	+25°C	-5	5	%
			2, 3	+125°C, -55°C	-5	5	%
Common Mode Range	+CMR		1	+25°C	5	-	V
			2, 3	+125°C, -55°C	5	-	V
	-CMR		1	+25°C	-	-5	V
			2, 3	+125°C, -55°C	-	-5	V
Input Offset Voltage (V <sub>Y</sub> )	V <sub>IO</sub> (V <sub>Y</sub> )	V <sub>CM</sub> = 0V	1	+25°C	-10	10	mV
			2, 3	+125°C, -55°C	-15	15	mV
Input Bias Current (V <sub>Y</sub> )	I <sub>B</sub> (V <sub>Y</sub> )	V <sub>CM</sub> = 0V	1	+25°C	-15	15	μA
			2, 3	+125°C, -55°C	-20	20	μA
Input Offset Current (V <sub>Y</sub> )	I <sub>IO</sub> (V <sub>Y</sub> )	V <sub>CM</sub> = 0V	1	+25°C	-2	2	μA
			2, 3	+125°C, -55°C	-3	3	μA
Common Mode (V <sub>Y</sub> ) Rejection Ratio	+CMRR(V <sub>Y</sub> )	V <sub>Y</sub> = 0 to +5V, V <sub>X</sub> = +2V	1	+25°C	60	-	dB
			2, 3	+125°C, -55°C	60	-	dB
	-CMRR(V <sub>Y</sub> )	V <sub>Y</sub> = 0 to -5V, V <sub>X</sub> = +2V	1	+25°C	60	-	dB
			2, 3	+125°C, -55°C	60	-	dB
Input Offset Voltage (V <sub>X</sub> )	V <sub>IO</sub> (V <sub>X</sub> )	V <sub>CM</sub> = 0V	1	+25°C	-2	2	mV
			2, 3	+125°C, -55°C	-15	15	mV
Input Bias Current (V <sub>X</sub> )	I <sub>B</sub> (V <sub>X</sub> )	V <sub>CM</sub> = 0V	1	+25°C	-2	2	μA
			2, 3	+125°C, -55°C	-5	5	μA
Input Offset Current (V <sub>X</sub> )	I <sub>IO</sub> (V <sub>X</sub> )	V <sub>CM</sub> = 0V	1	+25°C	-2	2	μA
			2, 3	+125°C, -55°C	-3	3	μA
Input Offset Voltage (V <sub>Z</sub> )	V <sub>IO</sub> (V <sub>Z</sub> )	V <sub>X</sub> = 0V, V <sub>Y</sub> = 0V	1	+25°C	-15	15	mV
			2, 3	+125°C, -55°C	-15	15	mV
Output Voltage Swing	+V <sub>OUT</sub>	V <sub>Y</sub> = +5V, V <sub>X</sub> = +2.5V	1	+25°C	5	-	V
			2, 3	+125°C, -55°C	5	-	V
	-V <sub>OUT</sub>	V <sub>Y</sub> = -5V, V <sub>X</sub> = +2.5V	1	+25°C	-	-5	V
			2, 3	+125°C, -55°C	-	-5	V
Output Current	+I <sub>OUT</sub>	V <sub>Y</sub> = +5V, V <sub>X</sub> = +2.5V	1	+25°C	20	-	mA
			2, 3	+125°C, -55°C	20	-	mA
	-I <sub>OUT</sub>	V <sub>Y</sub> = -5V, V <sub>X</sub> = +2.5V	1	+25°C	-	-20	mA
			2, 3	+125°C, -55°C	-	-20	mA

# Specifications HA2546/883

**TABLE 1. DC ELECTRICAL PERFORMANCE CHARACTERISTICS (Continued)**

Device Tested at:  $V_{SUPPLY} = \pm 15V$ ,  $R_{LOAD} = 1k\Omega$ ,  $C_{LOAD} = 50pF$ , Unless Otherwise Specified.

PARAMETERS	SYMBOL	CONDITIONS	GROUP A SUBGROU PS	TEMPERATURE	LIMITS		UNITS
					MIN	MAX	
Power Supply Rejection Ratio	+PSRR	$\Delta V_S = 3V$ , $V_+ = +15V$ , $V_- = -15V$ , $V_+ = +12V$ , $V_- = -15V$	1	+25°C	58	-	dB
			2, 3	+125°C, -55°C	58	-	dB
	-PSRR	$\Delta V_S = 3V$ , $V_+ = +15V$ , $V_- = -15V$ , $V_+ = +15V$ , $V_- = -12V$	1	+25°C	58	-	dB
			2, 3	+125°C, -55°C	58	-	dB
Quiescent Power Supply Current	+I <sub>CC</sub>	$V_X = V_Y = 0V$ , $I_{OUT} = 0mA$	1	+25°C	29	-	mA
			2, 3	+125°C, -55°C	29	-	mA
	-I <sub>CC</sub>	$V_X = V_Y = 0V$ , $I_{OUT} = 0mA$	1	+25°C	-	-29	mA
			2, 3	+125°C, -55°C	-	-29	mA

**TABLE 2. AC ELECTRICAL PERFORMANCE CHARACTERISTICS**

Table 2 Intentionally Left Blank. See AC Specifications in Table 3.

**TABLE 3. ELECTRICAL PERFORMANCE CHARACTERISTICS**

Device Tested at:  $V_{SUPPLY} = \pm 15V$ ,  $R_{LOAD} = 1k\Omega$ ,  $C_{LOAD} = 50pF$ , Unless Otherwise Specified.

PARAMETER	SYMBOL	CONDITIONS	NOTES	TEMPERATURE	LIMITS		UNITS
					MIN	MAX	
Slew Rate	+SR	$V_{OUT} = -5V$ to $+5V$ , $V_X = 2V_{DC}$	1	+25°C	300	-	V/ $\mu$ s
			1	+125°C, -55°C	300	-	V/ $\mu$ s
	-SR	$V_{OUT} = +5V$ to $-5V$ , $V_X = 2V_{DC}$	1	+25°C	300	-	V/ $\mu$ s
			1	+125°C, -55°C	300	-	V/ $\mu$ s
Rise and Fall Time	TR	$V_{OUT} = -100mV$ to $+100mV$ $V_X = 2V_{DC}$	1, 3	+25°C	-	15	ns
			1, 3	+125°C, -55°C	-	17	ns
	TF	$V_{OUT} = +100mV$ to $-100mV$ $V_X = 2V_{DC}$	1, 3	+25°C	-	15	ns
			1, 3	+125°C, -55°C	-	17	ns
Overshoot	+OS	$V_{OUT} = -100mV$ to $+100mV$ $V_X = 2V_{DC}$	1	+25°C	-	30	%
			1	+125°C, -55°C	-	30	%
	-OS	$V_{OUT} = +100mV$ to $-100mV$ $V_X = 2V_{DC}$	1	+25°C	-	30	%
			1	+125°C, -55°C	-	30	%
Full Power Bandwidth	FPBW	$V_{PEAK} = 5V$ , $V_X = 2V_{DC}$	1, 2	+25°C	9.5	-	MHz
			1, 2	+125°C, -55°C	9.5	-	MHz

NOTES:

- Parameters listed in Table 3 are controlled via design or process parameters and are not directly tested at final production. These parameters are lab characterized upon initial design release, or upon design changes. These parameters are guaranteed by characterization based upon data from multiple production runs which reflect lot to lot and within lot variation.
- Full Power Bandwidth guarantee based on Slew Rate measurement using  $FPBW = \text{Slew Rate}/(2\pi V_{PEAK})$ .
- Measured between 10% and 90% points.

**TABLE 4. ELECTRICAL TEST REQUIREMENTS**

MIL-STD-883 TEST REQUIREMENTS	SUBGROUPS (SEE TABLE 1)
Interim Electrical Parameters (Pre Burn-in)	1
Final Electrical Test Parameters	1(Note 1), 2, 3
Group A Test Requirements	1, 2, 3
Groups C and D Endpoints	1

NOTE:

- PDA applies to Subgroup 1 only.

**Die Characteristics**

**DIE DIMENSIONS:**

79.9mils x 119.7mils x 19mils ± 1mils

**METALLIZATION:**

Type: Al, 1%Cu  
 Thickness: 16kÅ ± 2kÅ

**GLASSIVATION:**

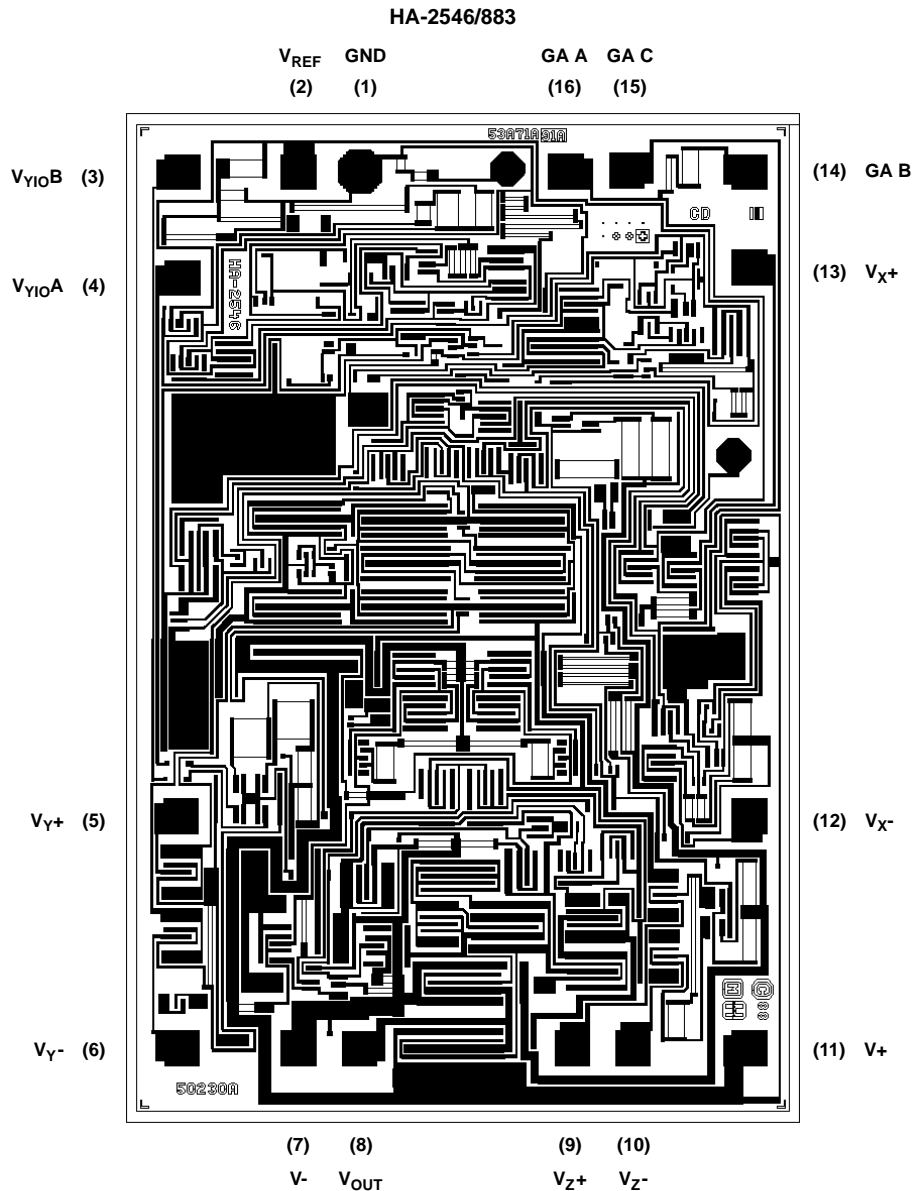
Type: Nitride (Si<sub>3</sub>N<sub>4</sub>) over Silox (SiO<sub>2</sub>, 5% Phos)  
 Silox Thickness: 12kÅ ± 1.5kÅ  
 Nitride Thickness: 3.5kÅ ± 1.5kÅ

**WORST CASE CURRENT DENSITY:**

0.72 x 10<sup>5</sup> A/cm<sup>2</sup>

TRANSISTOR COUNT: 87

**Metallization Mask Layout**

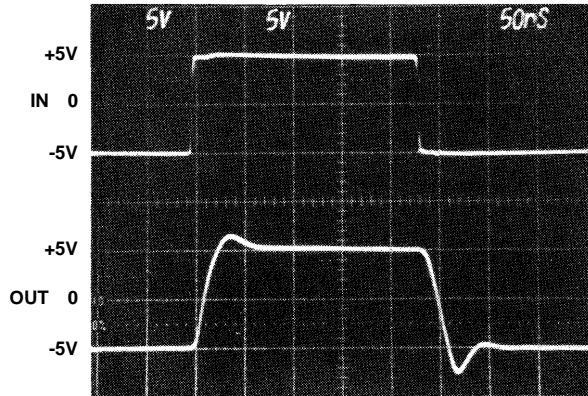




**Test Waveforms (Continued)**

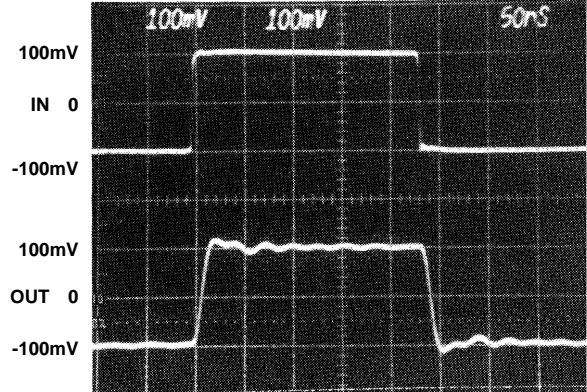
**V<sub>Y</sub> LARGE SIGNAL RESPONSE**

Vertical Scale: 5V/Div. Horizontal Scale: 50ns/Div.



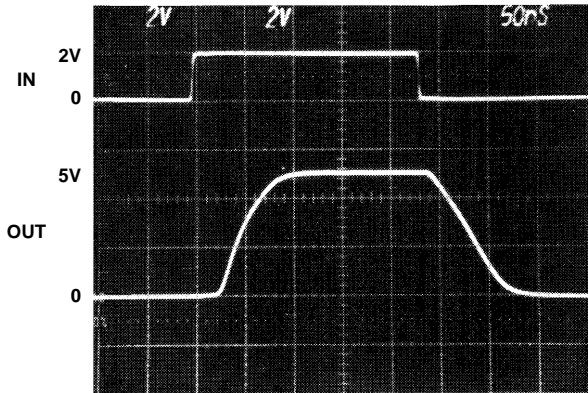
**V<sub>Y</sub> SMALL SIGNAL RESPONSE**

Vertical Scale: 100mV/Div. Horizontal Scale: 50ns/Div.



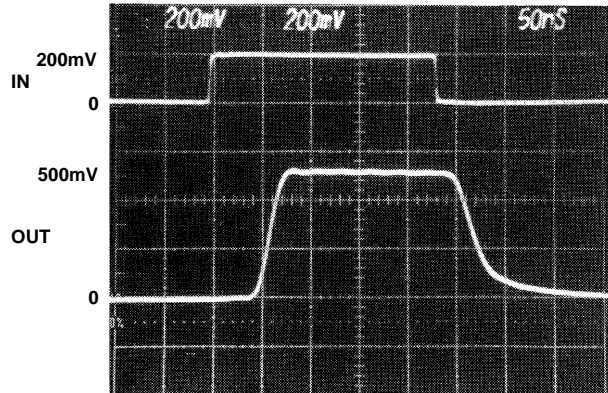
**V<sub>X</sub> LARGE SIGNAL RESPONSE**

Vertical Scale: 2V/Div. Horizontal Scale: 50ns/Div.



**V<sub>X</sub> SMALL SIGNAL RESPONSE**

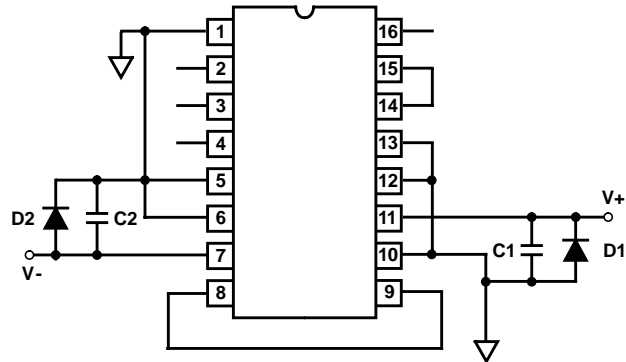
Vertical Scale: 200mV/Div. Horizontal Scale: 50ns/Div.



# HA2546/883

## Burn-In Circuits

HA-2546/883 CERDIP



NOTES:

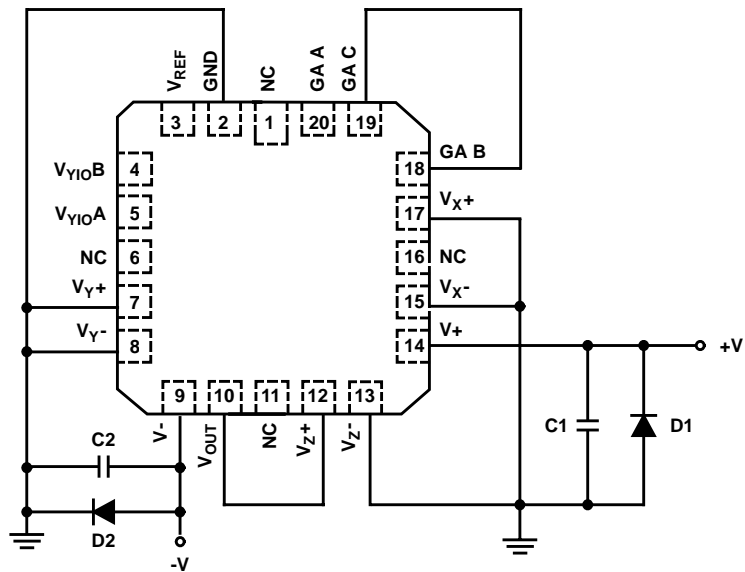
C1 = 0.01μF/Socket Min.

C2 = 0.01μF/Socket Min.

D1 = D2 = IN4002 or Equivalent/Board

| (V+) - (V-) | = 31V ± 1V

HA-2546/883 CERAMIC LCC



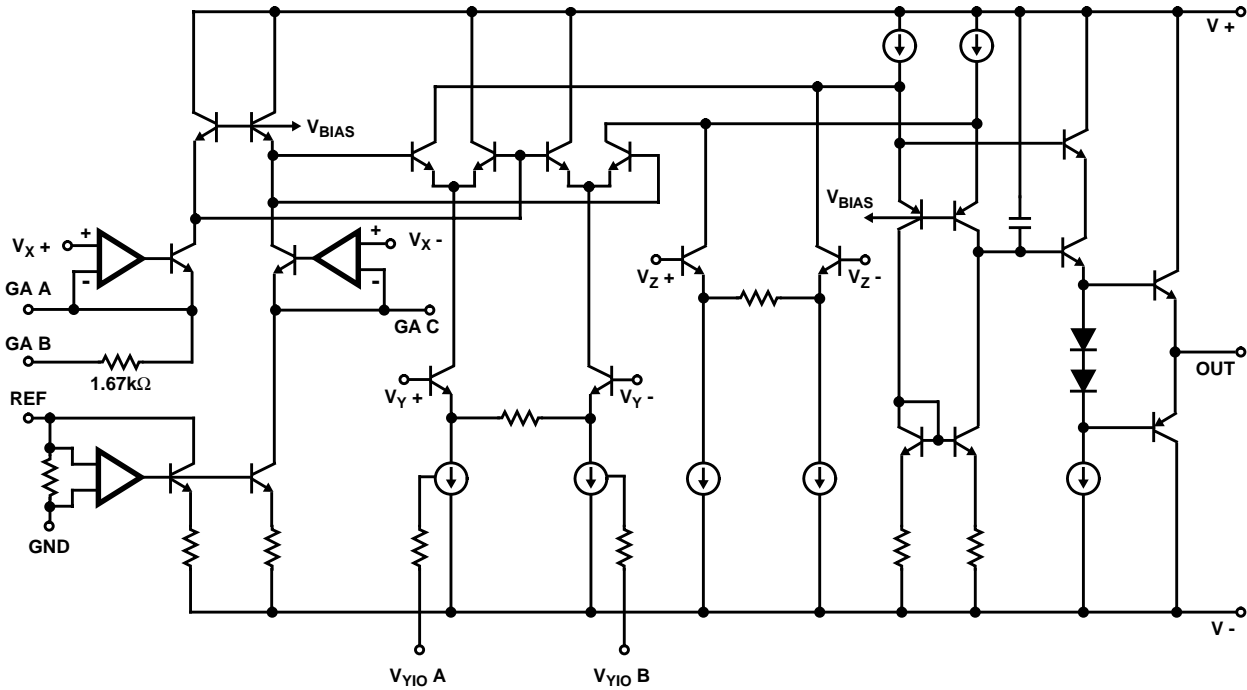
NOTES:

C1 = C2 = 0.01μF/Socket Min.

D1 = D2 = IN4002 or Equivalent/Board

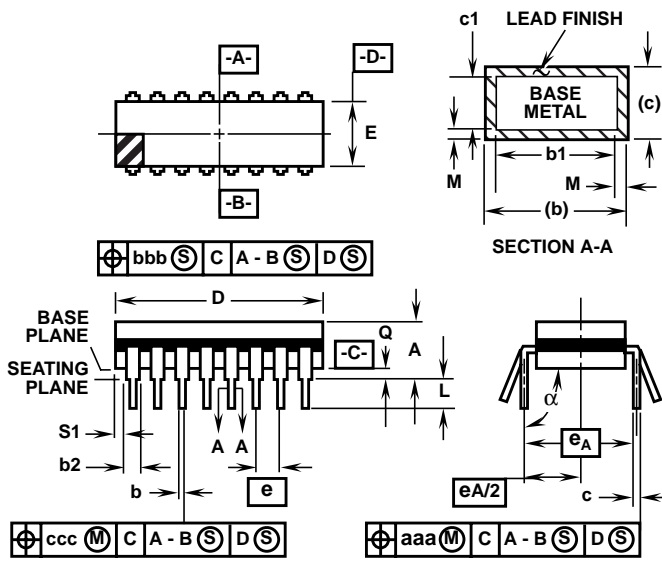
| (V+) - (V-) | = 31V ± 1V

Simplified Schematic





**Packaging**



**F16.3 MIL-STD-1835 GDIP1-T16 (D-2, CONFIGURATION A)  
16 LEAD DUAL-IN-LINE FRIT-SEAL CERAMIC PACKAGE**

SYMBOL	INCHES		MILLIMETERS		NOTES
	MIN	MAX	MIN	MAX	
A	-	0.200	-	5.08	-
b	0.014	0.026	0.36	0.66	2
b1	0.014	0.023	0.36	0.58	3
b2	0.045	0.065	1.14	1.65	-
b3	0.023	0.045	0.58	1.14	4
c	0.008	0.018	0.20	0.46	2
c1	0.008	0.015	0.20	0.38	3
D	-	0.840	-	21.34	5
E	0.220	0.310	5.59	7.87	5
e	0.100 BSC		2.54 BSC		-
eA	0.300 BSC		7.62 BSC		-
eA/2	0.150 BSC		3.81 BSC		-
L	0.125	0.200	3.18	5.08	-
Q	0.015	0.060	0.38	1.52	6
S1	0.005	-	0.13	-	7
S2	0.005	-	0.13	-	-
$\alpha$	90°	105°	90°	105°	-
aaa	-	0.015	-	0.38	-
bbb	-	0.030	-	0.76	-
ccc	-	0.010	-	0.25	-
M	-	0.0015	-	0.038	2
N	16		16		8

**NOTES:**

1. Index area: A notch or a pin one identification mark shall be located adjacent to pin one and shall be located within the shaded area shown. The manufacturer's identification shall not be used as a pin one identification mark.
2. The maximum limits of lead dimensions b and c or M shall be measured at the centroid of the finished lead surfaces, when solder dip or tin plate lead finish is applied.
3. Dimensions b1 and c1 apply to lead base metal only. Dimension M applies to lead plating and finish thickness.
4. Corner leads (1, N, N/2, and N/2+1) may be configured with a partial lead paddle. For this configuration dimension b3 replaces dimension b1.
5. This dimension allows for off-center lid, meniscus, and glass overrun.
6. Dimension Q shall be measured from the seating plane to the base plane.
7. Measure dimension S1 at all four corners.
8. N is the maximum number of terminal positions.
9. Dimensioning and tolerancing per ANSI Y14.5M - 1982.
10. Controlling Dimension: Inch.
11. Materials: Compliant to MIL-I-38535.



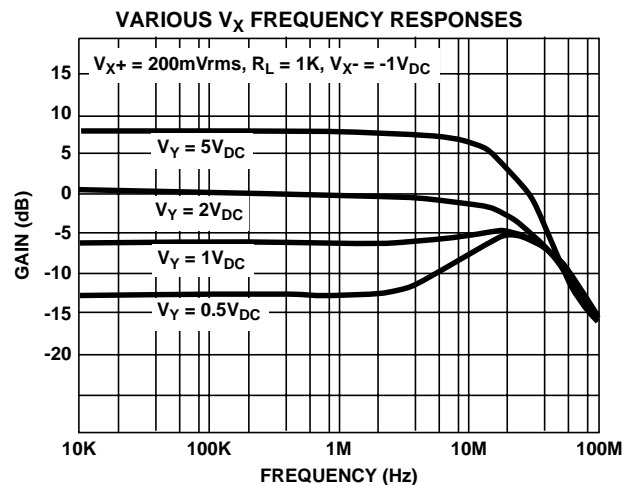
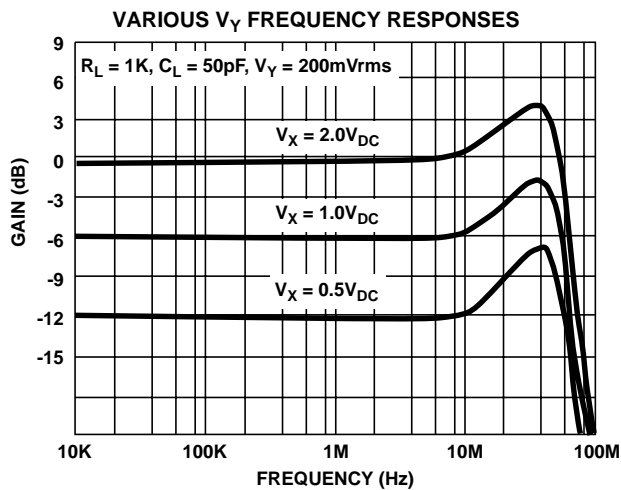
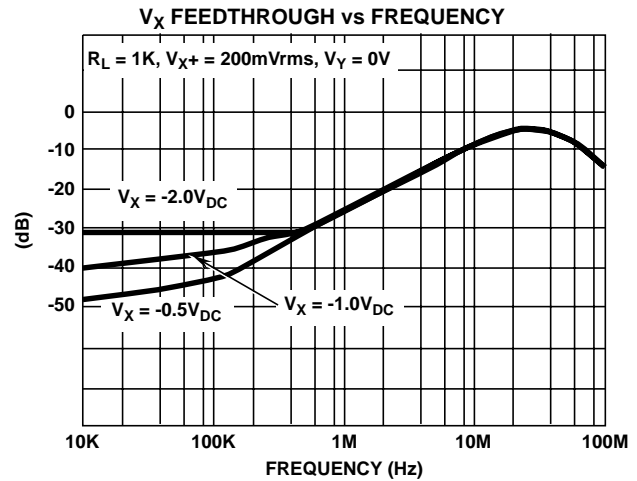
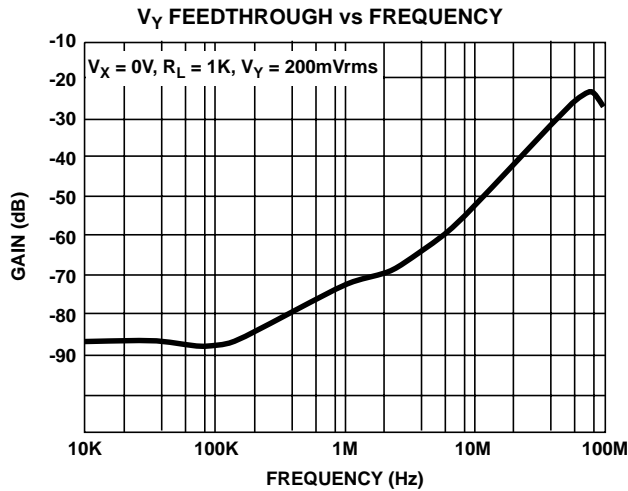
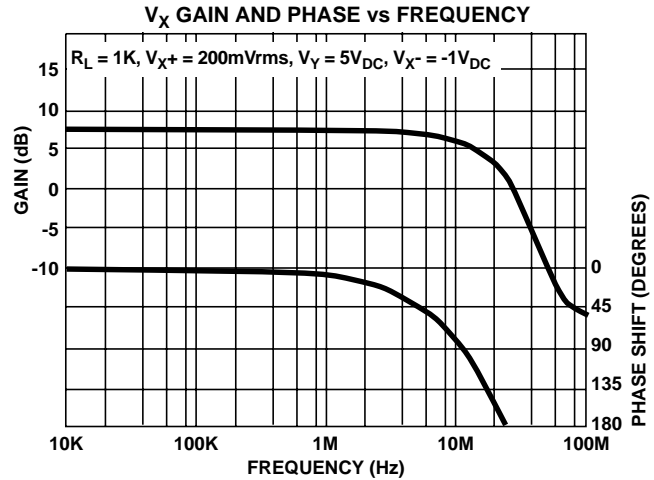
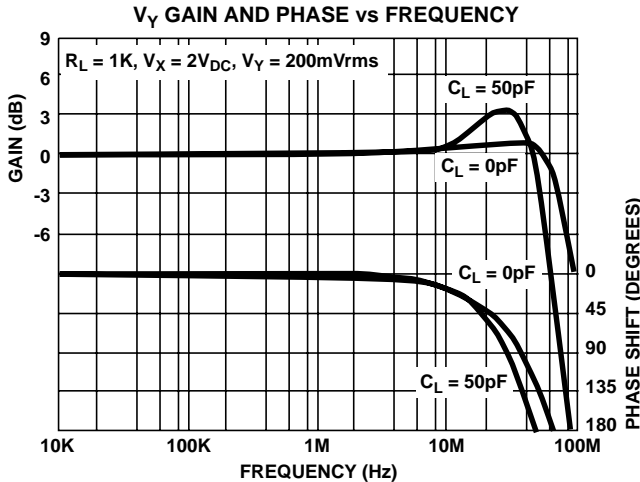
## DESIGN INFORMATION

## Wideband Two Quadrant Analog Multiplier

August 1999

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### Typical Performance Curves $V_S = \pm 15V$ , $T_A = +25^\circ C$ , See Test Circuit For Multiplier Configuration.

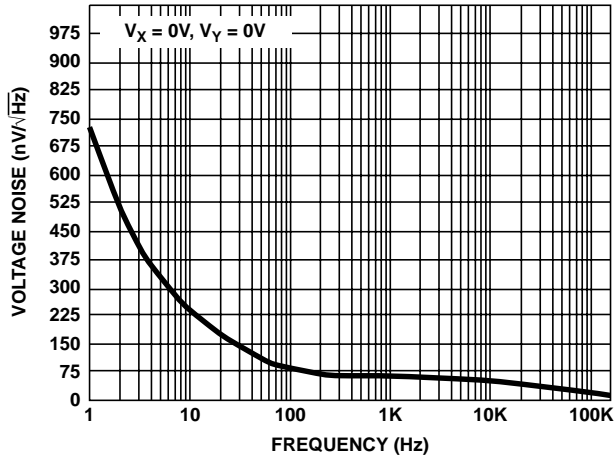


# DESIGN INFORMATION (Continued)

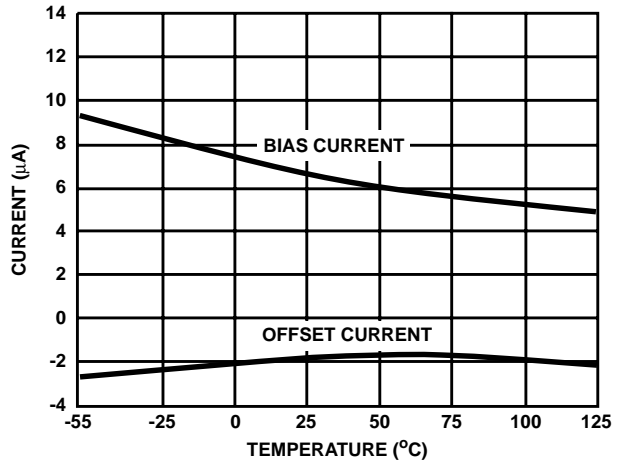
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## Typical Performance Curves $V_S = \pm 15V, T_A = +25^\circ C$ , See Test Circuit For Multiplier Configuration. (Continued)

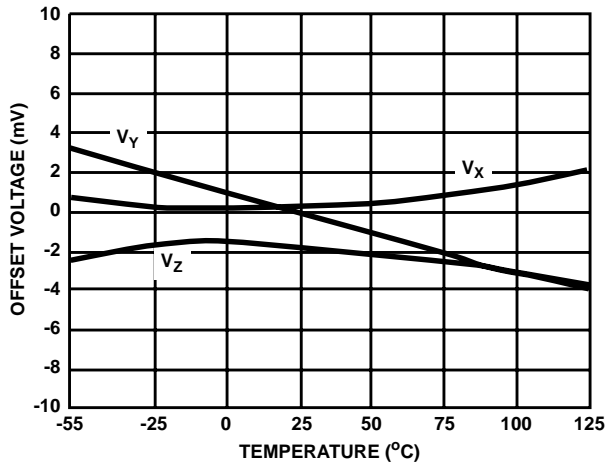
**NOISE CHARACTERISTICS**



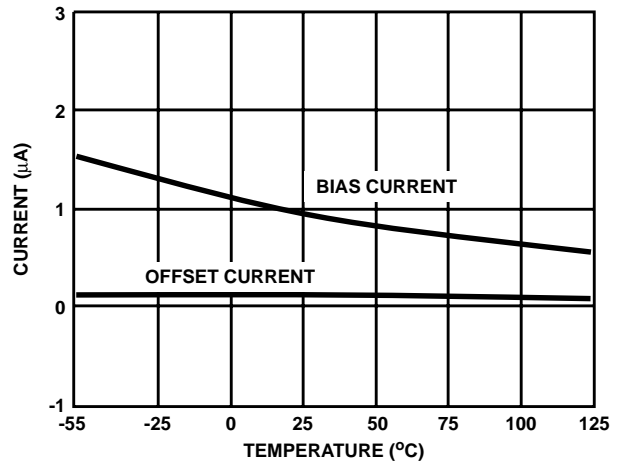
**$V_Y$  OFFSET AND BIAS CURRENT vs TEMPERATURE**



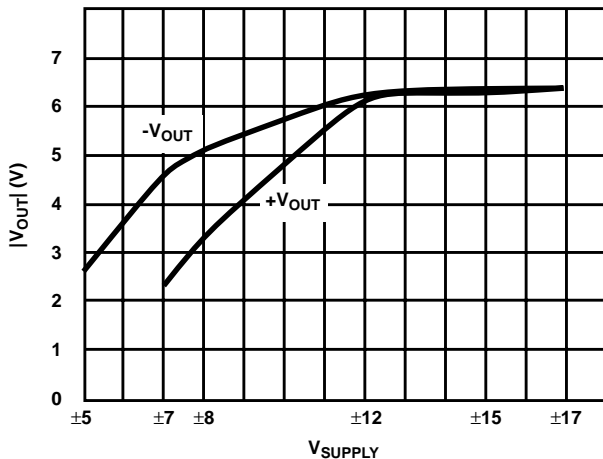
**OFFSET VOLTAGE vs TEMPERATURE**



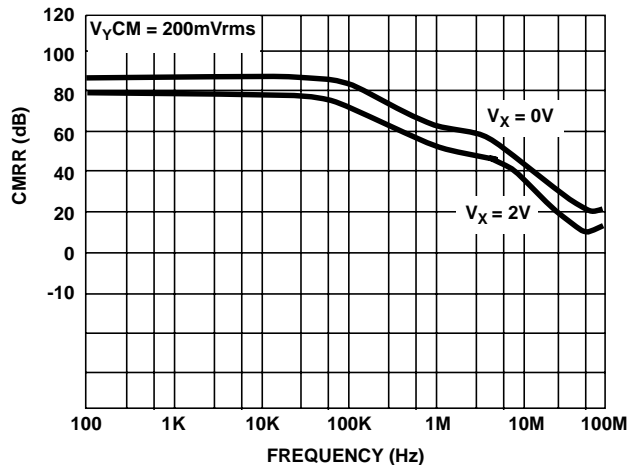
**$V_X$  OFFSET AND BIAS CURRENT vs TEMPERATURE**



**$V_{OUT}$  vs  $V_{SUPPLY}$**



**$V_Y$  CMRR vs FREQUENCY**

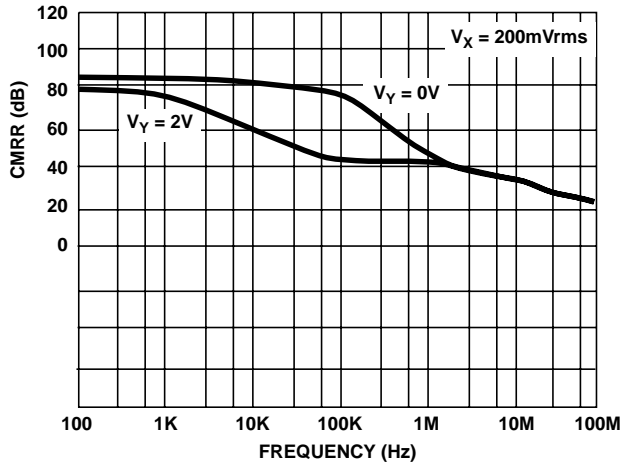


## DESIGN INFORMATION (Continued)

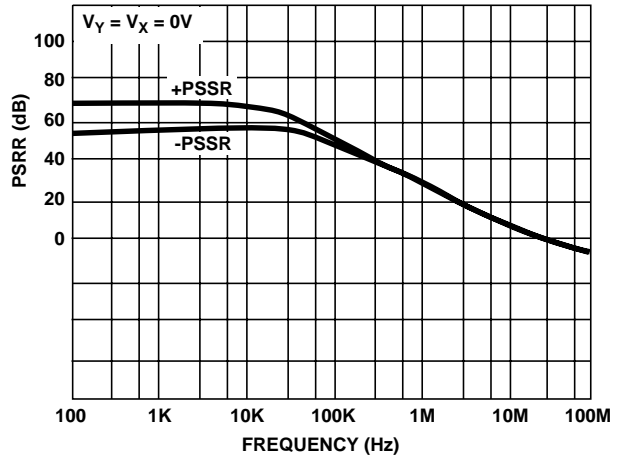
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### Typical Performance Curves $V_S = \pm 15V, T_A = +25^\circ C$ , See Test Circuit For Multiplier Configuration. (Continued)

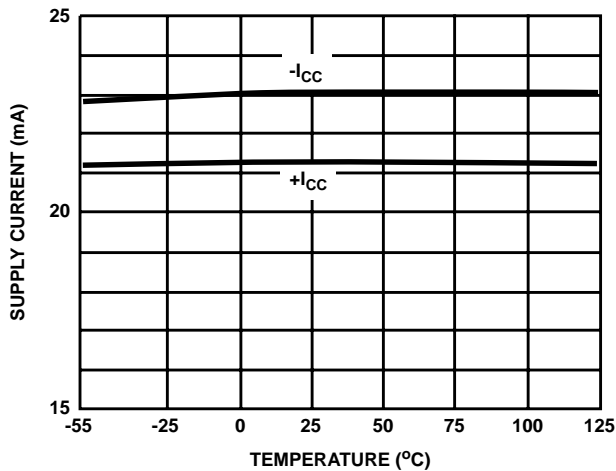
$V_X$  COMMON MODE REJECTION RATIO vs FREQUENCY



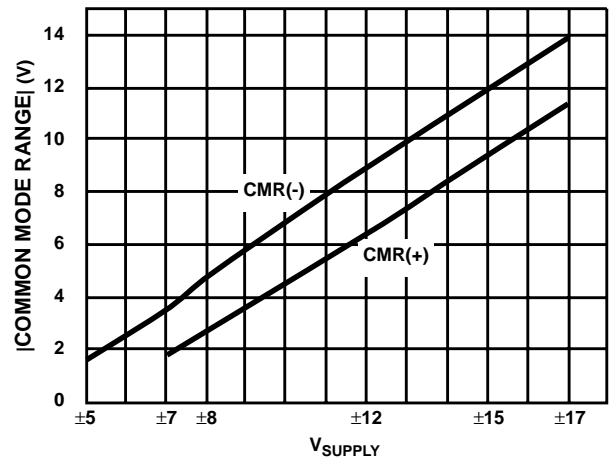
PSRR vs FREQUENCY



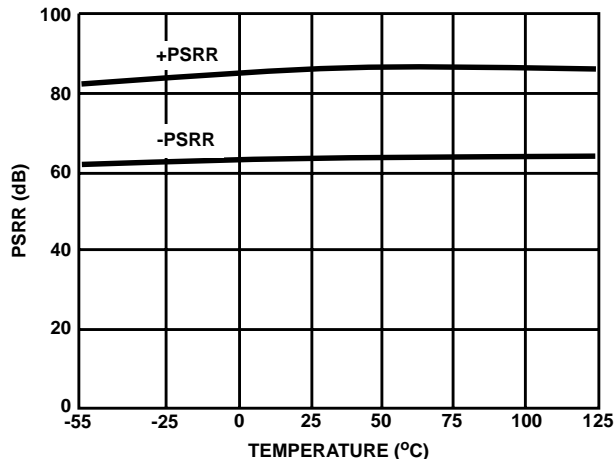
SUPPLY CURRENT vs TEMPERATURE



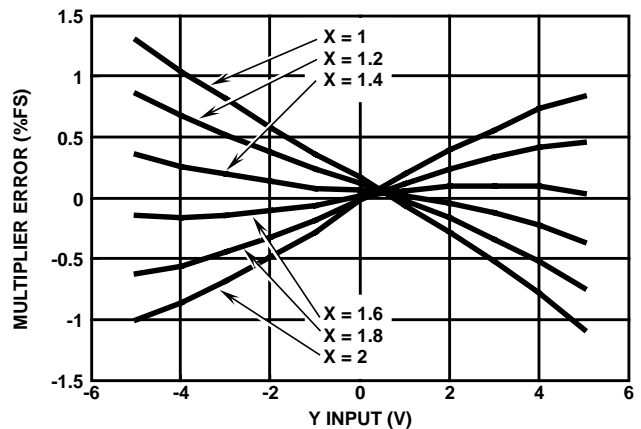
COMMON MODE RANGE vs  $V_{SUPPLY}$



PSRR vs TEMPERATURE



MULTIPLIER ERROR

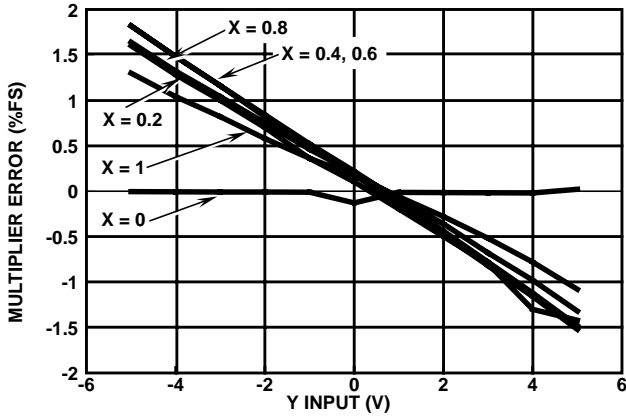


# DESIGN INFORMATION (Continued)

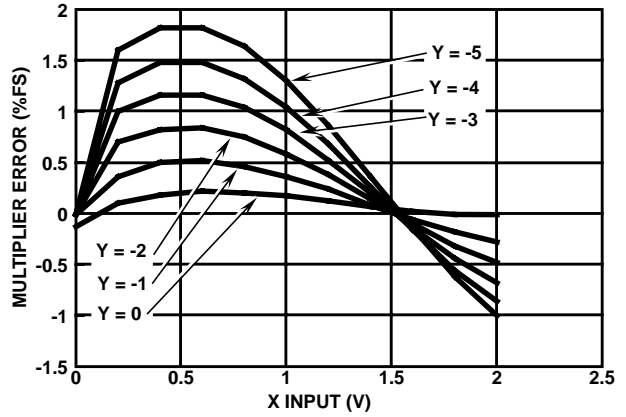
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## Typical Performance Curves $V_S = \pm 15V, T_A = +25^\circ C$ , See Test Circuit For Multiplier Configuration. (Continued)

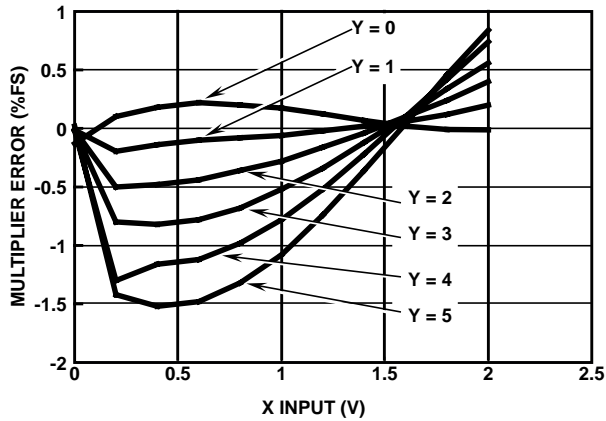
MULTIPLIER ERROR



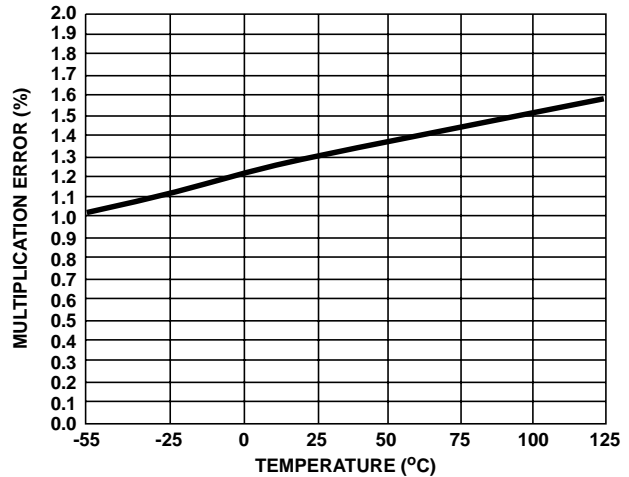
MULTIPLIER ERROR



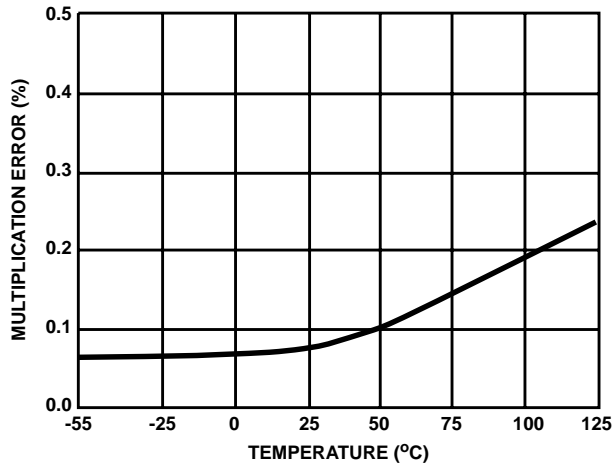
MULTIPLIER ERROR



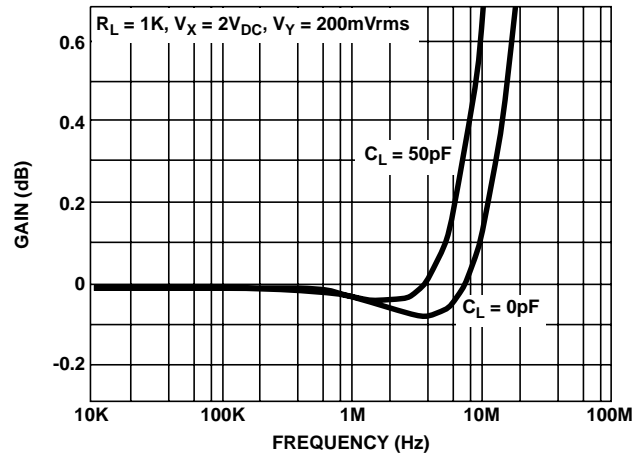
WORST CASE MULTIPLICATION ERROR vs TEMPERATURE



MULTIPLICATION ERROR vs TEMPERATURE



GAIN VARIATION vs FREQUENCY

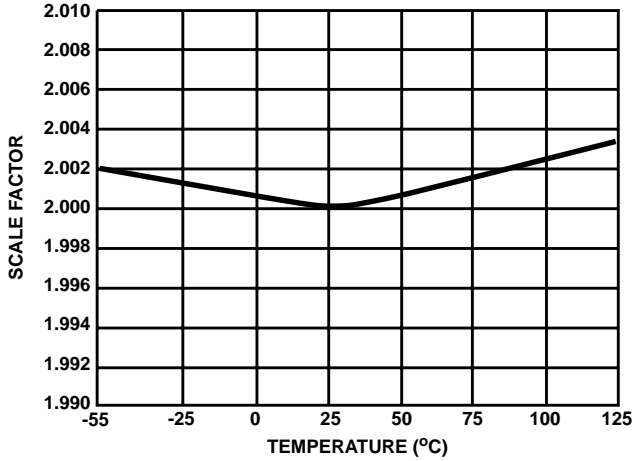


## DESIGN INFORMATION (Continued)

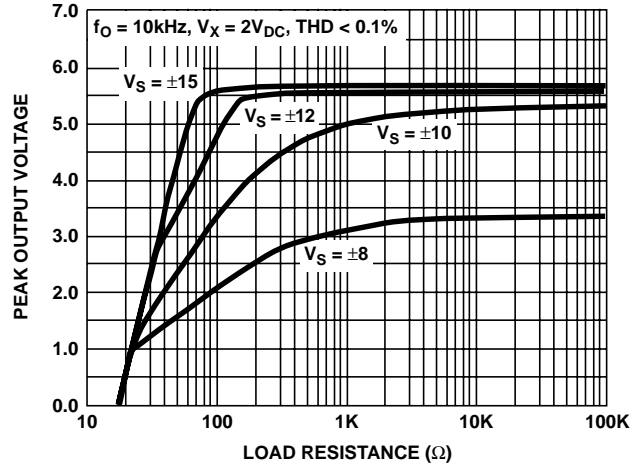
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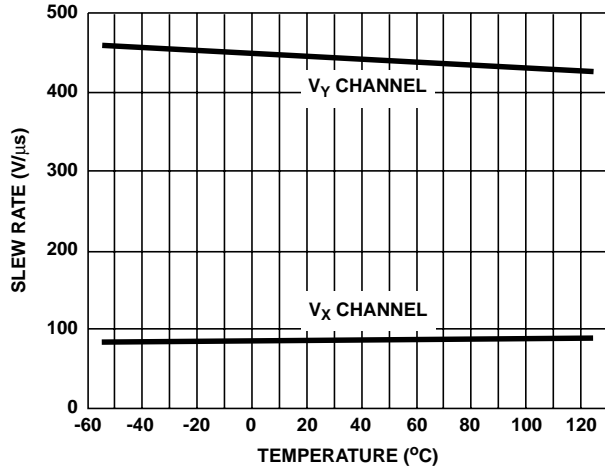
SCALE FACTOR vs TEMPERATURE



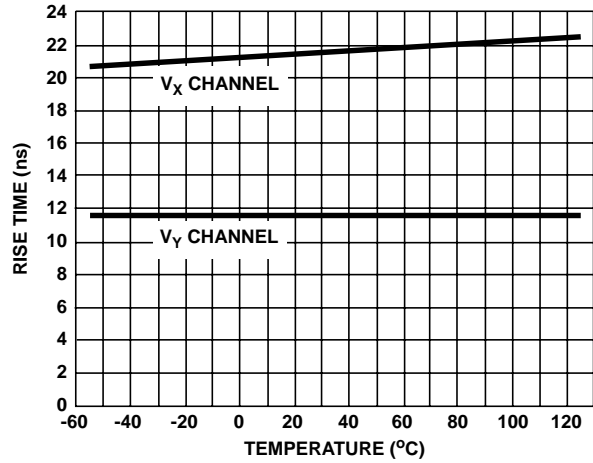
OUTPUT VOLTAGE SWING vs LOAD RESISTANCE



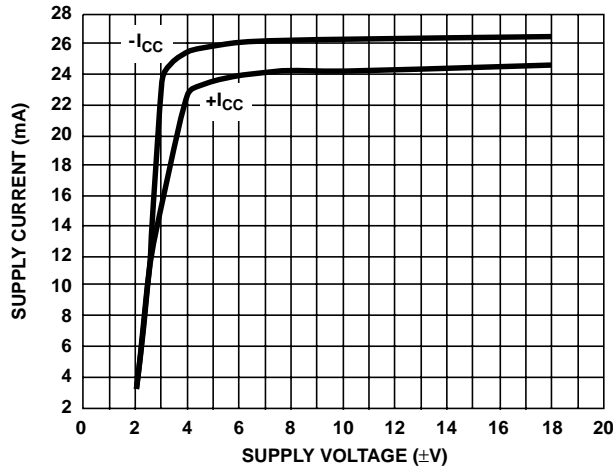
SLEW RATE vs TEMPERATURE



RISE TIME vs TEMPERATURE



SUPPLY CURRENT vs SUPPLY VOLTAGE



## DESIGN INFORMATION (Continued)

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### Application Information

#### Theory of Operation

The HA-2546 is a two quadrant multiplier with the following three differential inputs; the signal channel,  $V_{Y+}$  and  $V_{Y-}$ , the control channel,  $V_{X+}$  and  $V_{X-}$ , and the summed channel,  $V_{Z+}$  and  $V_{Z-}$ , to complete the feedback of the output amplifier. The differential voltages of channel X and Y are converted to differential currents. These currents are then multiplied in a circuit similar to a Gilbert Cell multiplier, producing a differential current product. The differential voltage of the Z channel is converted into a differential current which then sums with the products currents. The differential "product/sum" currents are converted to a single-ended current and then converted to a voltage output by a transimpedance amplifier.

The open loop transfer equation for the HA-2546 is:

$$V_{OUT} = A \left[ \frac{(V_{X+} - V_{X-})(V_{Y+} - V_{Y-})}{SF} - (V_{Z+} - V_{Z-}) \right]$$

where;

A = Output Amplifier Open Loop Gain

SF = Scale Factor

$V_{X+}$ ,  $V_{Y+}$ ,  $V_{Z+}$  = Differential Inputs

The scale factor is used to maintain the output of the multiplier within the normal operating range of  $\pm 5V$ . The scale factor can be defined by the user by way of an optional external resistor,  $R_{EXT}$ , and the Gain Adjust pins, Gain Adjust A (GA A), Gain Adjust B (GA B), and Gain Adjust C (GA C). The scale factor is determined as follows:

SF = 2, when GA B is shorted to GA C

SF  $\cong 1.2 R_{EXT}$ , when  $R_{EXT}$  is connected between GA A and GA C ( $R_{EXT}$  is in k $\Omega$ )

SF  $\cong 1.2 (R_{EXT} + 1.667k\Omega)$ , when  $R_{EXT}$  is connected to GA B and GA C ( $R_{EXT}$  is in k $\Omega$ )

The scale factor can be adjusted from 2 to 5. It should be noted that any adjustments to the scale factor will affect the AC performance of the control channel,  $V_{X-}$ . The normal input operating range of  $V_{X-}$  is equal to the scale factor voltage.

The typical multiplier configuration is shown in Figure 1. The ideal transfer function for this configuration is:

$$V_{OUT} = \begin{cases} \frac{(V_{X+} - V_{X-})(V_{Y+} - V_{Y-})}{2} + V_{Z-}, & \text{when } V_{X-} \geq 0V \\ 0, & \text{when } V_{X-} < 0V \end{cases}$$

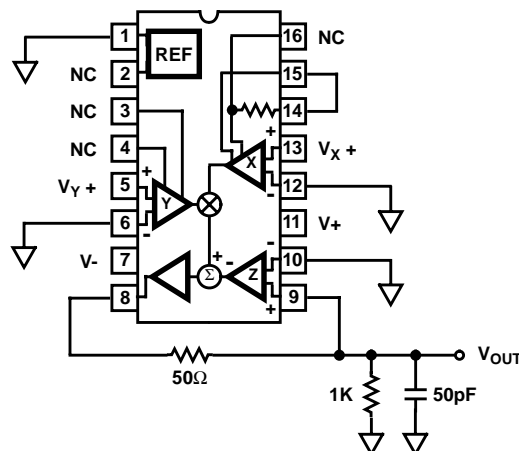


FIGURE 1.

The  $V_{X-}$  pin is usually connected to ground so that when  $V_{X+}$  is negative there is no signal at the output, i.e. two quadrant operation. If the  $V_{X-}$  input is a negative going signal the  $V_{X+}$  pin maybe grounded and the  $V_{X-}$  pin used as the control input.

The  $V_{Y-}$  terminal is usually grounded allowing the  $V_{Y+}$  to swing  $\pm 5V$ . The  $V_{Z+}$  terminal is usually connected directly to  $V_{OUT}$  to complete the feedback loop of the output amplifier while  $V_{Z-}$  is grounded. The scale factor is normally set to 2 by connecting GA B to GA C. Therefore the transfer equation simplifies to  $V_{OUT} = (V_{X-} V_{Y+}) / 2$ .

#### Offset Adjustment

The signal channel offset voltage may be nulled by using a 20k $\Omega$  potentiometer between  $V_{Y10}$  Adjust pins A and B and connecting the wiper to -V. Reducing the signal channel offset will reduce  $V_{X-}$  AC feedthrough. Output offset voltage can also be nulled by connecting  $V_{Z-}$  to the wiper of a 20k $\Omega$  potentiometer which is tied between +V and -V.

#### Capacitive Drive Capability

When driving capacitive loads  $>20pF$ , a 50 $\Omega$  resistor is recommended between  $V_{OUT}$  and  $V_{Z+}$ , using  $V_{Z+}$  as the output (See Figure 1). This will prevent the multiplier from going unstable.

#### Power Supply Decoupling

Power supply decoupling is essential for high frequency circuits. A 0.01 $\mu F$  high quality ceramic capacitor at each supply pin in parallel with a 1 $\mu F$  tantalum capacitor will provide excellent decoupling. Chip capacitors produce the best results due to the close spacing with which they may be placed to the supply pins minimizing lead inductance.

#### Adjusting Scale Factor

The HA-2546 two quadrant multiplier may be configured for many uses. Following are examples of a few typical applications.



## DESIGN INFORMATION (Continued)

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Adjusting the scale factor will tailor the control signal,  $V_X$ , input voltage range to match your needs. Referring to the simplified schematic and looking for the  $V_X$  input stage, you will notice the unusual design. The internal reference sets up a 1.2mA current sink for the  $V_X$  differential pair. The control signal applied to this input will be forced across the scale factor setting resistor and set the current flowing in the  $V_{X+}$  side of the differential pair. When the current through this resistor reaches 1.2mA, all the current available is flowing in the one side and full scale has been reached. Normally the 1.67k $\Omega$  internal resistor sets the scale factor to 2V when the Gain Adjust pins B and C are connected together, but you may set this resistor to any convenient value using pins 16 (GA A) and 15 (GA C).

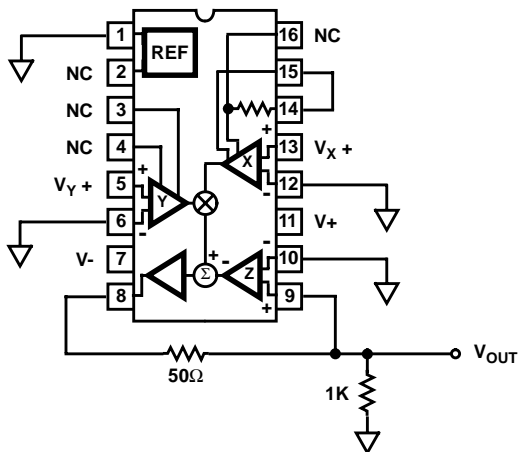


FIGURE 2A. MULTIPLIER,  $V_{OUT} = V_X V_Y / 2$ , SCALE FACTOR = 2V

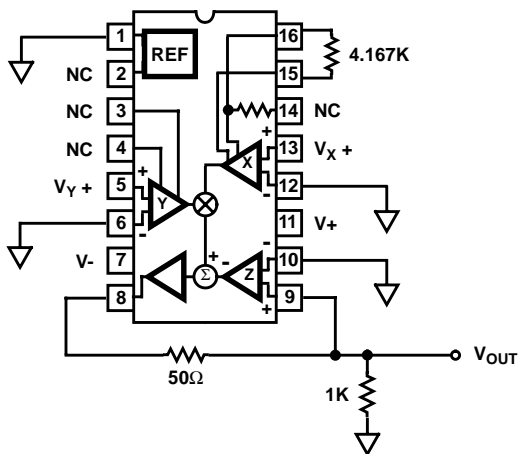


FIGURE 2B. MULTIPLIER,  $V_{OUT} = V_X V_Y / 5$ , SCALE FACTOR = 5V

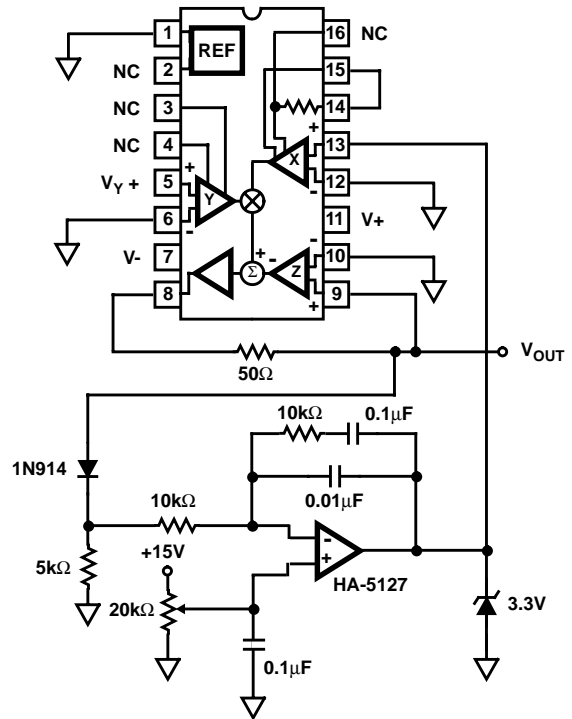


FIGURE 3. AUTOMATIC GAIN CONTROL

In Figure 3, the HA-2546 is configured in a true Automatic Gain Control or AGC application. The HA-5127, low noise op amp, provides the gain control level to the X input. This level will set the peak output voltage of the multiplier to match the reference level. The feedback network around the HA-5127 provides stability and a response time adjustment for the gain control circuit.

## DESIGN INFORMATION (Continued)

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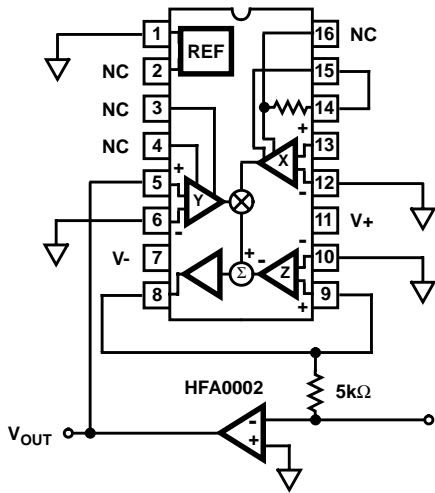


FIGURE 4A. VOLTAGE CONTROLLED AMPLIFIER

This multiplier has the advantage over other AGC circuits, in that the signal bandwidth is not affected by the control signal gain adjustment.

A wide range of gain adjustment is available with the Voltage Controlled Amplifier configuration shown in Figure 4A. and Figure 4B. Here the gain of the HFA0002 is swept from 20V/V at a control voltage of 0.902V to a gain of almost 1000V/V with a control voltage of 0.03V.

The Video Fader circuit provides a unique function as shown in Figure 5. Here Ch B is applied to the minus Z input in addition to the minus Y input.  $V_{MIX}$  will control the percentage of Ch A and Ch B that are mixed together to produce a resulting video image or other signal.

Many other applications are possible including division, squaring, square-root, percentage calculations, etc. Please refer to the HA-2556 four quadrant multiplier for additional applications.

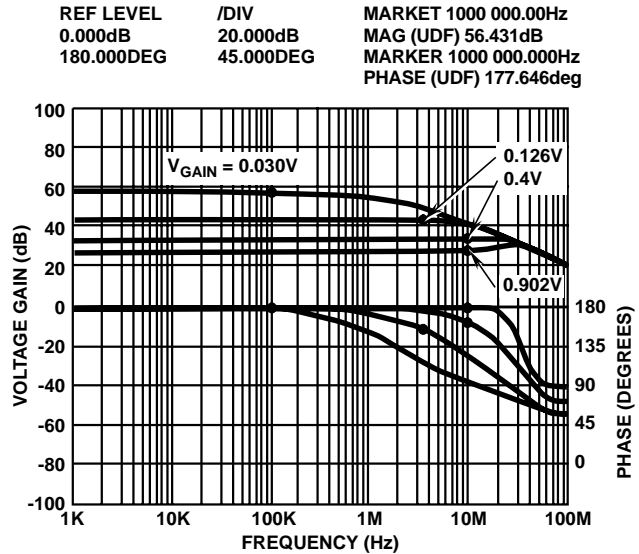
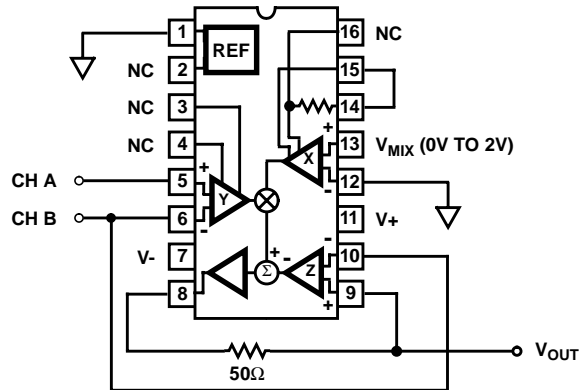


FIGURE 4B. VOLTAGE CONTROLLED AMPLIFIER



$$V_{OUT} = \text{Ch B} + (\text{Ch A} - \text{Ch B}) V_{MIX} / \text{Scale Factor}$$

$$\text{Scale Factor} = 2$$

$$V_{OUT} = \text{All Ch B; if } V_{MIX} = 0V$$

$$V_{OUT} = \text{All Ch A; if } V_{MIX} = 2V \text{ (Full Scale)}$$

$$V_{OUT} = \text{Mix of Ch A and Ch B; if } 0V < V_{MIX} < 2V$$

FIGURE 5. VIDEO FADER

**DESIGN INFORMATION** (Continued)

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**TYPICAL PERFORMANCE CHARACTERISTICS**

Device Tested at: Supply Voltage =  $\pm 15V$ ,  $R_{LOAD} = 1k\Omega$ ,  $C_{LOAD} = 50pF$ , Unless Otherwise Specified.

PARAMETER	CONDITIONS	TEMP	TYP	UNITS	
Multiplication Error Drift		Full	0.002	%/°C	
Differential Gain	$V_X = 2V$ , $V_Y = 300mV_{P-P}$ , $f_O = 3.58MHz$	+25°C	0.1	%	
Differential Phase	$V_X = 2V$ , $V_Y = 300mV_{P-P}$ , $f_O = 3.58MHz$	+25°C	0.1	Degrees	
Gain Tolerance	DC to 5MHz 5MHz to 8MHz	$V_X = 2V$	+25°C	0.1	dB
			+25°C	0.18	dB
1% Amplitude Error		+25°C	6	MHz	
1% Vector Error		+25°C	260	kHz	
THD+N	$f_O = 10kHz$ , $V_Y = 1V_{rms}$ , $V_X = 2V$	+25°C	0.03	%	
Voltage Noise	$f_O = 10Hz$ $f_O = 100Hz$ $f_O = 1kHz$	$V_X = 0V$ , $V_Y = 0V$	+25°C	400	nV/Hz
			+25°C	150	nV/Hz
			+25°C	75	nV/Hz
Common Mode Range		+25°C	$\pm 9$	V	
SIGNAL INPUT, $V_Y$					
Average Offset Voltage Drift		Full	45	$\mu V/°C$	
Differential Input Resistance		+25°C	720	k $\Omega$	
Small Signal Bandwidth (-3dB)	$V_X = 2V$	+25°C	40	MHz	
Feedthrough	$f_O = 5MHz$ , $V_X = 0V$ , $V_Y = 200mV_{rms}$	+25°C	-52	dB	
$V_Y$ TRANSIENT RESPONSE					
Propagation Delay		+25°C	25	ns	
Settling Time	$V_Y = \pm 5V$ , $V_X = 2V$	+25°C	200	ns	
CONTROL INPUT, $V_X$					
Average Offset Voltage Drift		Full	10	$\mu V/°C$	
Differential Input Resistance		+25°C	360	k $\Omega$	
Small Signal Bandwidth (-3dB)	$V_Y = 5V$ , $V_X = -1V$	+25°C	17	MHz	
Feedthrough	$f_O = 100kHz$ , $V_Y = 0V$ , $V_X = 200mV_{rms}$	+25°C	-40	dB	
Common Mode Rejection Ratio	$V_X = 0V$ to $2V$ , $V_Y = 5V$	+25°C	80	dB	
$V_X$ TRANSIENT RESPONSE					
Propagation Delay		+25°C	50	ns	
Settling Time	$V_X = 0$ to $2V$ , $V_Y = 5V$	+25°C	200	ns	
$V_Z$ CHARACTERISTICS					
Open Loop Gain		+25°C	70	dB	
Differential Input Resistance		+25°C	900	k $\Omega$	
OUTPUT CHARACTERISTICS					
Output Resistance		+25°C	1	$\Omega$	

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