

AK7714

20-Bit Audio Codec with DSP and Delay RAM

General Description

The AK7714 is a highly integrated audio processing IC with 20-bit stereo ADC and 4-channel DAC plus on-chip DSP with 64-kbit delay RAM. The built-in ADC has a dynamic range of 99 dB with S/(N+D) of 92 dB, and the DAC has a dynamic range of 98 dB with S/(N+D) of 89 dB, thereby ensuring high performances. The AK7714 is compatible with each of the sampling frequencies of 48 kHz, 44.1 kHz or 32 kHz, and a program of 512 steps (when sampling at 44.1 kHz or 32 kHz) can be executed in one sampling time. The AK7714 is a programmable audio DSP containing RAM for delay. This allows sound field control programs to be implemented for surround sound, echo 3D, parametric equalizer and the like.

Furthermore, the AK7714 uses a small 100-pin LQFP package, providing the optimum system for the sound field of car audio equipment which is required to meet space saving requirements.

Features

DSP:

- Word length: 24-bit (Data RAM)

Instruction cycle time: 44.2 ns
 Multiplier: 24 x 16 → 40-bit
 Divider: 24 / 24 → 16-bit

- ALU: 34-bit arithmetic operation (Overflow margin: 4 bits)

24-bit arithmetic and logic operation

- Shift+Register: 1, 2, 3, 4, 8 and 15 bits shifted left

1, 2, 3, 4, (6, 14), 8 and 15 bits shifted right

()Numbers in parentheses are restricted.

Provided with indirect shift function

Program RAM: 448 x 32-bit
 Coefficient RAM: 384 x 16-bit
 Data RAM: 128 x 24-bit

- Internal delay memory: 4096 x 16-bit or 2048 x 24-bit

- Sampling frequency: 32 kHz to 48 kHz

Serial micro controller interface

Master clock: 512/384/256 fs (512 fs for sampling at 44.1 kHz or 32 kHz)

Master/slave operation

- Serial signal input port (2 to 6 ch), output port (2 to 6 ch): 16/20/24-bit

ADC: 2 channels

- 20-bit 64 x Over-sampling delta sigma

DR, S/N: 99 dB
 S/(N+D): 92 dB
 Digital HPF (fc = 1 Hz)

DAC: 4 channels

20-bit 128 x Over-sampling delta sigma

DR, S/N: 98 dB S/(N+D): 89 dB

Others

Power voltage: + 5 V ±5%

Operating temperature range: -40°C to +85 °C

Package: 100-pin LQFP (0.5 mm pitch)

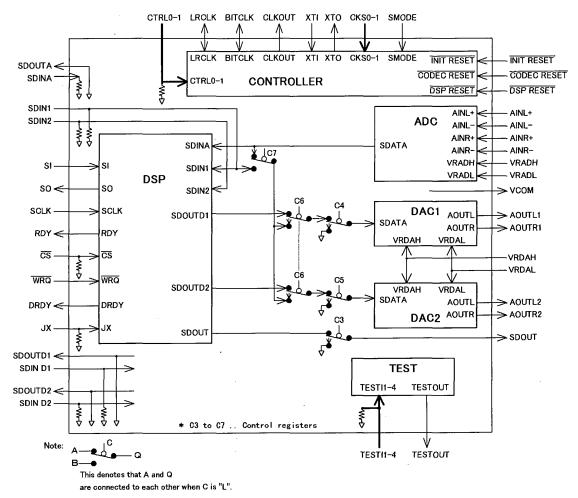
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Block diagram

☐ For standard operation (Internal connection mode, C3 to C7: initial settings)

• CPCL :L Internal connection mode

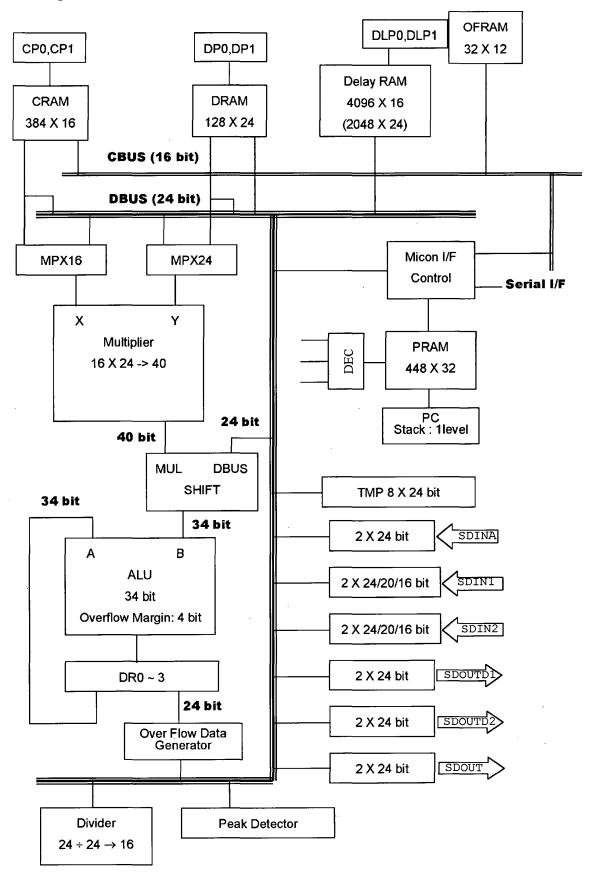
• C3 ~ C7 :All 0 (Initial settings)



This block diagram is a simplified illustration of the AK7714; it is not a circuit diagram.

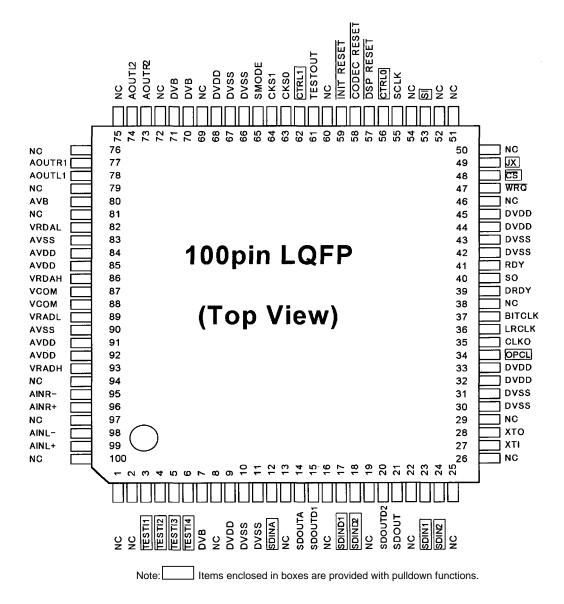
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☐ Block Diagram of AK7714 DSP Section



Description of Input/Output Pins

(1) Pin layout



(2) Pin function

Pin No.	Pin name	I/O	Function	Classification
3	TESTI1	I	Test pin: Leave open or connect to DVSS. (Pulldown)	Test
4	TESTI2	I	Test pin: Leave open or connect to DVSS. (Pulldown)	
5	TESTI3	ı	Test pin: Leave open or connect to DVSS. (Pulldown)	
6	TESTI4	ı	Test pin: Leave open or connect to DVSS. (Pulldown)	
7	DVB	-	+5 V power supply (Silicon substrate potential)	Power supply
9	DVDD	-	+5 V Digital power supply	
10,11	DVSS	-	Digital ground	
12	SDINA	I	DSP Serial data input pin (Pulldown) OPCL="L": Disabled. Leave open or connect to DVSS. OPCL="H": Compatible with MSB first 24 bits	Digital section Serial input/output data
14	SDOUTA	0	ADC Serial data output pin OPCL="L": Outputs "L". OPCL="H": Outputs MSB first 20-bit data.	
15	SDOUTD1	0	DSP Serial data output pin OPCL="L": Outputs "L". OPCL="H": Outputs MSB first 24-bit data.	
17	SDIND1	I	DAC1 Serial data input pin (Pulldown) OPCL="L": Disabled. Leave open or connect to DVSS. OPCL="H": Compatible with MSB first 20 bits	
18	SDIND2	I	DAC2 Serial data input pin (Pulldown) OPCL="L": Disabled. Leave open or connect to DVSS. OPCL="H": Compatible with MSB first 20 bits	
20	SDOUTD2	0	DSP Serial data output pin OPCL="L": Outputs "L". OPCL="H": Outputs MSB first 24-bit data.	
21	SDOUT	0	DSP Serial data output pin Outputs MSB first 24-bit data.	

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Pin No.	Pin name	I/O	Function	Classification
23	SDIN1	I	DSP Serial data input pin (Pulldown) Compatible with MSB first/LSB first 24, 20 and 16 bits	Digital section Serial input data
24	SDIN2	I	DSP Serial data input pin (Pulldown) Compatible with MSB first/LSB first 24, 20 and 16 bits	
27	ХТІ	I	Master clock input pin Connect a crystal oscillator between this pin and the XTO pin, or input the external CMOS clock signal XTI pin. The clock frequency can be selected by the CKS0 and CKS1 pins.	System clock
28	ХТО	0	Crystal oscillator output pin When a crystal oscillator is used, connect it between the XTI pin and this pin. When the external clock is used, keep this pin open.	
30,31	DVSS	-	Digital Ground	Power supply
32,33	DVDD	-	+5 V Digital Power supply	
34	OPCL	ı	ADC/DAC connection selector pin (Pulldown) OPCL="L" (Leave open or connect to DVSS.): Connected OPCL="H": Disconnected	Control
35	CLKO	0	Clock output pin Outputs the XTI clock. Allows the output to be set to "L" using the CTRL0 and CTRL1 pins.	Others
36	LRCLK	I/O	LR channel select Clock pin SMODE="L": Slave mode: Inputs the fs clock. SMODE="H": Master mode: Outputs the fs clock.	System clock
37	BITCLK	I/O	Serial bit clock pin SMODE="L": Slave mode: Inputs 64 fs or 48 fs clocks. SMODE="H": Master mode: Outputs 64 fs clocks.	
39	DRDY	0	Output data ready pin for microcomputer interface.	Microcomputer interface
40	SO	0	Serial data output pin for microcomputer interface (compatible with Hi-Z) Hi-Z state is obtained when \overline{CS} ="H".	
41	RDY	0	Data write ready output pin for microcomputer interface (compatible with Hi-Z) Hi-Z state is obtained when \overline{CS} ="H".	

Pin No.	Pin name	I/O			Function			Classification	
42,43	DVSS	1	Digital Ground	I				Power supply	
44,45	DVDD	-	+5 V Digital po	ower supply					
47	WRQ	ı	Microcompute	r interface Wri	te Request pin	1		Microcomputer interface	
48	CS	_	Microcompute Also used for S When only one unchanged. So	O control in add AK7714 is use O and RDY will	dition to chip se d and SO is n <u>ot</u> be Hi-z when C	lection <u>.</u> : used, CS="L" ı	•		
49	JX	I	External condi	ition jump pin	(Pulldown)			Program control	
53	SI	_	Microcompute	Microcomputer interface serial data input pin (Pulldown)					
55	SCLK	I	Microcompute	r interface ser	ial data clock i	nput pin			
56	CTRL0	ı	Clock output of Use of CTRL0 a outputs in the master m	and CTRL1 allow	ws the CLKO ou			Control	
			CTRL1	CTRL0	CLKO	LRCLK	BITCLK		
			0	0	Output	Output	Output	- -	
			1	0 1	"L"	Output "H"("L")	Output "L"	-	
57	DSP RESET	I	Reset pin_ Normally, DSP use.	Reset pin					
58	CODEC RESET	-							
59	INIT RESET	I	Reset pin (for Used to input "l		e AK7714 at po	wer-on.			

Pin No.	Pin name	I/O		Function					
			Test pin: Leav	e open.				Classification Test	
61	TESTOUT	0							
				ontrol pin (Pu				Control	
62	CTRL1	I		and CTRL1 allov	vs the CLKO	output and LRCLK/BI	rclk .		
			outputs in the master m	nde to be fixed	to "I " or "H"				
				oin 56) & Functi		on.			
			Clock selector		•				
63	CKS0	I	CKS1	CKS0	XTI	DSP			
			0	0	384fs	384fs			
			0	1	512fs	512fs			
64	CKS1	- 1	1	0	256fs	256fa			
			1	1	Test m	ode (disabled)			
			Slave/master i	node selector		·		1	
65	SMODE	- 1	Set LRCLK and	BITCLK to inp	ut or output n				
						CLK are set to input m			
			Digital ground		RCLK and BI	CLK are set to outpu	t mode.)	Power supply	
66.67	DVSS	_	Digital ground					Power supply	
00,01									
			Power supply		section 5 V (typ)			
68	DVDD	-	+5 V Digital p	ower supply					
			+5 V Power si	upply (normally	/ analog)	(Silicon substrate	notential)	-	
70,71	DVB	-			,	(00	,		
70	AOUTDO		DAC2 Rch an	alog output pi	n			Analog section	
73	AOUTR2	0							
			DAC2 Lch an	alog output pii	n			1	
74	AOUTL2	0		- · ·					
			DAGA Bala					-	
77	AOUTR1	0	DAC1 Rch an	alog output pi	n				
''	ACCINI								
			DAC1 Lch and	alog output pin	1			1	
78	AOUTL1	0							
			+5 V power su	ınnlı		(Silicon substrate	notontial	Power supply	
80	AVB	_	TO V POWERSU	ippiy		(Silicon substrate	potential)	r ower suppry	
				e voltage input				Analog section	
82	VRDAL	I	Normally, conn	ect to AVSS (pi	n 83).				
		<u> </u>						1	

Pin No.	Pin name	I/O	Function	Classification
83	AVSS	-	Analog ground 0V	Power supply
84,85	AVDD	-	Power supply pin for analog section 5 V (typ) (Silicon substrate potential) +5 V Analog power supply	
86	VRDAH	ı	DAC Reference voltage input pin Normally, connect to AVDD (pin 84,85), and connect 0.1 μ F and 10 μ F capacitors between this and VRDAL.	Analog section
87,88	VCOM	0	Common voltage pin for analog section Connect 0.1 μ F and 10 μ F capacitors between this and AVSS. Do not use for the external circuit.	
89	VRADL	1	ADC Reference voltage input pin Normally, connect to AVSS (pin 90).	
90	AVSS	-	Analog Ground 0V	Power supply
91,92	AVDD	-	+5 V Analog Power supply (Silicon substrate potential)	
93	VRADH	I	ADC Reference voltage input pin Normally, connect to AVDD (pin 91,92), and connect 0.1 μ F and 10 μ F capacitors between this and VRADL.	Analog section
95	AINR-	I	ADC Rch analog inverted input pin	
96	AINR+	I	ADC Rch analog non-inverted input pin	
98	AINL-	I	ADC Lch analog inverted input pin	
99	AINL+	I	ADC Lch analog non-inverted input pin	

Pin No.	Pin name	I/O	Function	Classification
1,2,8,13,16,19,22,25,26,29,38			NC (No connection)	NC
46,50,51,52,54,60,69,72,75,76	NC	-	These pins should be left floating.	
79,81,94,97,100				

Absolute maximum rating

(AVSS, DVSS = 0 V: All voltages indicated are relative to the ground.)

Item	Symbol	Min	Max	Unit
Power supply voltage				
Analog (AVDD), Boards (DVB and AVB)	VA	-0.3	6.0	V
Digital (DVDD) (Note 1)	VD	-0.3	6.0 or (VA+0.3)	V
Input current (except for power supply pin)	IIN	-10	+10	mA
Analog input voltage	VINA			V
AINL+, AINL-, AINR+, AINR-,		-0.3	VA+0.3	
VRADH, VRADL, VRDAH, VRDAL				
Digital input voltage (Note 1)	VIND	-0.3	VA+0.3	V
Operating ambient temperature	Ta	-40	85	°C
Storage temperature	Tstg	-65	150	°C

Note: 1. Must not exceed the maximum rating of 6.0 V (namely, $VD \le (VA+0.3 \text{ V}) \le 6.0 \text{ V}$).

(VA is a power supply to supply silicon substrate potential.)

WARNING: Operation at or beyond these limits may result in permanent damage of the device.

Normal operations are not guaranteed under these critical conditions in principle.

Recommended operating conditions

(AVSS, DVSS = 0 V: All voltages indicated are relative to the ground.)

Items	Symbol	Min	Тур	Max	Unit
Power supply voltage					
Board (AVB, DVB), AVDD	VA	4.75	5.0	5.25	V
DVDD	VD	4.75	5.0	VA	V
Reference voltage (VREF)					
VRADH, VRDAH	VRH		VA		V
VRADL, VRDAL	VRL		0.0		V

Note: 1. Start up VA simultaneously with or earlier than VD, and stop VD simultaneously with or earlier than VA.

- 2. When starting and stopping the power supply, meet the absolute maximum rating condition: VD ≤ (VA+0.3 V). It is generally recommended to use at VD ≤ VA. However, the VD must be 4.75 volts or more.
- 3. The analog input voltage and output voltage are proportional to the VRADH and VRDAH voltages.

Electric characteristics

(1) Analog characteristics

(Unless otherwise specified, $Ta = 25^{\circ}C$; AVDD, DVDD, AVB, DVB = 5.0 V; VRADH = AVDD, VRADL = AVSS, VRDAH = AVDD, VRDAL = AVSS; fs = 44.1 kHz; BITCLK = 64 fs; XTI = 256 fs; Signal frequency 1 kHz; measuring frequency = 10 Hz to 20 kHz; 20 bits; DSP section in the reset state; ADC with all ifferential inputs)

	Pa	rameter	Min	Тур	Max	Unit
ADC	Resolution				20	Bits
section	Dynamic characterist	ics				
	S/(N+D)	(-0.5 dB) (Note 1)	86	92		dB
	Dynamic range	(A filter) (Note 2)	94	99		dB
	S/N	(A filter)	94	99		dB
	Inter-channel isolation	(f =1 kHz)	90	105		dB
	DC accuracy					
	Inter-channel gain misr	natching		0.1	0.3	dB
	Gain drift			50		ppm/°C
	Analog input					
	Input voltage	(Note 3)	±1.9	±2.0	±2.1	Vp-p
	Input impedance			220		kΩ
DAC	Resolution				20	Bits
section	Dynamic characterist	ics			20	
	S/(N+D)	(0 dB)	83	89		dB
	Dynamic range	(-60 dB) (A filter) (Note 2)	93	98		dB
	S/N	(A filter)	93	98		dB
	Inter-channel isolation	(f = 1 kHz) (Note 4)	90	105		dB
	DC accuracy					
	Inter-channel gain misr	natching (Note 4)		0.2	0.5	dB
	Gain drift			50		ppm/°C
	Analog output					
	Output voltage	(Note 5)	2.70	2.95	3.20	Vp-p
	Load resistance		5			kΩ

Note:

- 1. Single end input will result in poorer characteristics.
- 2. Indicates S/(N+D) when -60 dB signal is input.
- 3. The full scale for analog input voltage ($\triangle AIN = (AIN+) (AIN-)$) can be represented By ($\pm FS = \pm (VRADH-VRADL) \times 0.4$).
- 4. Specified for L and R of each DAC.
- 5. The full-scale voltage (0 dB) and output voltage are proportional to VRDAH voltage. Analog output voltage (Typ. @ 0 dB) = 2.95 Vpp*VRDAH/5

(2) DC characteristics

 $(VDD = AVDD = DVDD = AVB = DVB = 5.0 V \pm 5\%, Ta = 25$ °C)

Parameter	Symbol	Min	Тур	Max	Unit
High level input voltage (Note 1)	VIH				٧
Input pins other than XTI and test pins		2.4			V
XTI and test pins		70% VDD			
Low level input voltage (Note 1)	VIL				
Input pins other than XTI and test pins				0.6	
XTI and test pins				30% VDD	
High level output voltage lout=-100μA	VOH	VDD-0.5			V
Low level output voltage lout=100µA	VOL			0.5	V
Input leak current (Note 2)	lin	-10		±10	μΑ
Input leak current Pulldown pins (Note 3)	lid		100		μ A

Note: 1. The test pins are as follows: TESTI1, TESTI2, TESTI3, TESTI4

- 2. The pulldown pins are not included.
- 3. The pulldown pins are as follows (Typ 50 k Ω): <u>TESTI1</u>, TESTI2, TESTI3, TESTI4, SDINA, SDIND1, SDIND2, SDIN1, SDIN2, OPCL, <u>CS</u>, JX, SI, CTRL0, CTRL1

Note: Regarding the input/output levels in the text, the low level is represented as "L" or 0, and the high level as "H" or 1. In principle, "0" and "1" are used for the bus (serial/parallel), such as registers.

(3) Current consumption

(AVDD = AVB, DVB, DVDD = $5.0 \text{ V}\pm5\%$, Ta = 25°C ; master clock (XTI) = 22.5792MHz = 512 fs [fs = 44.1kHz]; when operating for DAC 4 channel with 1 kHz sinusoidal wave full-scale input to each of ADC 2 ch analog input pins)

Power supply				
Parameter	Min	Тур	Max	Unit
Power supply current				
1) During operation				
a) AVDD + AVB + DVB		41		mA
b) DVDD (Note 1)		87		mA
c) To <u>tal (a+b)</u>		128	162	mA
2) When INIT RESET= "L"(reference value)		7		mA
Note 2				
Power consumption				
1) During operation				
a) AVDD+AVB+DVB		205		mW
b) DVDD (Note 1)		435		mW
c) Total(a+b)		640	850	mW
2) When INIT RESET ="L" (reference value)				mW
Note 2		35		

Note: 1. Varies slightly according to the frequency used and contents of the DSP program.

Note: 2. This is a reference value in case of using the crystal oscillator. But, varies slightly according to the types of crystal oscillators.

(4) Digital filter characteristics

Values described below are design values cited as references. These are not for guaranteeing the characteristics.

1) ADC Section:

(Ta = 25°C; AVDD, DVDD, AVB, DVB = 5.0 V $\pm 5\%$; fs = 44.1 kHz)

Parameter	Symbol	Min	Тур	Max	Unit
Pass band (-0.02 dB)	PB	0		20.00	kHz
(-6.0 dB)		0		22.05	kHz
Stop band (Note 1)	SB	24.35			kHz
Pass band ripple (Note 2)	PR			±0.005	dB
Stop band attenuation (Notes 3,4)	SA	80			dB
Group delay distortion	ΔGD			0	μS
Group delay (Ts = 1/fs)	GD		29.3		Ts

Note: 1. These frequencies scale with the sampling frequency (fs).

- 2. The pass band is from DC to 19.75 kHz when fs = 44.1 kHz.
- 3. The stop band is from 27.56 kHz to 2.795 MHz when fs = 44.1 kHz.
- 4. When fs = 44.1 kHz, the analog modulator samples analog input at 2.8224 MHz. The input signal is not attenuate dBy the digital filter in the multiple bands (n x 2.8224 MHz \pm 20.21 kHz; n = 0, 1, 2, 3...) of the sampling frequency.

2) DAC section

(Ta = 25°C; AVDD, DVDD, AVB, DVB = 5.0 V \pm 5%; fs = 44.1 kHz)

Parameter		Symbol	Min	Тур	Max	Unit
Digital filter						
Pass band ±0.07 dB	(Note 1)	PB	0		20.0	kHz
(-6.0 dB)			-	22.05	-	kHz
Stop band	(Note 1)	SB	24.1			kHz
Pass band ripple		PR			±0.07	dB
Stop band attenuation		SA	47			dB
Group delay (Ts = 1/fs)	(Note 2)	GD	-	15.3		Ts
Digital filter + Analog filter						
Amplitude characteristics				±0.5		dB
0 to	o 20.0 kHz					

Note: 1. The pass band and stop band frequencies are proportional to "fs" (system sampling rate), and are represente PB = 0.4535 fs(@ -0.06 dB) and SB = 0.546 fs, respectively.

The caluculating delay time which occurred by digital filtering.
 This time is from setting the 20-bit data of both channels on input register to the output of analog signal.

(5) Switching characteristics

1) System clock

(AVDD = AVB, DVB, DVDD = $5.0 \text{ V}\pm5\%$, Ta = 25°C , CL = 20 pF)

Parameter	Symbol	Min	Тур	Max	Unit
Master clock (XTI)					
a) With a crystal oscillator:					
256 fs: Frequency	fMCLK	11.0	11.2896	12.288	MHz
384 fs: Frequency	fMCLK	12.288	16.9344	18.432	MHz
512 fs: Frequency	fMCLK	16.384	22.5792	22.5792	MHz
b)With an external clock input:					
Duty factor (≤ 18.432 MHz)		40	50	60	%
(> 18.432MHz)		45	50	55	
256 fs: Frequency	fMCLK	11.0	11.2896	12.288	MHz
: High level width	tMCLKH	30			
: Low level width	tMCLKL	30			
384 fs: Frequency	fMCLK	12.288	16.9344	18.432	MHz
: High level width	tMCLKH	20			
: Low level width	tMCLKL	20			
512 fs: Frequency	fMCLK	16.384	22.5792	22.5792	MHz
: High level width	tMCLKH	17			
: Low level width	tMCLKL	17			
Clock rise time	tCR			6	ns
Clock fall time	tCF			6	ns
LRCLK Sampling frequency	fs	32	44.1	48	kHz
			1		fs
Slave mode: Clock rise time	tLR			10	ns
Slave mode: Clock fall time	tLF			10	ns
BITCLK	fBCLK	48	64		fs
Slave mode: High level width	tBCLKH	100			
Slave mode: Low level width	tBCLKL	100			
Slave mode: Clock rise time	tBR	•		6	ns
Slave mode: Clock fall time	tBF			6	ns

2) Reset

(AVDD = AVB, DVB, DVDD = 5.0 V \pm 5%, Ta = 25°C, CL = 20 pF)

Paramet	ter	Symbol	Min	Тур	Max	Unit
INIT RESET	(Note 1)	tRST	150			ns
DSP RESET	(Note 1)	tRST	150			ns
CODEC RESET	(Note 1)	tRST	150			ns

Note 1. "L" is acceptable when power is turned on.

3) Audio interface

(AVDD = AVB, DVB, DVDD = 5.0 V $\pm 5\%$, Ta = 25°C, CL = 20 pF)

Parameter	Symbol	Min	Тур	Max	Unit
Slave mode					
BITCLK frequency	fBCLK	48	64		fs
BITCLK low level width	tBCLKL	100			ns
BITCLK high level width	tBCLKH	100			ns
Delay time from BITCLK"↑" to LRCLK	tBLRD	40			ns
Delay time from LRCLK to BITCLK "↑"	tLRBD	40			ns
Delay time from LRCLK to serial data output	tLRD			70	ns
Delay time from BITCLK to serial data output	tBSOD			70	ns
Serial data input latch hold time	tBSIDS	40			ns
Serial data input latch setup time	tBSIDH	40			ns
Master mode					
BITCLK frequency	fBCLK		64		fs
BITCLK duty factor			50		
Delay time from BITCLK"↑" to LRCLK	tBLRD	40			ns
Delay time from LRCLK to BITCLK"↑"	tLRBD	40			ns
Delay time from LRCLK to serial data output	tLRD			70	ns
Delay time from BITCLK to serial data output	tBSOD			70	ns
Serial data input latch hold time	tBSIDS	40			ns
Serial data input latch setup time	tBSIDH	40			ns

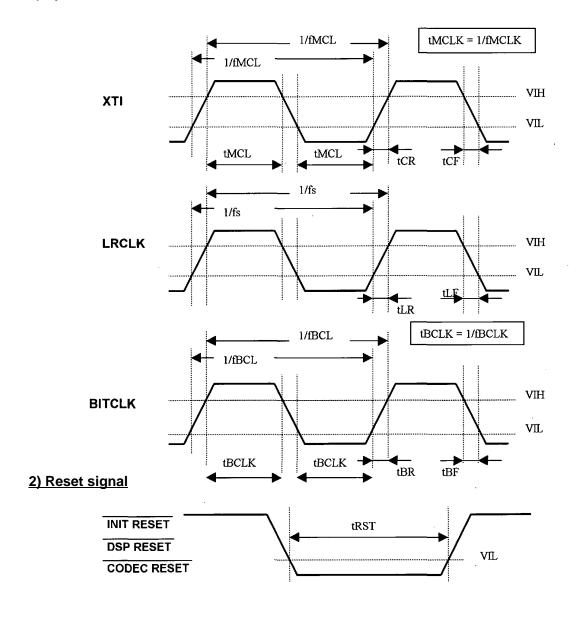
4) Microcomputer interface

(AVDD = AVB, DVB, DVDD = 5.0 V \pm 5%, Ta = 25°C, CL = 20 pF)

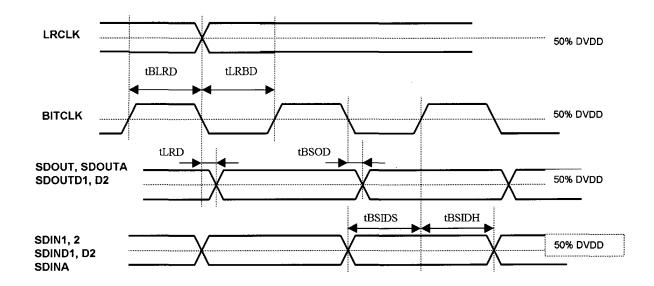
Parameter	Symbol	Min	Тур	Max	Unit
Microcomputer interface signal					
CS Fall time	tCSF			10	ns
CS Rise time	tCSR			10	ns
Rest time until RDY and SO Hi-Z states from CS"↓"	tCSHR			100	ns
Set time until RDY and SO Hi-Z states					
from CS"↑"	tCSHS			200	ns
WRQ fall time	tWRF			10	ns
WRQ rise time	tWRR			10	ns
SCLK fall time	tSF			10	ns
SCLK rise time	tSR			10	ns
SCLK low level width	tSCLKL	150			ns
SCLK high level width	tSCLKH	150			ns
Microcomputer to AK7714					
Time from RESET"↓" to WRQ"↓"	tREW	200			ns
Time from WRQ"↑" to RESET"↑"	tWRE	200			ns
WRQ high level width	tWRQH	200			ns
Time from WRQ "↓" to SCLK"↓"	tWSC	200			ns
Time from SCLK"↑" to WRQ"↑"	tSCW	6×tMCLK			ns
SI latch setup time	tSIS	100			ns
SI latch hold time	tSIH	100			ns
AK7714 to microcomputer					
CS high level width	tCSH	200			ns
Time from SCLK"↑" to CS"V"	tSCS	6×tMCLK			ns
Time from CS"↓" to SCLK"↓"	tCSC	200			ns
Time from CS"↑" to DRDY"↑"	tCSDR			100	ns
Delay time from SCLK "↓" to SO output	tSOS			100	ns
AK7714 to microcomputer					
(RAM DATA read-out)					
SI latch setup time (SI="H")	tRSISH	100			ns
SI latch setup time (SI="L")	tRSISL	100			ns
SI latch hold time	tRSIH	100			ns
Time from SCLK "↓" to SO (PRAM)	tSOPD			100	ns
Time from SCLK "↓" to SO (CRAM, OFRAM)	tSOCOD			100	ns

(6) Timing Diagram

1) System clock

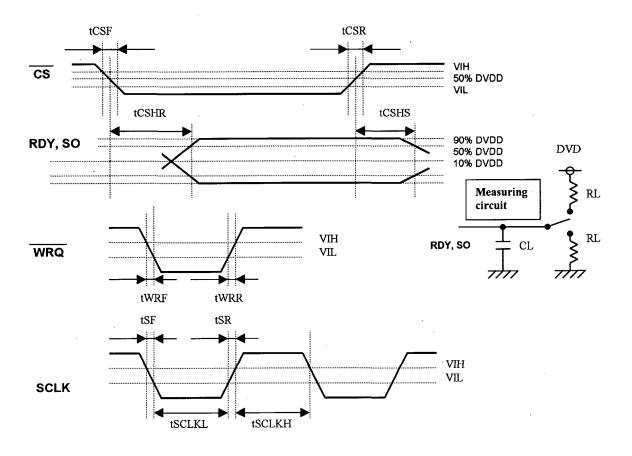


3) Audio interface



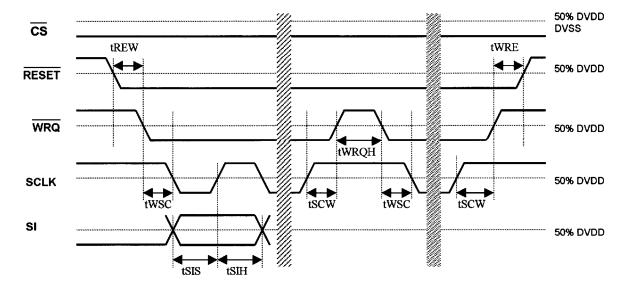
4) Microcomputer interface

■ Microcomputer interface signals



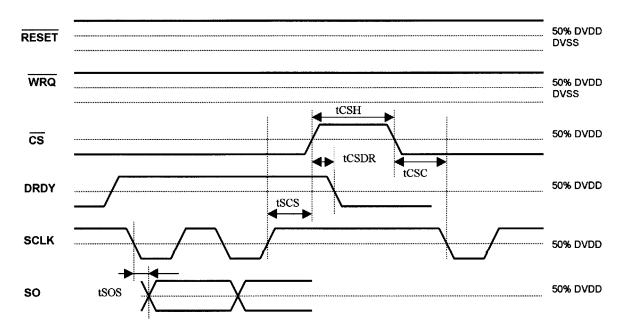
[ASAHI KASEI] [AK7714]

■ Microcomputer to AK7714

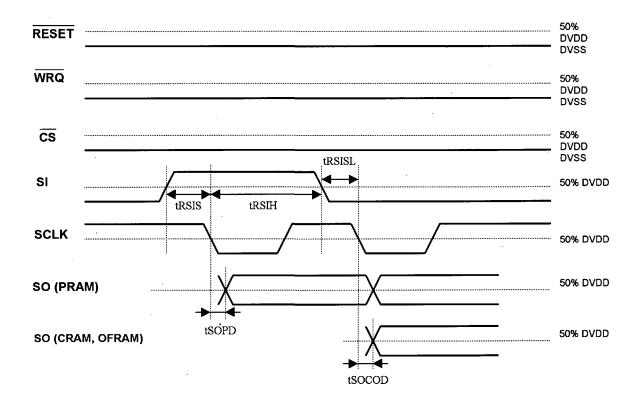


Note : Timing for RUN state is the same except that $\overline{\text{RESET}}$ is set to "H" $\overline{\text{RESET}}$ represents system reset in normal use.

■ AK7714 to Microcomputer



■ AK7714 to Microcomputer (RAM data read-out)



Function Description

(1) Various settings

1) OPCL (pin 34): ADC and DAC connection selector pin (See Block Diagram on page 2.)

□ Normally, OPCL is used in "L" or open. (Internal connection mode)
In this case, ADC output and DAC1/DAC2 inputs are directly connected to the DSP internally. At this time, leave the SDINA (pin 12), SDIND1 (pin 17) and SDIND2 (pin 18) open or set to "L".

It should be noted that "L" is output from the SDOUTA (pin 14), SDOUTD1 (pin 15) and SDOUTD2 (pin 20).

☐ When the OPCL is set to "H", the ADC output and DAC1/DAC2 inputs can be used independently from the DSP. (Input/output formats are restricted.)

Note: SDINA supports only MSB-first 24-bit input (including I²S compatibility). SDOUTA supports only MSB-first 20-bit output (including I²S compatibility).

SDIND1 and SDIND2 support only MSB-first 20-bit inputs (including I²S compatibility).

2) CTRL1 (pin 62) and CTRL0 (pin 56): clock output control pins

CLKO output and LRCLK and BITCLK outputs in the master mode can be fixed to "L" or "H" by setting these two pins.

Mode				Master mode	Master mode
	CTRL1	CTRL0	CLKO	LRCLK	BITCLK
1	0	0	Output	Output	Output
2	0	1	Disabled (This is a test mode, so do not use it.)		
3	1	0	"L"	Output	Output
4	1	1	"L"	"H" (For "L", see Note.)	"L"

Note: When CTRL1 and CTRL0 are used in the open state, Mode 1 will be selected.

Output is set to "L" when I2S compatible.

Mode 4 can be used only when the AK7714 is used "Analog to Analog".

3) CKS1 (pin 64) and CKS0 (pin 63): Clock selector pin

CKS1	CKS0	XTI	DSP
0	0	384fs	384fs
0	1	512fs	512fs
1	0	256fs	256fs
1	1	Test mode (disabled)	

fs: Sampling frequency

4) SMODE (pin 65): Slave and master mode selector pin

Sets LRCLK (pin 36) and BITCLK (pin 37) to either input or output.

a) Slave mode: SMODE = "L"

LRCLK (1 fs) and BITCLK (48 fs or 64 fs) become input.

b) Master mode: SMODE = "H"

LRCLK (1 fs) and BITCLK (64 fs) become output.

5) Control registers

The control registers (16 bits) can be set via the microcomputer interface in addition to the control pins. For the value to be written in the control registers, see the description of the microcomputer interface. The following describes the control register map.

Control register map (indicates the default values.)

Code	Function			
C15	Selects the data reset function after reset is rele	eased. 0: Used	1: Unused	
C14	Selects delay RAM sampling	0: 1Sampling	1: 2 Sampling	
C13	Selects DRAM addressing method	0: Ring addressing	1: Linear addressing	
C12	Random number generator circuit	0: Unused	1: Used	
C11	Test mode (Use at 0)	0: Normal operation	1: Test mode	
C10	Test mode (Use at 0)	0: Normal operation	1: Test mode	
C9	Test mode (Use at 0)	0: Normal operation	1: Test mode	
C8	Resets ADC section	0: Normal operation	1: Reset	
C7	Sets internal path	0: ADC serial data selected	1: SDIN1 selected	
C6	Sets internal path	0: Normal setting	1: Sets the path selected by C7	
C5	DAC2 section reset control	0: Normal operation	1: DAC2 section reset	
C4	DAC1 section reset control	0: Normal operation	1: DAC1 section reset	
C3	SDOUT output enable	0: SDOUT = Output	1: SDOUT = "L"	
C2 C1	2 0 1 LS 3 1 0 LS	SB first (24 bits) SB first (24 bits) SB first (20 bits) SB first (16 bit) independent of mode setting;	however, set to Mode 1.	
	1: I ² S compatible (In this case, all input/outpu	t pins are I ² S compatible.)		

Data can be loaded into the control registers only when \overline{DSP} RESET = "L" and \overline{CODEC} RESET = "L". If used otherwise, an operation error will occur. To avoid an operation error, do not use it.

a) C0, C1, C2

See (5) Audio data interface (internal connection mode)

b) C3

In case of not using the SDOUT, if this code is set "1" then the SDOUT outputs "L" level.

c) C4

In case of using only DAC2, this code can set to "1" and DAC1 will RESET. It can useful for saving the power consumption of DAC1. The output signals of AOUTL1 and AOUTR1 will be Hi-z.

d) C5

In case of using only DAC2, this code can set to "1" and DAC1 will RESET. It can useful for saving the power consumption of DAC1. The output signals of AOUTL1 and AOUTR1 will be Hi-z.

e) C6, C7

Normally C6 and C7 are "0" setting. In detail, please see (8) Special use, 2) Use as ADC and DAC (mainly for test).

f) C8

In case of not using the ADC part, this code can RESET the ADC part. In this case, the digital output from ADC will be "00000h" and it will save the power consumption of the ADC part.

g) C9, C10, C11

This is test mode. C9, C10 and C11 should be "0".

h) C12

This DSP has a single feedback type shift-register [24,21,19,18,17,16,15,14,14,19,9,5,1]s independently from calculation block.

This register change the data in every sampling time. And its output connected with DBUS, so in case of selected MSRG command at program code, then 24-bit random data will appear in every sampling . In case of using this register, please set this code to "1".

i) C13

This code sets the addressing method of DRAM (Data Ram).

C13 = 0 : Ring addressing

C13 = 1: Linear addressing.

DRAM has 128-word x 24-bit and has 2 addressing pointers (DP0, DP1).

The Ring addressing mode: Its start address increments 1 by every sampling time.

The Linear addressing mode: Its start address is always same, DP0 = 00h and DP1 = 40h.

j) C14

This code is setting for DLYRAM (internal 4k-word x 16-bit Delay RAM)sampling method.

Normally C14 = 0, this means its address pointer will work as ring addressing by every sampling.

If it set C14 = 1, this means the address pointer will work as ring addressing by every 2 sampling.

This is a decimation mode and it can extend delay time. But, it will appear aliasing.

When it is C14=1 mode, the Delay Ram area will consist 3 banks.

The bank 1 is from 000h to 3FFh: (The address of this area is always 1 sampling ring addressing.)

The bank 2, bank 3 and bank 4 (400h to 7FFh, 8FFh to BFFh and C00h to FFFh) can set to decimation mode.

k) C15

Normally it should be C15=0. At this time after release the system reset, DRAM (Data Ram) and DLYRAM (Internal delay Ram) will be clear to "0" .

It takes 8LRCLK at 512fs mode, 11LRCLK at 384fs mode and 16LRCLK at 256fs mode after the Reset pulse comes out. The Reset pulse comes out at first rising point of LRCLK at master mode and in case of slave mode, it comes out about after 3LRCLK.

(2) Power supply startup sequence

Turn on the power by setting to $\overline{\text{INIT RESET}} = \text{"L"}, \overline{\text{DSP RESET}} = \text{"L"}$ and $\overline{\text{CODEC RESET}} = \text{"L"}$. Then the AK7714 is initialized by setting to $\overline{\text{INIT RESET}} = \text{"H"}$. Note 1) Initialization by $\overline{\text{INIT RESET}}$ is sufficient if it is done only when the power is turned on.

Note1: Set to INIT RESET = "H" after setting the oscillation when a crystal oscillator is used.

This setting time may differ depending on the crystal oscillator and its external circuit.

2: Do not stop the system clock (slave mode: XTI, LRCLK, BITCLK, master mode: XTI) except when INIT RESET = "L".

If these clock signals are not supplied, too much current will flow because the dynamic logic is used internally, and an operation failure may result.

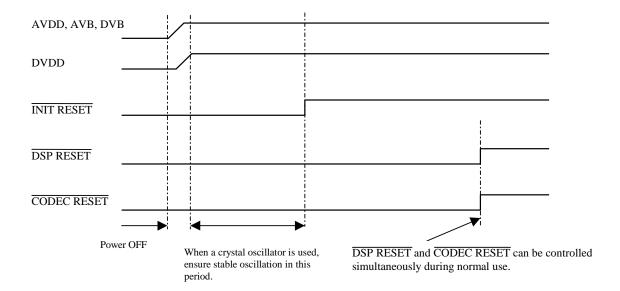


Fig. Power supply startup sequence

(3) Resetting

The AK7714 has three reset pins: INIT RESET, DSP RESET and CODEC RESET.

The INIT RESET pin is used to initialize the AK7714, as shown in "Power supply startup sequence above."

DSP RESET and CODEC RESET are normally controlled simultaneously. The system is reset when

DSP RESET = "L" and CODEC RESET = "L". (Description of "reset" is for "system reset".)

Under the condition of this system reset, the program write operation is performed (except for write operation during running).

During the system reset phase, the ADC and DAC sections are also reset. (The digital section of ADC output is MSB first 00000h and the analog section of DAC output is Hi-z.)

CLKO is output even during the system reset phase if CTRL (1:0) = 0h (Mode 1), but LRCLK and BITCLK in the master mode will stop.

The system reset is released by setting either DSP RESET or CODEC RESET to "H", and this will activate the internal counter. LRCLK and BITCLK in the master mode are generated by this counter: however, a hazard may occur when a clock signal is generated. When the system reset is released in the slave mode, internal timing will be actuated in synchronization with "↑" of LRCLK (when the standard input format is used). Timing between the external and internal clocks is adjusted at this time. If the phase difference in LRCLK and internal timing is within about -1/16 to 1/16 of the input sampling cycle (1/fs) during the operation, the operation is performed with internal timing remaining unchanged. If the phase difference exceeds the above range, the phase is adjusted by synchronization with "↑" of LRCLK (when the standard input format is used). This is a circuit to prevent failure of synchronization with the external circuit owing to noise and the like. For some time after returning to the normal state after loss of synchronization, normal data will not be output. If you want to change the clock, do so while the system is reset.

The ADC section can output 516-LRCLK after its internal counter started. (The internal counter starts at the first rising edge of LRCLK at master mode. In case of slave mode, it starts end of 2LRCLK after release of system reset.)

The AK7714 performs normal operation when both DSP RESET and CODEC RESET are set to "H".

(4) System clock

The required system clock is XTI (256 fs/384 fs/512 fs), LRCLK (fs) and BITCLK (64 fs) in the slave mode, and is XTI (256 fs/384 fs/512 fs) in the master mode.

LRCLK corresponds to the standard digital audio rate (32 kHz, 44.1 kHz, 48 kHz).

fs	XTI	(Master clock)		BITCLK
	256 fs	384 fs	512 fs	64 fs
32.0 kHz	- Note	12.2880 MHz	16.3840 MHz	2.0480 MHz
44.1 kHz	11.2896 MHz	16.9344 MHz	22.5792 MHz	2.8224 MHz
48.0 kHz	12.2880 MHz	18.4320 MHz	- Note	3.0720 MHz

Note: 256 fs is not supported at fs = 32.0 kHz. 512fs is not supported at fs = 48.0 kHz.

SMODE	CKS1	CKS0	XTI	LRCLK, BITCLK
L	L	L	384 fs	Input
L	L	Η	512 fs	Input
L	Η	L	256 fs	Input
Н	L	L	384 fs	Output
Н	L	Н	512 fs	Output
Н	Η	L	256 fs	Output

1) Master clock (XTI pin)

The master clock is obtained by connecting a crystal oscillator between the XTI pin and XTO pin or by inputting an external clock into the XTI pin while the XTO pin is open.

CLKO outputs a clock having the same frequency as the XTI. (The phase is different. Also, the phase difference varies slightly according to the product. When a crystal oscillator is used, the same frequency is obtained after the oscillation of the crystal oscillator has settled.)

In the AK7714, CLKO will be output when power is turned on and oscillation has occurred. If CLKO is not required, set to CTRL1 = "H".

2) Slave mode

The required system clock is XTI (256 fs/384 fs/512 fs), LRCLK (1 fs) and BITCLK (48/64 fs). The master clock (XTI) and LRCLK must be synchronized, but the phase does not have to be adjusted.

3) Master mode

The required system clock is XTI (256 fs/384 fs/512 fs). When the master clock (XTI) is input, LRCLK (1 fs) and BITCLK (64 fs) will be output from the internal counter synchronized with the XTI. (CTRL0 = "L" during normal operation.)

LRCLK and BITCLK will not be output during initial reset (INIT RESET = "L") and system reset (DSP RESET = CODEC RESET = "L").

(5) Audio data interface (internal connection mode)

The serial audio data pins SDIN1, SDIN2 and SDOUT (OPCL = L: Internal connection mode) are interfaced with the external system, using LRCLK and BITCLK. The data format is the first MSB of the 2's complement. Normally, the input/output format, in addition to the standard format used by AKM, can be changed to the I^2S compatible mode by setting the control register C0 to 1. (In this case, all input/output audio data pin interface is in the I^2S compatible mode.) The input SDIN1 and SDIN2 formats are adjusted (24 bits) at initialization. Setting the control registers C2 and C1 will cause them to be compatible with 24 bits, 20 bits and 16 bits. (Note: C0 = 0)

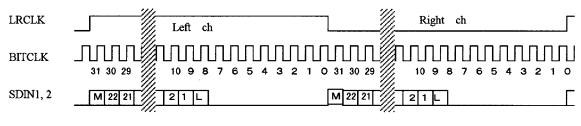
However, inidividul setting of SDIN1 and SDIN2 is not allowed.

The output SDOUT is fixed at 24 bits.

At slave mode BITCLK corresponds to not only 64fs but also 48fs. But, we recommend 64fs. Following formats describe 64fs examples.

1) Standard input format (C0 = 0: initial set value)

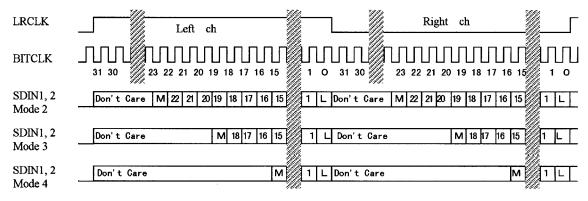
a) Mode 1 (C2, C1 = 0, 0: Initial set value)



M: MSB, L: LSB

• When you want to input the MSB-first 20-bit data into SDIN1 and 2, input four "0s" following the LSB.

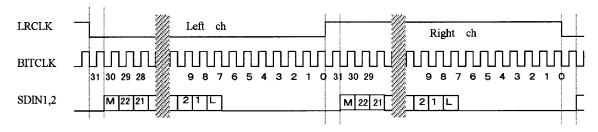
b) Mode 2, Mode 3 and Mode 4



M:MSB, L:LSB

SDIN1 and 2 Mode 2: (C2, C1) = (0, 1) 24 bits SDIN1 and 2 Mode 3: (C2, C1) = (1, 0) 20 bits SDIN1 and 2 Mode 4: (C2, C1) = (1, 1) 16 bits [ASAHI KASEI] [AK7714]

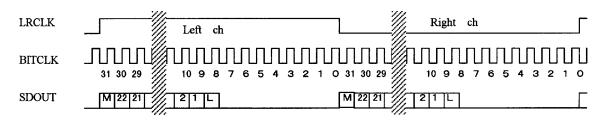
2) I²S compatible input format (C0 = 1)



M: MSB, L: LSB

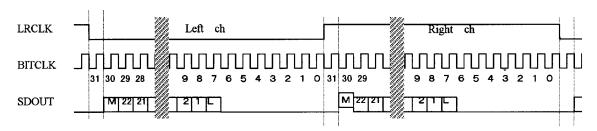
Mode 1: (C2, C1) = (0, 0) must be set.

3) Standard output format (C0 = 0: initial set value)



M: MSB, L: LSB

4) I^2 S compatible output format (C0 = 1)



M: MSB, L: LSB

(6) Interface with microcomputer

Interface with the microcomputer is provided by 7 control signals: \overline{CS} (Chip Select Bar), \overline{WRQ} (Write ReQuest Bar), SCLK (Serial data input CLocK), SI (Serial data Input), SO (Serial data Output), RDY (ReaDY) and DRDY (Data ReaDY).

In the AK7714, two types of operations are provided; writing and reading during the reset phase (namely, system reset) and those during the run phase. During the reset phase, data can be written to the control register, program RAM, coefficient RAM, offset RAM, and external conditional jump code, and data can be read from the program RAM, coefficient RAM and offset RAM. During the run phase, data can be written to the coefficient RAM, offset RAM and external conditional jump code, and data on the DBUS (data bus) can be read from SO.

 $\overline{\text{CS}}$ is used also for SO control in addition to the chip selection. When only one AK7714 is used without SO, $\overline{\text{CS}}$ = "L" is allowed at all times.

If there is no communication with the microcomputer, set SCLK to "H" and SI to "L" for use.

Data is serially input and output with the MSB first. The following 8-bit command data is sent first, and the specified work is performed.

Command code list

Contents of command	Condition	ns for use:	Code (C7C6	6C5C4C3C2C1C0)
Program RAM write	[RSPW]	During reset	phase:	Code (11000000)
External conditional JMP code write	[JCON]	both reset and	run phases:	Code (11000100)
Coefficient RAM write	[RSCW]	during reset	ohase:	Code (10100000)
Coefficient RAM write	[RNCW]	during run ph	nase:	Code (10100100)
Offset RAM write	[RSOW]	during reset	ohase:	Code (10010000)
Offset RAM write	[RNOW]	during run ph	nase:	Code (10010100)
CRAM/OFFRAM rewrite preparation	n[BNBW]	during run ph	iase:	Code (10001000)
Control register write	[CONW]	during reset	ohase:	Code (00000110)
Program RAM read		during reset p	ohase:	Code (11000011)
Program CRAM read		during reset	ohase:	Code (10100001)
Program OFRAM read		during reset p	ohase:	Code (10010001)

NOTE: Do not send codes other than the above command codes. Otherwise, an operation error may occur.

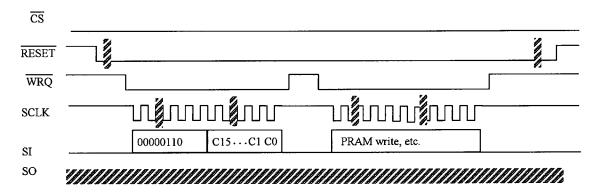
1) Write during reset phase

a) Control register write (during reset phase)

Data comprising a set of three bytes is used to perform the control register write operation (during the reset phase). When all data has been transferred, the RDY terminal goes to "L". It goes to "H" upon completion of the write operation.

Da	ta transfer procedure	
	① Command code	(00000110)
	② Control data	(C15C8)
	3 Control data	(C7C0)

The register to control the operation mode of this LSI comprises 16 bits. For the function of each bit, see the description of 5) "Control registers" on page 22.

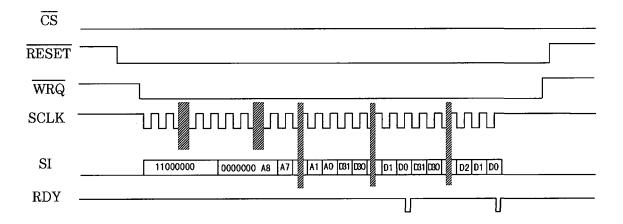


Control register write operation

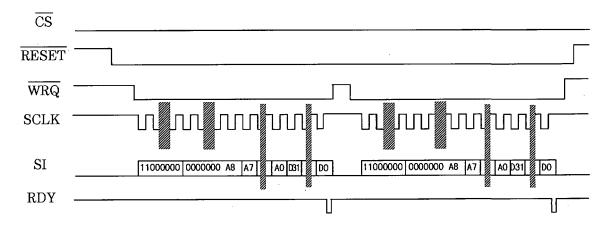
b) Program RAM write (during reset phase)

The program RAM write operation is performed during the reset phase according to data comprising a set of 7 bytes (in case of 1-word program RAM writting). When all data has been transferred, the RDY terminal is set to "L". Upon completion of writing into PRAM, it goes to "H" to allow the next data to be input. When data of continuous addresses is written, input the data(④ to ⑦) of after the next address as is. (No command code or address is required and RESET and WRQ hold to "L".) To write discontinuous data, shift the WRQ terminal from "H" to "L" again. Then input the command code, address and data in that order. (For RESET, operate both CODEC RESET and DSP RESET simultaneously.)

Data	Data transfer procedure				
	① Command code	(11000000)			
	② Address upper	(8A000000)			
	3 Address lower	(A7A0)			
	① Data	(D15D8)			
	⑤ Data	(D23D16)			
	6 Data	(D15D8))			
	⑦ Data	(D7D0)			



Input of continuous address data into PRAM



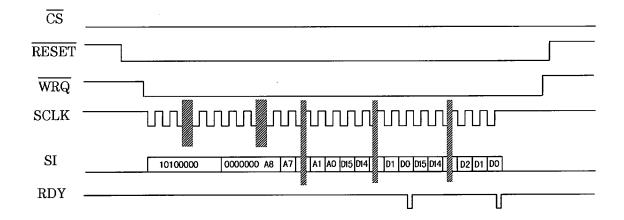
Input of discontinuous address data into PRAM

c) Coefficient RAM write (during reset phase)

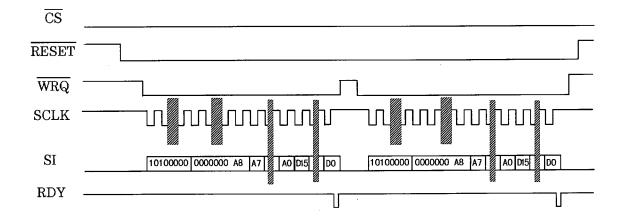
Data comprising a set of 5 bytes (in case of 1-word CRAM writting) is used to perform the coefficient RAM write operation (during the reset phase). When all data has been transferred, the RDY terminal goes to "L". Upon completion of writing into CRAM, it goes to "H" to allow the next data to be input. When data of continuous addresses is written, input the data(4) to 5) of after the next address as is. (No command code or address is required and RESET and WRQ hold to "L".) To write discontinuous data, shift the WRQ terminal from "H" to "L". Then input the command code, address and data in that order.

(For $\overline{\text{RESET}}$, operate both $\overline{\text{CODEC RESET}}$ and $\overline{\text{DSP RESET}}$ simultaneously.)

Data transfer procedure					
	① Command code	(10100000)			
	② Address upper	(8A000000)			
	3 Address lower	(A7A0)			
	① Data	(D15 D8)			
	⑤ Data	(D7D0)			



Input of continuous address data into CRAM

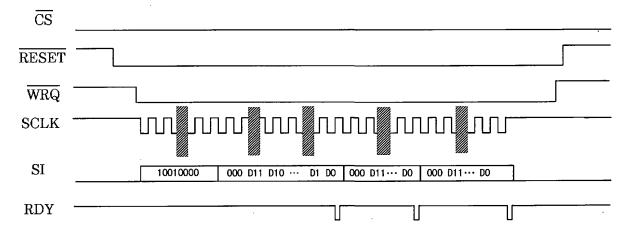


Input of discontinuous address data into CRAM

d) Offset RAM write (during reset phase)

Data comprising a set of 3 bytes (in case of 1-word Offset RAM writting) is used to perform offset RAM write operation (during the reset phase). In this case the operation must be started from address 0. When data of continuous addresses is written, input the data(② to ③) of after the next address as is. (No command code is required and RESET and WRQ hold to "L".) When all data has been transferred, the RDY terminal goes to "L". Upon completion of writing into the OFFRAM, it goes to "H" to allow the next data to be input. (For RESET, operate both CODEC RESET and DSP RESET simultaneously.)

Data	transfer procedure	
	① Command code	(10010000)
	② Data	(0000 D11 D10 D9 D8)
	3 Data	(D7D0)

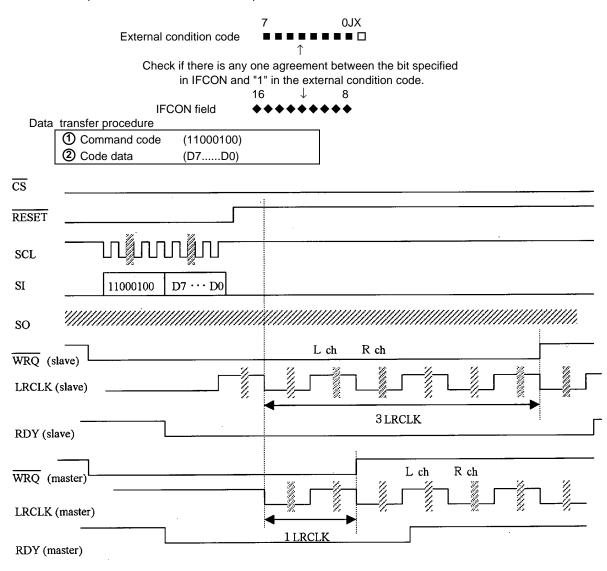


Input of data into OFFRAM

e) External conditional jump code write (during reset phase)

Data comprising a set of two bytes is used to perform the external conditional jump code write operation. The data can be input during both the reset and operation phases, and the input data is set to the specified register at the leading edge of LRCLK. When all data has been transferred, the RDY terminal goes to "L". Upon completion of writing, it goes to "H". A jump command will be executed if there is any one agreement between "1" of each bit of the external condition code 8 bits (soft set) plus 1 bit (hard set) at the external input terminal JX and "1" of each bit of the IFCON field. The data during the reset phase can be written only before release of the reset, after all data has been transferred. WRQ transition from "L" to "H" in the write operation during the reset phase must be executed after three LRCLK in the slave mode and one LRCLK in master mode, respectively, from the falling edge of LRCLK after release of the reset. Then RDY goes to "H" after capturing the rise of the next LRCLK. Write operation from the microcomputer is disabled until RDY goes to "H". The IFCON field provides external conditions written on the program.

Note: The LRCLK phase is inverted in the I2S-compatible state.



Timing for external conditional jump write operation (during reset phase)

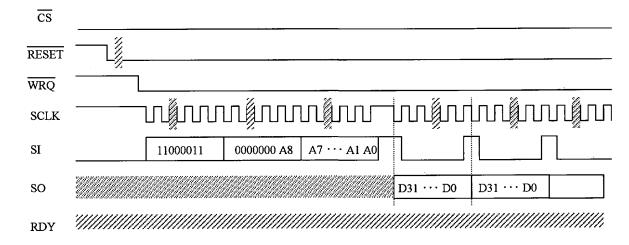
2) Read during reset phase

a) Program RAM read (during reset phase)

To read data written into PRAM, input the command code and the address you want to read out. After that, set SI to "H" and SCLK to "L". Then the data is output from SO in synchronization with the falling edge of SCLK. (Ignore the RDY operation that will occur in this case.)

If there are continuous addresses to be read, repeat the above procedure starting from the step where SI is set to "H".

Data	transfer procedure	
	① Command code input	(11000011)
	② Read address input	(00)
		(A7A0)

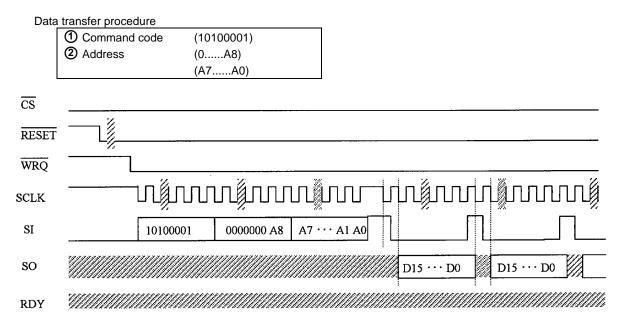


Reading of PRAM data

b) CRAM data read (during reset phase)

To read out the written coefficient data, input the command code and the address you want to read out. After that, set SI to "H" and SCLK to "L" as preparation. Then, when SI is set to "L", the data is output from SO in synchronization with the falling edge of SCLK.

If there are continuous addresses to be read, repeat the above procedure starting from the step where SI is set to "H".

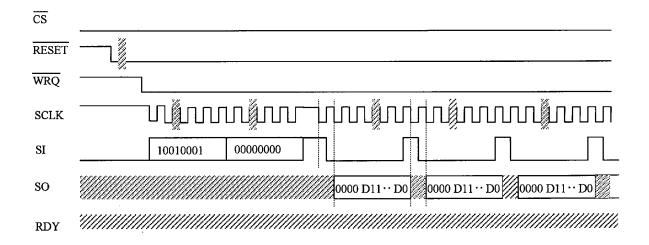


Reading of CRAM data

c) OFRAM data read (during reset phase)

The written offset data can be read out during the reset phase. To read it, input the command code and 8-bit "0". After that, set SI to "H" and SCLK to "L". This completes preparation for outputting the data. Then set SI to "L", and the data is output in synchronization with the falling edge of SCLK. In this case, OFRAM can be output only from the address data at address 0.

Data	Data transfer procedure				
	① Command code	(10010001)			
	② Data input	(0000000)			



Reading of OFRAM data

3) Write during RUN phase

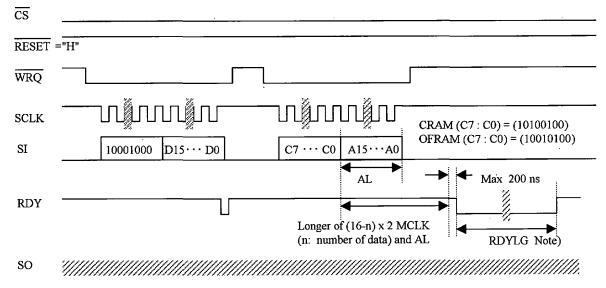
a) CRAM/OFRAM rewrite preparation and write (during RUN phase)

This function is used to rewrite CRAM (coefficient RAM) and OFRAM (offset RAM) during program execution. After inputting the command code, you can input a maximum of 16 data of the continuous addresses you want to rewrite. Then input the write command code and rewrite the leading address. Every time the RAM address to be rewritten is specified, the contents of RAM are rewritten. The following is an example to show how five data from address "10" of the coefficient RAM are rewritten:

Note that address "13" is not executed until address "12" is rewritten.

Data	Data_transfer procedure					
	*Preparation for rewrite		① Command code	(10001000)		
			② Data	(D15D8)		
			3 Data	(D7D0)		
	*Rewrite	1) CRAM	① Command code	(10100100)		
			② Address upper	(8A000000)		
			3 Address lower	(A7000A0)		
		2) OFRAM	① Command code	(10010100)		
			② Address upper	(0000000)		
			3 Address lower	(000A4A0)		

Note: CRAM ranges from A8 to A0. OFRAM ranges from A4 to A0.



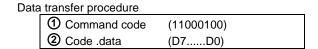
Note: The RDY signal will go to high within the maximum of two LRCLKs if the RDYLG width is programmed to ensure a new address to be rewritten within one sampling cycle.

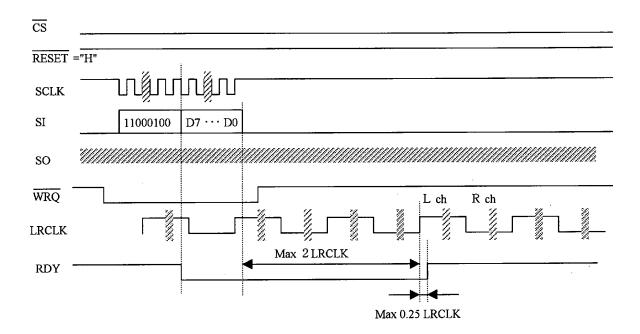
CRAM and OFRAM rewriting preparation and writing

b) External conditional jump code rewrite (during RUN phase)

Data comprising a set of two bytes is used to write the external conditional jump code. Data can be input during both the reset and operation phases, and input data is set to the specified register at the rising edge of LRCLK. When all data has been transferred, the RDY terminal goes to "L". Upon completion of writing, it goes to "H". A jump command will be executed if there is any one agreement between each bit of the 8-bit external condition code and "1" of each bit of the IFCON field. A write operation from the microcomputer is disabled until RDY goes to "H".

Note: The LRCLK phase is inverted in the I2S-compatible state.



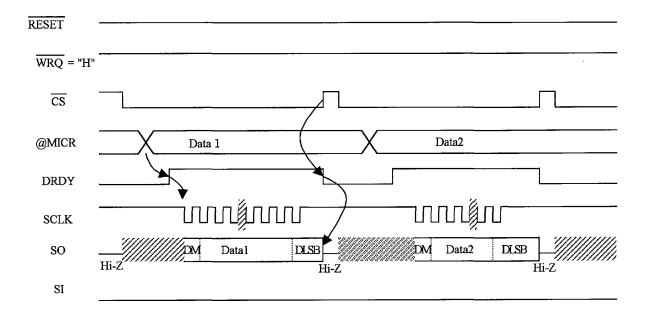


External condition jump write timing (during RUN phase)

4) Read-out during RUN phase (SO output)

SO outputs data on DBUS (data bus) of the DSP section. Data is set when @MICR is specified by the DST field. Setting of data allows DRDY to go to "H", and data is output synchronized with the falling edge of SCLK. When $\overline{\text{CS}}$ goes to "H", DRDY goes to "L" to wait for the next command. Once DRDY goes to "H", the data of the last @MICR command immediately before DRDY goes to "H" will be held until $\overline{\text{CS}}$ goes to "H", and subsequent commands will be rejected.

A maximum of 24 bits are output from SO. After the required number of data (not exceeding 24 bits) is taken out by SCLK, the next data can be output by setting $\overline{\text{CS}}$ to "H".



SO read (during RUN phase)

(7) ADC section high-pass filter

The AK7714 incorporates a digital high-pass filter (HPF) for cancelling the section DC offset in the ADC section. The HPF cut-off frequency is about 1 Hz (fs = 48 kHz). This cut-off frequency is proportional to the sampling frequency (fs).

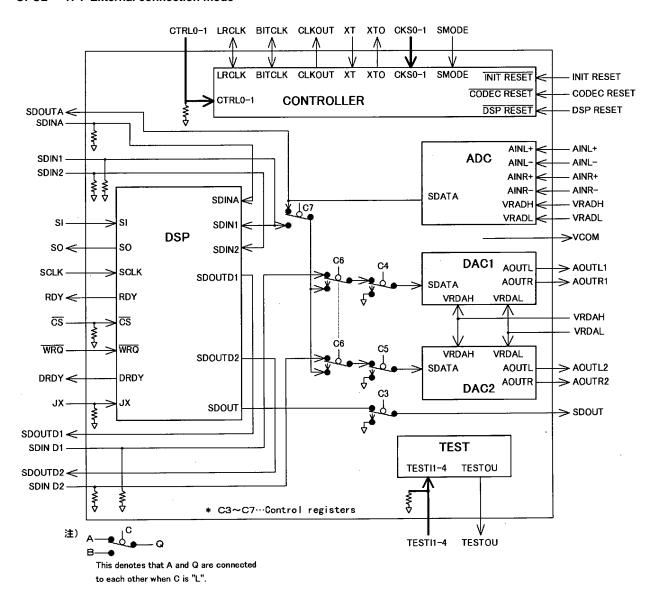
	48 kHz	44.1 kHz	32 kHz
Cut-off frequency	0.93 Hz	0.86 Hz	0.62 Hz

(8) Special use

1) External connection mode

Normally, OPCL is used at "L" (internal connection mode), but when OPCL is set to "H", the ADC output and DAC1/DAC2 inputs can be used independently form DSP. (External connection mode)

OPCL = "H": External connection mode



The following shows the input/output interface in external connection mode:

- SDINA for MSB-first 24-bit input (including I2S compatibility)
- ullet SDOUTA for MSB-first 20-bit output (including I²S compatibility)
- SDOUTD1 and SDOUTD2 for MSB-first 24-bit outputs (including 12S compatibility)
- SDIND1 and SDIND2 for MSB-first 20-bit inputs (including I2S compatibility)

Conversion between the input/output standard format and I2S is interlocked with the control register C0, similar to the case for internal connection mode.

2) Use as ADC and DAC (mainly for test)

Only the ADC and DAC sections can be operated while keeping the DSP section in the reset state with the independent control of DSP RESET and CODEC RESET. (When no DSP processing is required, power saving and noise reduction can be expected. However, the ADC data cannot be output in internal connection mode. In external connection mode, it is output from SDOUTA.)

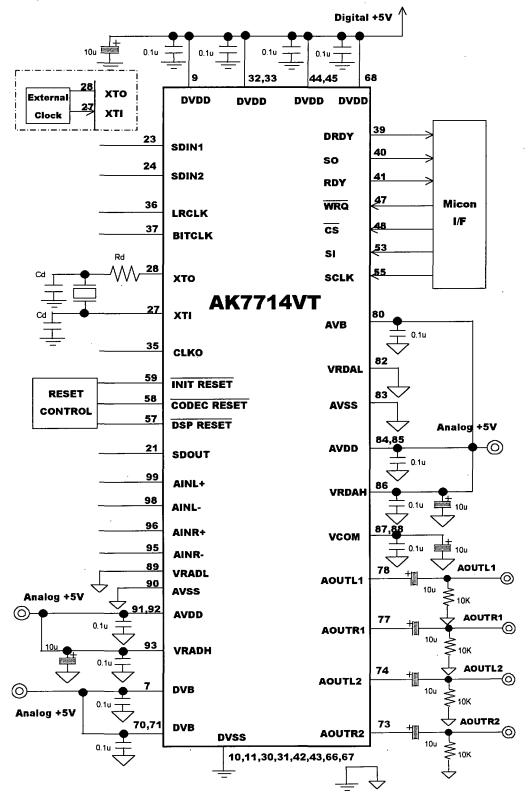
In internal connection mode, setting of the control registers allows the following operations to be performed:

```
a) ADC to DAC1 and DAC2 (Analog to Analog)
The ADC output data is directly connected over to DAC1 and DAC2. (C6 = 1, C7 = 0)
When input to the DAC2 is not required, set C5 = 1.
(When input to the DAC1 is not required, set C4 = 1.)
b) SDIN1 to DAC1 and DAC2
SDIN1 input data is directly connected to DAC1 and DAC2. (C6 = 1, C7 = 1)
In this case, only the MSB-first 20-bit input (including I²S compatibility: C0 = 1) is supported.
When input to DAC2 is not required, set C5 = 1.
(When input to DAC1 is not required, set C4 = 1.)
```

For this operation, set only CODEC RESET to "H" after setting the control registers during the system reset phase (DSP RESET = CODEC RESET = "L"). To make a new setting, be sure to perfom the system reset.

System Design

(1) Example circuit



(2) Peripheral circuit

1) Ground and power supply

To minimize digital noise coupling, AVDD and DVDD are individually decoupled in AK7714. System analog power is supplied to AVDD, AVB and DVB. AVB and DVB are connected to each other through the IC board, and eventually have several ohms of resistance. If the set of AVDD, AVB and DVB and DVDD are driven by individual power sources, start up AVDD, AVB and DVB simultaneously with DVDD, or start up AVDD, AVB and DVB first.

Generally, power supply and ground wires must be connected separately according to the analog and digital systems. Connect them at a position close to the power source on the PC board. Decoupling capacitors, and ceramic capacitors of small capacity in particular, should be connected at positions as close as possible to the AK7714.

If the absolute maximum rating conditions of a power supply cannot be maintained depending on the system, it is recommended to supply the AK7714 power from the same regulator. Power patterns must be separated into analog and digital patterns. For a digital pattern, connection must be made through an appropriate one-ohm resistor from the regulator. In this case, the capacitor with the larger capacity must be connected to the analog side.

2) Reference voltage

The input voltage difference between the VRADH pin and the VRADL pin determines the full scale of analog input, while the potentials difference between the VRDAH pin and the VRDAL pin determines the full scale of the analog output. Normally, connect AVDD to VRADH and VRDAH, and connect $0.1\mu\text{F}$ ceramic capacitors from them to AVSS. VCOM is used as the common voltage of the analog signal.

To shut out high frequency noise, connect a $0.1\mu F$ ceramic capacitor in parallel with an appropriate $10\mu F$ electrolytic capacitor between this pin and AVSS. The ceramic capacitor in particular should be connected at a position as close as possible to the pin. Do not lead current from the VCOM pin. To avoid coupling to the AK7714, digital signals and clock signals in particular should be kept away from the VRADH, VRADL, VRDAH, VRDAL and VCOM pins as far as possible.

3) Analog input

Analog input signals are applied to the modulator through the differential input pins of each channel. The input voltage is equal to the differential voltage between AIN+ and AIN- (Δ VAIN = (AIN+) - (AIN-)), and the input range is \pm FS = \pm (VRADH - VRADL) × 0.4.

When VRADH = 5V and VRADL = 0V, the input range is within ± 2.0 V. The output code format is given in terms of 2's complements. Table 1 shows the output code relative to input voltage.

Input voltage	Output code (hexadecimal)	
	20 bit	
> (+FS - 1.5 LSB)	3FFFF	
-0.5 LSB	00000 7FFFF	
> (-FS + 0.5 LSB)	80000	

Table 1. Output code relative to input voltage

When fs = 48 kHz, the AK7714 samples the analog input at 3.072 MHz. The digital filter eliminates noise from 30 kHz to 3.042 MHz. However, noise is not rejected in the bandwidth close to 3.072 MHz. Most audio signals do not have large noise in the vicinity of 3.072 MHz, so a simple RC filter is sufficient.

A/D converter reference voltage is applied to the VRADH and VRADL pins. Normally, connect AVDD to VRADH, and AVSS to VRADL. To eliminate high frequency noise, connect a $0.1\mu F$ ceramic capacitor in parallel with a $10\mu F$ electrolytic capacitor between the VRADH pin and VRADL.

The analog source voltage to the AK7714 is +5 V. Voltage of AVDD + 0.3 V or more, voltage of AVSS - 0.3 V or less, and current of 10 mA or more must not be applied to analog input pins (AINL and AINR). Excessive current will damage the internal protection circuit and will cause latch-up, thereby damaging the IC. Accordingly, if the surrounding analog circuit voltage is ±15 V, the analog input pins must be protected from signals with the absolute maximum rating or more.

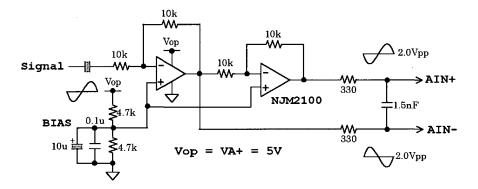


Fig. 1 Example of input buffer circuit (differential input)

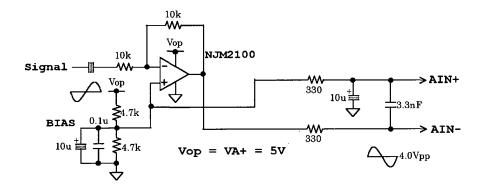


Fig. 2 Example of input buffer circuit (single end input)

An analog signal can be applied to the AK7714 is single end mode. In this case, apply the analog signal (the full scale is 4.0 Vpp when the internal reference voltage is used) to the AIN-input, and bias to the AIN+input. However, use of a low saturated operational amplifier is recommended if the operational amplifier is driven by the 5-volt power supply. The electrolytic capacitor connected to AIN+ is effective for reducing the second harmomics. (See Fig. 2.)

4) Analog output

Analog output is single-ended, and the output range is 2.95Vpp (typical) with respect to VCOM voltage. The out-of-band noise (shaping noise) produced by the built-in $\Delta \Sigma$ modulator is reduced by the built-in switched capacitor filter (SCF) and continuous filter (CTF). Therefore, it is not necessary to add an external filter for normal application. The input code format is given in terms of 2's complements with the positive full-scale output for the 3FFFFH(@ 20 bit) input code, and the negative full-scale output for the 80000H (@ 20 bit) input code. VCOM voltage is output as an ideal value for 00000H(@ 20 bit) input code.

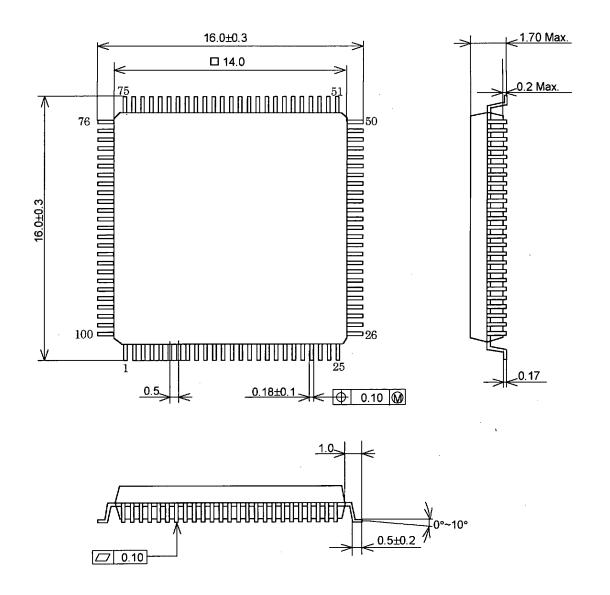
5) Connection to digital circuit

To minimize the noise resulting from the digital circuit, connect CMOS logic to the digital output. The applicable logic family includes the 4000B, 74HC, 74AC, 74ACT and 74HCT series.

Package

• 100-pin LQFP

(Unit: mm)



• Material & Lead finish

Package:

Ероху

Lead-frame:

Copper

Lead-finish:

Soldering plate

Marking



Meanings of XXXXAAA

XXXX: Time of manufacture (numeral)
AAA: Lot number (Alphabet)

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