

FAN8039D3

5-CH Motor Driver

Features

- 1 Phase, Full-wave, Linear DC Motor Driver
- Built-in 5-CH Balanced TransFormerless (BTL) Driver
- Built-in thermal shut down circuit (TSD)
- Built-in Variable Regulator With Power Tr.
- Built-in Power Save Circuit
- Built-in stand by mode circuit
- Wide Operating Supply Voltage : 4.5 ~ 13.2V

Description

The FAN8039D3 is a monolithic integrated circuit suitable for a 5-ch motor driver which drives the tracking actuator, focus actuator, sled motor, tray motor, spindle motor of the DVDP/CAR-CD systems.

28-SSOPH-375SG2



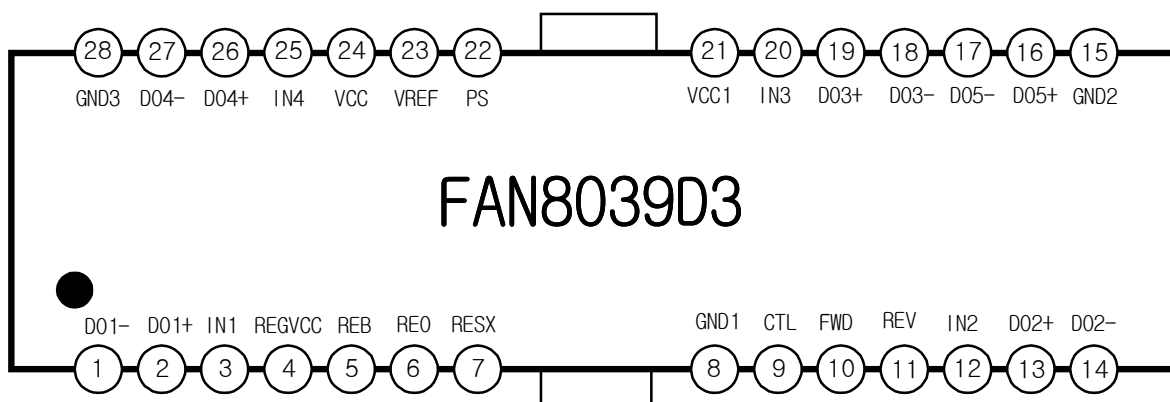
Typical Applications

- Compact disk player
- Video compact disk player
- Car compact disk player
- Mixing with compact disk player and mini disk player
- DVDP

Ordering Information

Device	Package	Operating Temp
FAN8039D3	28-SSOPH-375SG2	-35°C ~ 85°C
FAN8039D3TF	28-SSOPH-375SG2	-35°C ~ 85°C

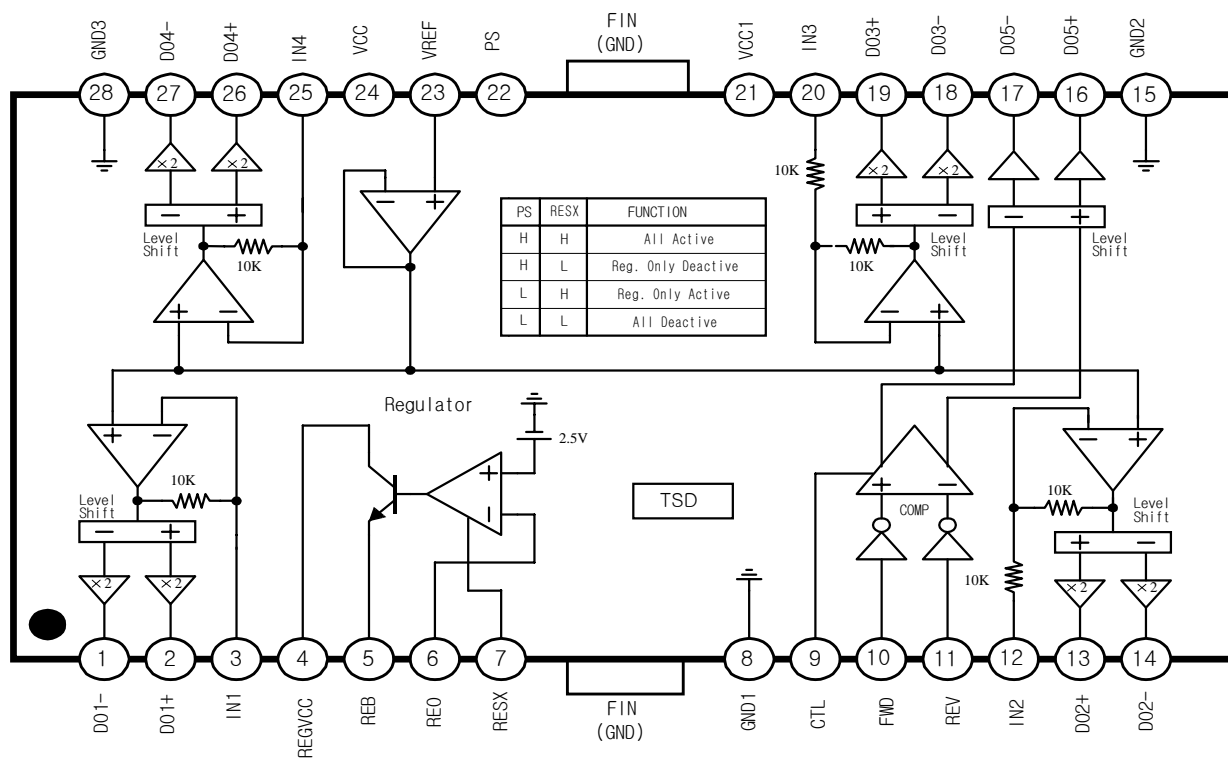
Pin Assignments



Pin Definitions

NO	Symbol	Description	NO	Symbol	Description
1	DO1-	CH1 Drive Output (-)	15	GND2	Power Ground1 (CH 2,3,5)
2	DO1+	CH1 Drive Output (+)	16	DO5+	CH5 Drive Output (+)
3	IN1	CH1 Drive Input	17	DO5-	CH5 Drive Output(-)
4	REGVCC	Regulator Supply Voltage	18	DO3-	CH3 Drive Output(-)
5	REB	Regulator Output	19	DO3+	CH3 Drive Output (+)
6	REO	Regulator Feedback Input	20	IN3	CH3 Drive Input
7	RESX	Regulator Reset	21	VCC1	Supply Voltage1
8	GND1	Signal Ground	22	PS	Power Save
9	CTL	CH5 Motor Speed Control	23	VREF	Bias Voltage
10	FWD	CH5 Forward Input	24	VCC	Supply Voltage
11	REV	CH5 Reverse Input	25	IN4	CH4 Drive Input
12	IN2	CH2 Drive Input	26	DO4+	CH4 Drive Output (+)
13	DO2+	CH2 Drive Output (+)	27	DO4-	CH4 Drive Output (-)
14	DO2-	CH2 Drive Output (-)	28	GND3	Power Ground2 (CH 1,4)

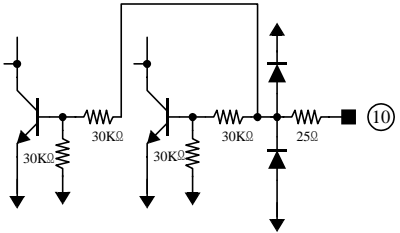
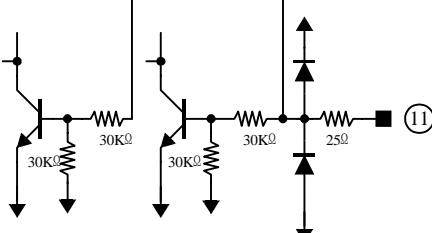
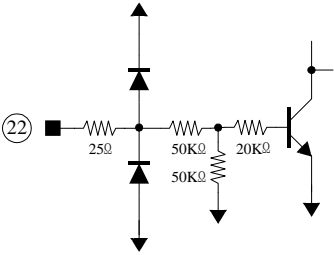
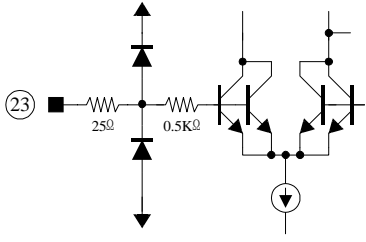
Internal Block Diagram



Equivalent Circuits

BTL DRIVER OUTPUT	BTL DRIVE INPUT
REGULATOR OUTPUT	REGULATOR FEEDBACK INPUT
REGULATOR RESET	MOTOR SPEED CONTROL

Equivalent Circuits (Continued)

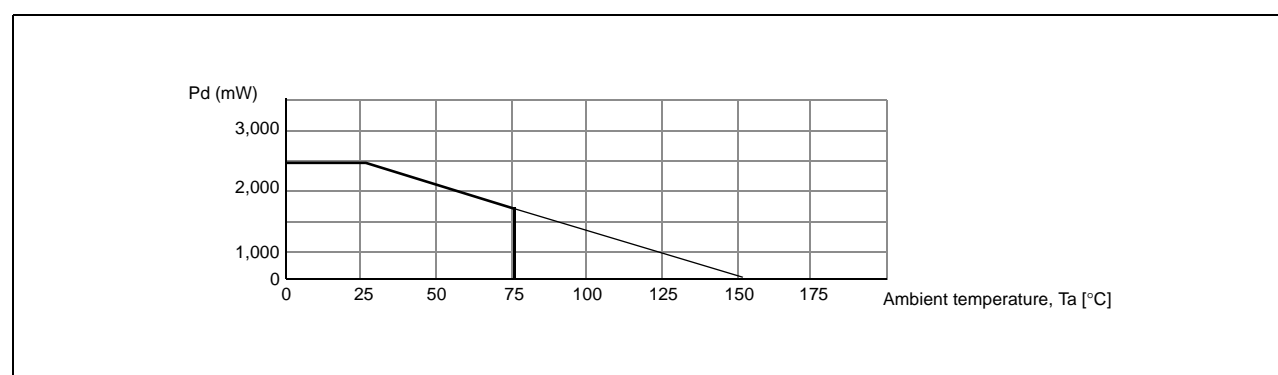
FOWARD INPUT	REVERSE INPUT
 <p>The diagram shows two transistors in a push-pull configuration. Each transistor's base is connected to a 30KΩ resistor, which is in turn connected to a common node. This node is also connected to a 25Ω resistor that leads to a square symbol labeled 10. The emitters of both transistors are connected to ground. The collector of the left transistor is connected to the collector of the right transistor through a feedback loop.</p>	 <p>The diagram is identical to the Forward Input circuit, but the 25Ω resistor is connected to a square symbol labeled 11.</p>
POWER SAVE	BIAS VOLTAGE
 <p>The diagram shows a single transistor with its emitter connected to ground. The base is connected to a 25Ω resistor, which is connected to a square symbol labeled 22. The base is also connected to a 50KΩ resistor that leads to another 50KΩ resistor connected to ground. A 20KΩ resistor is connected between the base and the emitter.</p>	 <p>The diagram shows a differential pair of transistors. The emitters are connected to a common node that is connected to ground through a current source. The base of the left transistor is connected to a 25Ω resistor, which is connected to a square symbol labeled 23. The base of the right transistor is connected to a 0.5KΩ resistor that leads to the common emitter node.</p>

Absolute Maximum Ratings (Ta = 25°C)

Parameter	Symbol	Value	Unit
Maximum Supply Voltage	VCC	18	V
Power Dissipation	PD	2.5 ^{note}	W
Operating Temperature	TOPR	-35 ~ +85	°C
Storage Temperature	TSTG	-55 ~ +150	°C
Maximum output current	IOMAX	1	A
Regulator Maximum output current	IROMAX	400	mA

Notes:

1. When mounted on 70mm × 70mm × 1.6mm PCB
2. Power dissipation reduces 20mW/°C for using above TA = 25°C
3. Do not exceed PD and SOA (Safe Operating Area)



Recommended Operating Conditions (Ta = 25°C)

Parameter	Symbol	Min.	Typ.	Max.	Unit
Operating Supply Voltage	VCC	4.5	-	13.2	V

Electrical Characteristics

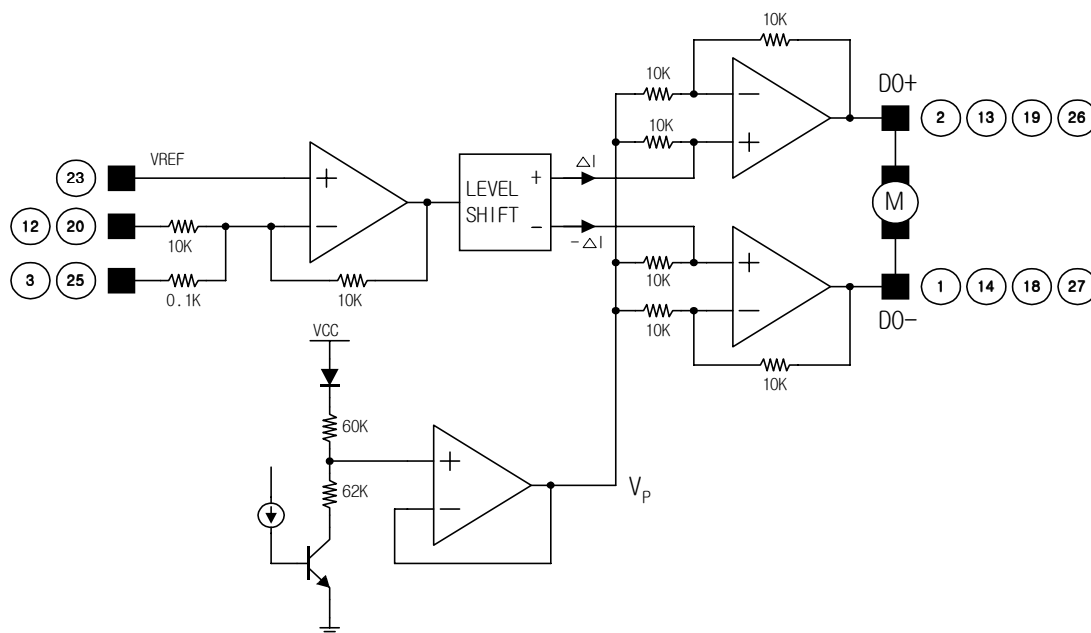
(VCC = VCC1 = 8V, TA = 25°C, unless otherwise specified)

Characteristics	Symbol	Condition	Min	Typ	Max	Unit
Quiescent Circuit Current	ICCQ	Under no - load	-	20	-	mA
Power Save On Current	IPS	Pin7=GND	-	1	2	mA
Power Save On Voltage	VPSON	Pin7=Variation	-	-	0.5	V
Power Save Off Voltage	VPSOFF	Pin7=Variation	2	-	-	V
VARIABLE REGULATOR PART						
Load Regulation	ΔV_{RL}	$I_L = 0\text{mA} \rightarrow 200\text{mA}$	-40	0	10	mV
Line Regulation	ΔV_{CC}	$I_L = 200\text{mA}, V_{CC}=6\text{V} \rightarrow 9\text{V}$	-20	0	30	mV
Regulator Output Voltage 1	VREG1	$I_L = 100\text{mA}$	4.75	5	5.25	V
Regulator Output Voltage 2	VREG2	$I_L = 100\text{mA}$	3.135	3.3	3.465	V
*Regulator Output Peak Current	IPK	$T_j = 25^\circ\text{C}$		700		mA
BLT DRIVER PART						
Output Offset Voltage	VOO	$V_{IN}=2.5\text{V}$	-40	0	40	mV
Maximum Output Voltage1	VOM1	$V_{CC}=V_{CC1}=8\text{V}, R_L = 12\Omega$	5.5	6.5	-	V
Maximum Output Voltage2	VOM2	$V_{CC}=V_{CC1}=13\text{V}, R_L = 24\Omega$	10.5	11.5	-	V
Close Loop Voltage Gain	AVF	$V_{IN}=0\text{V}, 1\text{V}_{rms}, f = 1\text{KHz}$	10.5	12	13.5	dB
Slew rate	SR	$V_{OUT}=4\text{VP-P}, f = 120\text{KHz}, \text{Square}$	-	2	-	V/ μs
LOADING MOTOR DRIVER PART						
Input High Level Voltage	V _{IH}	-	2	-	-	V
Input Low Level Voltage	V _{IL}	-	-	-	0.5	V
Output Voltage1	VO1	$V_{CC}=V_{CC1}=8\text{V}, V_{CTL}=\text{OPEN}, R_L=12\Omega$	5.5	6.5	-	V
Output Voltage2	VO2	$V_{CC}=V_{CC1}=12\text{V}, V_{CTL}=\text{OPEN}, R_L=24\Omega$	9.5	10.5	-	V
Output Offset Voltage1	VOO1	$V_{IN}=5\text{V}, 5\text{V}$	-40	-	40	mV
Output Offset Voltage2	VOO2	$V_{IN}=0\text{V}, 0\text{V}$	-40	-	40	mV

*.Pulse Testing with Low Duty.

Application Information

1. Driver (Except For Loading Motor Driver)



The voltage, V_{REF} is the reference voltage given by the external bias voltage of the pin #23. The input signal (V_{IN}) through pin #12,20 is amplified by 10k/10k times and then fed to the level shift. The level shift produces the current due to the difference between the input signal and the arbitrary reference signal. The current produced as $+\Delta I$ and $-\Delta I$ is fed into the driver Amp. The driver Amp. operates the power TR. of the output stage as the 2 times gain ($1+10k/10k$) according to the state of the input signal.

$$V_{IN} = V_{REF} + \Delta V \Delta I = \frac{\Delta V}{10K}$$

$$DO+ = V_P + \Delta I \cdot 10K \cdot \left(1 + \frac{10K}{10K}\right) = V_P + 2\Delta V$$

$$DO- = V_P - \Delta I \cdot 10K \cdot \left(1 + \frac{10K}{10K}\right) = V_P - 2\Delta V$$

$$V_{OUT} = (DO+) - (DO-) = 4\Delta V$$

$$GAIN = 20 \log \left(\frac{V_{OUT}}{\Delta V} \right) = 12 \text{dB}$$

If it is desired to change the gain, then the pin #3 or 25 can be used.

The output stage is the balanced transformerless (BTL) driver.

The bias voltage V_P is expressed as

$$\begin{aligned} V_P &= (V_{CC} - V_{BE} - V_{CE(SAT)}) \times \frac{62K}{60K + 62K} + V_{CE(SAT)} \\ &= \frac{V_{CC} - V_{BE} - V_{CE(SAT)}}{1.97} + V_{CE(SAT)} \end{aligned}$$

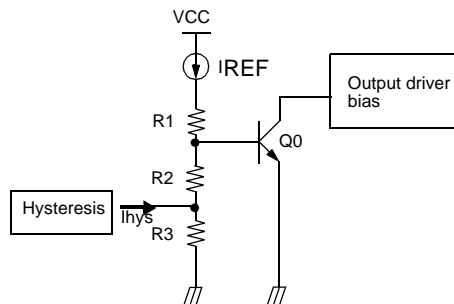
2. Thermal Shutdown

When the chip temperature reaches to 175°C, then the TSD circuit is activated.

This shuts down the bias current of the output drivers, and all the output drivers are in cut-off state. Thus the chip temperature begins to decrease.

when the chip temperature falls to 150°C, the TSD circuit is deactivated and the output drivers are normally operated.

The TSD circuit has the hysteresis temperature of 25°C.



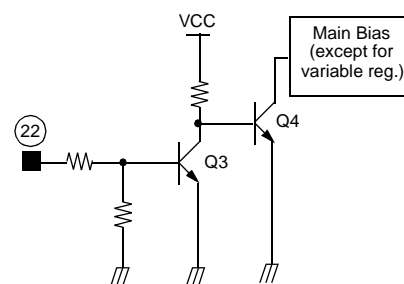
3. Power Save Function

When the pin22 is high, the TR Q3 is turned on and Q4 is off, so the bias circuit is enabled. On the other hand, when the pin22 is Low (GND), the TR Q3 is turned off and Q4 is on, so the bias circuit is disabled.

That is, this function will cause all the circuit blocks of the chip except for the variable regulator to be in the off state. thus the low power quiescent state is established

Truth table is as follows;

Pin#22	FAN8039D3
High	Power Save Off
Low	Power Save On



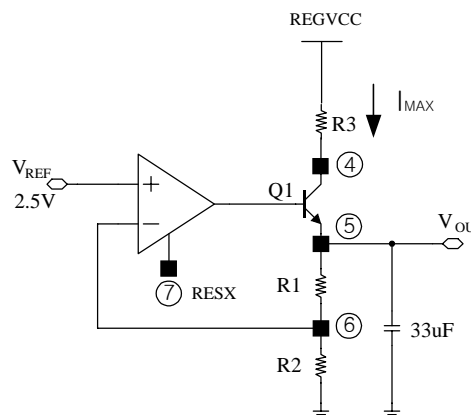
4. Variable Regulator

The VREF is the output voltage of the referenced biasing circuit and is the reference voltage of the regulator. (VREF=2.5V) The external circuit is composed of a capacitor, 33uF, which is used as a ripple eliminator. The output voltage, VOUT is decided as follows.

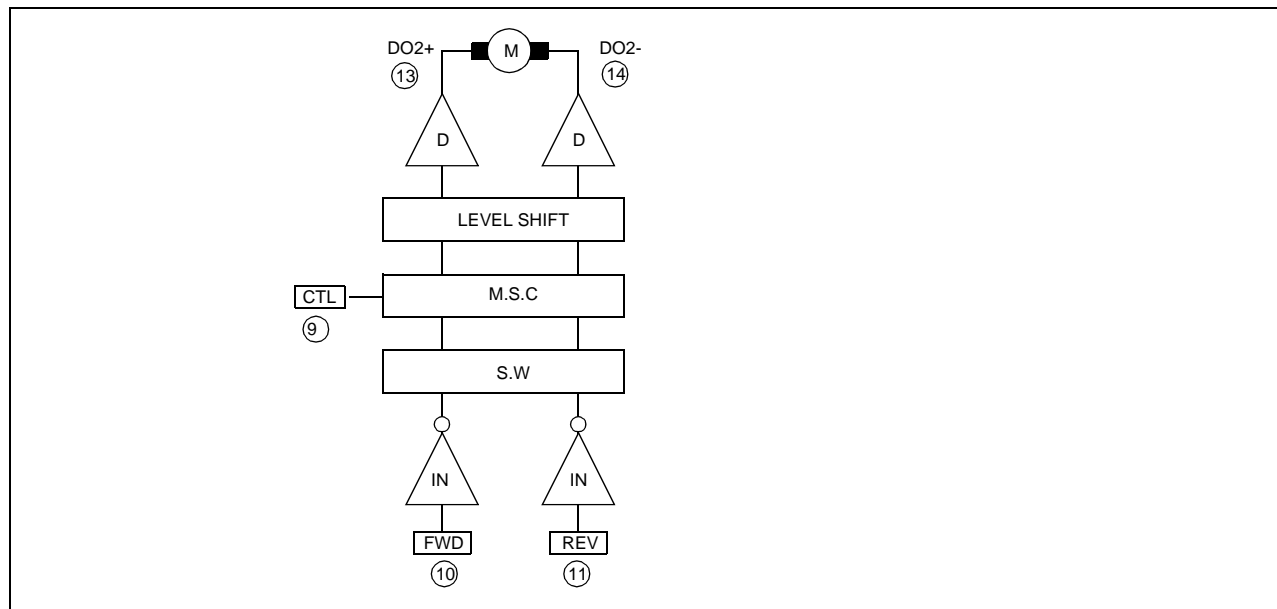
$$V_{OUT} = V_{REF} \cdot \left(1 + \frac{R_1}{R_2}\right) = 2.5 \times 2 = 5V (R_1 = R_2)$$

Resistor R3 should be used, it can reduce the heating problem of regulator output TR Q1. R3 value is decided as follows

$$R_3 = \frac{(REGVCC - (V_{out} + 1.5))}{I_{MAX}}$$



5. Loading Motor Driver

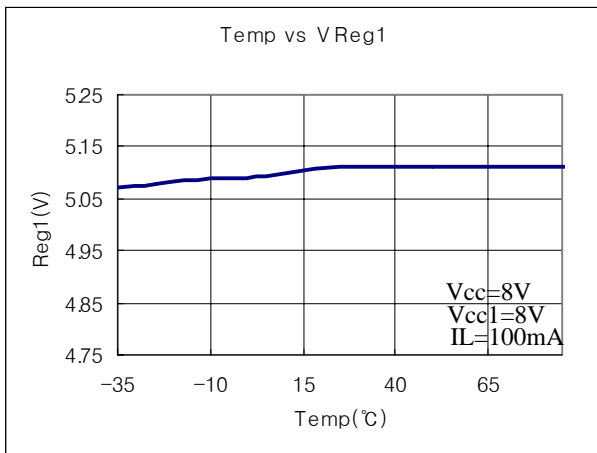
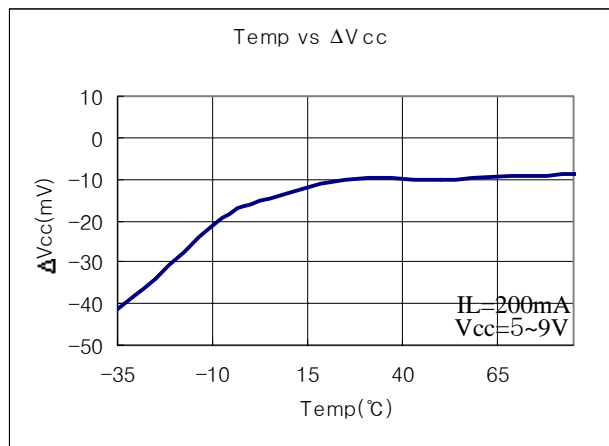
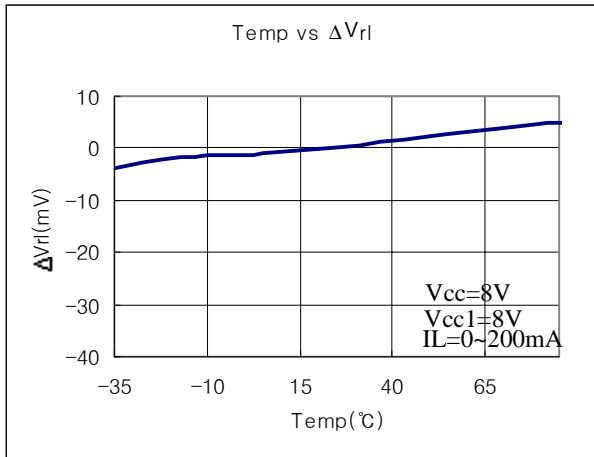
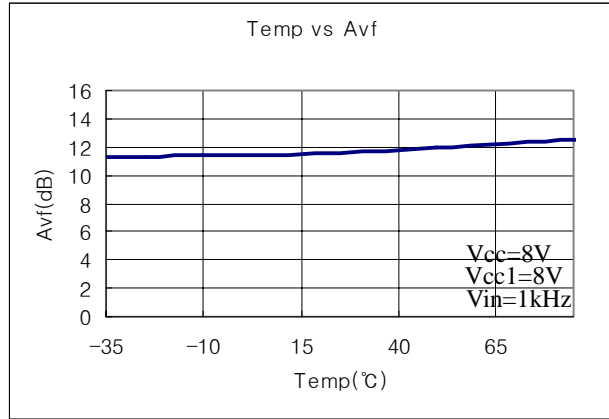
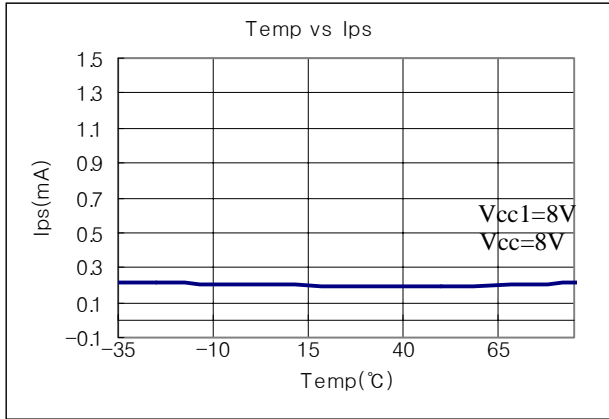


- Rotational direction control
The forward and reverse rotational direction is controlled by FWD (pin10) and REV (pin11) and the input conditions are as follows.

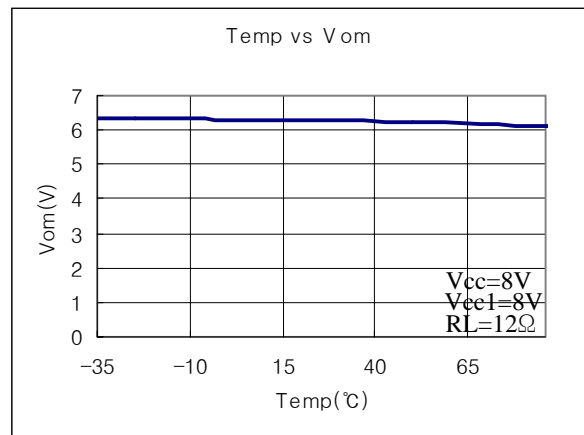
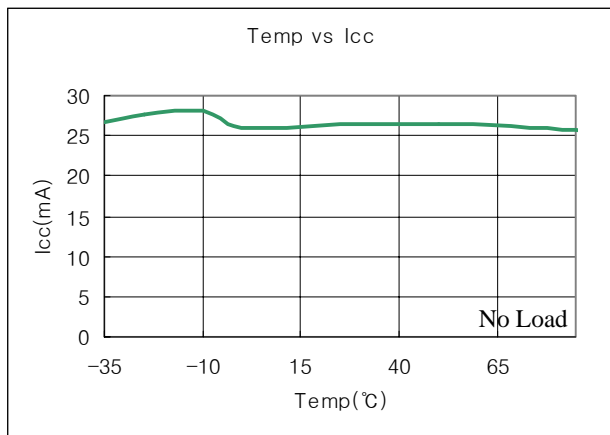
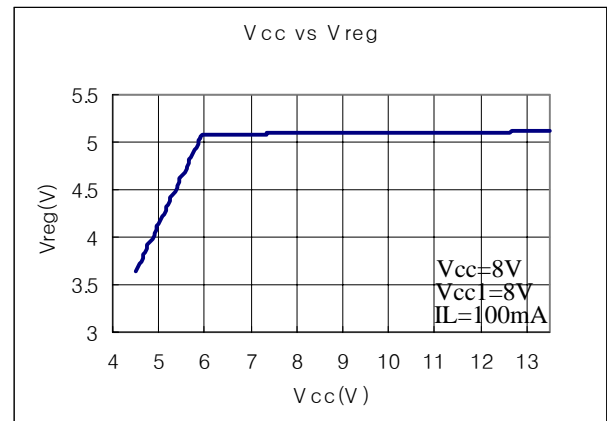
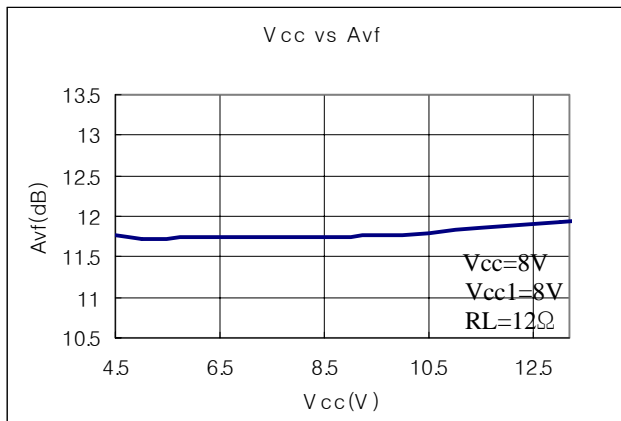
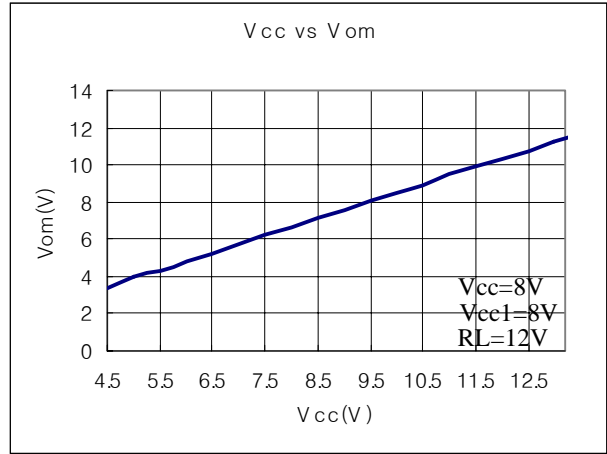
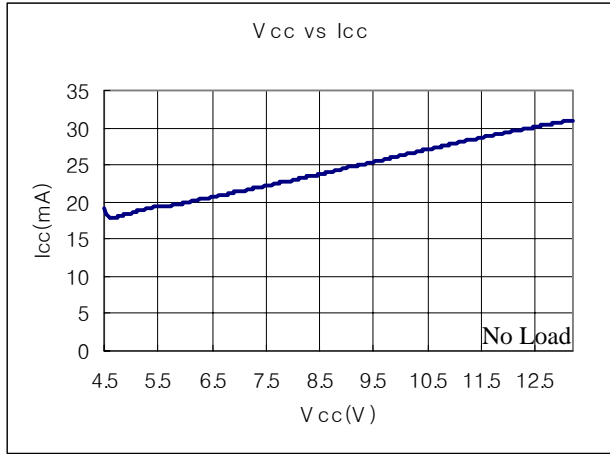
INPUT		OUTPUT		
FWD	REV	OUT 1	OUT 2	State
H	H	Vp	Vp	High Impedence
H	L	H	L	Forward
L	H	L	H	Reverse
L	L	-	-	Short Brake

- Where Vp(Power reference voltage) is approximately about 3.75V at VCC1, VCC = 8V
- Motor speed control (When VCC=VCC1=8V)
 - The almost maximum torque is obtained when the pin 9,(CTL) is open.
 - If the voltage of the pin 9,(CTL) is 0V, the motor will not operate.

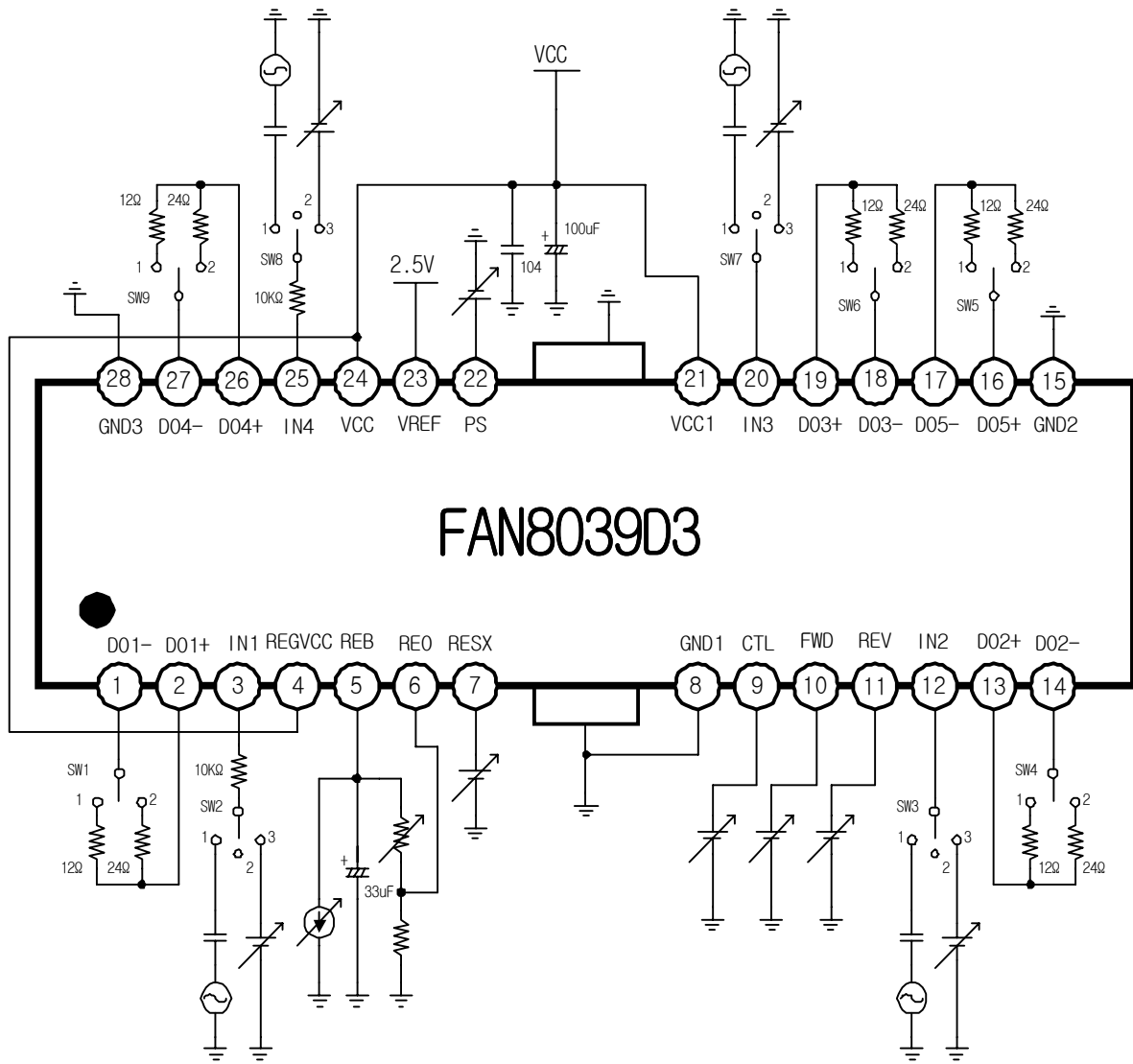
Typical Performance Characteristics



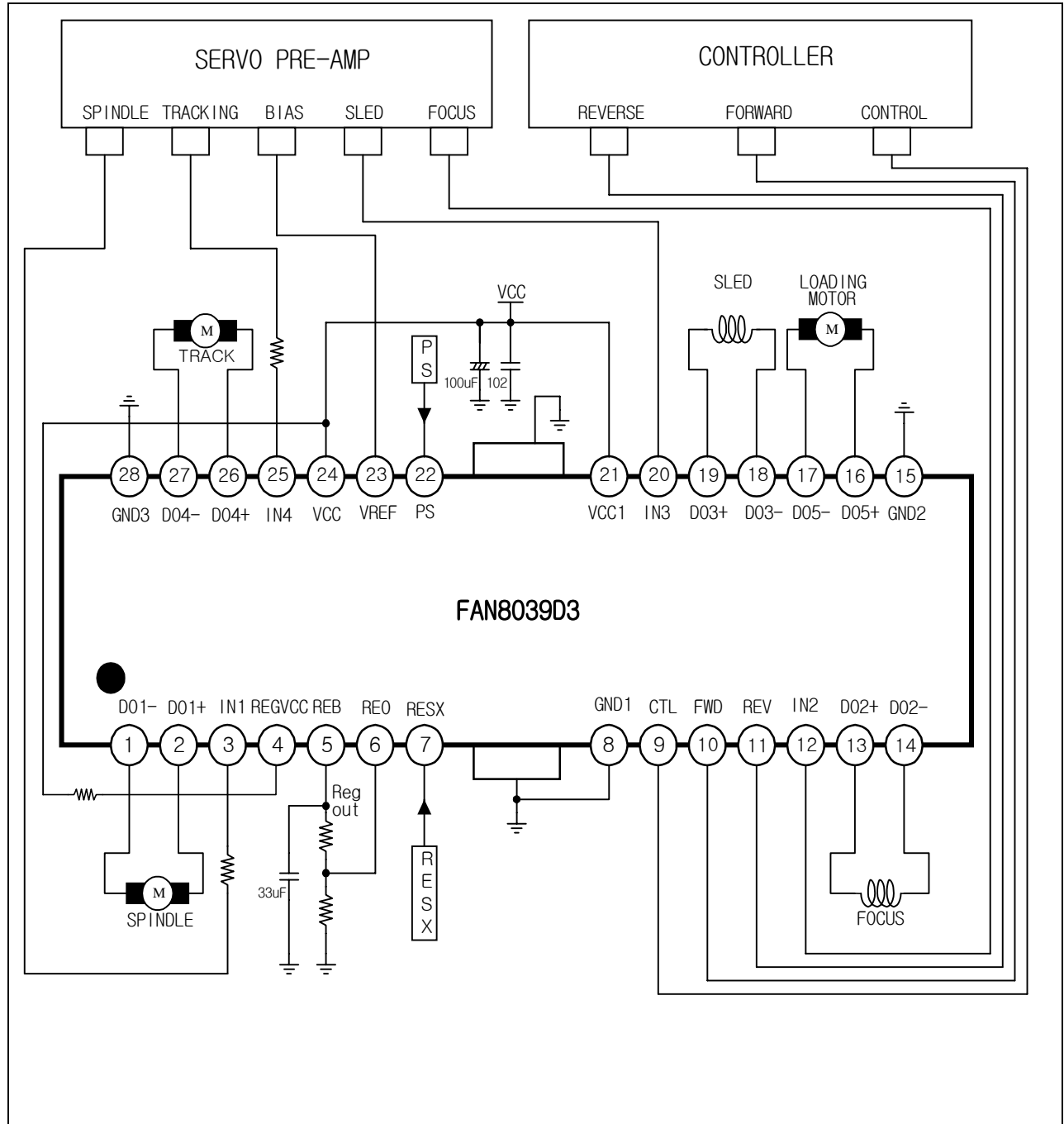
Typical Performance Characteristics



Test Circuit



Application Circuit



DISCLAIMER

FAIRCHILD SEMICONDUCTOR RESERVES THE RIGHT TO MAKE CHANGES WITHOUT FURTHER NOTICE TO ANY PRODUCTS HEREIN TO IMPROVE RELIABILITY, FUNCTION OR DESIGN. FAIRCHILD DOES NOT ASSUME ANY LIABILITY ARISING OUT OF THE APPLICATION OR USE OF ANY PRODUCT OR CIRCUIT DESCRIBED HEREIN; NEITHER DOES IT CONVEY ANY LICENSE UNDER ITS PATENT RIGHTS, NOR THE RIGHTS OF OTHERS.

LIFE SUPPORT POLICY

FAIRCHILD'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT OF FAIRCHILD SEMICONDUCTOR CORPORATION. As used herein:

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and (c) whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury of the user.
2. A critical component in any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.