## SKYWORIS®

## PRELIMINARY DATA SHEET

## AP147-320: ISM 900 MHz Band 2-Watt InGaP HBT Power Amplifier

## Features

- 902-928 MHz operation
- Output power greater than 33 dBm
- 3.2 V nominal operating voltage
- Integrated analog power control voltage, $\mathrm{V}_{\text {APC }}=0.1-2.8 \mathrm{~V}$
- High PAE of $50 \%$ at maximum output power
- Ultrasmall, thermally enhanced micro lead frame package.
- Low current in standby mode of $<10 \mu \mathrm{~A}$
- Available on tape and reel


## Description

The AP147-320 is a high-performance 3-stage, high-power amplifier IC designed for use in 900 MHz ISM band applications such as automatic meter readers and RFID. It has an integrated analog power control voltage for achieving the desired output power levels. The IC is manufactured on an advanced InGaP HBT process. The AP147-320 is packaged in a thermally enhanced, ultrasmall, micro lead frame package.

## Block Diagram



## Package Dimensions



## Pin Assignments

| Pin | Symbol | Description |
| :---: | :--- | :--- |
| 1,3, <br> 8,14 | GND | Equipotential point. Connect to the printed circuit <br> board common via the lowest possible impedance. |
| 2 | RF IN | RF input to the amplifier |
| 4 | $\mathrm{~V}_{\text {CC }} 1$ | DC power supply voltage input to the first <br> amplifier stage |
| 5 | $\mathrm{~V}_{\text {APC }} 1$ | Power control voltage input to the first and second <br> amplifier stages |
| 6 | $\mathrm{~V}_{\text {APC }} 2$ | Power control voltage input to the final amplifier stage |
| 9,10, <br> 11,12 | RF OUT $N_{\text {CC }} 3$ | RF output port and DC power supply voltage input <br> to the final amplifier stage. These pins should be <br> connected together directly at the pins for DC <br> current sharing. |
| 13 | $2_{\text {FC }}$ | (Optional) Harmonic trap. Connect a series resonant <br> circuit at the second harmonic of the RF input signal <br> to reduce second harmonic power at the RF output. |
| 15,16 | $\mathrm{~V}_{\text {CC }} 1$ | DC power supply voltage input to the second amplifier <br> stage. These pins should be connected together <br> directly at the pins for DC current sharing. |
| Exposed <br> Pad |  | Connect to the printed circuit board common via the <br> lowest possible electrical and thermal impedance. |

## Absolute Maximum Ratings

| Characteristic | Value |
| :--- | :---: |
| Supply voltage $\left(V_{\mathrm{CC}} \& \mathrm{~V}_{\text {REF }}\right)$ | 6.0 V |
| Power control voltage $\left(\mathrm{V}_{\text {APC }} 1 \& \mathrm{~V}_{\text {APC }} 2\right)$ | 4.2 V |
| RF input power | 15 dBm |
| Operating temperature | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage temperature | $-65^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |

Performance is guaranteed only under the conditions listed in the specifications table and is not guaranteed under the full range(s) described by the Absolute Maximum specifications. Exceeding any of the absolute maximum/minimum specifications may result in permanent damage to the device and will void the warranty.

CAUTION: Although this device is designed to be as robust as possible, ESD (Electrostatic Discharge) can damage this device. This device must be protected at all times from ESD. Static charges may easily produce potentials of several kilovolts on the human body or equipment, which can discharge without detection. Industry-standard ESD precautions must be employed at all times.

## General DC Electrical Specifications

| Parameter | Symbol | Condition | Min. | Typ. | Max. |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Unit |  |  |  |  |  |
| Supply voltage | $\mathrm{V}_{\mathrm{CC}}$ |  | 2.7 | 3.2 | 4.2 |
| Power control voltage | $\mathrm{V}_{\text {APC }}$ |  | V |  |  |
| Power control current | IVAPC |  | 0.1 | 2.6 | 2.8 |
| Leakage current | $\mathrm{S}_{21}$ | $\mathrm{P}_{\mathrm{IN}}<-30 \mathrm{dBm}, \mathrm{V}_{\text {APC }} 1,2=0.1 \mathrm{~V}$ | V |  |  |

## General RF Transmit Electrical Specifications

Conditions: $\mathbf{V}_{\text {CC }}=\mathbf{3 . 2} \mathbf{V}, \mathbf{V}_{\text {REF }}=\mathbf{3 . 2} \mathbf{V}, \mathrm{V}_{\text {APC }}=\mathbf{2 . 6} \mathbf{V}, \mathrm{T}_{\mathrm{A}}=25{ }^{\circ} \mathrm{C}$

| Parameter | Symbol | Condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Frequency range | F |  | 900 |  | 928 | MHz |
| Gain | $\mathrm{S}_{21}$ | Small signal |  | 31.6 |  | dB |
| Gain variation over frequency | $\Delta \mathrm{S}_{21}$ | Small signal |  | 0.26 |  | dB |
| Input return loss | $\mathrm{S}_{11}$ | Small signal |  | -14.7 |  | dB |
| Output return loss | $\mathrm{S}_{22}$ | Small signal |  | -5.7 |  | dB |
| Quiescent current | $I_{\text {CQ }}$ | (No RF signal) |  | 0.16 |  | A |
| Output $\mathrm{P}_{1 \mathrm{~dB}}$ | $\mathrm{P}_{1 \mathrm{~dB}}$ | CW |  | 33.1 |  | dBm |
| Current consumption | $I_{\text {CC }}$ | Output $\mathrm{P}_{1 \mathrm{~dB}}$ |  | 1.3 |  | A |
| Power added efficiency | PAE | Output $\mathrm{P}_{1 \mathrm{~dB}}$ |  | 50\% |  | \% |
| Second harmonic | $\mathrm{F}_{2}$ | Output P1dB |  | -55 |  | dBc |
| Third harmonic | $\mathrm{F}_{3}$ | Output P1dB |  | -66 |  | dBc |
| Ruggedness |  | Output VSWR $=10: 1$, All phase angles, $V_{C C}=4.2 \mathrm{~V}, \mathrm{P}_{\text {IN }}=10 \mathrm{dBm}, \mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}$ | No module damage or permanent performance degradation |  |  |  |
| Stability |  | Output VSWR = 10:1, All phase angles, $V_{\mathrm{CC}}=4.2 \mathrm{~V}, \mathrm{P}_{\text {IN }}=10 \mathrm{dBm}, \mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}$ |  | -36 |  | dBm |

## Typical Performance Data


$I_{C Q}$ and IVAPC vs. $V_{\text {APC }}$
Conditions: $\mathrm{V}_{\mathrm{CC}}=3.2 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.2 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$


$I_{\text {co }}$ vs. $V_{\text {cc }}$
Conditions: $\mathrm{V}_{\text {APC }}=3.2 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.2 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$


Forward Isolation
Conditions: $\mathrm{V}_{\mathrm{CC}}=3.2 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=3.2 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$P_{\text {IN }}$ vs. $P_{\text {OUt }}$ at $V_{\text {APC }}$
Conditions: CW, $\mathrm{V}_{\mathrm{CC}}=3.2 \mathrm{~V}$, $\mathrm{V}_{\text {REF }}=3.2 \mathrm{~V}$, $\mathrm{F}_{\mathrm{C}}=\mathbf{9 1 5} \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$P_{\text {Out }}$ vs. Gain at $V_{\text {APC }}$
Conditions: CW, $\mathrm{V}_{\text {CC }}=3.2 \mathrm{~V}$, $\mathrm{V}_{\text {REF }}=3.2 \mathrm{~V}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$P_{\text {out }}$ vs. IcC at $V_{\text {APC }}$ Conditions: CW, $\mathrm{V}_{\mathrm{CC}}=3.2 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.2 \mathrm{~V}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$P_{\text {OUt }}$ vs. PAE at $V_{\text {APC }}$
Conditions: CW, $\mathrm{V}_{\mathrm{CC}}=3.2 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.2 \mathrm{~V}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$P_{\text {IN }}$ vs. $P_{\text {OUt }}$ at $V_{\text {CC }}$
Conditions: CW, $\mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.2 \mathrm{~V}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$P_{\text {OUt }}$ vs. ICC at $V_{\text {CC }}$
Conditions: CW, $\mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=\mathrm{V}_{\mathrm{CC}}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$
$P_{\text {IN }}$ vs. Pout $_{\text {at }} V_{\text {CC }}$
Conditions: CW, $\mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.2 \mathrm{~V}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$P_{\text {OUT }}$ vs. Gain at $V_{C C}$
Conditions: CW, $\mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=\mathrm{V}_{\mathrm{CC}}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$


Conditions: CW, $\mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=\mathrm{V}_{\mathrm{CC}}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

$\mathrm{P}_{1 \mathrm{~dB}}$, Gain, and $\mathrm{I}_{\mathrm{CC}} \mathrm{VS} . \mathrm{V}_{\mathrm{CC}}$
Conditions: $\mathrm{CW}, \mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=\mathrm{V}_{\mathrm{CC}}$,
$P_{1 \mathrm{~dB}}$, Gain, and $\mathrm{I}_{\mathrm{CC}} \mathrm{VS} . \mathrm{V}_{\mathrm{CC}}$
Conditions: $\mathrm{CW}, \mathrm{V}_{\text {APC }}=2.6 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=\mathrm{V}_{\mathrm{CC}}$, $\mathrm{F}_{\mathrm{C}}=\mathbf{9 1 5 \mathrm { MHz } , \mathrm { T } _ { \mathrm { A } } = 2 5 ^ { \circ } \mathrm { C } , ~}$

$P_{1 d B}$, Gain, and ICC vs. $V_{\text {APC }}$ Conditions: CW, $\mathrm{V}_{\mathrm{CC}}=3.2 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=\mathrm{V}_{\mathrm{CC}}$, $\mathrm{F}_{\mathrm{C}}=915 \mathrm{MHz}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

## Bill of Material for Evaluation Board

| Part \# | ID | Size | Value | Units | Manufacturer | Product Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $\mathrm{L}_{1}$ | 0402 | 1.8 | nH | Johanson | L-07C1N8ST |
| 2 | $\mathrm{R}_{1}$ | 0402 | 180 | $\Omega$ | Panasonic* | ERJ2GEJ181X |
| 3 | $\mathrm{C}_{1}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 4 | $\mathrm{C}_{2}$ | 0402 | 18 | pF | Murata | GRM1555C1H180JZ35E |
| 5 | $\mathrm{C}_{3}$ | 0402 |  |  | DO NOT PLACE |  |
| 6 | $\mathrm{C}_{4}$ | 0402 |  |  | DO NOT PLACE |  |
| 7 | $\mathrm{C}_{5}$ | 0402 |  |  | DO NOT PLACE |  |
| 8 | $\mathrm{C}_{6}$ | 0402 |  |  | DO NOT PLACE |  |
| 9 | $\mathrm{C}_{7}$ | 1206 | 10 | $\mu \mathrm{F}$ | AVX | TAJA106M006R |
| 10 | $\mathrm{C}_{8}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 11 | $\mathrm{C}_{9}$ | 0402 | 1 | nF | Murata | GRM155R71H102KA01* |
| 12 | $\mathrm{C}_{10}$ | 0402 | 10 | nF | Murata | GRM155R71E103KA01* |
| 13 | $\mathrm{C}_{11}$ | 0402 | 100 | nF | Murata | GRM155R71A104KA01* |
| 14 | $\mathrm{C}_{12}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 15 | $\mathrm{C}_{13}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 16 | $\mathrm{C}_{14}$ | 0402 |  |  | DO NOT PLACE |  |
| 17 | $\mathrm{C}_{15}$ | 0402 |  |  | DO NOT PLACE |  |
| 18 | $\mathrm{C}_{16}$ | 0402 |  |  | DO NOT PLACE |  |
| 19 | $\mathrm{C}_{17}$ | 0402 | 68 | pF | Murata | GRM1555C1H680JD83* |
| 20 | $\mathrm{C}_{18}$ | 1206 | 10 | uF | AVX | TAJA106M006R |
| 21 | $\mathrm{C}_{19}$ | 0402 | 100 | nF | Murata | GRM155R71A104KA01* |
| 22 | $\mathrm{C}_{20}$ | 0402 | 10 | nF | Murata | GRM155R71E103KA01* |
| 23 | $\mathrm{C}_{21}$ | 0402 | 1 | nF | Murata | GRM155R71H102KA01* |
| 24 | $\mathrm{C}_{22}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 25 | $\mathrm{C}_{23}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 26 | $\mathrm{C}_{24}$ | 0402 | 10 | nF | Murata | GRM155R71E103KA01* |
| 27 | $\mathrm{C}_{25}$ | 1206 | 10 | uF | AVX | TAJA106M006R |
| 28 | $\mathrm{C}_{26}$ | 0402 | 100 | nF | Murata | GRM155R71A104KA01* |
| 29 | $\mathrm{C}_{27}$ | 0402 | 10 | nF | Murata | GRM155R71E103KA01* |
| 30 | $\mathrm{C}_{28}$ | 0402 | 1 | nF | Murata | GRM155R71H102KA01* |
| 31 | $\mathrm{C}_{29}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 32 | $\mathrm{C}_{30}$ | 0402 | 15 | pF (Hi-Q) | Murata | GJM1555C1H150JB01E |
| 33 | $\mathrm{C}_{31}$ | 0402 | 100 | pF | Murata | GRM1555C1H101JD83E |
| 34 | $\mathrm{C}_{32}$ | 0402 | 4.7 | pF ( $\mathrm{Hi}-\mathrm{Q}$ ) | Murata | GJM1555C1H4R7CB01E |
| 35 | SMA Connector, $50 \Omega$ (X2) |  |  |  | AEP | 9114-9113-000 |
| 36 | PCB |  |  |  |  |  |

## Application Circuit



## Component Placement Diagram



Coplanar spacing: 0.026 inches.
Transmission line width: 0.025 inches.

## Application Circuit Notes

RF Input, Pin 2, uses a 100 pf ( $\mathrm{C}_{1}$ ) DC blocking capacitor, a $180 \Omega\left(\mathrm{R}_{1}\right)$ resistor, and a $18 \mathrm{pF}\left(\mathrm{C}_{2}\right)$ tuning capacitor. The length of the line between the 18 pF capacitor and the amplifier is important to the input VSWR of the device. This distance is indicated below.
$V_{\text {CC }} 1$, Pin 4, is the collector bias for stage 1. Capacitors of $10 \mathrm{uF}\left(\mathrm{C}_{7}\right), 100 \mathrm{pF}\left(\mathrm{C}_{8}\right), 1.0 \mathrm{nF}\left(\mathrm{C}_{9}\right), 10 \mathrm{nF}\left(\mathrm{C}_{10}\right), 100 \mathrm{nF}\left(\mathrm{C}_{11}\right)$, and an inductor of $1.8 \mathrm{nH}\left(\mathrm{L}_{1}\right)$ are used for cleaning up the power supply input signal and RF bypassing.
$\mathrm{V}_{\mathrm{Cc}} 2$, Pin 15 , is the collector bias for stage 2 . Tuning is accomplished, in conjunction with pin 16 , with a $68 \mathrm{pF}\left(\mathrm{C}_{17}\right), 1.0 \mathrm{nF}$ $\left(\mathrm{C}_{21}\right)$, and a $100 \mathrm{pF}\left(\mathrm{C}_{22}\right)$ capacitor on the collector bias line. Capacitors of $10 \mu \mathrm{~F}\left(\mathrm{C}_{18}\right), 100 \mathrm{pF}\left(\mathrm{C}_{19}\right)$, and $10 \mathrm{nF}\left(\mathrm{C}_{20}\right)$ are used to clean up the power supply input signal.
$\mathrm{V}_{\mathrm{cc}} 2$, Pin 16 , is the collector bias for stage 2 . Tuning is accomplished, in conjunction with pin 15 , with a $68 \mathrm{pF}\left(\mathrm{C}_{17}\right), 1.0 \mathrm{nF}$ $\left(\mathrm{C}_{21}\right)$, and a $100 \mathrm{pF}\left(\mathrm{C}_{22}\right)$ capacitor on the collector bias line. Capacitors of $10 \mu \mathrm{~F}\left(\mathrm{C}_{18}\right), 100 \mathrm{pF}\left(\mathrm{C}_{19}\right)$, and $10 \mathrm{nF}\left(\mathrm{C}_{20}\right)$ are used to clean up the power supply input signal.

RF OUT $/{ }_{\text {CC }} 3$, Pins (9-12), Tuning is accomplished with a $15 \mathrm{pF}\left(\mathrm{C}_{30}\right)$ high- Q capacitor and a $7.0 \mathrm{pF}\left(\mathrm{C}_{32}\right)$ high- Q capacitor on the RF output line. A length of transmission line and a bank of bypassing capacitors of $10 \mathrm{uF}\left(\mathrm{C}_{25}\right), 100 \mathrm{nF}$ $\left(\mathrm{C}_{26}\right), 10 \mathrm{nF}\left(\mathrm{C}_{27}\right), 1.0 \mathrm{nF}\left(\mathrm{C}_{28}\right)$, and $100 \mathrm{pF}\left(\mathrm{C}_{29}\right)$ are required on the drain bias line for tuning and bias injection. A 100 pF $\left(C_{31}\right)$ DC blocking capacitor is needed on the output line.
VAPC1, Pin 5 , is the power control voltage for stages 1 and 2. A $100 \mathrm{pF}\left(\mathrm{C}_{13}\right)$ capacitor is required for biasing. Pin 5 and Pin 6 can be combined right at the amplifier by a jumper or a substrate connection between the two pads. In this configuration, a single 100 pF capacitor could be used for biasing.
VAPC6, Pin 6, is the power control voltage for stages 1 and 2. A $100 \mathrm{pF}\left(\mathrm{C}_{12}\right)$ capacitor is required for biasing. Pin 6 and pin 5 can be combined right at the amplifier by a jumper or a substrate connection between the two pads. In this configuration, a single 100 pF capacitor could be used for biasing.
$V_{\text {REF }}$, Pin 7 , is the control voltage for the bias control circuit. A combination of a $100 \mathrm{pF}\left(\mathrm{C}_{23}\right)$ and a $10 \mathrm{nF}\left(\mathrm{C}_{24}\right)$ capacitor is required for bypassing.

## Package Footprint



NOTE: All units in inches [mm].
NOTE 1: Length of all non-grounded lands underneath the package.
NOTE 2: Width between grounded lands and non-grounded lands.
NOTE 3: Length of from ground pad to edge of package.
NOTE 4: Width between non-grounded lands.
NOTE 5: Radius of the vias.
NOTE 6: Length of all lands from the edge of the package.
NOTE 7: Width of the ground lands.
NOTE 8: Distance betwwen all vias.
NOTE 9: $X$ and $Y$ dimension of the package.
NOTE 10: Width of the land for RFOUT.
NOTE 11: Width of the land for the ground pad.

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## Evaluation Board Stack-Up



## Application Board Bias Procedure

Step 1. Connect DC ground to the pad located in the top righthand corner of the board as indicated below.
Step 2. Connect Stage 1 collector bias voltage, $\mathrm{V}_{\mathrm{Cc}} 1$, to the pad at the bottom left side of the board as indicated below.
$\mathrm{V}_{\mathrm{CC}} 1=3.2 \mathrm{~V}$
Step 3. Connect Stage 2 collector bias voltage, $\mathrm{V}_{\text {cc }} 2$, to the second pad in from the top left side of the board as indicated below. $\mathrm{V}_{\mathrm{C}}{ }^{2}=3.2 \mathrm{~V}$
Step 4. Connect Stage 3 collector bias voltage, $\mathrm{V}_{\mathrm{cc}} 3$, to the first pad at the top right side of the board as indicated below. $\mathrm{V}_{\mathrm{CC}} 2=3.2 \mathrm{~V}$
Step 5 . Connect the bias control voltage, $V_{\text {REF }}$ to the pad at the bottom right of the board as indicated below. $\mathrm{V}_{\text {REF }}$ should be tied to the same line as $\mathrm{V}_{\mathrm{cc}} 1, \mathrm{~V}_{\mathrm{cc}} 2$, and $\mathrm{V}_{\mathrm{cc}} 3$ to obtain a single voltage supply.

NOTE: It is important that the $V_{C C} 1, V_{C C} 2, V_{C C} 3$, and $V_{\text {REF }}$ voltage source be adjusted such that 3.2 V is measured at the board. The high drain currents will drop the drain voltage significantly if long leads are used. Adjust the bias voltage to compensate.

## Recommended Solder Reflow Profiles

Refer to the "Recommended Solder Reflow Profile" Application Note.

## Tape and Reel Information

Refer to the "Discrete Devices and IC Switch/Attenuators
Tape and Reel Package Orientation" Application Note.

Step 6. Connect the power control voltages, $\mathrm{V}_{\text {APC }} 1$ and $\mathrm{V}_{\text {APC }} 2$, the middle two pads on the bottom of the board, as indicated below. This is done to obtain a single control voltage supply. $\mathrm{V}_{\text {APC }}=0 \mathrm{~V}$ at this point.
Step 7. Apply RF at RF IN port at -20 dBm to start, then adjust power level to 0 to 5 dBm .
Step 8. The $V_{\text {APC }}$ can be increased to 2.8 V max. or until desired output power level is achieved at RF OUT port.
Step 9. The pads in the upper left-hand corner, first from the left and third from the left, are to be grounded.


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