

### T1 / E1 / J1 OCTAL FRAMER

IDT82V2108

**PRELIMINARY** 

#### **FEATURES**

- Octal Framer supporting T1, E1 and J1 Formats
- Provides programmable system interface to support Mitel® ST-bus, AT&T® CHI and MVIP bus, supporting data rates of 1.544 / 2.048 / 8.192Mb/s; up to four links can be byte interleaved on one system bus without external logic
- Provides up to three internal floating HDLC controllers for each framer to support ISDN PRI and V5.X interface. Each HDLC contains 128-byte deep FIFOs in both the receive and transmit directions
- Provides jitter attenuation performance exceeding the requirements set by the associated standards for both Rx and Tx path
- Provides payload, line and digital loop-backs
- Provides a floating Pseudo Random Bit Sequence / repetitive pattern generator/detector, which can be assigned to any one of eight framers, the pattern may be inserted / detected in an unframed or Nx64K or Nx56K (T1 only) basis
- Provides signaling insertion / extraction for CCS / CAS and RBS signaling system
- Provides programmable codes insertion, data / sign inversion and digital milliwatt code insertion on a per channel / timeslot basis
- Supports automatic / manual alarming transmit and integration
- Provides performance monitor to counter CRC error, framing bit error, far end block CRC error (E1), out of frame event (T1/J1) and change of frame alignment event (T1/J1)
- Provides programmable In-band Loop-back Code transmitter/receiver, Bit Oriented Message generator / detector
- Supports polled or interrupt driven processing for all events
- Supports multiplexed or non-multiplexed address/data bus MPU interface for configuration, control and status monitoring
- JTAG boundary scan meets IEEE 1149.1
- Low power 3.3V CMOS technology with 5V tolerant inputs
- Operating industrial temperature range: -40°C to +85°C
- Package available: 128 pin PQFP 144 pin PBGA

#### **APPLICATIONS**

- High density internet E1 or T1 / J1 interface for routers, multiplexers, switches and digital modems.
- Frame relay switches and access devices (FRADS)
- SONET / SDH add drop multiplexers
- Digital private branch exchanges (PBX)
- Channel service units (CSU) and data service units (DSU)
- Channel banks and multiplexers
- Digital access and cross-connect systems (DACS)

### **STANDARDS**

#### E1 MODE:

ITU-T: G.704, G.706, G.732, G.802, G.737, G.738, G.739, G.742, G.823, G.964, G.965, I.431, O.151, O.152, O.153;

ETSI: ETS 300 011, ETS 300 233, ETS 324-1, ETS 347-1, TBR 4, TBR 12, TBR 13;

GO - MVIP

#### T1/J1 MODE:

ANSI: T1.107, T1.231, T1.403, T1.408;

TR: TSY-000147, TSY-000191, NWT-000303, TSY-000312, TSY-000-

499:

AT&T: TR 54016, TR 62411

TTC: JT-G 703, JT-G 704, JT-G706, JT-G 1431

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#### **DESCRIPTION**

The IDT82V2108 is a flexible feature-rich octal T1/E1/J1 Framer. Controlled by the software, the IDT82V2108 can be globally configured as an Octal E1 or T1/J1 Framer. When E1 or T1/J1 has been set globally, the operation mode of each of the eight framers can be configured independently. The configuration is performed through a parallel Multiplexed/Non-Multiplexed microprocessor interface.

The IDT82V2108 realizes frame synchronization, frame generating, signaling extraction and insertion, alarm and test signals generation and detection in a single chip. It also integrates up to three HDLC receivers and HDLC transmitters for each of the eight framers.

In E1 Mode, the receive path of each framer can be configured to frame to Basic Frame, CRC Multi-Frame and Signaling Multi-Frame. The framing can also be bypassed (unframed mode). It detects and indicates the event of out of Basic Frame Sync, out of CRC Multi-Frame, out of Signaling Multi-Frame, the Remote Alarm Indication signal and the Remote Signaling Multi-Frame Alarm Indication signal. It also monitors the Red and AIS alarms. Basic Frame Alignment Signal errors, Far End Block Errors (FEBE) and CRC errors are counted. Up to three HDLC links are provided to extract the HDLC message on TS16, the Sa National bits and/or any arbitrary timeslot. An Elastic Store Buffer that optionally supports slip buffering and adaptation to backplane timing is provided. In E1 receive path, signaling debounce, signaling freezing, idle code substitution, digital milliwatt code insertion, trunk conditioning, data inversion and pattern generation or detection are also supported on a per-timeslot basis.

In E1 mode, the transmit path of each framer can be configured to generate Basic Frame, CRC Multi-Frame and Signaling Multi-Frame. The framing can also be disabled (unframed mode). It can also transmit Remote Alarm Indication signal, the Remote Signaling Multi-Frame Alarm Indication signal, AIS signal and FEBE. Up to three HDLC links are provided to insert the HDLC message on TS16, the Sa National bits and/or any arbitrary timeslot. The signaling insertion, idle code substitution, data insertion, data inversion and test pattern generation or detection are also supported on a per-timeslot basis.

In E1 mode, any four of the eight framers can be multiplexed or demultiplexed to or from one of the two 8.192M bit/s buses.

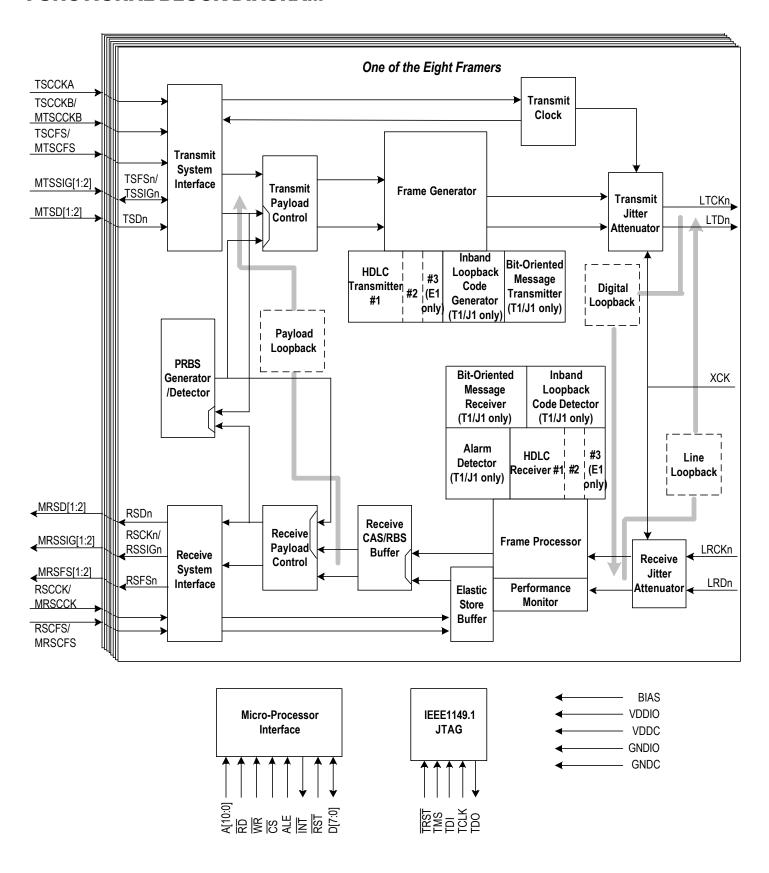
In T1/J1 mode, the receive path of each framer can be configured to frame to Super Frame (SF) or Extended Super Frame (ESF) formats. The framing can also be bypassed (unframed mode). It detects and indicates the out of SF/ESF sync event, the Yellow, Red and AIS alarms. It also detects the presence of inband loopback codes, bit oriented message. Frame Alignment Signal errors, CRC-6 errors, out of SF/ESF events and Frame Alignment position changes are counted. Up to two HDLC links are provides to extract the HDLC message on the F-bit or any arbitrary channels in ESF mode. An Elastic Store Buffer that optionally supports controlled slip and adaptation to backplane timing is provided. In T1/J1 receive path, signaling debounce, signaling freezing, idle code substitution, digital milliwatt code insertion, idle code insertion, data inversion and pattern generation or detection are also supported on a per-channel basis.

In T1/J1 mode, the transmit path of each framer can be configured to generates SF or ESF. The framing can also be disabled (unframed mode). It can also transmit Yellow signal and AIS signal. Inband loopback codes and bit oriented message can also be transmitted. Up to two HDLC links are provided to insert the HDLC message on the F-

bit or any arbitrary channels in ESF mode. The signaling insertion, idle code substitution, data insertion, data inversion and test pattern generation or detection are also supported on a per-channel basis.

In T1/J1 mode, the data stream of 1.544M bit/s can be converted to/ from the data stream of 2.048M bit/s on the system side by software configuration. In addition, any four of the eight framers can be multiplexed or de-multiplexed to or from one of the two 8.192M bit/s buses.

### **FUNCTIONAL BLOCK DIAGRAM**



## **CONTENTS**

1	PIN ASSIGNMENTS	. 13
	1.1 128 PIN PQFP PACKAGE (TOP VIEW)	13
	1.2 144 PIN PBGA PACKAGE (BOTTOM VIEW)	. 14
2	PIN DESCRIPTION	. 15
3	FUNCTIONAL DESCRIPTION	. 21
	3.1 T1/E1/J1 MODE SELECTION	
	3.2 FRAME PROCESSOR (FRMP)	
	3.2.1 E1 Mode	
	3.2.2 T1/J1 Mode	26
	3.3 PERFORMANCE MONITOR (PMON)	. 28
	3.3.1 E1 Mode	. 28
	3.3.2 T1/J1 Mode	. 28
	3.4 ALARM DETECTOR (ALMD) - T1 / J1 ONLY	. 28
	3.5 HDLC RECEIVER (RHDLC)	. 29
	3.5.1 E1 Mode	. 29
	3.5.2 T1 / J1 Mode	. 30
	3.6 BIT-ORIENTED MESSAGE RECEIVER (RBOM) - T1 / J1 ONLY	. 30
	3.7 INBAND LOOPBACK CODE DETECTOR (IBCD) - T1 / J1 ONLY	31
	3.8 ELASTIC STORE BUFFER (ELSB)	. 31
	3.8.1 E1 Mode	31
	3.8.2 T1 / J1 Mode	31
	3.9 RECEIVE CAS/RBS BUFFER (RCRB)	
	3.9.1 E1 Mode	
	3.9.2 T1 / J1 Mode	
	3.10 RECEIVE PAYLOAD CONTROL (RPLC)	
	3.10.1 E1 Mode	
	3.10.2 T1 / J1 Mode	. 34
	3.11 RECEIVE SYSTEM INTERFACE (RESI)	. 35
	3.11.1 E1 Mode	
	3.11.1.1 Receive Clock Slave Mode	
	3.11.1.1.1 Receive Clock Slave RSCK Reference Mode	
	3.11.1.2 Receive Clock Master Mode	
	3.11.1.2.1 Receive Clock Master Full E1 Mode	
	3.11.1.2.2 Receive Clock Master Fractional E1 (with F-bit) Mode	
	3.11.1.3 Receive Multiplexed Mode	
	3.11.1.4 Parity Check & Polarity Fix	
	3.11.1.5 Offset	
	3.11.1.6 Output On RSDn/MRSD & RSSIGn/MRSSIG	. 44

3.11.2 T1 / J1 Mode	47
3.11.2.1 Receive Clock Slave Mode	47
3.11.2.1.1 Receive Clock Slave RSCK Reference Mode	
3.11.2.1.2 Receive Clock Slave External Signaling Mode	
3.11.2.2 Receive Clock Master Mode	
3.11.2.2.1 Receive Clock Master Full 11/J1 Mode	
3.11.2.3 Receive Multiplexed Mode	
3.11.2.4 Parity Check	53
3.11.2.5 Offset	55
3.11.2.6 Output On RSDn/MRSD & RSSIGn/MRSSIG	55
3.12 PRBS GENERATOR / DETECT OR (PRGD)	56
3.12.1 E1 Mode	56
3.12.2 T1 / J1 Mode	57
3.13 TRANSMIT SYSTEM INTERFACE (TRSI)	57
3.13.1 E1 Mode	
3.13.1.1 Transmit Clock Slave Mode	58
3.13.1.1.1 Transmit Clock Slave TSFS Enable Mode	
3.13.1.1.2 Transmit Clock Slave External Signaling Mode	
3.13.1.2 Transmit Clock Master Mode	
3.13.1.3 Transmit Multiplexed Mode	
3.13.1.4 Parity Check	
3.13.1.5 Offset	
3.13.2 T1 / J1 Mode	
3.13.2.1 Transmit Clock Slave Mode	
3.13.2.1.1 Transmit Clock Slave TSFS Enable Mode	
3.13.2.2 Transmit Clock Master Mode	
3.13.2.3 Transmit Multiplexed Mode	72
3.13.2.4 Parity Check	75
3.13.2.5 Offset	
3.14 TRANSMIT PAYLOAD CONTROL (TPLC)	76
3.14.1 E1 Mode	
3.14.2 T1 / J1 Mode	
3.15 FRAME GENERATOR (FRMG)	77
3.15.1 E1 Mode	
3.15.2 T1 / J1 Mode	78
3.16 HDLC TRANSMITTER (THDLC)	
3.16.1 E1 Mode	
3.16.2 T1 / J1 Mode	
3.17 BIT-ORIENTED MESSAGE TRANSMITTER (TBOM) - T1 / J1 ONLY	
3.18 INBAND LOOPBACK CODE GENERATOR (IBCG) - T1 / J1 ONLY	

	3.19 JITTER ATTENUATOR (RJAT/TJAT)	80
	3.19.1 E1 Mode	80
	3.19.2 T1 / J1 Mode	83
	3.20 TRANSMIT CLOCK	85
	3.20.1 E1 Mode	85
	3.20.2 T1 / J1 Mode	85
	3.21 LINE INTERFACE	85
	3.21.1 E1 Mode	85
	3.21.2 T1 / J1 Mode	85
	3.22 INTERRUPT SUMMARY	86
	3.22.1 E1 Mode	86
	3.22.2 T1 / J1 Mode	86
	3.23 LOOPBACK MODE	86
	3.23.1 Line Loopback	86
	3.23.2 Digital Loopback	86
	3.23.3 Payload Loopback	86
	3.24 CLOCK MONITOR	86
4	OPERATION	90
	4.1 E1 MODE	90
	4.1.1 Default Setting	90
	4.1.2 Various Operation Modes Configuration	90
	4.1.3 Operation Example	95
	4.1.3.1 Using The HDLC Receiver	95
	4.1.3.2 Using The HDLC Transmitter	95
	4.1.3.3 Using The PRBS Generator / Detector	99
	4.1.3.4 Using Payload Control and Receive CAS/RBS Buffer	104
	4.1.3.5 Using TJAT / Timing Option	104
	4.2 T1/J1 MODE	105
	4.2.1 Default Setting	105
	4.2.2 OPERATION IN J1 MODE	105
	4.2.3 Various Operation Modes Configuration	105
	4.2.4 Operation Example	110
	4.2.4.1 Using The HDLC Receiver	110
	4.2.4.2 Using The HDLC Transmitter	112
	4.2.4.3 Using The PRBS Generator / Detector	114
	4.2.4.4 Using Payload Control and Receive CAS/RBS Buffer	118
	4.2.4.5 Using TJAT / Timing Option	118
5	PROGRAMMING INFORMATION	119
	5.1 REGISTER MAP	119

	5.1.1 E1 Mode Register Map	120
	5.1.2 T1 / J1 Mode Register Map	123
	5.2 REGISTER DESCRIPTION	126
	5.2.1 E1 Mode	126
	5.2.2 T1 / J1 Mode	198
6	IEEE STD 1149.1 JTAG TEST ACCESS PORT	. 259
	6.1 JTAG INSTRUCTIONS AND INSTRUCTION REGISTER (IR)	261
	6.2 JTAG DATA REGISTER	261
	6.2.1 Device Identification Register (IDR)	261
	6.2.2 Bypass Register (BYR)	261
	6.2.3 Boundary Scan Register (BSR)	261
	6.3 TESTACCESS PORT CONTROLLER	261
7	PHYSICAL AND ELECTRICAL SPECIFICATIONS	. 265
	7.1 ABSOLUTE MAXIMUM RATINGS	265
	7.2 OPERATING CONDITIONS	265
	7.3 D.C. CHARACTERISTICS	265
	7.4 CLOCK RESET TIMING	266
	7.4.1 Clock Parameters E1 Configuration	266
	7.4.2 Clock Parameters T1/J1 Configuration	266
	7.5 MICROPROCESSOR READACCESS TIMING	267
	7.6 MICROPROCESSOR WRITEACCESS TIMING	268
	7.7 I/O TIMING CHARACTERISTICS	269
	7.7.1 Transmit System Interface Timing	269
	7.7.2 Receive System Interface Timing	270
	7.7.3 Receive & Transmit Line Timing	271
	7.7.3.1 Receive Line Interface Timing	271
	7 7 3 2 Transmit Line Interface Timing	271

## **LIST OF FIGURES**

Figure - 1. E1 Frame Searching Process	22
Figure - 2. Basic Frame Searching Process	23
Figure - 3. HDLC Packet	29
Figure-4.TS16ArrangementinSignalingMulti-Frame	32
Figure - 5. Signaling Output in E1 Mode	32
Figure - 6. Signaling Output in T1/J1 Mode	32
Figure - 7. Receive Clock Slave RSCK Reference Mode	36
Figure - 8. E1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 1	37
Figure - 9. E1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 2	37
Figure-10. Receive Clock Slave External Signaling Mode	
Figure - 11. E1 Receive Clock Slave External Signaling Mode - Functional Timing Example 1	38
Figure - 12. E1 Receive Clock Slave External Signaling Mode - Functional Timing Example 2	39
Figure - 13. Receive Clock Master Full E1 or T1/J1 Mode	40
Figure - 14. E1 Receive Clock Master Full E1 Mode - Functional Timing Example	40
Figure - 15. Receive Clock Master Fractional E1 or T1/J1 Mode	41
Figure-16. E1 Receive Clock Master Fractional E1 Mode-Functional Timing Example	42
Figure-17. Receive Multiplexed Mode	43
Figure - 18. E1 Receive Multiplexed Mode - Functional Timing Example 1	43
Figure - 19. E1 Receive Multiplexed Mode - Functional Timing Example 2	44
Figure - 20. Receive Bit Offset - Between RSCFS & RSDn	
Figure - 21. Receive Bit Offset - Between RSFSn & RSDn	46
Figure - 22. T1/J1 To E1 Format Conversion	48
Figure - 23. T1/J1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 1	48
Figure - 24. T1/J1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 2	49
Figure - 25. T1/J1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 3	
Figure - 26. T1/J1 Receive Clock Slave External Signaling Mode - Functional Timing Example 1	50
Figure - 27. T1/J1 Receive Clock Slave External Signaling Mode - Functional Timing Example 2	51
Figure - 28. T1/J1 Receive Clock Slave External Signaling Mode - Functional Timing Example 3	51
Figure - 29. T1/J1 Receive Clock Master Full T1/J1 Mode - Functional Timing Example	52
Figure-30. T1/J1 Receive Clock Master Fractional T1/J1 Mode-Functional Timing Example	53
Figure-31. T1/J1 Receive Multiplexed Mode-Functional Timing Example 1	54
Figure-32. T1/J1 Receive Multiplexed Mode-Functional Timing Example 2	54
Figure - 33. Receive Bit Offset in T1/J1 Mode	
Figure - 34. PRBS Pattern Generator	56
Figure - 35. Transmit Clock Slave TSFS Enable Mode	
Figure - 36. E1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 1	59
Figure - 37. E1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 2	
Figure-38. Transmit Clock Slave External Signaling Mode	60

Figure - 39. E1 Transmit Clock Slave External Signaling Mode - Functional Timing Example 1	60
Figure - 40. E1 Transmit Clock Slave External Signaling Mode - Functional Timing Example 2	61
Figure - 41. Transmit Clock Master Mode	61
Figure - 42. E1 Transmit Clock Master Mode - Functional Timing Example	62
Figure-43. Transmit Multiplexed Mode	63
Figure - 44. E1 Transmit Multiplexed Mode - Functional Timing Example 1	64
Figure - 45. E1 Transmit Multiplexed Mode - Functional Timing Example 2	64
Figure - 46. Transmit Bit Offset in E1 Mode - 1	65
Figure - 47. Transmit Bit Offset in E1 Mode - 2	66
Figure - 48. Transmit Bit Offset in E1 Mode - 3	66
Figure - 49. Transmit Bit Offset in E1 Mode - 4	67
Figure - 50. Transmit Bit Offset in E1 Mode - 5	67
Figure - 51. E1 To T1/J1 Format Conversion	69
Figure - 52. T1/J1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 1	69
Figure - 53. T1/J1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 2	70
Figure - 54. T1/J1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 3	70
Figure-55. T1/J1 Transmit Clock Slave External Signaling Mode-Functional Timing Example 1	71
Figure-56. T1/J1 Transmit Clock Slave External Signaling Mode-Functional Timing Example 2	71
Figure-57. T1/J1 Transmit Clock Slave External Signaling Mode-Functional Timing Example 3	72
Figure-58. T1/J1 Transmit Clock Master Mode-Functional Timing Example	73
Figure-59. T1/J1 Transmit Multiplexed Mode-Functional Timing Example 1	74
Figure-60. T1/J1 Transmit Multiplexed Mode-Functional Timing Example 2	74
Figure - 61. Transmit Bit Offset in T1/J1 Mode - 1	75
Figure - 62. Transmit Bit Offset in T1/J1 Mode - 2	75
Figure - 63. E1 Mode Jitter Tolerance (N1 = N2 = 2fH)	82
Figure - 64. E1 Mode Jitter Transfer (N1 = N2 = 2fH)	82
Figure - 65. T1/J1 Mode Jitter Tolerance (N1 = N2 = 2fH)	84
Figure - 66. T1/J1 Mode Jitter Transfer (N1 = N2 = 2fH)	84
Figure - 67. Transmit Clock Select	85
Figure-68. Line Loopback	87
Figure-69. Digital Loopback	88
Figure-70. Payload Loopback	89
Figure - 71. Interrupt Service in E1 Mode HDLC Receiver	96
Figure - 72. Writing Data to E1 Mode THDLC FIFO	97
Figure - 73. Interrupt Service in E1 Mode HDLC Transmitter	98
Figure - 74. Polling Mode in E1 Mode HDLC Transmitter	99
Figure - 75. Writing Sequence of Indirect Register in E1 Mode	104
Figure - 76. Reading Sequence of Indirect Register in E1 Mode	104
Figure - 77. Interrupt Service in T1/J1 Mode HDLC Receiver	. 111
Figure - 78. Writing Data to T1/J1 Mode THDLC FIFO	. 112

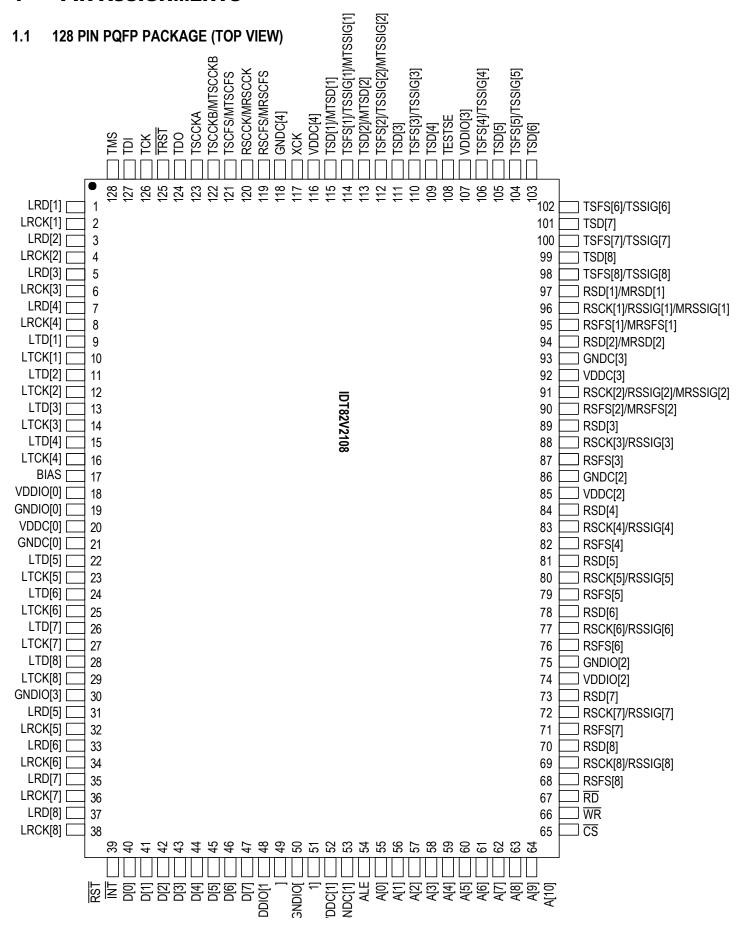
Figure - 79. Interrupt Service in T1/J1 Mode HDLC Transmitter	113
Figure - 80. Polling Mode in T1/J1 Mode HDLC Transmitter	114
Figure - 81. Writing Sequence of Indirect Register in T1/J1 Mode	118
Figure - 82. Reading Sequence of Indirect Register in T1/J1 Mode	118
Figure-83.JTAGArchitecture	259
Figure - 84. JTAG State Diagram	264
Figure - 85. Read Access Timing	267
Figure - 86. Write Access Timing	268
Figure - 87. Transmit Interface Timing (Transmit System Common Clock#B)	269
Figure-88. TransmitInterface Timing (Line Transmit Clock)	269
Figure - 89. Receive Interface Timing (Receive System Common Clock)	270
Figure - 90. Receive Interface Timing (Receive System Clock)	270
Figure-91. Receive Line Interface Timing	271
Figure - 92. Transmit Line Interface Timing	271

## **LIST OF TABLES**

Table - 1. The Structure Of TS0	23
Table - 2. Interrupt Sources In The E1 Frame Processor	25
Table - 3. The Structure Of SF Format	26
Table - 4. The Structure Of ESF Format	27
Table - 5. Interrupt Sources In The T1 / J1 Frame Processor	27
Table - 6. Basic Frame Alignment Pattern Error Counter	28
Table - 7. Alarm Summary in ALMD	29
Table - 8. A-Law Digital Milliwatt Pattern	34
Table - 9. u-Law Digital Milliwatt Pattern	34
Table - 10. E1 Mode Receive System Interface in Different Operation Modes	35
Table - 11. Operation Mode Selection in E1 Receive Path	35
Table - 12. Active Edge Selection of RSCCK (in E1 Receive Clock Slave RSCK Reference Mode)	36
Table - 13. Active Edge Selection of RSCCK (in E1 Receive Clock Slave External Signaling Mode)	38
Table - 14. Active Edge Selection of RSCK (in E1 Receive Clock Master Mode)	39
Table - 15. Active Edge Selection of MRSCCK (in E1 Receive Multiplexed Mode)	41
Table - 16. Offset in Different Operation Modes	45
Table - 17. Receive System Interface Bit Offset (FPMODE [b5, E1-011H] = 0)	45
Table - 18. Receive System Interface Bit Offset (FPMODE [b5, E1-011H] = 1)	45
Table - 19. Bit Offset Between RSFSn and RSDn When the BRXSMFP and the ALTIFP (b2, b0, E1-011H) are Both Set To Logical 1	45
Table - 20. T1/J1 Mode Receive System Interface in Different Operation Modes	47
Table - 21. Operation Mode Selection in T1/J1 Receive Path	47
Table - 22. Active Edge Selection of RSCCK (in T1/J1 Receive Clock Slave RSCK Reference Mode)	48
Table - 23. Active Edge Selection of RSCCK (in T1/J1 Receive Clock Slave External Signaling Mode)	50
Table - 24. Active Edge Selection of MRSCCK (in T1/J1 Receive Multiplexed Mode)	52
Table - 25. Receive System Interface Bit Offset	55
Table - 26. E1 Mode Transmit System Interface in Different Operation Modes	58
Table - 27. Operation Mode Selection in E1 Transmit Path	58
Table - 28. Active Edge Selection of TSCCKB (in E1 Transmit Clock Slave TSFS Enable Mode)	60
Table - 29. Active Edge Selection of TSCCKB (in E1 Transmit Clock Slave External Signaling Mode)	60
Table - 30. Active Edge Selection of MTSCCKB (in E1 Transmit Multiplexed Mode)	63
Table - 31. Transmit System Interface Bit Offset (CHI [b3, E1-01CH] = 1, CMS [b2, E1-018H] = 0)	65
Table - 32. Transmit System Interface Bit Offset (CHI [b3, E1-01CH] = 1, CMS [b2, E1-018H] = 1)	65
Table - 33. T1/J1 Mode Transmit System Interface in Different Operation Modes	68
Table - 34. Operation Mode Selection in T1/J1 Transmit Path	68
Table - 35. Active Edge Selection of TSCCKB (in T1/J1 Transmit Clock Slave TSFS Enable Mode)	69
Table - 36. Remote Alarm Indication	77
Table - 37. Content in International Bits (when the INDIS [b1, E1-040H] is logic 0)	78
Table - 38. Interrupt Summary	78

Table - 39. Default Setting in Receive Path	90
Table - 40. Default Setting in Transmit Path	90
Table - 41. Various Operation Modes in Receive Path for Reference	91
Table - 42. Various Operation Modes in Transmit Path for Reference	93
Table - 43. Example for Using HDLC Receiver	95
Table - 44. Example for Using HDLC Transmitter	97
Table - 45. Test Pattern	100
Table - 46. The Setting of PRGD	101
Table - 47. Initializtion of TPLC	101
Table - 48. Initializtion of RPLC	103
Table - 49. Error Insertion	103
Table - 50. Default Setting in Receive Path	105
Table - 51. Default Setting in Transmit Path	105
Table - 52. Various Operation Modes in Receive Path for Reference	106
Table - 53. Various Operation Modes in Transmit Path for Reference	108
Table - 54 . Example for Using HDLC Receiver	110
Table - 55. Example for Using HDLC Transmitter	112
Table - 56. Test Pattern	115
Table - 57. The Setting of PRGD	116
Table - 58. Initializtion of TPLC	116
Table - 59. Initializtion of RPLC	117
Table - 60. Error Insertion	118
Table - 61. T1/E1 Mode Selection Register	119
Table - 62a. E1 Mode Register Map - Direct Register	120
Table - 62b. E1 Mode Register Map - Indirect Register	123
Table - 63a. T1/J1 Mode Register Map - Direct Register	123
Table - 63b. T1/J1 Mode Register Map - Indirect Register	125
Table - 64. IR Code	260
Table - 65. IDR	261
Table - 66. Boundary Scan Sequence and the I/O Pad Cell Type	261
Table - 67. TAP Controller State Description	263

### 1 PIN ASSIGNMENTS



## 1.2 144 PIN PBGA PACKAGE (BOTTOM VIEW)

	12	11	10	9	8	7	6	5	4	3	2	1	-
Α	TSD[7]	TSD[6]	TSFS[4]/ TSSIG[4]	GNDC[4]	TSD[3]	TSFS[1]/ TSSIG[1]/ MTSSIG[1]/	VDDC[3]	RSCCK/ MRSCCK	TSCCKA	TCK	LRD[1]	LRD[3]	A
В	RSD[1]/ MRSD[1]	TSD[8]	TSFS[6]/ TSSIG[6]	TSD[5]	TSFS[2]/ TSSIG[2]/ MTSSIG[2]/	TSD[1]/ MTSD[1]	XCK	RSCFS/ MRSCFS	TSCCKB/ MTSCCKB	TMS	LRD[2]	LRD[4]	В
С	RSD[2]/ MRSD[2]	RSCK[1]/ RSSIG[1]/ MRSSIG[1]/	TSFS[8]/ TSSIG[8]	TSFS[7]/ TSSIG[7]	VDDC[4]	TSFS[3]/ TSSIG[3]	GNDC[3]	TSCFS/ MTSCFS	TDO	LRCK[1]	LRCK[3]	LTCK[1]	С
D	RSCK[2]\ RSSIG[2]\ MRSSIG[2]	RSFS[2]/ MRSFS[2]	(VDDIO[3])	RSFS[1]/ MRSFS[1]	TSFS[5]/ TSSIG[5]	TSD[4]	TSD[2]/ MTSD[2]	TRST	TDI	LRCK[2]	LTD[1]	LTCK[2]	D
Е	RSCK[3]/ RSSIG[3]	RSFS[3]	RSD[3]	GNDIO[3]	VDDC[5]	VDDC[6]	VDDC[7]	VDDC[8]	LRCK[4]	LTD[2]	LTCK[4]	LTD[4]	E
F	RSD[4]	VDDC[2]	GNDC[2]	RSD[5]	(VDDC[9]	VDDC[10]	VDDC[11]	VDDC[12]	LTD[3]	LTCK[3]	GNDIO[0]	(VDDIO[0])	F
G	RSFS[4]	RSCK[4]/ RSSIG[4]	RSD[6]	RSCK[6]/ RSSIG[6]	GNDC[5]	GNDC[6]	GNDC[7]	GNDC[8]	BIAS	LTD[5]	GNDC[0]	VDDC[0]	G
Н	RSFS[5]	RSCK[5]/ RSSIG[5]	TESTSE	RSCK[7]/ RSSIG[7]	GNDC[9]	GNDC[10]	GNDC[11]	GNDC[12]	LTCK[8]	LTCK[6]	LTCK[5]	LTD[6]	Н
J	RSFS[6]	RSD[7]	RSFS[8]	(A[9]	(A[7]	VDDIO[1]	D[4]	(INT)	LRD[5]	LTD[8]	LTD[7]	LTCK[7]	J
K	VDDIO[2]	RSD[8]	$\overline{\mathbb{WR}}$	(A[6]	A[3]	A[0]	D[5]	D[2]	LRCK[7]	LRCK[6]	LRCK[5]	GNDIO[1]	К
L	RSFS[7]	$\overline{\mathbb{R}\mathbb{D}}$	A[10]	(A[4]	A[1]	ALE	VDDC[1]	D[7]	D[0]	LRCK[8]	LRD[7]	LRD[6]	L
M	RSCK[8]/ RSSIG[8]	CS	A[8]	A[5]	A[2]	GNDC[1]	GNDIO[2]	D[6]	D[3]	D[1]	RST	LRD[8]	М
	12	11	10	9	8	7	6	5	4	3	2	1	J

## **2 PIN DESCRIPTION**

Nama Tima Pin No.			No	
		PBGA	Description	
		. 4		Line and System Interface
LRD[1]	Input	1	A2	LRD[1:8]: Line Receive Data for Framer 1 ~ 8
LRD[2]		3	B2	These pins receive the data stream from line interface units or from a higher demultiplex
LRD[3]		5	A1	interface. Data on these pins are sampled on the active edge of the corresponding
LRD[4]		7	B1	LRCKn.
		31	J4	LIVOIVII.
LRD[5]		33	L1	
LRD[6]		35	L2	
LRD[7]				
LRD[8]	1(	37	M1	LDOVA OLIVEDENIA OLIVETANIA O
LRCK[1]	Input	2	C3	LRCK[1:8]: Line Receive Clock for Framer 1 ~ 8
LRCK[2]		4	D3	These pins receive externally recovered line clock (2.048 or 1.544 MHz). The clock is
LRCK[3]		6	C2	used to sample the data on the corresponding LRDn.
LRCK[4]		8	E4	
LRCK[5]		32	K2	
LRCK[6]		34	K3	
LRCK[7]		36	K4	
LRCK[8]		38	L3	
RSCK[1]/RSSIG[1]/	Output	96	C11	RSCK[1:8]: Receive Side System Clock for Framer 1 ~ 8
MRSSIG[1]				In Receive Clock Master Full E1 or T1/J1 mode, the clock is a smoothed version of the
RSCK[2]/RSSIG[2]/		91	D12	corresponding 2.048 or 1.544 MHz Line Receive Clock (LRCK). The RSCKn is pulsed for
MRSSIG[2]				each bit in the 256-bit or 193-bit frame. The corresponding RSFSn and RSDn pins are
RSCK[3]/RSSIG[3]		88	E12	updated on the active edge of the RSCKn.
RSCK[4]/RSSIG[4]		83	G11	In Receive Clock Master Nx64K mode, the clock is a gapped version of the associated
RSCK[5]/RSSIG[5]		80	H11	smoothed LRCKn. The pulse number of the RSCKn in each frame is controllable from 0 to
RSCK[6]/RSSIG[6]		77	G9	255 or from 0 to 192 on a per-timeslot/channel basis. The corresponding RSFSn and
RSCK[7]/RSSIG[7]		72	H9	RSDn pins are updated on the active edge of the RSCKn.
RSCK[8]/RSSIG[8]		69	M12	In Receive Clock Slave RSCK Reference mode, the RSCKn can be selected to be either
'''				a 2.048/1.544 MHz jitter attenuated version of the corresponding LRCKn or an 8KHz clock
				divided down from the smoothed line clock LRCKn.
				RSSIG[1:8]: Receive Side System Signaling for Framer 1 ~ 8
				In Receive Clock Slave External Signaling mode, the extracted signaling is output on
				these pins. The signal on these pins is timeslot/channel-aligned with the data output on
				the corresponding RSDn pin and is updated on the active edge of the RSCCK. The
				extracted signaling is located in the lower nibble (b5 ~ b8). In E1 mode, the extracted
				signaling repeats during the entire Signaling Multi-Frame for the same timeslot. In T1/J1
				mode, the extracted signaling repeats during the entire SF/ESF for the same channel.
				MRSSIG[1:2]: Multiplexed Receive Side System Signaling
				When the multiplexed bus structure is configured, the extracted signaling data from the
				selected framers are multiplexed on these pins using a byte-interleaved multiplexing
				scheme. The data on the MRSSIG[1:2] are updated on the active edge of the MRSCCK.
RSD[1]/MRSD[1]	Output	97	B12	RSD[1:8]: Receive Side System Data for Framer 1 ~ 8
RSD[2]/MRSD[2]		94	C12	The processed data stream is output on these pins.
RSD[3]		89	E10	In Receive Clock Master mode, the RSDn is updated on the active edge of the
RSD[4]		84	F12	corresponding RSCKn.
RSD[5]		81	F9	In Receive Clock Slave mode, the RSDn is updated on the active edge of the RSCCK.
RSD[6]		78	G10	This toolive shock share mode, the result is appeared on the active edge of the resource.
RSD[7]		73	J11	MRSD[1:2]: Multiplexed Receive Side System Data
RSD[8]		70	K11	When the multiplexed bus structure is configured, the processed data stream from the
1,00[0]		, ,	'`''	selected framers is multiplexed on these pins using the byte-interleaved multiplexing
				scheme. The data on the MRSD[1:2] are updated on the active edge of the MRSCCK.
				Solicing. The data on the Mixod [1.2] are appared on the active edge of the Mixod N.

	_	Pin No.						
Name	Type	PQFP	PBGA	Description				
RSFS[1]/MRSFS[1] RSFS[2]/MRSFS[2] RSFS[3] RSFS[4] RSFS[5] RSFS[6] RSFS[7] RSFS[7] RSFS[8]	Output	95 90 87 82 79 76 71 68	D9 D11 E11 G12 H12 J12 L12 J10	RSFS[1:8]: Receive Side System Frame Pulse for Framer 1 ~ 8  In E1 mode, RSFSn can be configured to indicate the beginning of Basic Frame, or CRC Multi-Frame or/and Signaling Multi-Frame for data stream on RSDn. When configured for the Basic Frame, RSFSn will pulse high/low during the first bit of each Basic Frame. When configured for CRC Multi-Frame, RSFSn will pulse during the first bit of the first frame of the CRC Multi-Frame. When configured for the Signaling Multi-Frame, RSFSn will pulse during the first bit of the first frame of the Signaling Multi-Frame. When configured to indicate both Signaling and CRC Multi-Frame, RSFSn will go high/low on the first bit of the first frame of the Signaling Multi-Frame and go the opposite after the first bit of the first frame of the CRC Multi-Frame.  In T1/J1 mode, RSFSn can be configured to indicate each F-bit, or the first F-bit of every 12-frame SF / every 24-frame ESF. RSFSn pulses during the above F-bit.  In both E1 and T1/J1 mode, when Receive Clock Master mode is active, the RSFSn is updated on the active edge of the RSCCK.  MRSFS[1:2]: Multiplexed Receive Side System Frame Pulse  When the multiplexed bus structure is configured, the signals on these pins indicate the beginning of a multiplexed frame. The MRSFS[1:2] are undeted on the active edge of				
RSCCK/MRSCCK	Input	120	A5	beginning of a multiplexed frame. The MRSFS[1:2] are updated on the active edge of the MRSCCK.  RSCCK: Receive Side System Common Clock RSCCK is used only in Receive Clock Slave mode. In E1 mode, it is a 2.048 or 4.096 MHz clock. In T1 mode, it is a 1.544 or 2.048 or 4.096 MHz clock. In Receive Clock Slave RSCK Reference mode, the RSDn and the RSFSn are updated and the RSCFS is sampled on the active edge of the RSCCK. In Receive Clock Slave External Signaling mode, the RSDn, the RSFSn and the RSSIGn are updated and the RSCFS is sampled on the active edge of the RSCCK.				
				MRSCCK: Multiplexed Receive Side System Common Clock When the multiplexed bus structure is configured, MRSCCK is an 8.192 or 16.384 MHz clock for the receive system multiplexed bus. The MRSCFS is sampled and the MRSD[1:2], the MRSFS[1:2] and the MRSSIG[1:2] are updated on the active edge of the MRSCCK.				
RSCFS/MRSCFS	Input	119	B5	RSCFS: Receive Side System Common Frame Pulse In Receive Clock Slave mode, RSCFS can be selected as a frame alignment reference. It is asserted on the request of each Basic Frame or each Multi-Frame in E1 mode, or it is asserted on the request of F-bit in T1/J1 mode. The RSCFS is sampled on the active edge of the RSCCK.  MRSCFS: Multiplexed Receive Side System Common Frame Pulse When the multiplexed bus structure is configured, the signal on this pin aligns the multiplexed frame to the backplane timing. The MRSCFS is sampled on the active edge of the MRSCCK.				

	_	Pin	No.	<b>-</b>
Name	Type	PQFP	PBGA	Description
TSD[1]/MTSD[1]	Input	115	B7	TSD[1:8]: Transmit Side System Data for Framer 1 ~ 8
TSD[2]/MTSD[2]		113	D6	The data streams from the system backplane are input on these pins.
TSD[3]		111	A8	In Transmit Clock Master mode, the TSDn is sampled on the active edge of the
TSD[4]		109 105	D7 B9	corresponding LTCKn.
TSD[5] TSD[6]		103	A11	In Transmit Clock Slave mode, the TSDn is sampled on the active edge of the TSCCKB.
TSD[7]		103	A12	MTSD[1:2]: Multiplexed Transmit Side System Data
TSD[8]		99	B11	When the multiplexed bus structure is configured, the data stream from the backplane is
				carried on the multiplexed bus for the selected framers. The MTSD[1:2] are sampled on
				the active edge of the MTSCCKB.
TSFS[1]/TSSIG[1]/	Output	114	A7	TSFS[1:8]: Transmit Side System Frame Pulse for Framer 1 ~ 8
MTSSIG[1]	/Input			In Transmit Clock Master mode, the TSFSn indicates the beginning of each Basic Frame
TSFS[2]/TSSIG[2]/		112	B8	in E1 mode, or indicates the F-bit of SF/ESF in T1/J1 mode. The TSFSn is updated on the
MTSSIG[2]		110	07	active edge of the corresponding LTCKn.
TSFS[3]/TSSIG[3] TSFS[4]/TSSIG[4]		110 106	C7 A10	In Transmit Clock Slave TSFS Enabled mode, the TSFSn indicates the beginning of each Basic Frame in E1 mode, or indicates the F-bit of SF/ESF in T1/J1 mode. The TSFSn is
TSFS[5]/TSSIG[5]		104	D8	updated on the active edge of the TSCCKB.
TSFS[6]/TSSIG[6]		102	B10	apadiod on the doubte sage of the 1000NB.
TSFS[7]/TSSIG[7]		100	C9	TSSIG[1:8]: Transmit Side System Signaling for Framer 1 ~ 8
TSFS[8]/TSSIG[8]		98	C10	In Transmit Clock Slave External Signaling mode, these are the TSSIG inputs. The
				signaling is located in the lower nibble (b5 ~ b8) and sampled on the active edge of the
				TSCCKB. In E1 mode, the signaling repeats during the entire Signaling Multi-Frame for the
				same timeslot. In T1/J1 mode, the signaling repeats during the entire SF/ESF for the same
				channel.
				MTSSIG[1:2]: Multiplexed Transmit Side System Signaling
				When the multiplexed bus structure is configured, the signaling on the bus is organized in
				a byte-interleaved scheme for the selected framers. The MTSSIG[1:2] are sampled on the
				active edge of the MTSCCKB.
TSCCKA	Input	123	A4	TSCCKA: Transmit Side System Common Clock A
				TSCCKA is one of the reference clocks for the transmit jitter attenuator DPLL. TSCCKA
				can be configured to input the clock as:
				1. 16.384MHz clock;
				2. Line rate: 2.048MHz (for E1) or 1.544MHz (for T1); 3. Nx8KHz (N is from 1 to 256) so long as TSCCKA is a jitter-free clock.
				The IDT82V2108 can be configured to ignore the TSCCKA and utilize LRCK and TSCCKB
				instead. The TSCCKA is replaced by LRCK if line loopback is enabled.
TSCCKB/	Input	122	B4	TSCCKB: Transmit Side System Common Clock B
MTSCCKB	·			In E1 mode, the TSCCKB is a 2.048 or 4.096 MHz clock. In T1/J1 mode, the TSCCKB is a
				1.544 or 2.048 or 4.096 MHz clock.
				In Transmit Clock Slave TSFS mode, the TSDn and TSCFS are sampled and the TSFSn
				is updated on the active edge of the TSCCKB. In Transmit Clock Slave External Signaling
				mode, the TSDn, TSSIGn and TSCFS are sampled on the active edge of the TSCCKB.
				MTSCCKB: Multiplexed Transmit Side System Common Clock B
				When the multiplexed bus structure is configured, MTSCCKB is an 8.192 or 16.384 MHz
				reference clock for the transmit system multiplexed bus. The MTSCFS, the MTSD[1:2] and
				the MTSSIG[1:2] are sampled on the active edge of the MTSCCKB.

		Pin	No.	_
Name	Type	PQFP	PBGA	Description
TSCFS/ MTSCFS	Input	121	C5	TSCFS: Transmit Side System Common Frame Pulse In Transmit Clock Slave mode, TSCFS is used to frame align all the framers to the system backplane. In E1 mode, the pulse can be configured to indicate the first bit of a Basic Frame, CRC Multi-Frame / Signaling Multi-Frame. In T1/J1 mode, the pulse can be configured to indicate the first bit of SF/ESF. The width of the pulse must be at least 1 TSCCKB cycle wide. The TSCFS is sampled on the active edge of the TSCCKB.  MTSCFS: Multiplexed Transmit Side System Common Frame Pulse When the multiplexed bus structure is configured, MTSCFS is used to frame align the multiplexed frames to the system backplane. The MTSCFS is sampled on the active edge of
				the MTSCCKB.
LTD[1] LTD[2] LTD[3] LTD[4] LTD[5] LTD[6] LTD[7] LTD[7]	Output	9 11 13 15 22 24 26 28	D2 E3 F4 E1 G3 H1 J2 J3	LTD[1:8]: Line Transmit Data for Framer 1 ~ 8  These pins output the data stream to line interface units or a higher multiplex interface.  The data on the LTDn is updated on the active edge of the corresponding LTCKn.
LTCK[1] LTCK[2] LTCK[3] LTCK[4] LTCK[5] LTCK[6] LTCK[7] LTCK[8]	Output	10 12 14 16 23 25 27 29	C1 D1 F3 E2 H2 H3 J1	LTCKn: Line Transmit Clock for Framer 1 ~ 8 It is a nominal E1 (2.048MHz) or T1/J1 (1.544MHz) clock. The LTCK can be derived from TSCCKA, TSCCKB, LRCK or XCK. On the active edge of the LTCKn, the corresponding LTDn is updated.
XCK	Input	117	B6	XCK: Crystal Clock The clock frequency equals 49.152MHz ± 50 ppm 50% duty cycle for E1 and 37.056MHz ± 32 ppm 50% duty cycle for T1/J1.
				Microprocessor Interface
RST	Input	39	M2	RST: Reset (Active Low) A low signal for at least 100ns on this pin can reset the device anytime. The RST is a Schmitt-trigger input with weak pull-up.
CS	Input	65	M11	CS: Chip Select (Active Low)  This pin must be asserted low to enable the microprocessor interface. The signal must be asserted high at least once after power up to clear the internal test modes. A transition from high to low must occur on this pin for each Read/Write operation and cannot return to high until the operation is over.
ĪNT	Output	40	J5	INT: Open-Drain Interrupt Signal (Active Low)  This pin will keep low until all the active unmasked interrupt are acknowledged at their sources.
A[0] A[1] A[2] A[3] A[4] A[5] A[6] A[7] A[8] A[9]	Input	54 55 56 57 58 59 60 61 62 63 64	K7 L8 M8 K8 L9 M9 K9 J8 M10 J9	A[10:0]: Address Bus  The signals on these pins select the register for the microprocessor to access.

Pin No.		1 No.	D		
Name	Type	PQFP	PBGA	Description	
D[0]	I/O	41	L4	D[7:0]: Bi-directional Data Bus	
D[1]		42	М3	Signals on these pins are the data for Read/Write operation.	
D[2]		43	K5		
D[3]		44	M4		
D[4]		45	J6		
D[5]		46	K6		
D[6]		47	M5		
D[7]		48	L5		
RD	Input	67	L11	RD: Read Strobe (Active Low)	
				A low signal on this pin enables a read operation on the selected register.	
WR	Input	66	K10	WR: Write Strobe (Active Low)	
				A low signal on this pin enables a write operation on the selected register.	
ALE	Input	53	L7	ALE: Address Latch Enable	
	· ·			In non-multiplexed address/data bus, the ALE is connected to High.	
				It is internally pulled-up.	
				JTAG Signals (per IEEE 1149.1)	
TRST	Input	125	D5	TRST: Test Reset (Active Low)	
				A low signal on this pin will reset the JTAG test port anytime. This pin is a Schmitt-triggered	
				input with an internal pull-up resistor. It must be connected to the RST pin or ground when	
				JTAG is not used.	
TMS	Input	128	В3	TMS: Test Mode Select	
				The signal on this pin controls the JTAG test performance and is clocked into the device on	
				the rising edge of the TCK. This pin has an internal pull-up resistor.	
TCK	Input	126	A3	TCK: Test Clock	
				The clock for the JTAG test is input on this pin. The TDI and the TMS are clocked into the	
				device on the rising edge of the TCK and the TDO is clocked out of the device on the falling	
				edge of the TCK.	
TDI	Input	127	D4	TDI: Test Input	
				The test data are input on this pin. It is sampled on the rising edge of the TCK. This pin has an	
				internal pull-up resistor.	
TDO	Tri-State	124	C4	TDO: Test Output	
				The test data are output on this pin. It is sampled on the falling edge of the TCK. This pin is in	
				tri-state mode, except during the process of scanning of the data.	
				Supplies and Grounds	
BIAS	Power	17	G4	BIAS: +5V Bias	
				This pin enables +5V tolerance on the inputs. When +5V tolerance inputs are required, the	
				BIAS must be connected to a well-decoupled +5V rail. When +3V input is required, the BIAS	
				must be connected to a well-decoupled +3.3V DC supply.	
				During power up, the power should be applied to the BIAS pin before any of VDDC/VDDIO	
			_	pins is powered.	
VDDIO[0]	Power	18	F1	VDDIO[3:0]:	
VDDIO[1]		49	J7	These pins must be connected to a common, well-decoupled +3.3V DC supply together with	
VDDIO[2]		74	K12	the core power pins VDDC[4:0] externally.	
VDDIO[3]		107	D10		

Nama	Tura	Pin	No.	December 1997
Name	Type	PQFP	PBGA	Description
VDDC[0] VDDC[1] VDDC[2] VDDC[3] VDDC[4] VDDC[5:12]	Power	20 51 85 92 116	G1 L6 F11 A6 C8 E8, E7, E6, E5, F8, F7, F6,	VDDC[4:0]: These pins must be connected to a common, well-decoupled +3.3V DC supply together with the pad ring power pins VDDIO[3:0] externally. The VDDC[5:12] are extra power pins for PBGA.
GNDIO[0] GNDIO[1] GNDIO[2] GNDIO[3]	Ground	19 50 75 30	F2 K1 M6 E9	GNDIO[3:0]: These pins must be connected to a common ground together with the core ground pins GNDC[4:0].
GNDC[0] GNDC[1] GNDC[2] GNDC[3] GNDC[4] GNDC[5:12]	Ground	21 52 86 93 118 -	G2 M7 F10 C6 A9 G8, G7, G6, G5, H8, H7, H6,	GNDC[4:0]: These pins must be connected to a common ground together with the pad ring ground pins GNDIO[3:0]. The GNDC[5:12] are extra ground pins for PBGA.
TESTSE	Input	108	H10	This pin is connected to ground for normal operation and reserved for testing.

#### Notes:

- 1. All outputs have 4mA drive capability except for the D[7:0], the LTCK[1:8] and the RSCK[1:8] pins which have 6mA drive capability.
- 2. All input and bi-directional pins present minimum capacitive loading.

#### 3 FUNCTIONAL DESCRIPTION

#### 3.1 T1 / E1 / J1 MODE SELECTION

The IDT82V2108 can be configured as a duplex eight ports E1 framer, or a duplex eight ports T1 framer, or a duplex eight ports J1 framer. When the TMODE (b0, 400H)\* is set to 0, the device is in E1 mode. When the TMODE (b0, 400H) is set to 1, the device is in T1/J1 mode (default mode). In T1/J1 mode, when the JYEL (b3, T1/J1-020H) and the J1\_YEL (b5, T1/J1-02CH) are both set to 0, the receive path of the corresponding framer is in T1 mode; when the JYEL (b3, T1/J1-020H) and the J1\_YEL (b5, T1/J1-02CH) are both set to 1, the receive path of the corresponding framer is in J1 mode; when the J1\_CRC (b6, T1/J1-044H) and the J1\_YEL (b5, T1/J1-044H) are both set to 0, the transmit path of the corresponding framer is in T1 mode; when the J1\_CRC (b6, T1/J1-044H) and the J1\_YEL (b5, T1/J1-044H) are both set to 1, the transmit path of the corresponding framer is in J1 mode.

#### 3.2 FRAME PROCESSOR (FRMP)

The Frame Processor of each framer operates independently.

#### 3.2.1 E1 MODE

In E1 mode, the Frame Processor searches for Basic Frame synchronization, CRC Multi-frame synchronization, and Channel Associated Signaling (CAS) Multi-frame synchronization in the received data stream. Figure - 1 shows the searching process.

Once the frame is synchronized, the Frame Processor keeps on monitoring the received data stream. If there are any framing bit errors, CAS Multi-Frame alignment pattern errors, CRC Multi-Frame alignment pattern errors or CRC errors, the Frame Processor will indicate these errors. The status of loss of frame, loss of Signaling Multi-Frame and loss of CRC Multi-Frame can also be detected and declared based on user-selectable criteria. The reframe operation can be initiated by excessive CRC errors, or the CRC Multi-Frame alignment is not found within 400ms. A software reset can also make the Frame Processor reframe.

The Frame Processor can extract the data stream in TS16, and output the extracted data on a separate pin. The Frame Processor also extracts the contents of the International bits (from both the FAS and the NFAS frames), the National bits and the Extra bits (from TS16 in the frame 0 of the Signaling Multi-Frame), and stores these data in registers. The CRC Sub Multi-Frame alignment 4 bit codeword in the National bit positions Sa4 to Sa8 can also be extracted and stored in registers, and updated every CRC Sub Multi-Frame.

The Framer Processor identifies the Remote Alarm bit (bit 3 of TS0 of NFAS frames) and Remote Signaling Multi-Frame Alarm (bit 6 of TS16 of the frame 0 of the Signaling Multi-Frame). The "de-bounced" Remote Alarm and Remote Signaling Multi-Frame Alarm can be indicated if the corresponding bit has been a certain logic for consecutive 2 or 3 times. The AIS (Alarm Indication Signal) can also be detected, and if the AIS

#### Note:

\* The contents in the brackets indicate the position of this bit and the address of the register. If more than one register contains the same bit, the address is only for the first register, the addresses of the remaining registers are listed together with the first register in the Register Description paragraph. condition has persisted for at least 100 ms, an AIS Alarm is declared. The Frame Processor can also declare a Red Alarm if the out-of-frame condition has persisted for at least 100 ms.

An interrupt output is provided to indicate status changes and the occurrence of some events. The interrupts may be generated every Basic Frame, CRC Sub Multi-Frame, CRC Multi-Frame or Signaling Multi-Frame.

The Frame Processor can also be bypassed to receive unframed data.

#### **Basic Frame**

The algorithm of searching for the E1 Basic Frame alignment pattern (as shown in Figure - 2) meets the ITU-T Recommendation G.706 4.1.2 and 4.2

Generally, it is performed by detecting a successive FAS/NFAS/FAS sequence. If STEP 2 is not met, a new search will start after the following frame is skipped. If STEP 3 is not met, a new search will start immediately in the next frame. Once the Basic Frame alignment pattern is detected in the received PCM data stream, the Basic Frame synchronization is found and the OOFV (b6, E1-036H) will be set to logic 0 for indication. Then, this block goes on monitoring the received data stream. If the received Basic Frame alignment signal does not meet its pattern, it will be indicated by setting the FERI (b2, E1-034H). The criteria of out of Basic Frame synchronization are selected by the BIT2C (b6, E1-031H). If one of the conditions that set in the BIT2C (b6, E1-031H) is met, the search process will restart when the REFRDIS (b0, E1-030H) is 0. Excessive CRC errors will also lead to re-searching for the Basic Frame (refer to "CRC Multi-Frame" for details).

However, the Basic Frame synchronization can also be forced to research for a new Basic Frame any time when there is a transition from 0 to 1 on the REFR (b2, E1-030H).

#### **CRC Multi-Frame**

The CRC Multi-Frame is provided to enhance the ability of verifying the data stream. The structure of TS0 of CRC Multi-Frame is illustrated in Table - 1:

A CRC Multi-Frame consists of 16 continuous Basic Frames (No. 0 - 15) which are numbered from a Basic Frame with FAS. Each CRC Multi-Frame can be divided into two Sub Multi-Frames (SMF I & SMF II).

The first bit of TS0 of each frame is called International (Si) bit. The Si bit in each even frame is the CRC bit. Thus, there are C1, C2, C3, C4 in each SMF. The C1 is the most significant bit, while the C4 is the least significant bit. The Si bit in the first six odd frames is the CRC Multi-Frame alignment pattern. Its pattern is '001011'. The Si bit in the Frame 13 and the Frame 15 are E1 and E2 bits. The E bits' value can indicate the Far End Block Errors (FEBE).

After the Basic Frame has been synchronized, the Frame Processor initiates an 8 and 400ms timer to check the CRC Multi-Frame alignment signal if the CRCEN (b7, E1-030H) is 1. The CRC Multi-Frame synchronization is declared with a logic 0 in the OOCMFV (b4, E1-036H) only if at least two CRC Multi-Frame alignment patterns are found within 8ms, with the interval time of each pattern being a multiple of 2ms. Then if the received CRC Multi-Frame alignment signal does not meet its pattern, it will be indicated by the CMFERI (b0, E1-034H). The Frame Processor calculates the data in the SMF(N) per the algorithm in the G.704 and the G.706 to get a four-bit remainder, then compares the four-bit remainder with the C1, C2, C3, C4 in the next SMF. If there is a difference between

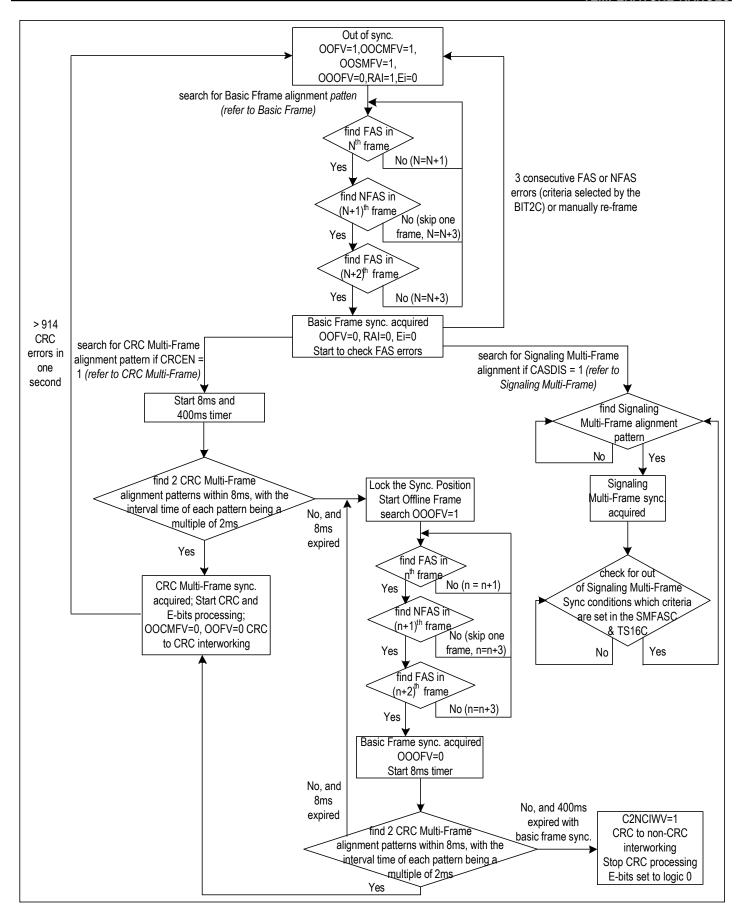


Figure - 1. E1 Frame Searching Process

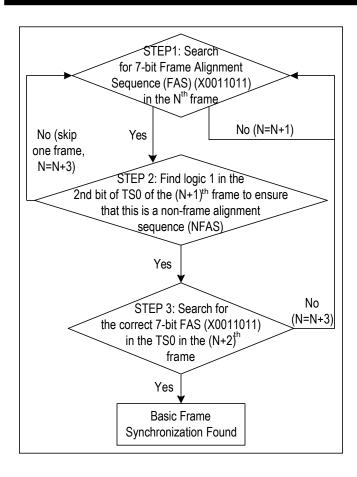


Figure - 2. Basic Frame Searching Process

them, bit errors exist in SMF(N) and a CRC error is counted. The CRCERR[9:0] (b7~0, E1-039H & b1~0, E1-03AH) are used to indicate the CRC error numbers and are updated every second. Once the CRCERR[9:0] (b7~0, E1-039H & b1~0, E1-03AH) are updated, a logic 1 will be set in the NEWDATA (b6, E1-03AH) for indication. If the CRCERR[9:0] are over-written, the OVR (b7, E1-03AH) will be asserted. When more than 914 CRC errors occur in one second which is indicated in the EXCRCERR (b0, E1-031H, a new search for the Basic Frame alignment pattern will start if the REFCRCE (b1, E1-030H) is set to 1 and the REFRDIS (b0, E1-030H) is set to 0.

If the 2 CRC Multi-Frame alignment patterns can not be found within 8ms with the interval time being a multiple of 2ms, an offline search for the Basic Frame alignment pattern will start which is indicated in the OOOFV (b3, E1-036H). The process is the same as shown in Figure - 2. This offline operation searches in parallel with the pre-found Basic Frame synchronization searching process. After the new Basic Frame synchronization is found by this offline search, the 8ms timer is restarted to check whether the two CRC Multi-Frame alignment patterns are found within 8ms, with the interval time of each pattern being a multiple of 2ms again. If the condition can not be met, the procedure will go on until the 400ms timer ends. If the condition still can not be met at that time and the Basic Frame is still synchronized, the device declares by the C2NCIWV (b7, E1-036) to run under the CRC to non-CRC interworking process. In this process, the CRC Multi-Frame alignment pattern can still be searched if the C2NCIWCK (b5, E1-030H) is logic 1.

#### **CAS Signaling Multi-Frame**

If the CRCEN (E1-030H) is logic 1, after the CRC Multi-Frame has been found, the Frame Processor starts to search for Signaling Multi-Frame alignment pattern when the CASDIS (E1-030H) is logic 0. If the CRCEN is logic 0, after the Basic Frame has been found, the Frame Processor starts to search for Signaling Multi-Frame alignment signal when the CASDIS (E1-030H) is logic 0. Refer to Figure - 1.

The Signaling Multi-Frame alignment pattern is located in the 1-4 bits of TS16 of Frame 0 of Signaling Multi-Frame. The pattern is '0000'. Once the pattern is detected, the Signaling Multi-Frame synchronization is acquired which is indicated with a logic 0 in the OOSMFV (b5, E1-

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	CME	Basic Frame			the E	Eight Bits in Ti	meslot 0			
	SMF	No. / Type	1 (Si bit)	2	3	4	5	6	7	8
		0 /FAS	C1	0	0	1	1	0	1	1
		1 / NFAS	0	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8
		2/FAS	C2	0	0	1	1	0	1	1
	SMFI	3 / NFAS	0	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8
	SIVIF	4 / FAS	C3	0	0	1	1	0	1	1
		5 / NFAS	1	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8
		6/FAS	C4	0	0	1	1	0	1	1
CRC-4		7 / NFAS	0	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8
Multi-Frame		8/FAS	C1	0	0	1	1	0	1	1
		9/NFAS	1	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8
		10 / FAS	C2	0	0	1	1	0	1	1
	SMF II	11 / NFAS	1	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8
	SIVIE	12 / FAS	C3	0	0	1	1	0	1	1
		13 / NFAS	E1	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8
		14 / FAS	C4	0	0	1	1	0	1	1
		15 / NFAS	E2	1	Α	Sa4	Sa5	Sa6	Sa7	Sa8

036H). If the received Signaling Multi-Frame alignment signal does not meet its pattern, the SMFERI (b1, E1-034H) is set to logic 1. The entire content in TS16 of Frame 0 of Signaling Multi-Frame is '0000XYXX'. Y is for remote Signaling Multi-Frame alarm indication and X are extra bits.

A new search of Signaling Multi-Frame alignment pattern is initiated when the out of the Signaling Multi-Frame criteria set in the SMFASC (b5, E1-031H) and the TS16C (b4, E1-031H) is met or when it is out of Basic Frame synchronization.

All the above frame synchronization functions can only be executed when the UNF (b6, E1-000H) is logic 0.

#### **RED Alarm**

RED alarm is declared when the out of Basic Frame sync condition has persisted for 100ms. RED alarm is removed when the out of Basic Frame sync condition has been absent for 100ms. The RED alarm status is reflected in the RED (b3, E1-037H).

The received data stream is out of Basic Frame sync when: 1) The Basic Frame has not been synchronized; 2) The received data stream meets the out of Basic Frame sync criteria set in the BIT2C (b6, E1-031H); 3) There are excessive CRC errors in the received data stream. Any one of the three conditions will be indicated by the OOFV (b6, E1-036H).

The integration of RED alarm uses the following algorithm: The algorithm monitors the occurrence of out of Basic Frame over a 4ms interval. A valid out of Basic Frame presence is accumulated when one or more out of Basic Frame indications occurred during the 4ms interval. Each valid out of Basic Frame presence increases one accumulation. An invalid out of Basic Frame presence is also accumulated when there is no out of Basic Frame indication occurring during the 4ms inverval. Each invalid out of Basic Frame indication decreases one accumulation (until the accumulation is zero). The RED alarm is declared when 25 valid out of Basic Frame presences have been accumulated. The RED alarm is removed when the out of Basic Frame presences reaches 0.

#### AIS Alarm

The AIS density criteria are selected in the AISC (b1, E1-031H). That is, if it is out of Basic Frame synchronization and less than 3 zeros are detected in a 512-bit stream, or if it is out of Basic Frame synchronization and less than 3 zeros are detected in each of 2 consecutive 512-bit streams, the status will be reported by the AISD (b5, E1-037H).

When the above status has lasted for 100ms, AIS alarm is declared with a logic 1 in the AIS (b2, E1-037H). However, in unframed mode, the detection of AIS alarm is disabled.

#### **Bits Extraction**

The Frame Processor extracts the National Bit codeword (Sa4[1:4] to Sa8[1:4] in the CRC Sub Multi-frame), the International bit (Si), the National bit (Sa), the Remote Alarm Indication bit (A), the Extra bits (X) and the Remote Signaling Multi-Frame Alarm Indication bit (Y).

#### - National Bit Codeword

The Frame Processor extracts one of the National Bit codeword (Sa4[1:4] to Sa8[1:4] in the CRC Sub Multi-frame) to the SaX[1:4] (b3~0, E1-03DH). Here the 'X' is selected from 4 to 8 by the SaSEL[2:0] (b7~5, E1-03BH). The SaX[1:4] (b3~0, E1-03DH) are debounced. They are updated only when two consecutive codewords are the same.

#### - International Bit

The International bits (Si) are extracted to the Si[1:0] (b7~6, E1-038H). The Si[1:0] (b7~6, E1-038H) are updated on the boundary of the associated FAS/NFAS frame and are not updated when out of frame is reported.

#### - National Bit

The National bits (Sa) are extracted to the Sa[4:8] (b4~0, E1-038H). The Sa[4:8] (b4~0, E1-038H) are updated on the boundary of the associated NFAS frame and are not updated when out of frame is reported.

#### - Remote Alarm Indication Bit

The Remote Alarm Indication bit (A) is extracted to the A (b5, E1-038H). The A (b5, E1-038H) is updated on the boundary of the associated NFAS frame.

#### - Extra Bit

The Extra bits (X) are extracted to the X[0:2] (b5 & b3~2, E1-03AH). The X[0:2] (b5 & b3~2, E1-03AH) are updated on the beginning of the Frame1 (next NFAS frame).

#### - Remote Signaling Multi-Frame Alarm Indication Bit

The Remote Signaling Multi-Frame Alarm Indication bit (Y) is extracted to the Y (b4, E1-03AH). The Y (b4, E1-03AH) is updated on the beginning of the Frame1 (next NFAS frame).

#### V5.2 Link

The V5.2 link ID signal, i.e. 2 out of 3 Sa7 bits being logic 0, is detected with the indication in the V52LINKV (b0, E1-036H).

#### Interrupt Sources

24 kinds of interrupts are derived from this block as shown in Table -2. When there are conditions meeting the interrupt sources, the corresponding Status bit will be asserted high. When there is a transition (logic 1 to 0 or logic 0 to 1) on the Status bit, the corresponding Status Interrupt Indication bit will be set to logic 1 (If the Status bit does not exist, the source will cause its Status Interrupt Indication bit to logic 1 directly) and the Status Interrupt Indication bit will be cleared when it is read. A logic 1 in the Status Interrupt Indication bit means an interrupt occurred. The interrupt will be reported by the INT pin if its Status Interrupt Enable bit is logic 1.

Table - 2. Interrupt Sources In The E1 Frame Processor

No.	Sources	Status Bit	Interrupt Indication Bit	Interrupt Enable Bit
1	The received Basic Frame alignment signal does not meet its	_	FERI (b2, E1-034H)	FERE (b2, E1-032H)
	pattern once Basic Frame sync is achieved.	_	1 LIVI (02, L1-00-11)	, ,
2	The received CRC Multi-Frame alignment signal does not meet its	-	CMFERI (b0, E1-034H)	CMFERE
3	pattern once CRC Multi-Frame sync is achieved.		, , ,	(b0, E1-032H) SMFERE
3	The received Signaling Multi-Frame alignment signal does not meet its pattern once Signaling Multi-Frame sync is achieved.	-	SMFERI (b1, E1-034H)	(b1, E1-032H)
4	The received data stream is out of Basic Frame sync, that is, when:			(61, £1-03211)
· I	The Basic Frame has not been synchronized; 2. The received	0051/		
	data stream meets the out of Basic Frame sync criteria set in the	OOFV (b6, E1-036H)	OOFI (b6, E1-034H)	OOFE (b6, E1-032H)
	BIT2C (b6, E1-031H); 3. There are excessive CRC errors in the	(50, £1-03011)		
	received data stream.			
5	The received data stream is out of CRC Multi-Frame sync, that is,	OOCMFV	OOCMEL(b4 E4 034U)	OOCMFE
	when: 1. The CRC Multi-Frame has not been synchronized; 2. There are excessive CRC errors in the received data stream.	(b4, E1-036H)	OOCMFI (b4, E1-034H)	(b4, E1-032H)
6	The received data stream is out of Signaling Multi-Frame sync, that			
	is, when: 1. The received data stream is out of Basic Frame sync; 2.	0001451/		0001455
	The received data stream meets the out of Signaling Multi-Frame	OOSMFV (b5, E1-036H)	OOSMFI (b5, E1-034H)	OOSMFE (b5, E1-032H)
	sync criteria set in the SMFASC (b5, E1-031H) and the TS16C (b4,	(00, E1-0001)		(03, E1-03211)
	E1-031H).	555		
7	The out of Basic Frame sync condition has lasted for 100ms. (e.g. the condition in item No.4 has lasted for 100ms.)	RED (b3, E1-037H)	REDI (b3, E1-035H)	REDE (b3, E1-033H)
8	The calculated CRC remainder is not equal to the received C[1:4]	(D3, E1-U3/H)		CRCEE
0	bits.	-	CRCEI (b0, E1-035H)	(b0, E1-033H)
9	It is out of Basic Frame sync and less than 3 zeros are detected in a			(50, 21 0001)
	512-bit stream, or it is out of Basic Frame sync and less than 3	AISD	VIODI (PE E1 USER)	VIGUE (PE E1 USSEI)
	zeros are detected in each of 2 consecutive 512-bit stream. These	(b5, E1-037H)	AISDI (b5, E1-035H)	AISDE (b5, E1-033H)
40	two criteria are selected in the AISC (b1, E1-031H).			
10	The condition in item No.9 has lasted for 100ms.	AIS	AISI (b2, E1-035H)	AISE (b2, E1-033H)
11	Logic 1 is received in the A bit for a certain period. The criteria are	(b2, E1-037H) RAIV		,
''	defined in the RAIC (b3, E1-031H).	(b7, E1-037H)	RAII (b7, E1-035H)	RAIE (b7, E1-033H)
12	Logic 0 is received in any of the CRC error indication (E1 or E2) bit.	(01, =1 00111,	FEDEL/64 E4 005U)	FEBEE
	, , ,	-	FEBEI (b1, E1-035H)	(b1, E1-033H)
13	A logic 1 is received in the A bit and a logic 0 is received in any of	RAICCRCV	RAICCRCI	RAICCRCE
	the E1 or E2 bits for 10ms.	(b2, E1-036H)	(b6, E1-03FH)	(b6, E1-03EH)
14	A logic 0 is received in any of the E1 or E2 bits on ≥ 990 occasions	CFEBEV (b1, E1-036H)	CFEBEI (b5, E1-03FH)	CFEBEE (b5, E1-03EH)
15	per second for the latest 5 consecutive seconds.  Logic 1 is received in the Y bit for 3 consecutive Signaling Multi-	RMAIV		
10	Frames.	(b6, E1-037H)	RMAII (b6, E1-035H)	RMAIE (b6, E1-033H)
16	The device is operating in the CRC to non-CRC inter working mode.	C2NCIWV	OONOWA!! (1.7. E4.004!)	C2NCIWE
		(b7, E1-036H)	C2NCIWI (b7, E1-034H)	(b7, E1-032H)
17	The position of the Basic Frame alignment pattern is changed.		COFAI (b3, E1-034H)	COFAE
4.5		-	001 AI (00, L1-00411)	(b3, E1-032H)
18	The device is in the procedure of the offline Basic Frame searching.	000FV	OOOFI (b7, E1-03FH)	000FE
19	The first bit of each Basic Frame is received.	(b3, E1-036H)	IFPI (b3, E1-03FH)	(b7, E1-03EH) IFPE (b3, E1-03EH)
20	The first bit of each CRC Sub Multi-Frame is received.	-		ICSMFPE
-		-	ICSMFPI (b2, E1-03FH)	(b2, E1-03EH)
21	The first bit of each CRC Multi-Frame is received.		ICMEDI (b1 E1 02ELI)	ICMFPE
		-	ICMFPI (b1, E1-03FH)	(b1, E1-03EH)
22	The first bit of each Signaling Multi-Frame is received.	_	ISMFPI (b0, E1-03FH)	ISMFPE
00	2 and of 2 Co7 bits are received as leads 0	\/E0LINIA/	, ,	(b0, E1-03EH)
23	2 out of 3 Sa7 bits are received as logic 0.	V52LINKV	V52LINKI	V52LINKE
24	There is change in the corresponding SaX[1:4] (b3~0, E1-03DH).	(b0, E1-036H)	(b4, E1-03FH) Sa4I, Sa5I, Sa6I,	(b4, E1-03EH) Sa4E, Sa5E, Sa6E,
_ <del></del> _	Thore is change in the corresponding Sax[1.4] (bb 10, L1-03D11).	_	Sa7l, Sa8l	Sa7E, Sa8E
			(b4~0, E1-03CH)	(b4~0, E1-03BH)

#### 3.2.2 T1/J1 MODE

In T1/J1 Mode, the Frame Processor searches for the frame alignment patterns in the standard Super-Frame (SF) or in the Extended Super-Frame (ESF) framing formats. The format is selected by the ESF (b4, T1/J1-020H). The searching algorithm of T1 or J1 is selected by the JYEL (b3, T1/J1-020H). The Frame Processor acquires frame alignment per ITU-T requirement.

When frame alignment is achieved, the Framer Processor continues to monitor the received data stream. If there are any of framing bit errors or bit error events, the Frame Processor declares these events. The Frame Processor can also detect out-of-frame events based on selectable criteria.

The Frame Processor can also be disabled to receive unframed data.

#### Super Frame (SF) Format

The structure of T1/J1 SF format is illustrated in the Table - 3. The SF is made up of 12 frames. Each frame consists of a one bit overhead - F Bit and 24 8-bit channels. If two consecutive valid Frame Alignment Patterns - '100011011100' for T1 / '10001101110X' for J1 - are received in the F-Bit in the received data stream, the SF synchronization is acquired. If the same pattern is received in the data stream other than in the F-bit, it is a mimic pattern. If a mimic pattern exists during the frame searching procedure, the synchronization will not be declared and the MFP (b1, T1/J1-022H) will be set to indicate the presence of a mimic pattern.

#### **Extended Super Frame (ESF) Format**

The structure of T1/J1 ESF format is illustrated in the Table-4. The ESF is made up of 24 frames. Each frame consists of a one bit overhead - F Bit and 24 8-bit channels.

The pattern of the Frame Alignment Pattern is '001011', which is located from Frame 4 in every 4th F-bit position. When the ESFFA (b5, T1/J1-020H) is set to 0, if four consecutive Frame Alignment Patterns are detected in the F-Bit in the received data stream, the ESF synchronization is acquired. If the same pattern is received in the data stream other than in the F-bit, it is a mimic pattern. If a mimic pattern exists during the frame searching procedure, the synchronization will not be declared and the MFP (b1, T1/J1-022H) will be set to indicate the presence of a mimic pattern. When the ESFFA (b5, T1/J1-020H) is set to 1, the synchronization will be declared when 6 consecutive Frame Alignment Patterns are received error free and the CRC-6 checksum is also error free. In this condition, the existance of mimic patterns will not be considered.

A 4-frame capacity buffer is used to store the data when the Frame Processor is searching for SF/ESF synchronization. Once the SF/ESF is synchronized, the buffer is relinquished by the Frame Processor and the Frame Processor continues to monitor the errors. When the FAS error ratio exceeds the criteria configured in the M2O[1:0] (b7~6, T1/J1-020H), it is out of frame.

All the frame sync function can only be executed when the UNF (b6, T1/J1-000H) is logic 0.

The interrupt sources in this block are summarized in Table - 5. When there are conditions meeting the interrupt sources, the corresponding Status bit will be asserted high. When there is a transition (logic 1 to 0 or logic 0 to 1) on the Status bit, the corresponding Status Interrupt Indication bit will be set to logic 1 (If the Status bit does not exist, the source will cause its Status Interrupt Indication bit to be logic 1 directly) and the Status Interrupt Indication bit will be cleared when it is read. A logic 1 in the Status Interrupt Indication bit indicates an interrupt occurred. The interrupt is reported by the  $\overline{\text{INT}}$  pin if its Status Interrupt Enable bit was set to logic 1.

Table - 3. The Structure Of SF Format

Frame No. In The CF	F Dit (Frame Alignment)	The Bit In	Each Channel
Frame No. In The SF	F-Bit (Frame Alignment)	Data Bit	Signaling Bit
1	1	1 - 8	-
2	0	1 - 8	-
3	0	1 - 8	-
4	0	1 - 8	-
5	1	1 - 8	-
6	1	1 - 7	A (bit 8)
7	0	1 - 8	-
8	1	1 - 8	-
9	1	1 - 8	-
10	1	1 - 8	-
11	0	1 - 8	-
12	X	1 - 7	B (bit 8)

X should be logic 0 in T1 FAS.

X can be logic 0 or 1 in J1 FAS because this position is used as Yellow Alarm Indication bit.

Table - 4. The Structure Of ESF Format

F No. 1 The FOE		F-Bit Assignment		The Bit In Each Channel		
Frame No. In The ESF	FAS	DL	CRC	Data Bit	Signaling Bit	
1	-	DL	-	1 - 8	-	
2	-	-	C1	1 - 8	-	
3	-	DL	-	1 - 8	-	
4	0	-	-	1 - 8	-	
5	-	DL	-	1 - 8	-	
6	-	-	C2	1 - 7	A (bit 8)	
7	-	DL	-	1 - 8	-	
8	0	-	-	1 - 8	-	
9	-	DL	-	1 - 8	-	
10	-	-	C3	1 - 8	-	
11	-	DL	-	1 - 8	-	
12	1	-	-	1 - 7	B (bit 8)	
13	-	DL	-	1 - 8	-	
14	-	-	C4	1 - 8	-	
15	-	DL	-	1 - 8	-	
16	0	-	-	1 - 8	-	
17	-	DL	-	1 - 8	-	
18	-	-	C5	1 - 7	C (bit 8)	
19	-	DL	-	1 - 8	-	
20	1	-	-	1 - 8	-	
21	-	DL	-	1 - 8	-	
22	-	-	C6	1 - 8	-	
23	-	DL	-	1 - 8	-	
24	1	-	-	1 - 7	D (bit 8)	

Table - 5. Interrupt Sources In The T1 / J1 Frame Processor

No.	Sources	Status Bit	Interrupt Indication Bit	Interrupt Enable Bit
1	The frame alignment mimic pattern is detected in the received data stream.	MFP (b1, T1/J1-022H)	MFPI (b3, T1/J1-022H)	MFPE (b1, T1/J1-021H)
2	The SF / ESF synchronization is found.	INFR (b0, T1/J1-022H)	INFRI (b2, T1/J1-022H)	INFRE (b0, T1/J1-021H)
3	The position of the new frame alignment pattern differs from the position of the previous one.	-	COFAI (b7, T1/J1-022H)	COFAE (b5, T1/J1-021H)
4	One bit error is detected in frame alignment pattern.	-	FERI (b6, T1/J1-022H)	FERE (b4, T1/J1-021H)
5	Two or more bit errors are detected in one frame alignment pattern.	-	SFEI (b4, T1/J1-022H)	SFEE (b2, T1/J1-021H)
6	In the SF format, one bit error is detected in frame alignment pattern. In the ESF format, the received CRC-6 is not equal to the local calculated CRC-6.	-	BEEI (b5, T1/J1-022H)	BEEE (b3, T1/J1-021H)

### 3.3 PERFORMANCE MONITOR (PMON)

The Performance Monitor is used to count various performance events in the received data stream within defined intervals. The Performance Monitor of each framer operates independently.

#### 3.3.1 E1 MODE

The PMON block counts the Basic Frame alignment pattern errors. The method of counting the errors is defined by the WORDERR (b5, E1-000H) and CNTNFAS (b4, E1-000H) as shown in Table - 6. The number of the Basic Frame alignment pattern errors counted during the interval is reflected in the FER[6:0] (b6~0, E1-069H).

Table - 6. Basic Frame Alignment Pattern Error Counter

WORDERR (b5, E1-000H)	<b>CNTNFAS</b> (b4, E1-000H)	One Error is Counted
0	0	One bit error in FAS
0	1	One bit error in FAS or a logic 0 in bit 2 of TS0 of NFAS
1	0	One or more than one bit error in a FAS
1	1	Any bit error in a FAS and the 2nd bit of TS0 of the following NFAS

The PMON block counts the Far End Block Error (FEBE) which is detected in the E1 and E2 bits. The number of the FEBE counted during the interval is stored in the FEBE[9:0] (b1~0, E1-06BH & b7~0, E1-06AH).

The block also counts the CRC errors which mean the local calculated CRC remainders are not equal to the received CRC. The number of the CRC errors counted during the interval is indicated in the CRCE[9:0] ( $b1\sim0$ , E1-06DH &  $b7\sim0$ , E1-06CH).

The above three kinds of PMON Error Count registers are deactivated when the framer is out of Basic Framer synchronization. The latter two kinds of PMON Error Count registers are also deactivated when it is out of CRC Multi-Frame synchronization.

These PMON Error Count registers in a framer can be updated as a group. The intervals are typically 1 second when the AUTOUPDATE (b0, E1-000H) of the current framer is set. They can also be updated by writing to any of these PMON Error Count registers. The PMON Error Count registers in eight framers can also be updated together by writing to the Revision / Chip ID / Global PMON Update register (E1-009H). Once the PMON Error Count registers are updated, the XFER (b1, E1-068H) will be set to logic 1 and an interrupt can be asserted on the  $\overline{\text{INT}}$  pin if the INTE (b2, E1-068H) is logic 1.

If the performance number counted in the next interval is latched in its PMON Error Count register without the previous one being read, the PMON Error Count register is over-written. Any over-writing of the three kinds of PMON Error Count registers will be indicated in the OVR (b0, E1-068H).

#### 3.3.2 T1/J1 MODE

In SF format, the Performance Monitor counts three kinds of events:

1. Every one-bit error in a frame alignment pattern is counted. The number of the errors counted during the interval is reflected in the

FER[8:0] (b0, T1/J1-04DH & b7~0, T1/J1-04CH) (In SF format, the usage of the BEE[11:0] (b3~0, T1/J1-04BH & b7~0, T1/J1-04AH) is the same as that of the FER[8:0]);

- 2. The out of SF synchronization event is counted. The number counted during the interval is reflected in the OOF[4:0] (b4~0, T1/J1-04EH):
- 3. The number of changes of the frame alignment position during the interval is counted and is reflected in the COFA[2:0] (b2~0, T1/J1-04FH).

In ESF format, the Performance Monitor counts four kinds of events:

- 1. The block counts the CRC-6 errors which mean the local calculated CRC-6 remainders are not equal to the received CRC-6. The number of the errors counted during the interval is indicated in the BEE[11:0] (b3~0, T1/J1-04BH & b7~0, T1/J1-04AH);
- 2 ~ 4. (The same events as 1 ~ 3 in the SF format, described above.) These PMON Error Count registers in a framer can be updated as a group. The intervals are typically 1 second when the AUTOUPDATE (b0, T1/J1-000H) of the framer is set. They can also be updated by writing to any of these PMON Error Count registers. The PMON Error Count registers of eight framers can also be updated together by writing to the Revision / Chip ID / Global PMON Update register (T1/J1-00CH). Once the PMON Error Count registers are updated, the XFER (b1, T1/J1-049H) will be logic 1 and an interrupt can be asserted on the INT pin if the INTE (b2, T1/J1-049H) is logic 1.

If the performance number counted in the next interval is latched in its PMON register without the previous one being read, the PMON Error Count register is over-written. Any over-writing of the four kinds of PMON Error Count registers will be indicated in the OVR (b0, T1/J1-049H).

#### 3.4 ALARM DETECTOR (ALMD) - T1 / J1 ONLY

The Alarm Detector block exists in T1/J1 mode only. It detects the Yellow signal and the AlS (Blue Alarm) signal in SF/ESF in T1/J1 data stream and declares the Yellow alarm, the Red alarm and the AlS alarm. The T1 or J1 mode is selected by the J1\_YEL (b5, T1/J1-02CH) while the SF or ESF format is selected by the ESF (b4, T1/J1-02CH).

The Yellow signal is declared differently in each format:

- In T1 SF format: The Yellow signal occupies the 2nd bit of each channel. When the occurence of logic 1 in this bit position is less than 16 times (including 16 times) during a 40ms period, the Yellow signal is declared.
- In J1 SF format: The Yellow signal occupies the F-bit of the 12th frame. When the occurence of logic 0 on this bit position is less than 2 times (including 2 times) during a 40ms period, the Yellow signal is declared.
- In T1/J1 ESF format: The Yellow signal occupies the DL of the F-bit (refer to Table-4). The pattern is 'FF00' in T1 mode and 'FFFF' in J1 mode. When the AVC (b1, T1/J1-02AH) is logic 0, the Yellow signal is declared if 8 out of 10 successive patterns match the 'FF00' (in T1) or 'FFFF' (in J1) on the DL bit position. When the AVC (b1, T1/J1-02AH) is logic 1, the Yellow signal is declared if 4 out of 5 successive patterns match the 'FF00' (in T1) or 'FFFF' (in J1) on the DL bit position

Any of the above conditions will be indicated by the YELD (b1, T1/J1-02FH).

The AIS signal is declared when the received data are out of SF/ESF synchronization for 60ms and the received logic 0 is less than 127 times in the same period. Then the AIS signal will be indicated by the AISD (b0, T1/J1-02FH).

The Red signal is declared when one or more out of SF/ESF sync event occurs in 40ms. Then the Red signal will be indicated by the REDD (b2, T1/J1-02FH).

The Yellow alarm, AIS alarm and Red alarm will be declared or cleared when the corresponding alarm signal is present or absent for a certain period as summarized in Table - 7.

The Yellow alarm, AIS alarm and Red alarm are also the interrupt sources in the ALMD block. When the alarm occurs, the corresponding Interrupt Status Register (YEL, AIS or RED in b2, b0, b1 of T1/J1-02EH respectively) will indicate it. When there is any transition (from 0 to 1 or from 1 to 0) on the Interrupt Status Register, its corresponding Interrupt Indication Register (YELI, AISI or REDI in b5, b3, b4 of T1/J1-02EH respectively) will be logic 1. A transition on the Interrupt Indication Register can cause an interrupt on the  $\overline{\text{INT}}$  pin if the corresponding Interrupt Enabled Register (YELE, AISE or REDE in b2, b0, b1 of T1/J1-02DH respectively) is logic 1.

#### 3.5 HDLC RECEIVER (RHDLC)

The HDLC extraction is performed in this block. The HDLC Receiver #1, #2 and #3 in E1 mode or the HDLC Receiver #1 and #2 in T1/J1 ESF mode of each framer operate independently.

#### 3.5.1 E1 MODE

Three HDLC Receiver blocks (#1, #2 and #3) are employed for each framer to extract the HDLC link from the received data stream. Before selecting the HDLC link, the TXCISEL (b3, E1-00AH) should be set to 0. Thus, the configuration of the Link Control and Bits Select registers (addressed from 028H to 02DH) is for RHDLC. Next, select one of the three HDLC Receiver blocks by selecting the appropriate bits in the RHDLCSEL[1:0] (b7~6, E1-00AH). The #2 and #3 blocks can also be disabled by setting the V52DIS (b3, E1-007H). Then the position of the HDLC link can be defined as follows:

- 1. Set the DL\_EVEN (b7, E1-028H or b7, E1-02AH or b7, E1-02CH) and/or the DL\_ODD (b6, E1-028H or b6, E1-02AH or b6, E1-02CH) to select the even and/or odd frames (the even frames are FAS frames while the odd frames are NFAS frames):
- 2. Set the DL\_TS[4:0] (b4~0, E1-028H or b4~0, E1-02AH or b4~0, E1-02CH) to define the timeslot of the assigned frame (or to select the TS16\_EN (b5, E1-028H) to select the TS16 of the assigned frame);
- 3. Set the DL\_BIT[7:0] (b7~0, E1-029H or b7~0, E1-02BH or b7~0, E1-02DH) to select the bits of the assigned timeslot.

Three HDLC standards for E1 are defined and one is selected as follows:

- Common Channel Signaling (CCS) data link (extract the bits in the TS16):
  - 2. V5.1 / V5.2 D-channel and C-channels (extract the bits in any

timeslot except TS16);

3. Sa-bit data link.

All the functions of the selected HDLC Receiver block can be enabled only if the EN (b0, E1-048H) is set to logic 1.

A normal HDLC packet consists of the following parts as shown in Figure - 3:

Flag (7E)	HDLC Data	FCS	Flag (7E)
one byte 01111110	n bytes n ≥ 2	two bytes	one byte 01111110
(re	move the stuffed zero)		•

Figure - 3. HDLC Packet

Every HDLC packet starts with a 7E (Hex) opening flag sequence and ends with the same flag. Before the closing flag sequence, two bytes of CRC-CCITT frame check sequences (FCS) are provided to check all the HDLC data. The received FCS will be compared with the local calculated FCS.

A FIFO buffer is used to store the HDLC packet, that is, to store the data whose stuffed zeros have been removed and the FCS. However, when the address matching is enabled, the first and/or second byte compares with the address setting in the PA[7:0] (b7~0, E1-04CH) and the SA[7:0] (b7~0, E1-04DH), and only the data matching the selection in the MEN (b3, E1-048H) and the MM (b2, E1-048H) are stored into the FIFO. When address matching is disabled, all the HDLC data are stored. The first 7E opening flag which activates the HDLC link and the 7F (Hex) abort sequence which deactivates the HDLC link will also be converted to dummy bytes and stored into the FIFO regardless if the address is matching or not. These two types of flags will also assert the COLS (b5, E1-04AH) to indicate the HDLC link status change. The content in the FIFO is read in the RD[7:0] (b7~0, E1-04BH), and the status of the bytes will be reflected in the PBS[2:0] (b3~1, E1-04AH). Both of these registers (RD[7:0] (b7~0, E1-04BH) and PBS[2:0] (b3~1, E1-04AH)) can not be accessed at a rate greater than 1/15 of the XCK rate.

The depth of the FIFO is 128 bytes. When the FIFO is empty, the FE (b7, E1-04AH) will be set. If data are still written into the FIFO when the FIFO is already full, the FIFO will be over-written. The over-written condition will be indicated by the OVR (b6, E1-04AH) and forces the FIFO to be cleared.

A logic 1 in the PKIN (b4, E1-04AH) indicates a non-abort HDLC

Table - 7. Alarm Summary in ALMD

	Declaring	Clearing		
Yellow Alarm	the Yellow signal is present for 425ms (± 50ms)	the Yellow signal is absent for 425ms (± 50ms)		
AIS Alarm	the AIS signal is present for 1.5sec (± 100ms)	the AIS signal is absent for 16.8sec (± 500ms); or the AIS signal is absent for 180ms if the FASTD (b4, T1/J1-02DH) is set		
Red Alarm	the Red signal is present for 2.55sec (± 40ms)	the Red signal is absent for 16.6sec (± 500ms); or the Red signal is absent for 120ms if the FASTD (b4, T1/J1-02DH) is set		

packet was received. This bit (PKIN) is set regardless of the status of the FCS condition or if there are an integer or non-integer number of bytes stored in the FIFO.

The HDLC packet can be forced to terminate for four reasons:

- 1. The 7F abort sequence is received;
- 2. More than 15 successive logic ones are received in the data stream:
  - 3. Set the TR (b1, E1-048H) to logic 1;
- 4. Set the EN (b0, E1-048H) from logic 1 to logic 0 and back to logic

All the above methods can deactivate the HDLC link immediately and the latter two means can also clean the FIFO and interrupts. A new search for the 7E opening flag is also initiated.

The interrupt sources in this block are:

- 1. Receiving the first 7E opening flag sequence which terminates all ones data and activates the HDLC link;
  - 2. Receiving the 7E closing flag sequence;
  - 3. Receiving the abort sequence;
- 4. Exceeding the set point of the FIFO which is defined in the INTC[6:0] (b6~0, E1-049);
  - 5. Over-writting the FIFO.

Any one of the interrupt sources will assert the INTR (b0, E1-04AH) high. Then the  $\overline{\text{INT}}$  pin will be low to report the interrupt if the INTE (b7, E1-049H) is logic 1.

#### 3.5.2 T1 / J1 MODE

In the SF format, there is no HDLC link.

In the ESF format, two HDLC Receiver blocks (#1 and #2) are employed for each framer to extract the HDLC link. Before selecting the HDLC link, the TXCISEL (b3, T1/J1-00DH) should be set to 0. Thus, the configuration of the Link Control and Bits Select registers (addressed from 070H to 071H) is for the RHDLC. Then, selected by the RHDLCSEL[1:0] (b7~6, T1/J1-00DH), one of the two HDLC Receiver blocks are accessable to the microprocessor. The HDLC#1 extracts the HDLC link in the DL of the F-bit (its position is shown in Table - 4). The HDLC#2 extracts the HDLC link from one of the channels which position is selected as follows:

- 1. Set the DL2\_EVEN (b7, T1/J1-070H) and/or the DL2\_ODD (b6, T1/J1-070H) to select the even and/or odd frames;
- 2. Set the DL2\_TS[4:0] (b4~0, T1/J1-070H) to select the channel of the assigned frame;
- 3. Set the DL2\_BIT[7:0] (b7~0, T1/J1-071H) to select the bits of the assigned channel.

All the functions of the selected HDLC Receiver block can be enabled only if the EN (b0, T1/J1-054H) is set to logic 1.

The structure of the HDLC packet is the same as it is described in the E1 mode (refer to Figure - 3).

A FIFO buffer is used to store the HDLC packet, that is, to store the data whose stuffed zeros have been removed and the FCS. However, when the address matching is enabled, the first and/or second byte compares with the address setting in the PA[7:0] (b7~0, T1/J1-058H) and the SA[7:0] (b7~0, T1/J1-059H) and only the data matching the selection in the MEN (b3, T1/J1-054H) and the MM (b2, T1/J1-054H) are stored into the FIFO. When the address matching is disabled, the entire HDLC packet is stored. The first 7E opening flag which activates the HDLC link and the 7F abort sequence which deactivates the HDLC link will also be converted into dummy bytes and stored in the FIFO. These

two types of flags will also assert the COLS (b5, T1/J1-056H) to indicate the HDLC link status change. The content in the FIFO is read in the RD[7:0] (b7 $\sim$ 0, T1/J1-057H), and the status of the bytes will be reflected in the PBS[2:0] (b3 $\sim$ 1, T1/J1-056H). Both of the two registers can't be accessed at a rate greater than 1/15 of the XCK rate.

The depth of the FIFO is 128 bytes. When the FIFO is empty, the FE (b7, T1/J1-056H) will be set. If data are still written into the FIFO when the FIFO is already full, the FIFO will be over-written. The over-written condition will be indicated by the OVR (b6, T1/J1-056H) and force the FIFO to be cleared.

A logic 1 in the PKIN (b4, T1/J1-056H) indicates a non-abort HDLC packet was received whether there were FCS errors or non-integer number if bytes errors in it or not.

The HDLC packet can be forced to terminate by four means:

- 1. The 7F abort sequence is received;
- 2. More than 15 successive logic ones are received in the data stream:
  - 3. Set the TR (b2, T1/J1-054H) to logic 1;
- 4. Set the EN (b1, T1/J1-054H) from logic 1 to logic 0 and back to logic 1.

All the above methods can deactivate the HDLC link immediately and the latter two methods can also clear the FIFO and interrupts. A new search for the 7E opening flag is also initiated.

The interrupt sources in this block are:

- 1. Receiving the first 7E opening flag sequence which terminates the all ones data and activates the HDLC link;
  - 2. Receiving the 7E closing flag sequence;
  - 3. Receiving the abort sequence;
- 4. Exceeding the set point of the FIFO which is defined in the INTC[6:0] (b6~0, T1/J1-055H);
  - 5. Over-writting the FIFO.

Any one of the interrupt sources will assert the INTR (b0, T1/J1-056H) high. Then the  $\overline{\text{INT}}$  pin will be driven low to report the interrupt if the INTE (b7, T1/J1-055H) is logic 1.

# 3.6 BIT-ORIENTED MESSAGE RECEIVER (RBOM) - T1 / J1 ONLY

The Bit Oriented Message (BOM) can only be received in the ESF format in T1/J1 mode. The standard of the BOM is defined in ANSI T1.403 and in TR-TSY-000194. This block of each framer operates independently.

The BOM pattern is '111111110XXXXXXX0' which occupies the DL of the F-bit in the ESF format (refer to Table-4). The six 'X's represent the message. The BOM is declared only when the pattern is matched and the received message is identical 4 out of 5 times or 8 out of 10 times. The identification time is selected by the AVC (b1, T1/J1-02AH). After the BOM is declared, the BOM is loaded into the BOC[5:0] (b5~0, T1/J1-02BH). However, the BOM does not include all ones code in both T1 and J1 mode.

When the BOM is converted into non-BOM, the received data will be idle code. The pattern of the idle code is 'FFFF' in T1 mode and 'FF7E' in J1 mode. When the received data is 4 out of 5 times or 8 out of 10 times identical with the pattern, the idle code is declared. The identification time is selected by the AVC (b1, T1/J1-02AH).

There are two interrupt sources in this block. When the BOM is declared, the BOCI (b6, T1/J1-02BH) will indicate it. When the idle code is declared, the IDLEI (b7, T1/J1-02BH) will indicate it. If the BOCE (b0,

T1/J1-02AH) and IDLE (b2, T1/J1-02AH) are set to 1 respectively, the corresponding condition will cause an interrupt on the INT pin.

# 3.7 INBAND LOOPBACK CODE DETECTOR (IBCD) - T1 / J1 ONLY

The Inband Loopback Code Detector can track loopback activate/deactivate codes only in framed or unframed T1/J1 data stream. The Inband Loopback Code Detector of each framer operates independently.

The received data stream is compared with the target activate/deactivate code whose length and the content are programmed in the ASEL[1:0] (b1~0, T1/J1-03CH) / DSEL[1:0] (b3~2, T1/J1-03CH) and the ACT[7:0] (b7~0, T1/J1-03EH) / DACT[7:0] (b7~0, T1/J1-03FH) respectively. In framed mode, the F-bit can be selected by the IBCD\_IDLE (b5, T1/J1-000H) to compare with the target activate/deactivate code or not. In unframed mode, all 193 bits are compared with the target activate/deactivate code. Whether the received data stream matches the target activate or deactivate code and repeats for a 39.8ms period, the LBACP (b7, T1/J1-03DH) or LBDCP (b6, T1/J1-03DH) will indicate the appearance of the corresponding code. 2, 20 or 200 bit-error tolerance within each 39.8ms period is permitted by setting the IBCD\_ERR[1:0] (b5~4, T1/J1-03CH). However, even if the F-bit is compared, whether it is matched or not, the result will not cause bit errors, that is, the comparison result of the F-bit is passed over.

When the received data stream matches the target activate/deactivate code and repeats for 5.1 sec, the LBA (b1, T1/J1-03DH) / LBD (b0, T1/J1-03DH) will indicate the detection of the inband loopback code. The code sequences detection and timing is compatible with the specifications in T1.403, TA-TSY-000312 and TR-TSY-000303.

The status changes of the activate or deactivate code are the interrupt sources in the IBCD block. That is, when the value in the Status Register (LBA [b1, T1/J1-03DH] or LBD [b0, T1/J1-03DH]) changes, its corresponding Interrupt Indication Register (LBAI [b3, T1/J1-03DH]) or LBDI [b2, T1/J1-03DH]) will be logic 1. A logic 1 on the Interrupt Indication Register will cause an interrupt on the INT pin if the corresponding Interrupt Enable Register (LBAE [b5, T1/J1-03DH]) or LBDE [b4, T1/J1-03DH]) is logic 1.

#### 3.8 ELASTIC STORE BUFFER (ELSB)

The Elastic Store Buffer of each framer operates independently.

#### 3.8.1 E1 MODE

In Receive Clock Slave mode, a 2-basic-frame depth Elastic Store Buffer is used to synchronize the incoming frames to the Receive Side System Common Clock derived from the RSCCK pin, and to the Receive Side System Common Frame Pulse derived from the RSCFS pin. A write pointer is used to write the data to the Elastic Store Buffer, while a read pointer is used to read the data from the Elastic Store Buffer.

When the average frequency of the incoming data is greater than the average frequency of the Receive Side System Common Clock (RSCCK), the write pointer will be faster than the read pointer and the Elastic Store Buffer will be filled. So a frame will be deleted after its prior frame is read. When the read pointer crosses the frame boundary, a controlled slip will occur with a logic 1 indicated in the SLIPD (b1, E1-059H).

When the average frequency of the incoming data is less than the average frequency of the RSCCK, the write pointer will be slower than the read pointer and the Elastic Store Buffer will be empty. The frame

will be repeated after it is read. When the read pointer crosses the next frame boundary, a controlled slip will occur with a logic 0 indicated in the SLIPD (b1, E1-059H).

When the slip occurs, the SLIPI (b0, E1-059H) will indicate. An interrupt on the  $\overline{\text{INT}}$  pin will also occur if the SLIPE (b2, E1-059H) is set to logic 1.

In Receive Clock Slave mode, if it is out of Basic Frame synchronization, the idle code programmed in the D[7:0] (b7~0, E1-05AH) in the Elastic Store Buffer can be set to replace the data if the TRKEN (b1, E1-001H) is set to logic 1.

In Receive Clock Master mode, the Elastic Store Buffer is bypassed unless the device is in the Payload Loopback diagnosis mode. (Refer to Payload Loopback Mode for details).

#### 3.8.2 T1 / J1 MODE

In Receive Clock Slave mode, a 2-basic-frame depth Elastic Store Buffer is used to synchronize the incoming frames to the Receive Side System Common Clock derived from the RSCCK pin, and to the Receive Side System Common Frame Pulse derived from the RSCFS pin. A write pointer is used to write the data to the Elastic Store Buffer, while a read pointer is used to read the data from the Elastic Store Buffer.

When the average frequency of the incoming data is greater than the average frequency of the Receive Side System Common Clock (RSCCK), the write pointer will be faster than the read pointer and the Elastic Store Buffer will be filled. So a frame will be deleted after its prior frame is read. When the read pointer crosses the frame boundary, a controlled slip will occur with a logic 1 indicated in the SLIPD (b1, T1/J1-01DH).

When the average frequency of the incoming data is less than the average frequency of the RSCCK, the write pointer will be slower than the read pointer and the Elastic Store Buffer will be empty. The frame will be repeated after it is read. When the read pointer crosses the next frame boundary, a controlled slip will occur with a logic 0 indicated in the SLIPD (b1, T1/J1-01DH).

When the slip occurs, the SLIPI (b0, T1/J1-01DH) will indicate. An interrupt on the  $\overline{\text{INT}}$  pin will also occur if the SLIPE (b2, T1/J1-01DH) is logic 1.

In Receive Clock Slave mode, if it is out of SF/ESF sync, the idle code programmed in the D[7:0] (b7~0, T1/J1-01EH) in the Elastic Store Buffer will replace the data of all channels automatically.

In Receive Clock Master mode, the Elastic Store Buffer is bypassed unless the device is in the payload loopback diagnosis mode. (Refer to Payload Loopback Mode for details).

#### 3.9 RECEIVE CAS/RBS BUFFER (RCRB)

The Receive CAS/RBS Buffer of each framer operates independently.

#### 3.9.1 E1 MODE

In the Signaling Multi-Frame synchronization condition, the signaling bits are located in TS16, which is Channel Associated Signaling (CAS). Their arrangement in TS16 is shown in Figure - 4.

When the RSCKn/RSSIGn/MRSSIG[1:2] pins are used as the signaling output, i.e. in Receive Clock Slave External Signaling mode or in Receive Multiplex mode, the signaling codeword (ABCD) are clocked out in the lower nibble of the timeslot with its corresponding data serializing on the RSDn/MRSD[1:2] pins (as shown in the Figure - 5).

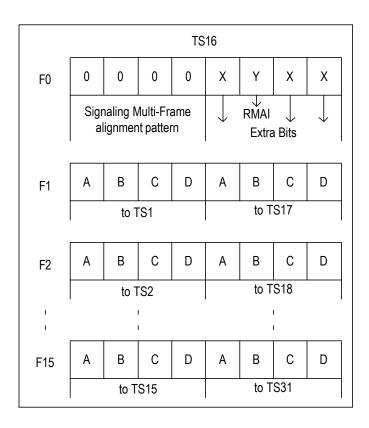


Figure - 4. TS16 Arrangement in Signaling Multi-Frame

When the COSS (b6, E1-064H) is logic 1, all the COSS[30:1] (b5~0, E1-064H and b7~0, E1-065H and b7~0, E1-066H and b7~0, E1-067H) in the Receive CAS/RBS Buffer registers will reflect the change of the signaling of each timeslot respectively (excluding the TS0 and TS16).

When the COSS (b6, E1-064H) is logic 0, the Receive CAS/RBS Buffer indirect registers (from 01H to 5FH of RCRB indirect registers) can be accessed by the microprocessor. The address of the indirect register is specified by the A[6:0] (b6~0, E1-066H). Whether the data are read from or written into the specified indirect register is determined by the R/WB (b7, E1-066H) and the data are in the D[7:0] (b7~0, E1-067H). The indirect registers have a read/write cycle. Before the read/write operation is completed, the BUSY (b7, E1-065H) will be set. New operations on the indirect registers can only be implemented when the BUSY (b7, E1-065H) is cleared. The read/write cycle is 490ns.

The indirect registers are divided into three segments: two segments (from 01H to 1FH & from 21H to 3FH) contain the signaling bits of each timeslot; another segment (from 40H to 5FH) contains the signaling debounce configuration of each timeslot.

A three-Signaling-Multi-Frame capacity buffer is used for signaling debounce and signaling freezing.

Signaling debounce can be set by the DEB (b0, E1-RCRB-indirect registers - 41~5FH) which is activated by the PCCE (b0, E1-064H). It updates the signaling bits only when 2 consecutive signaling of a timeslot are the same.

Signaling freeze will remain the signaling automatically when it is out of Signaling Multi-Frame synchronization or in unframed mode.

The signaling bits are extracted to the A, B, C, D (b3~0, E1-RCRB-indirect registers - 01~1FH or b3~0, E1-RCRB-indirect registers - 21~3FH).

There is a maximum 2 ms delay between the transition of the COSS[n] (E1-064H and E1-065H and E1-066H and E1-067H) and the

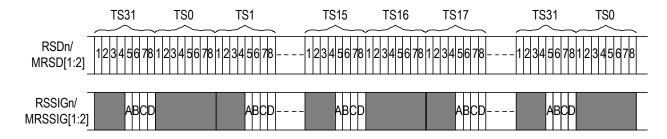


Figure - 5. Signaling Output in E1 Mode

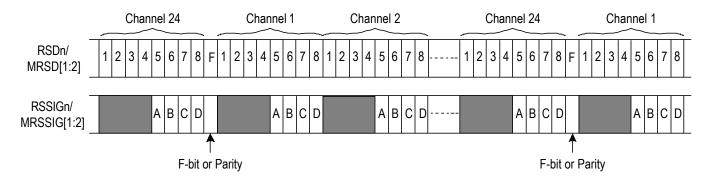


Figure - 6. Signaling Output in T1 / J1 Mode

updating of the A, B, C, D code in the corresponding indirect registers (b3~0, E1-RCRB-indirect registers - 21~3FH). To avoid this 2 ms delay, users can read the corresponding b3~0 in the E1-RCRB-indirect registers - (01~1FH) first. If the value of these four bits are different from the previous A, B, C, D code, then the content of b3~0 in the E1-RCRB-indirect registers - (01~1FH) is the updated A, B, C, D code. If the content of the four bits is the same as the previous A, B, C, D code, then users should read the b3~0 in the E1-RCRB-indirect registers - (21~3FH) to get the updated A, B, C, D code.

Any one of the 30-timeslot's signaling change will cause an interrupt on the  $\overline{\text{INT}}$  pin if the SIGE (b5, E1-064H) is set.

#### 3.9.2 T1 / J1 MODE

When the frame is synchronized, the signaling is located in the Bit 8 of Frame 6 (A bit) and Frame 12 (B bit) in SF format, and is located in the Bit 8 of Frame 6 (A bit), Frame 12 (B bit), Frame 18 (C bit) and Frame 24 (D bit) in ESF format (refer to Table-3 & 4). The SF/ESF signaling format is selected by the ESF (b2, T1/J1-040H).

When the RSCKn/RSSIGn/MRSSIG[1:2] pins are used as the signaling output, i.e. in Receive Clock Slave External Signaling mode or in Receive Multiplex mode, the signaling codeword (AB or ABCD) are clocked out in the lower nibble of the channel with its corresponding data serializing on the RSDn/MRSD[1:2] pins (as shown in the Figure - 6). However, in SF format, the signaling C and D are the repitition of signaling A and B.

When the COSS (b6, T1/J1-040H) bit is logic 1, all the COSS[24:1] (b7~0, T1/J1-041H and b7~0, T1/J1-042H and b7~0, T1/J1-043H) in the Receive CAS/RBS Buffer registers will reflect the change of the signaling of each channel respectively.

When the COSS (b6, T1/J1-040H) bit is logic 0, the Receive CAS/RBS Buffer indirect registers (from 01H to 58H of RCRB indirect registers) can be accessed by the microprocessor. The address of the indirect register is specified by the A[6:0] (b6~0, T1/J1-042H). Whether the data are read from or written into the specified indirect register is determined by the R/WB (b7, T1/J1-042H) and the data is in the D[7:0] (b7~0, T1/J1-043H). The indirect registers have a read/write cycle. Before the read/write operation is completed, the BUSY (b7, T1/J1-041H) will be set. New operations on the indirect registers can only be done when the BUSY (b7, T1/J1-041H) is cleared. The read/write cycle is 650ns.

The indirect registers are devided into three segments: two segments (from 01H to 18H & from 21H to 38H) contain the signaling bits of each channel; another segment (from 41H to 58H) contains the signaling debounce configuration of each channel.

A three-superframe capacity buffer is used for signaling debounce and signaling freezing.

Signaling debounce can be set by the DEB (b0, T1/J1-RCRB-indirect registers - 41~58H) which is activated by the PCCE (b0, T1/J1-040H). It updates the signaling bits only when 2 consecutive SF/ESF signaling bits of a channel are the same.

Signaling freeze will remain the signaling automatically when it is out of SF/ESF sync or in unframed mode.

The signaling bits are extracted to the A, B, C, D (b3~0, T1/J1-RCRB-indirect registers - 01~18H or b3~0, T1/J1-RCRB-indirect registers - 21~38H).

There is a maximum 2 ms delay between the transition of the COSS[n] (T1/J1-041H and T1/J1-042H and T1/J1-043H) and the updating of the A, B, C, D code in the corresponding indirect registers (b3~0,

T1/J1-RCRB-indirect registers -  $21\sim38H$ ). To avoid this 2 ms delay, users can read the corresponding  $b3\sim0$  in the T1/J1-RCRB-indirect registers - (01 $\sim18H$ ) first. If the value of these four bits are different from the previous A, B, C, D code, then the content of  $b3\sim0$  in the T1/J1-RCRB-indirect registers - (01 $\sim18H$ ) is the updated A, B, C, D code. If the content of the four bits is the same as the previous A, B, C, D code, then users should read the  $b3\sim0$  in the T1/J1-RCRB-indirect registers - (21 $\sim38H$ ) to get the updated A, B, C, D code.

Any one of the 24-channel's signaling change will cause an interrupt on the  $\overline{\text{INT}}$  pin if the SIGE (b5, T1/J1-040H) is set.

#### 3.10 RECEIVE PAYLOAD CONTROL (RPLC)

Different test patterns can be inserted in the received data stream or the received data stream can be extracted to the PRBS Generator/Detector for test in this block. The Receive Payload Control of each framer operates independently.

#### 3.10.1 E1 MODE

To enable the test for the received data stream, the PCCE (b0, E1-05CH) must be set to activate the setting in the indirect registers (from 20H to 7FH of RPLC indirect registers). The following methods can be executed for test on a per-TS basis:

- Selected by the PRGDSEL[2:0] (b7~5, E1-00CH), the received data of one of the eight framers will be extracted to the PRBS Generator/Detector when the RXPATGEN (b2, E1-00CH) is 0. The received data can be extracted in framed or unframed mode. The selection is made by the UNF\_DET (b0, E1-00CH). In unframed mode, all the 32 timeslots are extracted and the per-timeslot configuration in the TEST (b7, E1-RPLC-indirect registers 20~3FH) is ignored. In framed mode, the received data will only be extracted on the timeslot configured by the TEST (b7, E1-RPLC-indirect registers 20~3FH). Refer to the section of PRBS GENERATOR / DETECTOR (PRGD) for details.
- Replace the data that will be output on the RSDn/MRSD pin with the value in the DTRK[7:0] (b7~0, E1-RPLC-indirect registers 40~5FH) when the DTRKC/NxTS (b6, E1-RPLC-indirect registers 20~3FH) of the corresponding timeslot is logic 1. (When it is out of Basic Frame synchronization, the value in the DTRK[7:0] [b7~0, E1-RPLC-indirect registers 40~5FH] can replace the data automatically if the AUTOOOF [b1, E1-000H] is set. Or, when it is out of Basic Frame synchronization for 100ms, the value in the DTRK[7:0] [b7~0, E1-RPLC-indirect registers 40~5FH] can replace the data automatically if the AUTORED (b2, 000H) is set. These two kinds of data replacements can be executed even if the PCCE [b0, E1-05CH] is disabled and they replace all the timeslots.)
- Replace the data that will be output on the RSDn/MRSD pin with the  $\mu$ -law or A-law milliwatt pattern when the DMW (b4, E1-RPLC-indirect register 20~3FH) of the corresponding timeslot is logic 1. (The milliwatt pattern is selectable between A-law and  $\mu$ -law by the DMWALAW [b3, E1-RPLC-indirect register 20~3FH]. Refer to Table 8 & Table 9.)
- Selected by the PRGDSEL[2:0] (b7~5, E1-00CH), the test pattern from the PRBS Generator/Detector will replace the received data of one of the eight framers when the RXPATGEN (b2, E1-00CH) is 1. The test pattern can replace the received data in framed or unframed mode. The selection is made by the UNF\_GEN (b1, E1-00CH). In unframed mode, all 32 timeslots are replaced and the per-timeslot configuration in the TEST (b7, E1-RPLC-indirect registers 20~3FH) is ignored. In framed mode, the received data will only replace the timeslot configured by the TEST (b7, E1-RPLC-indirect registers 20~3FH). Refer to the section of

Table - 8. A-Law Digital Milliwatt Pattern

Bit 2 Bit 5 Bit 1 Bit 3 Bit 4 Bit 6 Bit 7 

Table - 9. u-Law Digital Milliwatt Pattern

Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7
0	0	0	1	1	1	1	0
0	0	0	0	1	0	1	1
0	0	0	0	1	0	1	1
0	0	0	1	1	1	1	0
1	0	0	1	1	1	1	0
1	0	0	0	1	0	1	1
1	0	0	0	1	0	1	1
1	0	0	1	1	1	1	0

PRBS GENERATOR / DETECTOR (PRGD) for details.

 Invert the most significant bit, the even bits and/or odd bits that will be output on the RSDn pin when the SIGNINV and the RINV[1:0] (b2~0, E1-RPLC-indirect registers - 20~3FH) of the corresponding timeslot is set

(The above methods are arranged from highest to lowest in priority.)

- Replace the signaling that will be output on the RSSIGn pin with the value in the A, B, C, D (b3~0, E1-RPLC-indirect registers - 61~7FH) when the STRKC (b5, E1-RPLC-indirect registers - 20~3FH) of the corresponding timeslot allows.

The data and signaling of all timeslots can be replaced with the setting in the DTRK[7:0] (b7~0, E1-RPLC-indirect registers - 40~5FH) and the A, B, C, D (b3~0, E1-RPLC-indirect registers - 61~7FH) respectively when the RXMTKC (b0, E1-001H) is set. To enable this function, PCCE (b0, E1-05CH) must be set to 1.

Addressed by the A[6:0] (b6~0, E1-05EH), the data read from or written into the indirect registers are in the D[7:0] (b7~0, E1-05FH). The read or write operation is determined by the R/WB (b7, E1-05EH). The indirect registers have a read/write cycle. Before the read/write operation is completed, the BUSY (b7, E1-05DH) will be set. New operations on the indirect registers can only be implemented when the BUSY (b7, E1-05DH) is cleared. The read/write cycle is 490ns.

#### 3.10.2 T1 / J1 MODE

To enable the test for the received data stream, the PCCE (b0, T1/J1-050H) must be set to activate the setting in the indirect registers (from 01H to 48H). The following methods can be used for test on a perchannel basis:

- Selected by the PRGDSEL[2:0] (b7~5, T1/J1-00FH), the received data of one of the eight framers will be extracted to the PRBS Generator/Detector when the RXPATGEN (b2, T1/J1-00FH) is 0. The received data can be extracted in framed or unframed mode. The selection is made by the UNF\_DET (b0, T1/J1-00FH). In unframed mode, all the 24 channels and the F-bit are extracted and the per-channel configuration in the TEST (b3, T1/J1-RPLC-indirect registers 01~18H) is ignored. In framed mode, the received data will only be extracted on the channel specified by the TEST (b3, T1/J1-RPLC-indirect registers 01~18H). Fractional T1/J1 signal can also be extracted in the specified channel when the Nx56k\_DET (b3, T1/J1-00FH) is set. Refer to the section of PRBS GENERATOR / DETECTOR (PRGD) for details.
- Replace the data that will be output on the RSDn/MRSD pin with the value in the DTRK[7:0] (b7~0, T1/J1-RPLC-indirect registers -

19~30H) when the DTRKC (b6, T1/J1-RPLC-indirect registers - 01~18H) of the corresponding channel is logic 1. (When it is out of SF/ESF synchronization, the value in the DTRK[7:0] (b7~0, T1/J1-RPLC-indirect registers - 19~30H) can replace the data automatically if the AUTOOOF (b1, T1/J1-000H) is set. Or, when the RED alarm is declared, the value in the DTRK[7:0] (b7~0, T1/J1-RPLC-indirect registers - 19~30H) can replace the data automatically if the AUTORED (b2, T1/J1-000H) is set. These two kinds of data replacements can be executed even if the PCCE (b0, T1/J1-050H) is disabled and they replace all the channels.)

- Replace the data that will be output on the RSDn/MRSD pin with the milliwatt pattern when the DMW (b5, T1/J1-RPLC-indirect register 01~18H) of the corresponding channel allows. (The milliwatt is  $\mu$ -law. Refer to Table 9.)
- Selected by the PRGDSEL[2:0] (b7~5, T1/J1-00FH), the test pattern from the PRBS Generator/Detector will replace the received data of one of the eight framers when the RXPATGEN (b2, T1/J1-00FH) is 1. The test pattern can replace the received data in framed or unframed mode. The selection is made by the UNF\_GEN (b1, T1/J1-00FH). In unframed mode, all the 24 channels and the F-bit are replaced and the per-channel configuration in the TEST (b3, T1/J1-RPLC-indirect registers 01~18H) is ignored. In framed mode, the received data will only be replaced on the channel specified by the TEST (b3, T1/J1-RPLC-indirect registers 01~18H). Fractional T1/J1 signal can also be replaced in the specified channel when the Nx56k\_GEN (b4, T1/J1-00FH) is set. Refer to the section of PRBS GENERATOR / DETECTOR (PRGD) for details.
- Invert the most significant bit and/or the other bits in a channel that will be output on the RSDn/MRSD pin when the SIGNINV and the IN-VERT (b4 & b7, T1/J1-RPLC-indirect registers 01~18H) of the corresponding channel are set.
- Fix the signaling bit with the value in the POL (b0, T1/J1-RPLC-indirect registers 01~18H) when the FIX (b1, T1/J1-RPLC-indirect registers 01~18H) of the corresponding channel is logic 1.

(The above methods are arranged from highest to lowest in priority.)

- Replace the signaling that will be output on the RSSIGn/MRSSIG pin with the value in the A, B, C, D (b3~0, T1/J1-RPLC-indirect registers - 31~48H) when the STRKC (b7, T1/J1-RPLC-indirect registers - 31~48H) of the corresponding channel is logic 1.

The data and signaling of all channels can be replaced with the setting in the DTRK[7:0] (b7~0, T1/J1-RPLC-indirect registers - 19~30H) and the A, B, C, D (b3~0, T1/J1-RPLC-indirect registers - 31~48H) respectively when the IMTKC (b0, T1/J1-001H) is set. To enable this function, the PCCE (b0, T1/J1-050H) must be set to 1.

Addressed by the A[6:0] (b6~0, T1/J1-052H), the data read from or written into the indirect registers are in the D[7:0] (b7~0, T1/J1-053H). The read or write operation is determined by the R/WB (b7, T1/J1-052H). Before the read/write operation is completed, the BUSY (b7, T1/J1-051H) will be set. New operations on the indirect registers can only be implemented when the BUSY (b7, T1/J1-051H) is cleared. The read/write cycle is 650ns.

#### 3.11 RECEIVE SYSTEM INTERFACE (RESI)

The Receive System Interface determines how to output the received data to the system back-plane. The data from the eight framers can be aligned with each other or be output independently. The timing clocks and framing pulses can be provided by the system back-plane common to eight framers, or obtained from the far end of the individual eight framers. The Receive System Interface supports various configurations to meet various requirements in different applications.

#### 3.11.1 E1 MODE

In E1 mode, the Receive System Interface can be set in Non-multiplexed Mode or Multiplexed Mode. In Non-multiplexed Mode, the RSDn pin is used to output the received data from each framer at the bit rate of 2.048 Mb/s. While in the Multiplexed Mode, the received data from the eight framers are byte interleaved to form two high speed data streams and output on the MRSD1 and MRSD2 pins at the bit rate of 8.192 Mb/s.

In the Non-multiplexed Mode, if the timing signal for clocking data on RSDn pin is provided by the system side and shared by all eight framers, the Receive System Interface should be set in Receive Clock Slave mode. If the timing signal for clocking data on each RSDn pin is received from each line side, the Receive System Interface should be set in Receive Clock Master mode.

In the Receive Clock Slave mode, if the multi-function pin RSCKn/RSSIGn is used to output a reference clock, the Receive System Interface is in Receive Clock Slave RSCK Reference Mode. If the RSCKn/RSSIGn pin is used to output the extracted signaling bits, the Receive System Interface is in Receive Clock Slave External Signaling mode.

In the Receive Clock Master mode, if the data in all 32 timeslots in an E1 basic frame is clocked out by the RSCKn, the Receive System Interface is in Receive Clock Master Full E1 mode. If the data in only some of the timeslots in an E1 frame are clocked out by the RSCKn, then the Receive System Interface is in Receive Clock Master Fractional E1 (with F-bit) Mode.

Table - 10 summarizes the receive system interface in different operation modes. To set the receive system interface of each framer into various operation modes, the registers must be configured as Table - 11.

Table - 10. E1 Mode Receive System Interface in Different Operation Modes

Operation Mode		Data Pin	Clock Pin	Framing Pin	Signaling Pin	Reference Clock	
Non-	Clock Slave	RSCK Reference	RSDn	RSCCK	RSCFS & RSFSn *	No	RSCKn
Multiplexed	Mode	External Signaling	RSDn	RSCCK	RSCFS & RSFSn *	RSSIGn	No
Mode	Clock Master	Full E1	RSDn	RSCKn	RSFSn	No	No
	Mode	Fractional E1 (with F-bit)	RSDn	RSCKn	RSFSn	No	No
Multiplexed Mode		MRSD	MRSCCK	MRSCFS & MRSFS *	MRSSIG	No	

#### Note:

Table - 11. Operation Mode Selection in E1 Receive Path

<b>RATE[1:0]</b> (b1~0, E1-010H)	<b>RSCKSLV</b> (b5, E1-010H)	<b>RSSIG_EN</b> (b6, E1-001H)	FRACTN[1:0] (b7~6, E1-010H)	Operation Mode
	1	0	-	Receive Clock Slave RSCK Reference
	l	1	-	Receive Clock Slave External Signaling
01	0	<u>-</u>	00	Receive Clock Master Full E1
			10	Receive Clock Master Fractional E1
			11	Receive Clock Master Fractional E1 with F-bit
11 (All the eight framers should be set)	1	1	-	Receive Multiplexed

<sup>\*</sup> In Receive Clock Slave mode and Receive Multiplexed mode, there are two framing signals. In Receive Clock Slave mode, the framing pulses on RSCFS can be ignored for some framers by setting the FPMODE (b5, E1-011H) to 0. However, in Receive Multiplexed mode, when the FPMODE (b5, E1-011H) of any of the eight framers is configured as logic 1, all the others are taken as logic 1. Only when all the FPMODE (b5, E1-011H) of the eight framers are configured as logic 0, the frame pulses on MRSCFS can be ignored. That is, the FPMODE (b5, E1-011H) should be configured to the same value in Receive Multiplexed mode.

#### 3.11.1.1 Receive Clock Slave Mode

In the Receive Clock Slave Mode, the Receive Side System Common Clock (RSCCK) is provided by the system side. It is used as a common timing clock for all eight framers. The speed of the RSCCK can be selected by the CMS (b2, E1-010H) to be the same as the received data (2.048MHz), or double of the received data (4.096 MHz). The CMS (b2, E1-010H) of the eight framers should be set to the same value. If the speed of the RSCCK is double that of the received data stream, there will be two active edges in one bit duration. In this case, the RSD\_RSCFS\_EDGE (b5, E1-014H) determines the active edge to update the signal on the RSDn, RSSIGn and RSFSn pins; however, the pulse on the RSCFS (if exists) is always samples on its first active edge.

In the Receive Clock Slave Mode, the Receive Side System Common Frame Pulse (RSCFS) is used as a common framing signal to align the data streams for all eight framers. The RSCFS asserts on each Basic Frame and its valid polarity is configured by the FPINV (b6, E1-011H). The framing signals on RSCFS can also be ignored by setting the FPMODE (b5, E1-011H) to 0.

In the Receive Clock Slave Mode, the bit rate on the RSDn pin is 2.048Mb/s.

In the Receive Clock Slave Mode, the Receive Side System Frame Pulse (RSFSn) can be configured by the PERTS\_RSFS (b3, E1-00EH) and REF\_MRSFS (b2, E1-00EH) to output all zeros, to indicate the frame position or to output the same pulse as the RSCFS. When it is defined to indicate the frame position, it can indicate the first bit of a Basic Frame, Signaling Multi-frame, CRC-Multiframe, or both the Signaling and CRC-multiframe. This selection is made by the ROHM, BRXSMFP, BRXCMFP, ALTIFP (b3, b2, b1, b0, E1-011H). When the RSFSn is for framing pulse indication, the valid polarity of it is configured by the FPINV (b6, E1-011H). In this case, if the FPMODE (b5, E1-011H) is low, the RSFSn can only indicate the Basic Frame no matter what the setting in the ROHM, BRXSMFP, BRXCMFP, ALTIFP (b3, b2, b1, b0, E1-011H).

The Receive Clock Slave Mode includes two sub-modes: Receive Clock Slave RSCK Reference Mode and Receive Clock Slave External Signaling Mode.

#### 3.11.1.1.1 Receive Clock Slave RSCK Reference Mode

In this mode (refer to Figure - 7), the data on the system interface is clocked by the RSCCK. The active edge of the RSCCK to sample the pulse on the RSCFS or to update the data on the RSDn and RSFSn pins is determined by the following bits in the registers (refer to Table - 12).

Table - 12. Active Edge Selection of RSCCK (in E1 Receive Clock Slave RSCK Reference Mode)

	Bit Determining the Active Edge of the RSCCK
RSCFS	FF (h2 F1 010H)
RSFSn	FE (b3, E1-010H)
RSDn	DE (b4, E1-010H)

**Note:** If the setting of the FE (b3, E1-010H) and DE (b4, E1-010H) is different, the RSFSn will be one clock edge ahead of RSDn.

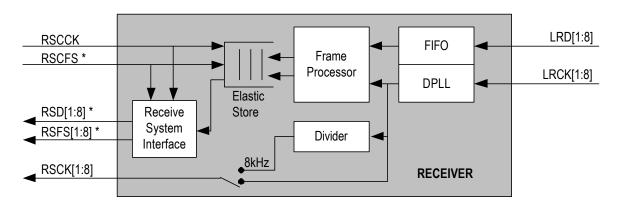
The FE (b3, E1-010H) of the eight framers should be set to the same value to ensure the RSCFS for the eight framers is sampled on the same active edge.

Note:

There is a special case when the CMS (b2, E1-010H) is logic 1 and the DE (b4, E1-010H) is equal to FE (b3, E1-010H). The RSD\_RSCFS\_EDGE (b5, E1-014H) is invalid and the signal on the RSDn and the RSFSn pins are updated on the first active edge of RSCCK.

Figure - 8 & 9 show the functional timing examples. Bit 1 of each timeslot is the first bit to be output.

Besides all the common functions described in the Receive Clock Slave mode, the special feature in this mode is that the multi-functional pin RSCKn/RSSIGn is used as RSCKn to output a reference clock. The RSCKn can be selected by the RSCKSEL (b5, E1-001H) to output a jitter attenuated 2.048MHz (i.e., smoothed LRCKn) or 8KHz clock (smoothed LRCKn divided by 256).

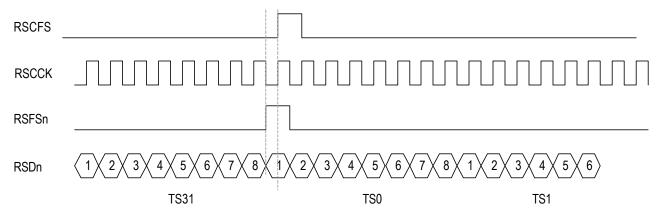


Note: \* RSCFS, RSD, RSFS are timed to RSCCK

Figure - 7. Receive Clock Slave RSCK Reference Mode

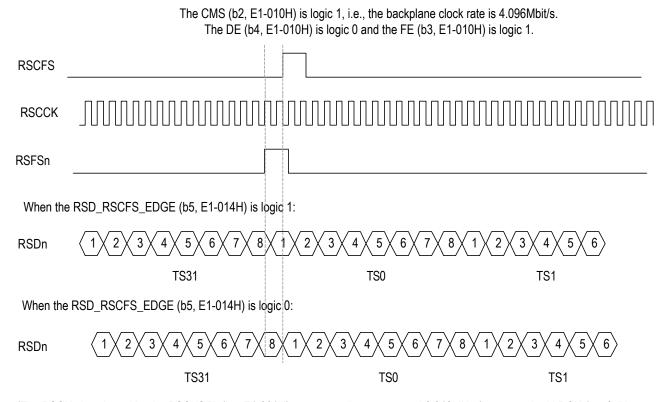
The CMS (b2, E1-010H) is logic 0, i.e., the backplane clock rate is 2.048Mbit/s.

The DE (b4, E1-010H) is logic 0 and the FE (b3, E1-010H) is logic 0. The timeslot offset and the bit offset enable are both 0:



(The RSCKn is selected by the RSCKSEL (b5, E1-001H) to output a jitter attenuated 2.048MHz (i.e., smoothed LRCKn) or 8KHz clock (smoothed LRCKn divided by 256).)

Figure - 8. E1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 1



(The RSCKn is selected by the RSCKSEL (b5, E1-001H) to output a jitter attenuated 2.048MHz (i.e., smoothed LRCKn) or 8KHz clock (smoothed LRCKn divided by 256).)

Figure - 9. E1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 2

# 3.11.1.1.2 Receive Clock Slave External Signaling Mode

In this mode (refer to Figure - 10), the data on the system interface is clocked by the RSCCK. The active edge of the RSCCK used to sample the pulse on the RSCFS or to update the data on the RSDn, RSFSn and RSSIGn is determined by the following bits in the registers (refer to Table - 13).

Figure - 11 & 12 show the functional timing examples. Bit 1 of each timeslot is the first bit to be output.

Besides all the common functions described in the Receive Clock Slave mode, the special feature in this mode is that the multi-functional pin RSCKn/RSSIGn is used as RSSIGn to output the extracted signaling bits. The extracted signaling bits are timeslot aligned with the data on the RSDn pin (refer to Figure - 5).

In the Out of Signaling Multi-Frame condition, the output signaling bits ABCD on the RSSIGn pin can be forced to be all ones if the OOSMFAIS (b2, E1-001H) is set to 1.

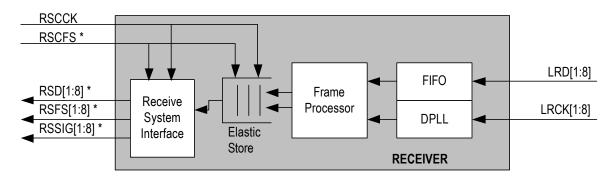
Table - 13. Active Edge Selection of RSCCK (in E1 Receive Clock Slave External Signaling Mode)

	Bit Determining the Active Edge of the RSCCK	
RSCFS	FF (h2 F1 010H)	
RSFSn	FE (b3, E1-010H)	
RSDn	DE /b4 E4 040U\	
RSSIGn	DE (b4, E1-010H)	

**Note:** If the setting of the FE (b3, E1-010H) and DE (b4, E1-010H) is different, the RSFSn will be one clock edge ahead of RSDn.

The FE (b3, E1-010H) of the eight framers should be set to the same value to ensure the RSCFS for the eight framers is sampled on the same active edge.

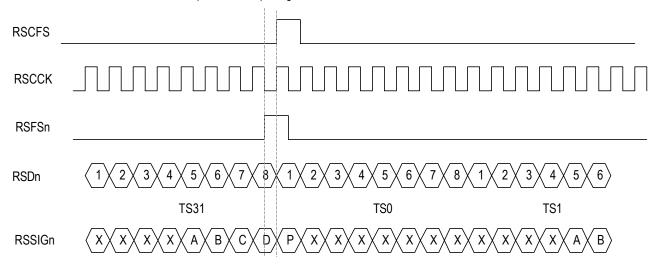
There is a special case when the CMS (b2, E1-010H) is logic 1 and the DE (b4, E1-010H) is equal to FE (b3, E1-010H). The RSD\_RSCFS\_EDGE (b5, E1-014H) is invalid and the signal on the RSDn, RSSIGn and RSFSn pins are updated on the first active edge of RSCCK.



Note: \* RSCFS, RSD, RSSIG, RSFS are timed to RSCCK

Figure - 10. Receive Clock Slave External Signaling Mode

The CMS (b2, E1-010H) is logic 0, i.e., the bankplane rate is 2.048Mbit/s. The DE (b4, E1-010H) is logic 1 and the FE (b3, E1-010H) is logic 0. The timeslot offset and the bit offset enable are both 0:

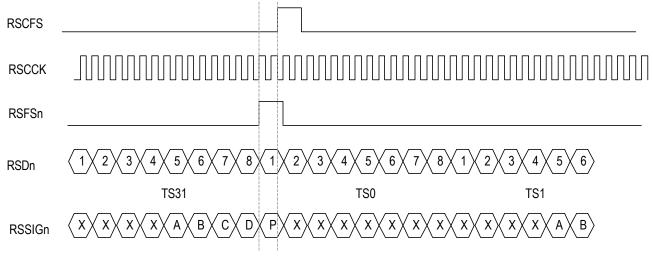


(The 'X' represents the filled bits and has no meaning. The 'P' represents the Parity bit..)

Figure - 11. E1 Receive Clock Slave External Signaling Mode - Functional Timing Example 1

The CMS (b2, E1-010H) is logic 1, i.e., the bankplane rate is 4.096Mbit/s.

The DE (b4, E1-010H) is logic 1 and the FE (b3, E1-010H) is logic 1. \* The timeslot offset and the bit offset enable are both 0:



(The 'X' represent the filled bits and has no meaning. The 'P' represents the Parity bit..)

#### Note:

Figure - 12. E1 Receive Clock Slave External Signaling Mode - Functional Timing Example 2

#### 3.11.1.2 Receive Clock Master Mode

In the Receive Clock Master mode, each framer uses its own clock signal on RSCKn pin and framing signal on RSFSn pin to output the data on each RSDn pin. As the common framing signal RSCFS is not used, the FPMODE bit (b5, E1-011H) must be set to 0.

In the Receive Clock Master Mode, the bit rate on the RSDn pin is 2.048Mb/s.

In the Receive Clock Master Mode, the RSFSn can be configured by the PERTS\_RSFS (b3, E1-00EH) and REF\_MRSFS (b2, E1-00EH) to output all zeros, to indicate the frame position or to output the same pulse as the RSCFS. When it is defined to indicate the frame position, it can indicate the first bit of a Basic Frame, Signaling Multi-frame, CRC-Multiframe, or both the Signaling and CRC-multiframe. This selection is made by the ROHM, BRXSMFP, BRXCMFP, ALTIFP (b3, b2, b1, b0, E1-011H). When the RSFSn is used for framing pulse indication, the valid polarity of it is configured by the FPINV (b6, E1-011H). In this case, if the FPMODE (b5, E1-011H) is low, the RSFSn can only indicate the Basic Frame no matter what the setting in the ROHM, BRXSMFP, BRXCMFP, ALTIFP (b3, b2, b1, b0, E1-011H).

In the Receive Clock Master Mode, the data on the system interface is clocked by the RSCKn. The active edge of the RSCKn used to update the data on the RSDn and RSFSn is determined by the DE (b4, E1-010H) and the FE (b3, E1-010H) respectively as shown in Table - 14.

The Receive Clock Master Mode includes two sub-modes: Receive Clock Master Full E1 Mode and Receive Clock Master Fractional E1 (with F-bit) Mode.

Table - 14. Active Edge Selection of RSCK (in E1 Receive Clock Master Mode)

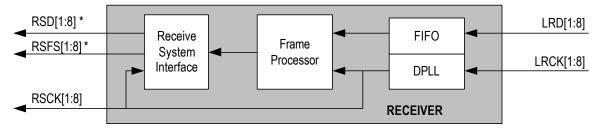
	the Bit Determining the Active Edge of the RSCKn
RSFSn	FE (b3, E1-010H)
RSDn	DE (b4, E1-010H)
Note: If the setting in the FE (b3, E1-010H) and DE (b4, E1-010H) is different, the RSFSn will be one clock edge ahead of RSDn.	

## 3.11.1.2.1 Receive Clock Master Full E1 Mode

Besides all the common functions described in the Receive Clock Master mode, the special feature in this mode (refer to Figure - 13) is that the RSCKn is a standard 2.048MHz clock, and the data in all 32 timeslots in a standard E1 frame is clocked out by the RSCKn.

Figure - 14 shows the functional timing examples. Bit 1 of each timeslot is the first bit to be output.

<sup>\*</sup> It is a special case when the CMS (b2, E1-010H) is logic 1 and the DE (b4, E1-010H) is equal to FE (b3, E1-010H). The signal on the RSDn, RSSIGn and RSFSn are updated on the first active edge of RSCCK.



Note: \* RSD, RSFS are timed to RSCK

Figure - 13. Receive Clock Master Full E1 or T1/J1 Mode

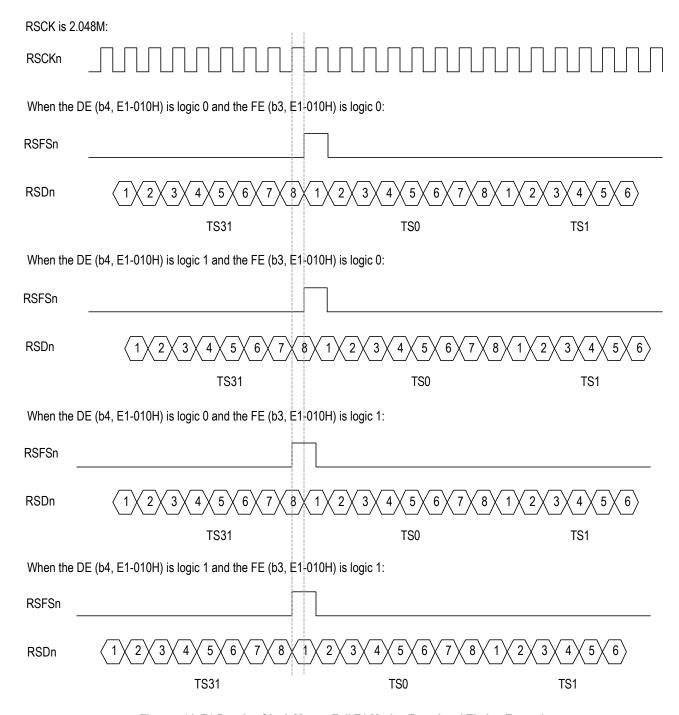


Figure - 14. E1 Receive Clock Master Full E1 Mode - Functional Timing Example

## 3.11.1.2.2 Receive Clock Master Fractional E1 (with F-bit) Mode

Besides all the common functions described in the Receive Clock Master mode, the special feature in this mode (refer to Figure - 15) is that the RSCKn is a gapped 2.048MHz clock (no clock signal during the selected timeslot).

In the Receive Clock Master Fractional E1 mode, the RSCKn is gapped during those timeslots with their DTRKC/NxTS (b6, E1-RPLC-indirect register - 20~3FH) in Receive Payload Control are logic 1. It clocks out during those timeslots with their DTRKC/NxTS\_IDLE (b6, E1-RPLC-indirect register - 20~3FH) set to logic 0. The data in the corresponding gapped timeslot is a don't-care. Figure - 16 shows the functional timing examples. Bit 1 of each timeslot is the first bit to be output.

The Receive Clock Master Fractional E1 with F-bit mode supports ITU recommendation G.802 where an E1 clock is output as a 193-bit T1 clock. In this mode, the RSCKn that starts from the 2nd bit of TS26 and ends at the last bit of the same Basic Frame are gapped, and the TS16 is also gapped. Thus, the DTRKC/NxTS (b6, E1-RPLC-indirect register - 20~3FH) of the timeslots of which the clock are gapped are invalid. The gapping of the remaining timeslots is still determined by the DTRKC/NxTS (b6, E1-RPLC-indirect register - 20~3FH), and the data in the corresponding gapped timeslot is a don't-care.

## 3.11.1.3 Receive Multiplexed Mode

In this mode (refer to Figure - 17), two multiplexed buses are used to receive the data from all eight framers. The data from up to four framers is byte-interleaved and output on one of the two multiplexed buses. The multiplexed bus is selected by the MRBS (b4, E1-001H). When the data from four framers are output on one multiplexed bus, the sequence of data is arranged by setting the timeslot offset TSOFF[6:0] (b6~0, E1-013H). The data from different framers on one multiplexed bus must be shifted by a different timeslot offset to avoid data mixing. Then the received data of each framer can be controlled by the MRBC (b3, E1-001H) to output to the selected multiplexed bus or not.

In the Receive Multiplexed mode, the data on the system interface are clocked by the MRSCCK. The active edge of the MRSCCK to sample the pulse on the MRSCFS and to update the data on the MRSD, MRSFS and MRSSIG are determined by the following bits in the registers (refer to Table - 15).

In the Receive Multiplexed mode, the Multiplexed Receive Side System Common Clock (MRSCCK) is provided by the system side. It is used as a common timing clock for all eight framers. The frequency of the RSCCK can be selected by the CMS (b2, E1-010H) to be the same as the bit rate of the received data stream (8.192Mb/s), or double the

Table - 15. Active Edge Selection of MRSCCK (in E1 Receive Multiplexed Mode)

	Bit Determining the Active Edge of the MRSCCK			
MRSCFS	EE /h2 E1 010U\			
MRSFS	FE (b3, E1-010H)			
MRSD	DE (b/ E1 010H)			
MRSSIG	DE (b4, E1-010H)			

**Note:** if the setting in the FE (b3, E1-010H) and DE (b4, E1-010H) is different, the MRSFS will be one clock edge ahead of MRSD.

The Æ (b3, E1-010H) and DE (b4, E1-010H) of all eight framers should be configured to the same value.

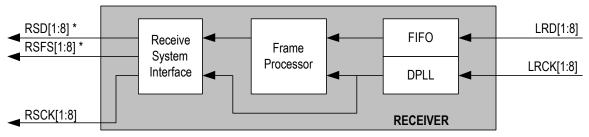
There is a special case when the CMS (b2, E1-010H) is logic 1 and the DE (b4, E1-010H) is equal to FE (b3, E1-010H). The RSD\_RSCFS\_EDGE (b5, E1-014H) is invalid and the signal on the MRSD, MRSSIG and MRSFS pins are updated on the first active edge of MRSCCK.

bit rate of the received data stream (16.384 Mb/s). If the frequency of the RSCCK is double the bit rate of the received data stream, there will be two active edges in one bit time. In this case, the RSD\_RSCFS\_EDGE (b5, E1-014H) determines the active edge to update the signal on the MRSD, MRSSIG and MRSFS pin; however, the pulse on the MRSCFS (if it exists) is always samples on its first active edge. However, if the CMS (b2, E1-010H) or the RSD\_RSCFS\_EDGE (b5, E1-014H) of any of the eight framers is configured as logic 1, all the others are taken as logic 1. That is, the CMS (b2, E1-010H) and the RSD\_RSCFS\_EDGE (b5, E1-014H) of the eight framers should be configured to the same value in the Receive Multiplexed mode.

In the Receive Multiplexed mode, the Multiplexed Receive Side System Common Frame Pulse (MRSCFS) is used as a common framing signal to align the data streams on the two multiplexed buses. The MRSCFS asserts on each first bit of Basic Frame of the selected first framer. The valid polarity of the MRSCFS is configured by the FPINV (b6, E1-011H). The framing signals on MRSCFS can also be ignored by setting the FPMODE (b5, E1-011H) to 0. The FPINV (b6, E1-011H) and the FPMODE (b5, E1-011H) of the eight framers should be set to the same value.

In the Receive Multiplexed mode, the bit rate on the MRSD pin is 8.192Mb/s.

In the Receive Multiplexed mode, the MRSFS can be configured by the PERTS\_RSFS (b3, E1-00EH) and REF\_MRSFS (b2, E1-00EH) to output all zeros, to indicate the frame position or to output the same



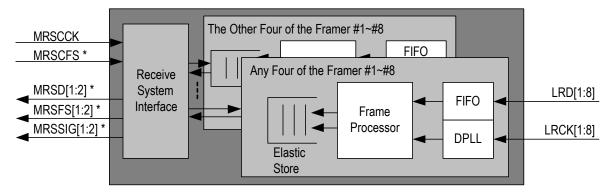
Note: \* RSD, RSFS are timed to gapped RSCK

Figure - 15. Receive Clock Master Fractional E1 or T1/J1 Mode

RSCK is 2.048M. In this example, RSCK is supposed to be held in an inactive state during TS0.

When the DE (b4, E1-010H) is logic 0 and the FE (b3, E1-010H) is logic 0: **RSCKn RSFSn RSDn** Don't Care **TS31** TS1 When the DE (b4, E1-010H) is logic 1 and the FE (b3, E1-010H) is logic 0: **RSCKn RSFSn RSDn** 8 Don't Care **TS31** TS1 When the DE (b4, E1-010H) is logic 0 and the FE (b3, E1-010H) is logic 1: **RSCKn RSFSn RSDn** Don't Care **TS31** TS1 When the DE (b4, E1-010H) is logic 1 and the FE (b3, E1-010H) is logic 1: **RSCKn RSFSn** RSDn Don't Care **TS31** TS1

Figure - 16. E1 Receive Clock Master Fractional E1 Mode - Functional Timing Example



Note: \* MRSCFS, MRSD, MRSFS, MRSSIG are timed to MRSCCK

Figure - 17. Receive Multiplexed Mode

pulse as the MRSCFS. The PERTS\_RSFS (b3, E1-00EH) and REF\_MRSFS (b2, E1-00EH) of the eight framers should be set to the same value. When it is defined to indicate the frame position, it can only indicate the first bit of a Basic Frame of the selected first framer no matter what is set in the ROHM, BRXSMFP, BRXCMFP, ALTIFP (b3, b2, b1, b0, E1-011H). When the MRSFS is for framing pulse indication, the valid polarity of it is configured by the FPINV (b6, E1-011H). The FPINV (b6, E1-011H) of the eight framers should be set to the same value.

In the Receive Multiplexed mode, the MRSSIG outputs extracted signaling. The extracted signaling bits are timeslot aligned with the data outputted on the MRSD. In the Out of Signaling Multi-Frame condition,

the output signaling ABCD on the MRSSIG pin can be forced to be all ones if the OOSMFAIS (b2, E1-001H) is set.

Figure - 18 & 19 show the functional timing examples. Bit 1 of each timeslot is the first bit to be output.

## 3.11.1.4 Parity Check & Polarity Fix

In all the above modes except for the Receive Clock Slave Fractional E1 (with F-bit) mode, if the RPTYE (b6, E1-012H) is logic 1, parity check can be conducted over the bits in the previous Basic Frame and the result is inserted into the first bit (MSB) of TS0 on RSDn/MRSD pin. The even parity or odd parity is selected by the RPTYP (b7, E1-012H) and

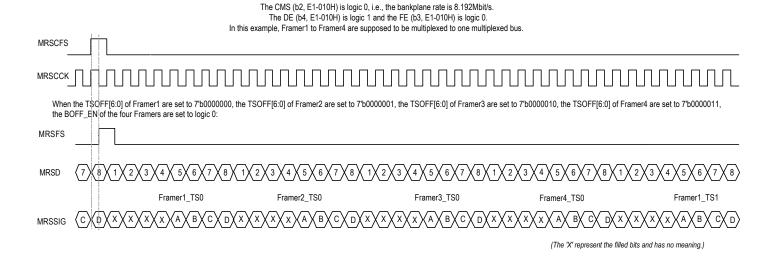


Figure - 18. E1 Receive Multiplexed Mode - Functional Timing Example 1

The CMS (b2, E1-010H) is logic 1, i.e., the bankplane rate is 16.384Mbit/s.

The DE (b4, E1-010H) is logic 0 and the FE (b3, E1-010H) is logic 0.

In this example, Framer1 to Framer4 are supposed to be multiplexed to one multiplexed bus

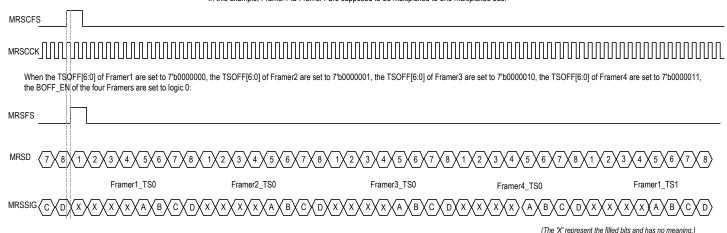


Figure - 19. E1 Receive Multiplexed Mode - Functional Timing Example 2

whether the first bit of TS0 is calculated or not is determined by the PTY\_EXTD (b3, E1-012H). Alternatively this first bit of TS0 can be forced to be logic 0 or 1 by setting the value in the FIXPOL (b4, E1-012H), when the FIXF (b5, E1-012H) is set. The priority of the FIXF (b5, E1-012H) is lower than the RPTYE (b6, E1-012H).

## 3.11.1.5 Offset

In the above five modes, timeslot offset and/or bit offset can be configured. If the offset is configured, the offset between different operation modes is summarized in Table - 16. Bit offset is disabled when the CMS (b2, E1-010H) is logic 1.

The timeslot offset is configured in the TSOFF[6:0] (b6~0, E1-013H). The TSOFF[6:0] (b6~0, E1-013H) give a binary representation.

Enabled by the BOFF\_EN (b3, E1-014H), the bit offset is configured in the BOFF[2:0] (b2~0, E1-014H). The bit offset follows the Concentration Highway Interface (CHI) specification (refer to Table - 17 & 18). When the bit offset is between the RSCFS/MRSCFS and the start of the corresponding frame on the RSDn/MRSD, the CET (clock edge transmit) is counted from the active edge of the RSCFS/MRSCFS (refer to the example in Figure - 20). The pulse on the RSFSn/MRSFS and the signal on the RSSIGn/MRSSIG (if it exists) are aligned to the RSDn/MRSD. When the bit offset is between the RSFSn/MRSFS and the start of the corresponding frame on the RSDn/MRSD, the CET is counted from the active edge of the RSFSn/MRSFS (refer to the example in Figure - 21). The signal on the RSSIGn/MRSSIG (if it exists) is aligned to the RSDn/MRSD.

Note that it is a special case when the BRXSMFP and the ALTIFP (b2, b0, E1-011H) are both set to logical 1. In this case, there is bit offset between the output on the RSFSn and RSDn. Refer to Table - 19 for the details.

## 3.11.1.6 Output On RSDn/MRSD & RSSIGn/MRSSIG

In all the five modes, the RSDn/MRSD and the RSSIGn/MRSSIG pin can be configured by the TRI[1:0] (b1~0, E1-012H) of the corresponding framer to be in high impedance state or to output the processed data stream

The data output on the RSDn/MRSSIG pin can also be forced to be all ones, and the signaling output on the RSSIGn/MRSSIG pin (if it exists) can be forced to be frozen at the current valid signaling when the RAIS (b1, E1-007H) is set.

Table - 16. Offset in Different Operation Modes

Operation Mode	<b>FPMODE</b> (b5, E1-011H)	Offset
Receive Clock Slave mode	1	The offset is between the RSCFS and the start of the corresponding frame on the RSDn (and RSSIGn).
Receive Clock Slave Illoue	0	The offset is between the RSFSn and the start of the corresponding frame on the RSDn (and RSSIGn).
Receive Clock Master mode	0 (must be zero)	The offset is between the RSFSn and the start of the corresponding frame on the RSDn.
Dessive Multiplexed made	1 (in any of the eight framers)	The offset is between the MRSCFS and the start of the corresponding frame on the MRSD and MRSSIG.
Receive Multiplexed mode	0	The offset is between the MRSFS and the start of the corresponding frame on the MRSD and MRSSIG.

Table - 17. Receive System Interface Bit Offset (FPMODE [b5, E1-011H] = 0)

FE	DE		BOFF[2:0] (b2~0, E1-014H)							
(b3, E1-010H)	(b4, E1-010H)	000	001	010	011	100	101	110	111	
0	0	4	6	8	10	12	14	16	18	
0	1	3	5	7	9	11	13	15	17	CET
1	0	3	5	7	9	11	13	15	17	
1	1	4	6	8	10	12	14	16	18	

Table - 18. Receive System Interface Bit Offset (FPMODE [b5, E1-011H] = 1)

FE	DE		BOFF[2:0] (b2~0, E1-014H)							
(b3, E1-010H)	(b4, E1-010H)	000	001	010	011	100	101	110	111	
0	0	2	4	6	8	10	12	14	16	
0	1	1	3	5	7	9	11	13	15	CET
1	0	1	3	5	7	9	11	13	15	
1	1	2	4	6	8	10	12	14	16	

Table - 19. Bit Offset Between RSFSn and RSDn When the BRXSMFP and the ALTIFP (b2, b0, E1-011H) are Both Set To Logical 1

<b>BOFF_EN</b> (b3, E1-014H)	<b>FPMODE</b> (b5, E1-011H)	<b>DE</b> (b4, E1-010H) & <b>FE</b> (b3, E1-010H)	Bit Offset Between the RSFSn and the RSDn	
0	v	Same	1 bit offset. RSFSn is ahead.	
U	^	Difference	1.5 bit offset. RSFSn is ahead.	
	1	4	Same	1 bit offset. RSFSn is ahead.
1	I	Difference	1.5 bit offset. RSFSn is ahead.	
'	0	Same	CHI encoification (Table 17)	
	U	Difference	CHI specification (Table – 17)	

For example: when DE (b4, E1-010H) = 0, FE (b3, E1-010H) = 0

For example: when DE (b4, E1-010H) = 1, FE (b3, E1-010H) = 0

**TS31** 

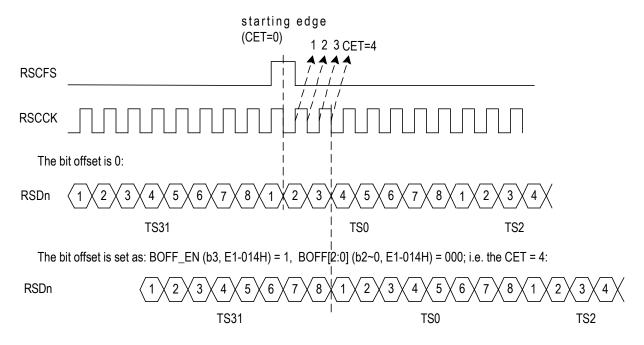


Figure - 20. Receive Bit Offset - Between RSCFS & RSDn

RSFSn

RSCCK/RSCKn

The bit offset is 0:

RSDn

Told Triangledge
(CET=0)

Told Triangledge
(CET=0)

Told Triangledge
(CET=0)

Told Triangledge
(CET=3)

Told Triangledge
(CET=3)

The bit offset is 0:

The bit offset is 0:

TS31

TS0

TS2

The bit offset is set as: BOFF\_EN (b3, E1-014H) = 1, BOFF[2:0] (b2~0, E1-014H) = 001; i.e. the CET = 3:

RSDn

Told Triangledge
(CET=0)

Told Triangledge
(CET=0)

Told Triangledge
(CET=3)

The bit offset is 0:

TS31

TS0

TS2

The bit offset is set as: BOFF\_EN (b3, E1-014H) = 1, BOFF[2:0] (b2~0, E1-014H) = 001; i.e. the CET = 3:

Figure - 21. Receive Bit Offset - Between RSFSn & RSDn

TS0

TS2

## 3.11.2 T1 / J1 MODE

In T1/J1 mode, the Receive System Interface can be set in Non-multiplexed Mode or Multiplexed Mode. In Non-multiplexed Mode, the RSDn pin is used to output the received data from each framer at the bit rate of 1.544 Mb/s or 2.048 Mb/s (T1/J1 mode E1 rate). While in the Multiplexed Mode, the received data from the eight framers are converted to 2.048 Mb/s format and byte interleaved to form two high speed data streams and output on the MRSD1 and MRSD2 pins at the bit rate of 8.192 Mb/s.

In the Non-multiplexed Mode, if the timing signal for clocking data on the RSDn pin is provided by the system side and shared by all eight framers, the Receive System Interface should be set in Receive Clock Slave mode. If the timing signal for clocking data on each RSDn pin is received from each line side, the Receive System Interface should be set in Receive Clock Master mode.

In the Receive Clock Slave mode, if the multi-function pin RSCKn/RSSIGn is used to output a reference clock, the Receive System Interface is in Receive Clock Slave RSCK Reference Mode. If the RSCKn/RSSIGn pin is used to output the extracted signaling bits, the Receive System Interface is in Receive Clock Slave External Signaling mode.

The T1/J1 mode E1 rate, which means the system clock rate is 2.048 MHz in T1/J1 mode, can only be supported in the Receive Clock Slave mode.

In the Receive Clock Master mode, if the data in all 24 channels of a T1/J1 basic frame is clocked out by the RSCKn signal, the Receive System Interface is in Receive Clock Master Full T1/J1 mode. If the data in only some timeslots of a T1/J1 frame is clocked out by the RSCKn, then the Receive System Interface is in Receive Clock Master Fractional T1/J1 Mode.

Table - 20 summarizes the receive system interface in different operating modes. To set the receive system interface of each framer into various operating modes, the registers must be configured as Table - 21.

## 3.11.2.1 Receive Clock Slave Mode

In the Receive Clock Slave Mode, the bit rate on the RSDn pin is 1.544Mb/s. However, if the system clock rate is 2.048MHz, the received data stream (1.544 Mb/s) should be converted to the same rate as the system side, that is, to work in T1/J1 mode E1 rate. Thus, the RSCCK2M (b4, T1/J1-001H) and the RSCCK8M (b3, T1/J1-001H) should be set to logic 1 and 0 respectiively. The conversion complies as follows: One dummy byte is inserted in the system side before 3 bytes of Frame N from the device are converted. This process repeats 8 times and the conversion of Frame N of 1.544M bit/s data rate to 2.048M bit/s data rate is completed. However, the F-bit of Frame N of the 1.544M bit/s data rate is inserted as the 8th bit of the N of the 2.048M bit/s data rate (refer to Figure - 22).

In the Receive Clock Slave Mode, the Receive Side System Common Clock (RSCCK) is provided by the system side. It is used as a common timing clock for all eight framers. The speed of the RSCCK can be 1.544MHz or 2.048MHz. When it is 2.048MHz, the RSCCK can be selected by the CMS (b4, T1/J1-078H) to be the same as the received data (2.048Mb/s), or double the received data (4.096 Mb/s). The CMS (b4, T1/J1-078H) of the eight framers should be set to the same value. If the speed of the RSCCK is double the received data stream, there will be two active edges in one bit duration. In this case, the RSD\_RSCFS\_EDGE (b5, T1/J1-078H) determines the active edge to update the signal on the RSDn, RSSIGn and RSFSn pins; however, the pulse on the RSCFS is always sampled on its first active edge.

Table - 20. T1/J1 Mode Receive System Interface in Different Operation Modes

Operation Mode		Data Pin	Clock Pin	Framing Pin	Signaling Pin	Reference Clock	
Non-	Clock Slave	RSCK Reference	RSDn	RSCCK	RSCFS/RSFSn	No	RSCKn
Multiplexed	Mode	External Signaling	RSDn	RSCCK	RSCFS/RSFSn	RSSIGn	No
Mode	Clock Master	Full T1/J1	RSDn	RSCKn	RSFSn	No	No
	Mode	Fractional T1/J1	RSDn	RSCKn	RSFSn	No	No
ľ	Multiplexed Mode		MRSD	MRSCCK	MRSCFS/MRSFS	MRSSIG	No

Table - 21. Operation Mode Selection in T1/J1 Receive Path

RSCCK2M / RSCCK8M (b4, T1/J1-001H) / (b3, T1/J1-001H)	IMODE[1:0] (b7~6, T1/J1-001H)	Operation Mode
00 / 10 *	10	Receive Clock Slave RSCK Reference
00 / 10	11	Receive Clock Slave External Signaling
00	01	Receive Clock Master Full T1/J1
00	00	Receive Clock Master Fractional T1/J1
01 (in any of the eight framers)	11	Receive Multiplexed

#### Note:

<sup>\*</sup> When the RSCCK2M / RSCCK8M are '00', the system clock rate is 1.544MHz.

When the RSCCK2M / RSCCK8M are '10', the system clock rate is 2.048MHz, i.e., T1/J1 mode E1 rate.

Figure - 22. T1/J1 To E1 Format Conversion

In the Receive Clock Slave Mode, the Receive Side System Common Frame Pulse (RSCFS) is used as a common framing signal to align the data stream for all eight framers. The RSCFS asserts on each F-bit and its valid polarity is configured by the FPINV (b6, T1/J1-078H).

In the Receive Clock Slave Mode, the RSFSn can indicate each F-bit of SF/ESF, every second F-bit or the first F-bit of every 12 frames (in SF format) / every 24 frames (in ESF format). All the indications are selected by the RSFSP (b2, T1/J1-001H) and ALTIFP (b1, T1/J1-001H). The valid polarity of the RSFSn is configured by the FPINV (b6, T1/J1-078H).

The Receive Clock Slave Mode includes two sub-modes: Receive Clock Slave RSCK Reference Mode and Receive Clock Slave External Signaling Mode. Note that if the receive system interface is configured to operate in T1/J1 mode E1 rate, framer 1, 3, 5, 7 must be configured in the same sub-mode and framer 2, 4, 6, 8 must be configured in the same sub-mode.

## 3.11.2.1.1 Receive Clock Slave RSCK Reference Mode

In this mode (refer to Figure - 7), the data on the system interface is clocked by the RSCCK. The active edge of the RSCCK to sample the data on the RSCFS pin or to update the data on the RSDn and RSFSn

pins is determined by the following bits in the registers (refer to Table - 22).

Table - 22. Active Edge Selection of RSCCK (in T1/J1 Receive Clock Slave RSCK Reference Mode)

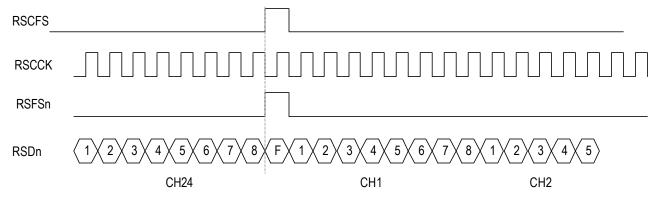
	the Bit Determining the Active Edge of the RSCCK
RSCFS	RSCFSFALL (b1, T1/J1-003H)
RSFSn	DCCCKDISE (PU 11/11 003H)
RSDn	RSCCKRISE (b0, T1/J1-003H)

**Note:** The RSCFSFALL (b1, T1/J1-003H) of the eight framers should be set to the same value to ensure the RSCFS for the eight framers is sampled on the same active edge.

It is a special case when the CMS (b4, T1/J1-078H) is logic 1 and the RSCFSFALL (b1, T1/J1-003H) is not equal to RSCCKRISE (b0, T1/J1-003H). The RSD\_RSCFS\_EDGE (b5, T1/J1-078H) is invalid and the signals on the RSDn and the RSFSn pins are updated on the first active edge of RSCCK.

Figure - 23 to 25 show the functional timing examples. Bit 1 of each channel is the first bit to be output.

The CMS (b4, T1/J1-078H) is logic 0 and the bankplane rate is 1.544Mbit/s. The RSCFSFALL (b1, T1/J1-003H) is logic 0 and the RSCCKRISE (b0, T1/J1-003H) is logic 0. The channel offset and the bit offset enable are both 0:



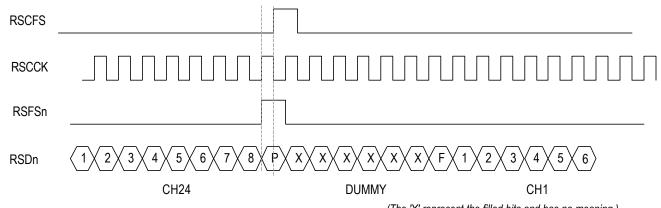
(The RSCKn is selected by the RSCKSEL (b5, T1/J1-001H) to output a jitter attenuated 1.544MHz (i.e., smoothed LRCKn) or 8KHz clock (smoothed LRCKn divided by 193).)

Figure - 23. T1/J1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 1

The CMS (b4, T1/J1-078H) is logic 0 and the bankplane clock rate is 2.048Mbit/s.

The RSCFSFALL (b1, T1/J1-003H) is logic 0 and the RSCCKRISE (b0, T1/J1-003H) is logic 1.

The channel offset and the bit offset enable are both 0:

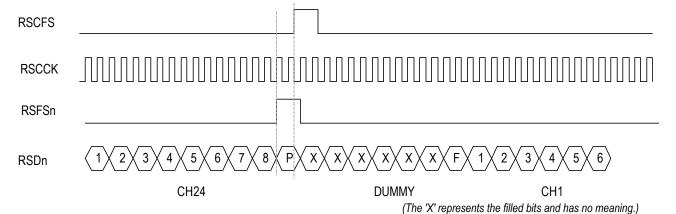


(The 'X' represent the filled bits and has no meaning.)

(The RSCKn is selected by the RSCKSEL (b5, T1/J1-001H) to output a jitter attenuated 1.544MHz (i.e., smoothed LRCKn) or 8KHz clock (smoothed LRCKn divided by 193).)

Figure - 24. T1/J1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 2

The CMS (b4, T1/J1-078H) is logic 1, i.e., the bankplane clock rate is 4.096Mbit/s. The RSCFSFALL (b1, T1/J1-003H) is logic 0 and the RSCCKRISE (b0, T1/J1-003H) is logic 1. When the channel offset and the bit offset enable are both 0:



(The RSCKn is selected by the RSCKSEL (b5, T1/J1-001H) to output a jitter attenuated 1.544MHz (i.e., smoothed LRCKn) or 8KHz clock (smoothed LRCKn divided by 193).)

Figure - 25. T1/J1 Receive Clock Slave RSCK Reference Mode - Functional Timing Example 3

Besides all the common functions described in the Receive Clock Slave mode, the special feature in this mode is that the multi-functional pin RSCKn/RSSIGn is used as RSCKn to output a reference clock. The RSCKn can be selected by the RSCKSEL (b5, T1/J1-001H) to output a jitter attenuated 1.544MHz (i.e., smoothed LRCKn) or 8KHz clock (smoothed LRCKn divided by 193).

# 3.11.2.1.2 Receive Clock Slave External Signaling Mode

In this mode (refer to Figure - 10), the data on the system interface is clocked by the RSCCK. The active edge of the RSCCK to sample the pulse on the RSCFS or to update the data on the RSDn, RSFSn and RSSIGn pins is determined by the following bits in the registers (refer to Table - 23).

Table - 23. Active Edge Selection of RSCCK (in T1/J1 Receive Clock Slave External Signaling Mode)

	the Bit Determining the Active Edge of the RSCCK
RSCFS	RSCFSFALL (b1, T1/J1-003H)
RSFSn	
RSDn	RSCCKRISE (b0, T1/J1-003H)
RSSIGn	•

**Note:** The RSCFSFALL (b1, T1/J1-003H) of the eight framers should be set to the same value to ensure the RSCFS for the eight framers is sampled on the same active edge.

It is a special case when the CMS (b4, T1/J1-078H) is logic 1 and the RSCFSFALL (b1, T1/J1-003H) is not equal to RSCCKRISE (b0, T1/J1-003H). The RSD\_RSCFS\_EDGE (b5, T1/J1-078H) is invalid and the signal on the RSDn, RSSIGn and the RSFSn pins are updated on the first active edge of RSCCK.

Figure - 26 to 28 show the functional timing examples. Bit 1 of each channel is the first bit to be output.

Besides all the common functions described in the Receive Clock Slave mode, the special feature in this mode is that the multi-functional pin RSCKn/RSSIGn is used as RSSIGn to output the extracted signaling bits. The extracted signaling bits are channel aligned with the data on the RSDn pin (refer to Figure - 6).

## 3.11.2.2 Receive Clock Master Mode

In the Receive Clock Master mode, each framer uses its own clock signal on RSCKn pin and framing signal on RSFSn pin to output the data on each RSDn pin.

In the Receive Clock Master Mode, the bit rate on the RSDn pin is 1.544Mb/s.

In the Receive Clock Master Mode, the RSFSn can indicate each F-bit of SF/ESF, every second F-bit or the first F-bit of every 12 frames (in SF format) / every 24 frames (in ESF format). All the indications are selected by the RSFSP (b2, T1/J1-001H) and ALTIFP (b1, T1/J1-001H). The valid polarity of the RSFSn is configured by the FPINV (b6, T1/J1-078H).

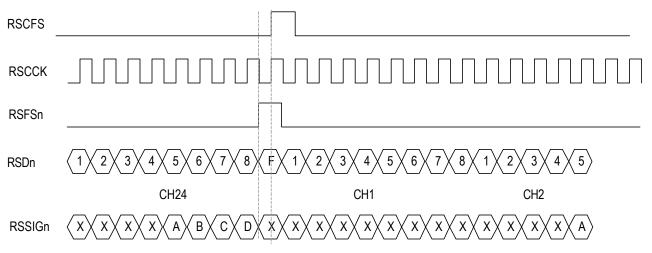
In the Receive Clock Master Mode, the data on the system interface is clocked by the RSCKn. The active edge of the RSCKn to update the data on the RSDn and RSFSn is determined by the RSCKRISE(b3, T1/J1-003H).

The Receive Clock Master Mode includes two sub-modes: Receive Clock Master Full T1/J1 Mode and Receive Clock Master Fractional T1/J1 Mode.

The CMS (b4, T1/J1-078H) is logic 0 and the bankplane clock rate is 1.544Mbit/s.

The RSCFSFALL (b1, T1/J1-003H) is logic 1 and the RSCCKRISE (b0, T1/J1-003H) is logic 0.

The channel offset and the bit offset enable are both 0:



(The 'X' represent the filled bits and has no meaning.)

Figure - 26. T1/J1 Receive Clock Slave External Signaling Mode - Functional Timing Example 1

The CMS (b4, T1/J1-078H) is logic 0 and the bankplane clock rate is 2.048Mbit/s.

The RSCFSFALL (b1, T1/J1-003H) is logic 1 and the RSCCKRISE (b0, T1/J1-003H) is logic 1.

The channel offset and the bit offset enable are both 0:

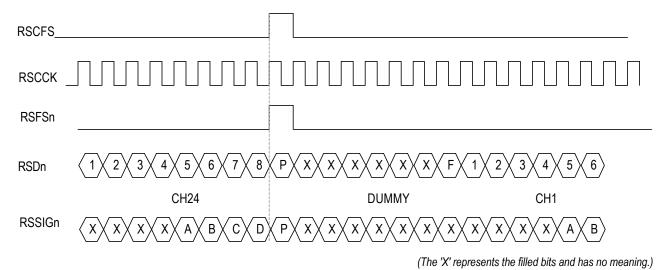


Figure - 27. T1/J1 Receive Clock Slave External Signaling Mode - Functional Timing Example 2

The CMS (b4, T1/J1-078H) is logic 1, i.e., the bankplane clock rate is 4.096Mbit/s. The RSCFSFALL (b1, T1/J1-003H) is logic 1 and the RSCCKRISE (b0, T1/J1-003H) is logic 1.

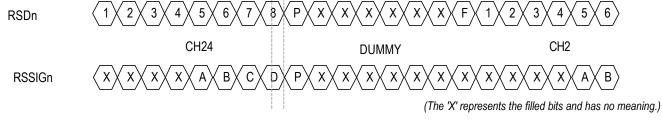


Figure - 28. T1/J1 Receive Clock Slave External Signaling Mode - Functional Timing Example 3

# 3.11.2.2.1 Receive Clock Master Full T1/J1 Mode

Besides all the common functions described in the Receive Clock Master mode, the special feature in this mode (refer to Figure - 13) is that the RSCKn is a standard 1.544MHz clock, and the data in all the 24 channels in a standard T1/J1 frame is clocked out by the RSCKn.

Figure - 29 shows the functional timing examples. Bit 1 of each channel is the first bit to be output.

# 3.11.2.2.2 Receive Clock Master Fractional T1/J1 Mode

Besides all the common functions described in the Receive Clock Master mode, the special feature in this mode (refer to Figure - 15) is that the RSCKn is a gapped 1.544MHz clock (no clock signal during the selected channel).

The RSCKn is gapped during those channels with their EXTRACT (b2, T1/J1-RPLC-indirect registers - 01~18H) in Receive Payload Control are logic 0 and clocks out during those channels with their EXTRACT (b2, T1/J1-RPLC-indirect registers - 01~18H) are logic 1. The data in the corresponding gapped channel is a don't care condition.

Figure - 30 shows the functional timing examples. Bit 1 of each channel is the first bit to be output.

# 3.11.2.3 Receive Multiplexed Mode

In this mode (refer to Figure - 17), two multiplexed buses are used to receive the data from all eight framers. The data from up to four framers is byte-interleaved output on one of the two multiplexed buses. The multiplexed bus is selected by the MRBS (b7, T1/J1-003H). When the data from four framers are output on one multiplexed bus, the sequence of data is arranged by setting the channel offset TSOFF[6:0] (b6~0, T1/J1-077H). The data from different framers on one multiplexed bus must be shifted at a different channel offset to avoid data mixing. Then the re-

ceived data of each framer can be controlled by the MRBC (b6, T1/J1-003H) to output to the selected multiplexed bus or not.

In the Receive Multiplexed mode, the data on the system interface are clocked by the MRSCCK. The active edge of the MRSCCK to sample the pulse on the MRSCFS and to update the data on the MRSD, MRSFS and MRSSIG are determined by the following bits in the registers (refer to Table - 24).

Table - 24. Active Edge Selection of MRSCCK (in T1/J1 Receive Multiplexed Mode)

	the Bit Determining the Active Edge of the MRSCCK
MRSCFS	RSCFSFALL (b1, T1/J1-003H)
MRSFS	
MRSD	RSCCKRISE (b0, T1/J1-003H)
MRSSIG	

**Note:** When the RSCFSFALL/RSCCKRISE of any of the eight framers is configured as logic 1, all the others are taken as logic 1. That is, the RSCFSFALL/RSCCKRISE should be configured to the same value in Receive Multiplexed mode.

It is a special case when the CMS (b4, T1/J1-078H) is logic 1 and the RSCFSFALL (b1, T1/J1-003H) is not equal to RSCCKRISE (b0, T1/J1-003H). The RSD\_RSCFS\_EDGE (b5, T1/J1-078H) is invalid and the signal on the MRSD, MRSSIG and the MRSFS pins are updated on the first active edge of RSCCK.

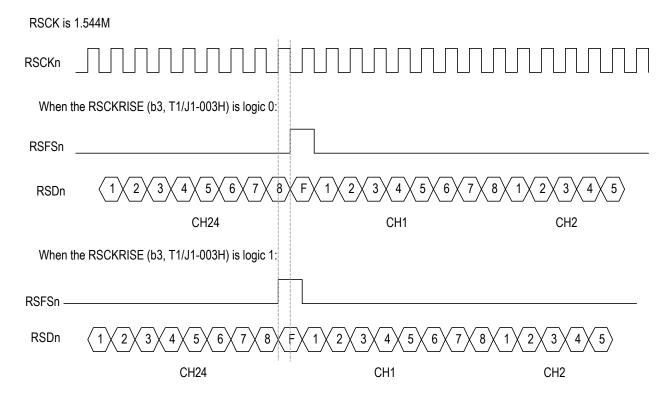


Figure - 29. T1/J1 Receive Clock Master Full T1/J1 Mode - Functional Timing Example

RSCK is 1.544M. In this example, RSCK is supposed to be held in inactive state during CH2.

Figure - 30. T1/J1 Receive Clock Master Fractional T1/J1 Mode - Functional Timing Example

In the Receive Multiplexed mode, the Multiplexed Receive Side System Common Clock (MRSCCK) is provided by the system side. It is used as a common timing clock for all eight framers. The frequency of the MRSCCK can be selected by the CMS (b4, T1/J1-078H) to be the same as the bit rate of the received data stream (8.192Mb/s), or double the bit rate of the received data stream (16.384 Mb/s). If the frequency of the RSCCK is double the bit rate of the received data stream, there will be two active edges in one bit duration. In this case, the RSD\_RSCFS\_EDGE (b5, T1/J1-078H) determines the active edge to update the signal on the MRSD, MRSSIG and MRSFS pin; however, the pulse on the MRSCFS is always sampled on its first active edge. If the CMS (b4, T1/J1-078H) or the RSD\_RSCFS\_EDGE (b5, T1/J1-078H) of any of the eight framers is configured as logic 1, all the others are taken as logic 1. That is, the CMS (b4, T1/J1-078H) and the RSD\_RSCFS\_EDGE (b5, T1/J1-078H) of the eight framers should be configured to the same value in the Receive Multiplexed mode.

In the Receive Multiplexed mode, the Multiplexed Receive Side System Common Frame Pulse (MRSCFS) is used as a common framing signal to align the data streams on the two multiplexed buses. The MRSCFS is asserted on the F-bit. The valid polarity of the MRSCFS is configured by the FPINV (b6, T1/J1-078H). The FPINV (b6, T1/J1-078H) of the eight framers should be set to the same value.

In the Receive Multiplexed mode, the bit rate on the MRSD pin is 8.192Mb/s.

In the Receive Multiplexed mode, regardless of the setting in the RSFSP (b2, T1/J1-001H) and ALTIFP (b1, T1/J1-001H), the MRSFS can only indicate each F-bit of SF/ESF of the selected first framer. The valid

polarity of the RSFSn is configured by the FPINV (b6, T1/J1-078H). The FPINV (b6, T1/J1-078H) of the eight framers should be set to the same value.

In the Receive Multiplexed mode, the MRSSIG outputs extracted signaling. The extracted signaling bits are channel aligned with the data outputted on the MRSD.

Figure - 31 & 32 show the functional timing examples. Bit 1 of each channel is the first bit to be output.

# 3.11.2.4 Parity Check

In all the above modes except for the Receive Clock Slave Fractional T1/J1 mode, if the RPRTYE (b0, T1/J1-002H) is logic 1, parity check can be conducted over the bits in the previous frame and the result is inserted into the F-bit on the RSDn/MRSD and RSSIGn/MRSSIG pin. The even parity or odd parity is selected by the RPTYP (b1, T1/J1-002H) and whether the F-bit is calculated or not is determined by the PTY\_EXTD (b3, T1/J1-002H).

(The 'X' represents the filled bits and has no meaning.)

The CMS (b4, T1/J1-078H) is logic 0, i.e., the bankplane clock rate is 8.192Mbit/s. The RSCCKRISE(b0, T1/J1-003H) is logic 1 and the RSCFSFALL (b1, T1/J1-003H) is logic 0. In this example, Framer1 to Frame4 are supposed to be multiplexed to one multiplexed bus.

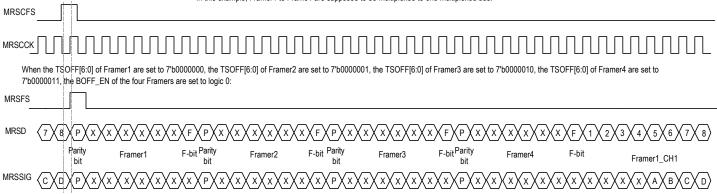


Figure - 31. T1/J1 Receive Multiplexed Mode - Functional Timing Example 1

The CMS (b4, T1/J1-078H) is logic 1, i.e., the bankplane clock rate is 16.384Mbit/s.

The RSCCKRISE(b0, T1/J1-003H) is logic 1 and the RSCFSFALL (b1, T1/J1-003H) is logic 1.

In this example, Framer1 to Frame4 are supposed to be multiplexed to one multiplexed bus.

The TSOFF[6:0] of Framer1 are set to 7'b0000000, the TSOFF[6:0] of Framer2 are set to 7'b0000001, the TSOFF[6:0] of Framer3 are set to 7'b0000010, the TSOFF[6:0] of Framer4 are set to 7'b0000011, the BOFF\_EN of the four Framers are set to logic 0:



Figure - 32. T1/J1 Receive Multiplexed Mode - Functional Timing Example 2

## 3.11.2.5 Offset

When the system clock rate is 2.048MHz (in Receive Clock Slave T1/J1 mode E1 rate mode) or 8.192MHz (in Receive Multiplexed mode), channel offset and/or bit offset between the RSCFS and the start of the corresponding frame on the RSDn/MRSD (and RSSIGn/MRSSIG) can be configured. Bit offset is disabled when the CMS (b4, T1/J1-078H) is logic 1.

The channel offset is configured in the TSOFF[6:0] (b6~0, T1/J1-077H). The TSOFF[6:0] (b6~0, E1-013H) give a binary representation.

Enabled by the BOFF\_EN (b3, T1/J1-078H), the bit offset is configured in the BOFF[2:0] (b2~0, T1/J1-078H). The bit offset follows the Concentration Highway Interface (CHI) specification (refer to Table - 25). The CET (clock edge transmit) is counted from the active edge of the RSCFS/MRSCFS (refer to the example in Figure - 33). The pulse on the RSFSn/MRSFS and the signal on the RSSIGn/MRSSIG (if exists) are aligned to the RSDn/MRSD.

## 3.11.2.6 Output On RSDn/MRSD & RSSIGn/MRSSIG

In all the five modes, the RSDn/MRSD and the RSSIGn/MRSSIG pin can be configured by the TRI[1:0] (b5~4, T1/J1-003H) of the corresponding framer to be in high impedance state or to output the processed data stream.

Table - 25. Receive System Interface Bit Offset

RSCFSFALL	RSCCKRISE		BOFF[2:0] (b2~0, T1/J1-078H)							
(b1, T1/J1-003)	(b0, T1/J1-003H)	000	001	010	011	100	101	110	111	
1	0	2	4	6	8	10	12	14	16	
1	1	1	3	5	7	9	11	13	15	CET
0	0	1	3	5	7	9	11	13	15	
0	1	2	4	6	8	10	12	14	16	

For example: when RSCFSFALL (b1, T1/J1-003H) = 1, RSCCKRISE (b0, T1/J1-003H) = 0

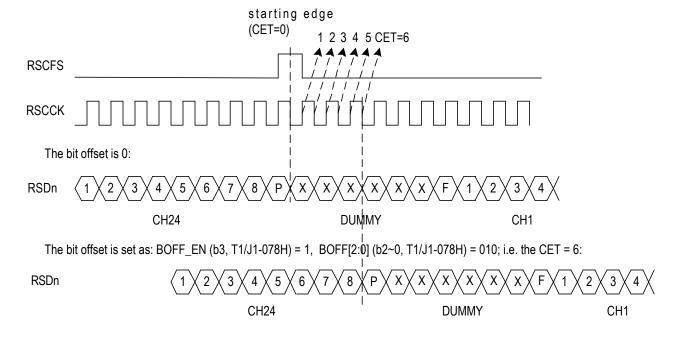


Figure - 33. Receive Bit Offset in T1/J1 Mode

# 3.12 PRBS GENERATOR / DETECTOR (PRGD)

The PRBS Generator/Detector is shared by eight framers. It can be assigned to any of the 8 framers at one time. The PRGD, together with the Receive / Transmit Payload Control blocks, is used to test the data stream.

## 3.12.1 E1 MODE

The PRBS Generator/Detector is a global control block. Any one of the eight framers can be linked to the pattern generator or detector by the PRGDSEL[2:0] (b7~5, E1-00CH). The pattern can be inserted in either the transmit or receive direction, and detected in the opposite direction. The direction is determined by the RXPATGEN (b2, E1-00CH). The pattern can be generated or detected in unframed or framed mode. The selection is made by the UNF\_GEN (b1, E1-00CH) or UNF\_DET (b0, E1-00CH) respectively. In unframed mode, all the 32 timeslots are replaced or extracted and the specification of the TEST (b7, E1-RPLC-indirect registers - 20~3FH) in Receive / Transmit Payload Control blocks are ignored. In framed mode, the timeslot is specified by the TEST (b7, E1-RPLC-indirect registers - 20~3FH) or b3, E1-TPLC-indirect registers - 20~3FH).

#### **Pattern Generator**

The repetitive or pseudo-random pattern selected by the PS (b4, E1-070H) is located in the PI[31:0] (b7~0, E1-078H & b7~0, E1-079H & b7~0, E1-07AH & b7~0, E1-07BH). However, the length of the valid data in the PI[31:0] is determined by the PL[4:0] (b4~0, E1-072H). If the repetitive pattern is selected, the valid PI[X:0] (X is equal to one number of the 31 to 1) reflect its content directly. If the pseudo-random pattern is selected, the valid PI[X:0] are its initial value and the feedback tap position (refer to Figure - 34) is determined by the PT[4:0] (b4~0, E1-073H). A single bit error can be inserted by setting the EVENT (b3, E1-074H) to 1, or continuous bit errors can be inserted at a bit error rate determined by the EIR[2:0] (b2~0, E1-074H). Before replacing the data in the assigned direction, the pattern can be inverted with the TINV (b3, E1-070H) enabled.

## **Pattern Detector**

The extracted data from the assigned direction are compared with a repetitive or pseudo-random pattern selected by the PS (b4, E1-070H). Before being compared, the data can be inverted with the RINV (b2, E1-070H) enabled. The extracted data are then compared with a 48-bit fixed window loaded with the pattern. This process goes on until the data coincide with the pattern. When they are synchronized, it is indicated by the SYNCV (b4, E1-071H). Bit errors in the synchronized data are indicated in the BEI (b2, E1-071H). When there are more than 10-bit errors in the fixed 48-bit window, the extracted data are out of synchronization. Automatic search for the re-synch will be done with the AUTOSYNC (b1, E1-070H) configured, or manual search can be done when there is a transition from low to high on the MANSYNC (b0, E1-070H). A manual search is recommended to execute to ensure the PRGD operates correctly when there is any setting change of the PRGD registers or the detector data source changes.

Selected by the PDR[1:0] (b7~6, E1-070H), the PD[31:0] (b7~0, E1-07CH & b7~0, E1-07DH & b7~0, E1-07EH & b7~0, E1-07FH) can contain the received pattern, the total error count or the total number of received bits. They update when the defined intervals are initiated. The intervals equal 1 second when the AUTOUPDATE (b0, E1-000H) is set in the corresponding framer. They can also be updated by writing to any of the PD[31:0] (b7~0, E1-07CH & b7~0, E1-07DH & b7~0, E1-07FH), or to the E1 Chip ID / Global PMON Update register (E1-009H). The update will be indicated by the XFERI (b1, E1-071H). If they are not read in the defined intervals, the PD[31:0] (b7~0, E1-07CH & b7~0, E1-07DH & b7~0, E1-07FH) will be overwritten with new data. The overwritten condition is indicated by the OVR (b0, E1-071H).

3 kinds of interrupts can be generated by this block:

- bit errors;
- synchronization status change (indicated in the SYNCI [b3, E1-071H]);
- the PD[31:0] (b7~0, E1-07CH & b7~0, E1-07DH & b7~0, E1-07EH & b7~0, E1-07FH) are updated.

When the interrupts are enabled by the BEE (b6, E1-071H), SYNCE

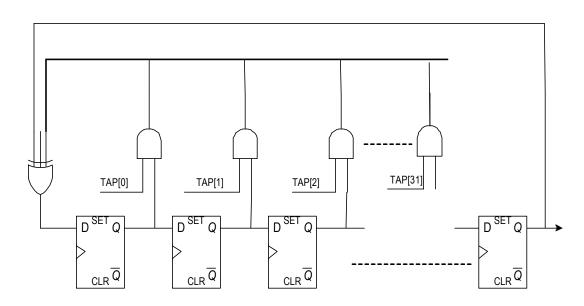


Figure - 34. PRBS Pattern Generator

(b7, E1-071H) and XFERE (b5, E1-071H) respectively, the  $\overline{\text{INT}}$  pin is asserted.

# 3.12.2 T1 / J1 MODE

The PRBS Generator/Detector is a global control block. Any one of the eight framers can be linked to pattern generator or detector by the PRGDSEL[2:0] (b7~5, T1/J1-00FH). The pattern can be inserted in either the transmit or receive direction, and detected in the opposite direction. The direction is determined by the RXPATGEN (b2, T1/J1-00FH). The pattern can be generated or detected in unframed or framed mode. The selection is made by the UNF\_GEN (b1, T1/J1-00FH) or UNF\_DET (b0, T1/J1-00FH) respectively. In unframed mode, all the 24 channels are replaced or extracted and the specification of the TEST (b3, T1/J1-RPLC-indirect registers - 01~18H or b3, T1/J1-TPLC-indirect registers -01~18H) in Receive / Transmit Payload Control blocks are ignored. In framed mode, the channel is specified by the TEST (b3, T1/J1-RPLC-indirect registers - 01~18H or b3, T1/J1-TPLC-indirect registers - 01~18H). However, fractional T1/J1 signal can be replaced or extracted in the specified channel when the Nx56k\_GEN (b4, T1/J1-00FH) or Nx56k\_DET (b3, T1/J1-00FH) is set respectively.

## **Pattern Generator**

The repetitive or pseudo-random pattern selected by the PS (b4, T1/J1-060H) is located in the PI[31:0] (b7~0, T1/J1-068H & b7~0, T1/J1-069H & b7~0, T1/J1-06AH & b7~0, T1/J1-06BH). However, the length of the valid data in the PI[31:0] is determined by the PL[4:0] (b4~0, T1/J1-062H). If the repetitive pattern is selected, the valid PI[X:0] (X is valid for 1 to 31) reflect its content directly. If the pseudo-random pattern is selected, the valid PI[X:0] are its initial value and the feedback tap position (refer to Figure - 34) is determined by the PT[4:0] (b4~0, T1/J1-063H). A single bit error can be inserted by setting the EVENT (b3, T1/J1-064H), or continuous bit errors can be inserted at a bit error rate determined by the EIR[2:0] (b2~0, T1/J1-064H). Before replacing the data in the assigned direction, the pattern can be inverted with the TINV (b3, T1/J1-060H) enabled.

# **Pattern Detector**

The extracted data from the assigned direction are compared with a repetitive or pseudo-random pattern selected by the PS (b4, T1/J1-060H). Before being compared, the data can be inverted with the RINV (b2, T1/J1-060H) enabled. The extracted data are then compared with a 48-bit fixed window loaded with the pattern. This process continues until the data coincide with the pattern. They are then synchronized with an indication in the SYNCV (b4, T1/J1-061H). Bit errors in the synchronized data are indicated in the BEI (b2, T1/J1-061H). When there are more than 10-bit errors in the fixed 48-bit window, the extracted data are out of synchronization. Automatic search for the re-synch will be done with the AUTOSYNC (b1, T1/J1-060H) configured, or manual search can be done when there is a transition from low to high on the MANSYNC (b0, T1/J1-060H). A manual search is recommended to execute to ensure the PRGD operates correctly when there is any setting change of the PRGD registers or the detector data source changes.

Selected by the PDR[1:0] (b7~6, T1/J1-060H), the PD[31:0] (b7~0, T1/J1-06CH & b7~0, T1/J1-06DH & b7~0, T1/J1-06EH & b7~0, T1/J1-06FH) can contain the received pattern, the total error count or the total number of received bits. They update when the defined intervals are initiated. The intervals equal 1 second when the AUTOUPDATE (b0, T1/

J1-000H) is set in the corresponding framer. They can also be updated by writing to any of the PD[31:0] (b7~0, T1/J1-06CH & b7~0, T1/J1-06DH & b7~0, T1/J1-06EH & b7~0, T1/J1-06FH), or to the T1/J1 Chip ID / Global PMON Update register (T1/J1-00CH). The update will be indicated by the XFERI (b1, T1/J1-061H). If they are not read in the defined intervals, the PD[31:0] (b7~0, T1/J1-06CH & b7~0, T1/J1-06DH & b7~0, T1/J1-06EH & b7~0

3 kinds of interrupts can be generated by this block:

- hit errors:
- synchronization status change (indicated in the SYNCI [b3, T1/J1-061H]);
- the PD[31:0] (b7~0, T1/J1-06CH & b7~0, T1/J1-06DH & b7~0, T1/J1-06EH & b7~0, T1/J1-06FH) are updated.

When the interrupts are enabled by the BEE (b6, T1/J1-061H) SYNCE (b7, T1/J1-061H) and XFERE (b5, T1/J1-061H) respectively, the  $\overline{\text{INT}}$  pin is asserted.

# 3.13 TRANSMIT SYSTEM INTERFACE (TRSI)

The Transmit System Interface determines how to input the data to the chip. The input data to the eight framers can be aligned with each other or inputted independently. The timing clocks and framing pulses can be provided by the system back-plane common to eight framers or provided for eight framers individually. The Transmit System Interface supports various configurations to meet various requirements in different applications.

## 3.13.1 E1 MODE

In E1 mode, the Transmit System Interface can be set in Non-multiplexed Mode or Multiplexed Mode. In Non-multiplexed Mode, the TSDn pin is used to input the data to each framer at the bit rate of 2.048 Mb/s. While in the Multiplexed Mode, the data input to the eight framers are byte interleaved from two high speed data streams and input on the MTSD1 and MTSD2 pins at the bit rate of 8.192 Mb/s.

In the Non-multiplexed Mode, if the timing signal for clocking data on TSDn pin is provided by the system side and shared by all eight framers, the Transmit System Interface should be set in Transmit Clock Slave mode. If the timing signal for clocking data on each TSDn pin is provided from each line side (processed timing signal), the Transmit System Interface should be set in Transmit Clock Master mode.

In the Non-multiplexed Mode, if there is a common framing pulse provided by the system side for the eight framers, the Transmit System Interface should be set in Transmit Clock Slave mode. If there is no common framing pulse, the Transmit System Interface should be set in Transmit Clock Master mode.

In the Transmit Clock Slave mode, if the multi-function pin TSFSn/TSSIGn is used to output the framing indication pulse, the Transmit System Interface is in Transmit Clock Slave TSFS Enable mode. If the TSFSn/TSSIGn is used to input the signaling bits to be inserted, the Transmit System Interface is in Transmit Clock Slave External Signaling mode.

In the Transmit Clock Master mode, the multi-function pin TSFSn/TSSIGn is used as TSFSn to input the framing indication pulse.

Table - 26 summarizes the transmit system interface in different operation modes. To set the transmit system interface of each framer into various operation modes, the registers must be configured as Table - 27.

Table - 26. E1 Mode Transmit System Interface in Different Operation Modes

Operation Mode		Data Pin	Clock Pin	Framing Pin	Signaling Pin	Reference Clock	
Non-	Clock Slave	TSFS Enable	TSDn	TSCCKB	TSCFS & TSFSn	No	TSCCKA
Multiplexed	Mode	External Signaling	TSDn	TSCCKB	TSCFS	TSSIGn	TSCCKA
Mode	ode Clock Master Mode		TSDn	LTCKn	TSFSn	No	TSCCKA & TSCCKB
Multiplexed Mode		MTSD	MTSCCKB	MTSCFS	MTSSIG	TSCCKA	

Table - 27. Operation Mode Selection in E1 Transmit Path

RATE[1:0] (b1~0, E1-018H)	TSCKSLV (b5, E1-018H)	TSSIG_EN (b6, E1-003H)	Operation Mode
	1	0	Transmit Clock Slave TSFS Enable
01	l	1	Transmit Clock Slave External Signaling
	0	-	Transmit Clock Master
11(All the eight framers should be set)	1	1	Transmit Multiplexed

## 3.13.1.1 Transmit Clock Slave Mode

In the Transmit Clock Slave mode, the Transmit Side System Common Clock B (TSCCKB) is provided by the system side. It is used as a common timing clock for all eight framers. The speed of the TSCCKB can be selected by the CMS (b2, E1-018H) to be the same as the data to be transmitted (2.048MHz), or twice the data (4.096MHz). The CMS (b2, E1-018H) of the eight framers should be set to the same value. If the speed of the TSCCKB is twice the data to be transmitted, there will be two active edges in one bit time. In this case, the COFF (b4, E1-01CH) determines the active edge to sample the signal on the TSDn and TSSIGn pins and the active edge to update the pulse on the TSFSn pin; however, the pulse on the TSCFS is always sampled on its first active edge.

In the Transmit Clock Slave mode, the Transmit Side System Common Clock A (TSCCKA) is provided by the system side. It is used as one of the reference clocks for the transmit jitter attenuator DPLL for all eight framers (refer to the Transmit Clock for details).

In the Transmit Clock Slave mode, the Transmit Side System Common Frame Pulse (TSCFS) is used as a common framing signal to align the data streams for the eight framers. The TSCFS is asserted on each

Basic Frame or Multi-Frame indicated by the FPTYP (b1, E1-019H). The valid polarity is congifured by the FPINV (b3, E1-019H).

In the Transmit Clock Slave mode, the bit rate on the TSDn pin is 2.048Mb/s.

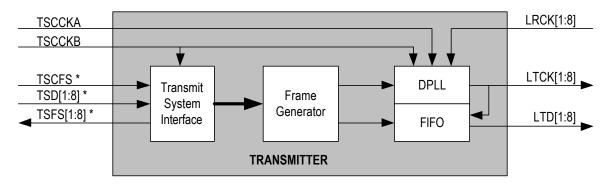
The Transmit Clock Slave Mode includes two sub-modes: Transmit Clock Slave TSFS Enable Mode and Transmit Clock Slave External Signaling Mode.

# 3.13.1.1.1 Transmit Clock Slave TSFS Enable Mode

In this mode (refer to Figure - 35), the data on the system interface are clocked by the TSCCKB. The active edge of the TSCCKB used to sample the pulse on the TSCFS and the data on the TSDn and TSFSn are determined by the following bits in the registers (refer to Table - 28).

Figure - 36 & 37 show the functional timing examples. Bit 1 of each timeslot is the first bit to be transmitted.

Besides all the common functions described in the Transmit Clock Slave mode, the special feature in this mode is that the multi-functional pin TSFSn/TSSIGn is used as TSFSn to output a framing pulse to indicate the first bit of each Basic Frame.



Note: \* TSCFS, TSD, TSFS are timed to TSCCKB

Figure - 35. Transmit Clock Slave TSFS Enable Mode

The CMS (b2, E1-018H) is logic 0, i.e., the backplane clock rate is 2.048Mbit/s.

The DE (b4, E1-018H) is logic 0 and the FE(b3, E1-018H) is logic 0.

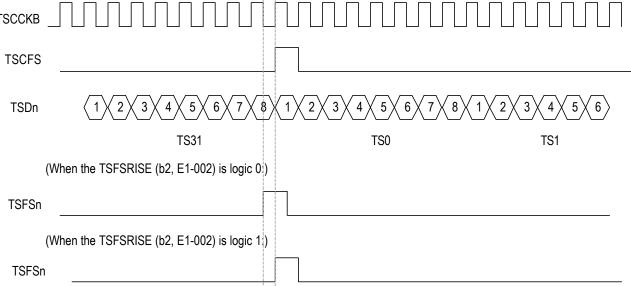


Figure - 36. E1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 1

The CMS (b2, E1-018H) is logic 1, i.e., the backplane clock rate is 4.096Mbit/s. The FE(b3, E1-018H) is logic 0 and the DE (b4, E1-018H) is logic 1.

The COFF (b4, E1-01CH) is in its default value.

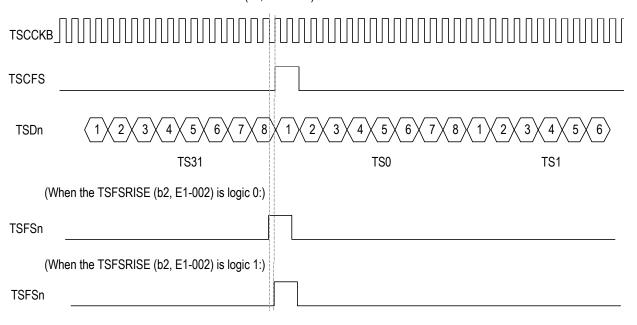


Figure - 37. E1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 2

Table - 28. Active Edge Selection of TSCCKB (in E1 Transmit Clock Slave TSFS Enable Mode)

	the Bit Determining the Active Edge of the TSCCKB
TSCFS	FE (b3, E1-018H)
TSDn	DE (b4, E1-018H)
TSFSn	TSFSRISE (b2, E1-002H)

**Note:** If the FE is not equal to the DE, the active edge decided by the FE is one clock edge before the active edge decided by the DE.

The FE (b3, E1-018H) of the eight framers should be set to the same value to ensure the TSCFS for the eight framers is sampled on the same active edge.

Table - 29. Active Edge Selection of TSCCKB (in E1 Transmit Clock Slave External Signaling Mode)

	the Bit Determining the Active Edge of the TSCCKB
TSCFS	FE (b3, E1-018H)
TSDn	DE /b/ E1 010U\
TSSIGn	DE (b4, E1-018H)

**Note:** If the FE is not equal to the DE, the active edge decided by the FE is one clock edge before the active edge decided by the DE.

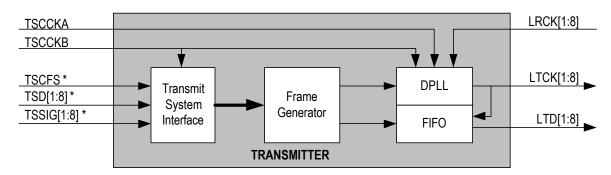
The FE (b3, E1-018H) of the eight framers should be set to the same value to ensure the TSCFS for the eight framers is sampled on the same active edge.

# 3.13.1.1.2 Transmit Clock Slave External Signaling Mode

In this mode (refer to Figure - 38), the data on the system interface are clocked by the TSCCKB. The active edge of the TSCCKB used to sample the pulse on the TSCFS and the data on the TSDn and TSSIGn is determined by the following bits in the registers (refer to Table - 29).

Figure - 39 & 40 show the functional timing examples. Bit 1 of each timeslot is the first bit to be transmitted.

Besides all the common functions described in the Transmit Clock Slave mode, the special feature in this mode is that the multi-functional pin TSFSn/TSSIGn is used as TSSIGn to input the signaling. The signaling on the TSSIGn pin may replace the data on TS16 when the CCS is disabled and the SIGSRC (b4, E1-TPLC-indirect registers -61~7FH) in the TPLC block is logic 0.



Note: \* TSCFS, TSD, TSSIG are timed to TSCCKB

Figure - 38. Transmit Clock Slave External Signaling Mode

The CMS (b2, E1-018H) is logic 0, i.e., the bankplane clock rate is 2.048Mbit/s. The DE (b4, E1-018H) is logic 0 and the FE(b3, E1-018H) is logic 1.

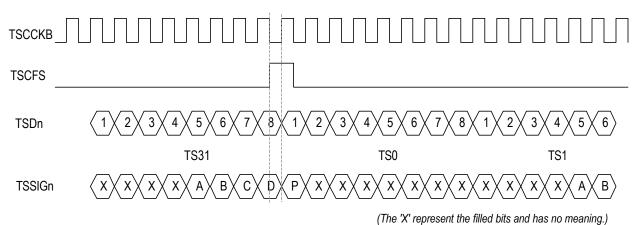


Figure - 39. E1 Transmit Clock Slave External Signaling Mode - Functional Timing Example 1

The CMS (b2, E1-018H) is logic 1, i.e., the bankplane clock rate is 4.096Mbit/s. The FE(b3, E1-018H) is logic 1 and the DE (b4, E1-018H) is logic 1.

The COFF (b4, E1-01CH) is in its default value.

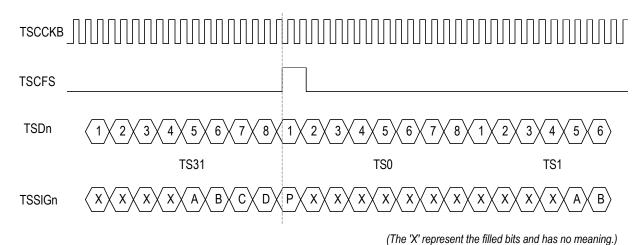


Figure - 40. E1 Transmit Clock Slave External Signaling Mode - Functional Timing Example 2

## 3.13.1.2 Transmit Clock Master Mode

In the Transmit Clock Master mode (refer to Figure - 41), the Transmit Side System Common Clock A (TSCCKA) and Transmit Side System Common Clock B (TSCCKB) provided by the system side are used as one of the reference clocks for the transmit jitter attenuator DPLL for all eight framers (refer to the Transmit Clock for details).

In the Transmit Clock Master mode, the multi-functional pin TSFSn/TSSIGn is used as TSFSn to output a framing pulse to indicate the first bit of each Basic Frame.

In the Transmit Clock Master mode, the bit rate on the TSDn pin is 2.048Mb/s.

In the Transmit Clock Master mode, each framer uses its own processed clock signal on LTCKn pin to sample/update the data on the system interface. The active edge of the LTCKn to sample the data on the TSDn pin is determinded by the DE (b4, E1-018H). The active edge of the LTCKn to update the pulse on the TSFSn pin is determinded by the TSFSRISE (b2, E1-002H).

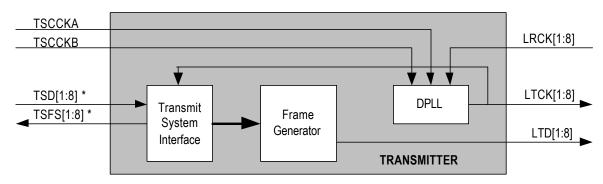
Figure - 42 shows the functional timing examples. Bit 1 of each

timeslot is the first bit to be transmitted.

# 3.13.1.3 Transmit Multiplexed Mode

In this mode (refer to Figure - 43), two multiplexed buses are used to input the data to all eight framers. Selected by the MTBS (b4, E1-003H) in each framer, the data on one of the two multiplexed buses is byte-interleaved input to up to four framers. When each group of four framers are selected, the input sequence of the data on the multiplexed bus is arranged by setting the timeslot offset TSOFF[6:0] (b6~0, E1-01BH). The data to different framers from one multiplexed bus must be shifted to a different timeslot offset to avoid data mixing. Then the data on the multiplexed bus will be input to each of the four selected framers with a byte-interleaved manner.

In the Transmit Multiplexed mode, the data on the system interface are clocked by the MTSCCKB. The active edge of the MTSCCKB to sample the data on the MTSCFS, MTSD and MTSSIG is determined by the following bits in the registers (refer to Table - 30).



Note: \* TSD, TSFS are timed to LTCK

Figure - 41. Transmit Clock Master Mode

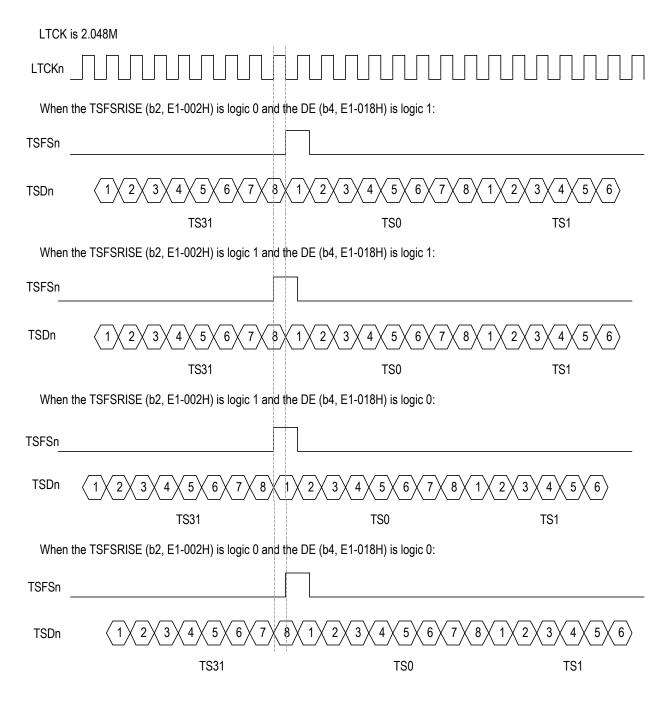


Figure - 42. E1 Transmit Clock Master Mode - Functional Timing Example

Table - 30. Active Edge Selection of MTSCCKB (in E1 Transmit Multiplexed Mode)

	the Bit Determining the Active Edge of the MTSCCKB
MTSCFS	FE (b3, E1-018H)
MTSD	DE /b4 E4 040U\
MTSSIG	DE (b4, E1-018H)

**Note:** If the FE is not equal to the DE, the active edge decided by the FE is one clock edge before the active edge decided by the DE.

The FE and the DE of the eight framers should be set to the same value respectively.

In the Transmit Multiplexed mode, the Multiplexed Transmit Side System Common Clock B (MTSCCKB) is provided by the system side. It is used as a common timing clock for all eight framers. The speed of the MTSCCKB can be selected by the CMS (b2, E1-018H) to be the same as the data to be transmitted (8.192MHz), or double the data (16.384MHz). If the speed of the MTSCCKB is double the data to be transmitted, there will be two active edges in one bit duration. In this case, the COFF (b4, E1-01CH) determines the active edge to sample the signal on the MTSD and MTSSIG pins and the active edge to update the pulse on the MTSFS pin; however, the pulse on the MTSCFS is always sampled on its first active edge. However, if the CMS (b2, E1-018H) or the COFF (b4, E1-01CH) of any of the eight framers is configured as logic 1, all the others are taken as logic 1. That is, the CMS (b2, E1-018H) and the COFF (b4, E1-01CH) of the eight framers should be configured to the same value in the Transmit Multiplexed mode.

In the Transmit Multiplexed mode, the Transmit Side System Common Clock A (TSCCKA) is provided by the system side. It is used as one of the reference clocks for the transmit jitter attenuator DPLL for all eight framers (refer to the Transmit Clock for details).

In the Transmit Multiplexed mode, the Multiplexed Transmit Side System Common Frame Pulse (MTSCFS) is used as a common framing signal to align data streams on the two multiplexed buses. The MTSCFS is asserted on each Basic Frame of the selected first framer. The valid polarity of the MTSCFS is congifured by the FPINV (b3, E1-019H). The FPINV (b3, E1-019H) of the eight framers should be set to the same value.

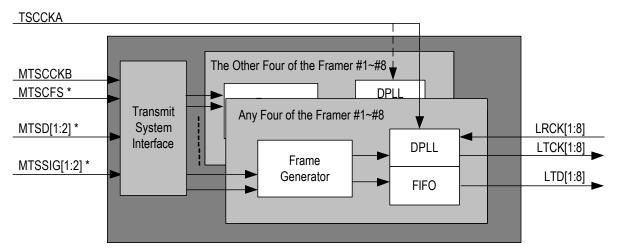
In the Transmit Multiplexed mode, the bit rate on the MTSD pin is 8.192Mb/s.

In the Transmit Multiplexed mode, the MTSSIG inputs the signaling bits to be inserted. The signaling bits are timeslot aligned with the data input from the MTSD. The signaling bits may replace the data on TS16 when the CCS is disabled and the SIGSRC (b4, E1-TPLC-indirect registers - 61~7FH) in the TPLC block is logic 0.

Figure - 44 & 45 show the functional timing examples. Bit 1 of each timeslot is the first bit to be transmitted.

# 3.13.1.4 Parity Check

In all the above four modes, parity check is calculated over the bits in the previous Basic Frame and the result is inserted into the first bit (MSB) of TS0 on the TSDn/MTSD pin. The even parity or odd parity is selected by the TPTYP (b7, E1-01AH) and whether the first bit of TS0 is calculated or not is determined by the PTY\_EXTD (b3, E1-01AH). The parity error event will be captured by the TDI (b5, E1-01AH). The parity error will cause an interrupt on the  $\overline{\text{INT}}$  pin if the TPTYE (b6, E1-01AH) is enabled.



Note: \* MTSCFS, MTSD, MTSSIG are timed to MTSCCKB

Figure - 43. Transmit Multiplexed Mode

The CMS (b2, E1-018H) is logic 0, i.e., the bankplane clock rate is 8.192Mbit/s.

The FE (b3, E1-018H) is logic 0 and the DE (b4, E1-018) is logic 0.

In this example, Framer1 to Framer4 are supposed to be demultiplexed from one multiplexed bus.

The TSOFF[6:0] of Framer1 are set to 7\*b0000010, the TSOFF[6:0] of Framer2 are set to 7\*b0000001, the TSOFF[6:0] of Framer4 are set to 7\*b0000011, the CHI and the BOFF[2:0] of the four Framers are set to logic 0:

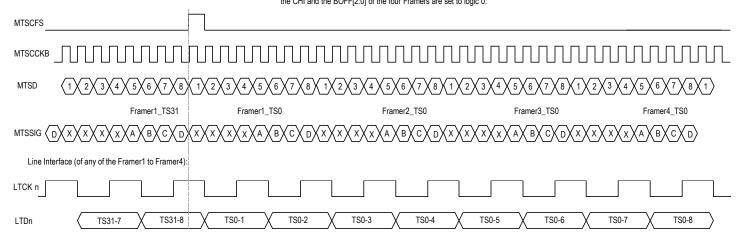


Figure - 44. E1 Transmit Multiplexed Mode - Functional Timing Example 1

The CMS (b2, E1-018H) is logic 1, i.e., the bankplane clock rate is 16.384Mbit/s.

The FE (b3, E1-018H) is logic 1 and the DE (b4, E1-018) is logic 0. The COFF (b4, E1-01CH) is in its default value. In this example, Framer1 to Framer4 are supposed to be demultiplexed from one multiplexed bus.

The TSOFF[6:0] of Framer1 are set to 7'b0000000, the TSOFF[6:0] of Framer2 are set to 7'b0000001, the TSOFF[6:0] of Framer4 are set to 7'b0000001, the TSOFF[6:0] of Framer4 are set to 7'b0000011, the CHI and the BOFF[2:0] of the four Framers are set to logic 0:

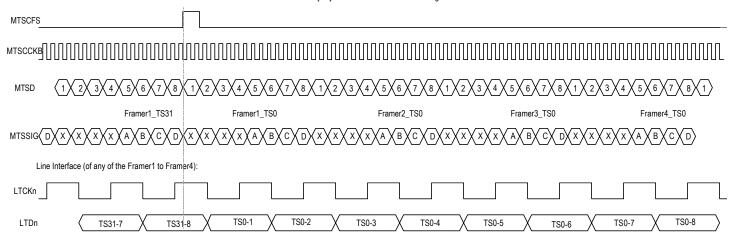


Figure - 45. E1 Transmit Multiplexed Mode - Functional Timing Example 2

## 3.13.1.5 Offset

In the Transmit Clock Slave mode and Transmit Multiplexed mode, timeslot offset is enabled by setting a non-zero value into the TSOFF[6:0] ( $b6\sim0$ , E1-01BH). The timeslot offset is between the TSCFS/MTSCFS and the start of the corresponding frame to be transmitted on the TSDn/MTSD. The timeslot offset can be set in both single clock mode (CMS [b2, E1-018H] = 0) and double clock mode (CMS [b2, E1-018H] = 1).

In all the above four modes, bit offset is enabled by setting a non-zero value into the BOFF[2:0] (b2~0, E1-01CH). In Transmit Clock Slave mode and Transmit Multiplexed mode, the bit offset is between the TSCFS/MTSCFS and the start of the corresponding frame to be transmitted on the TSDn/MTSD. The bit offset can be set in both single clock mode (CMS [b2, E1-018H] = 0) and double clock mode (CMS [b2, E1-018H] = 1). However, if the CHI (b3, E1-01CH) is logic 0, the bit offset

value equals the setting in the BOFF[2:0] (b2~0, E1-01CH). That is, '000' in the BOFF[2:0] (b2~0, E1-01CH) means no bit offset; '001' in the BOFF[2:0] (b2~0, E1-01CH) means one bit offset, and so on (refer to the examples in Figure - 46 and Figure - 47). If the CHI (b3, E1-01CH) is logic 1, the bit offset configured in the BOFF[2:0] (b2~0, E1-01CH) meets the Concentration Highway Interface (CHI) specification (refer to Table - 31 and Table - 32). The CER (clock edge receive) is counted from the active edge of the TSCFS/MTSCFS (refer to the examples in Figure - 48 and Figure - 49). When the bit offset is configured, the signal on the TSSIGn/MTSSIG or the pulse on the TSFSn is aligned to the RSDn/MRSD. In Transmit Clock Master mode, the bit offset is between the TSFsn and the start of the corresponding frame to be transmitted on the TSDn. In this case, the CHI specification is not supported and the bit offset value equals the setting in the BOFF[2:0] (b2~0, E1-01CH) (refer to the example in Figure - 50).

Table - 31. Transmit System Interface Bit Offset (CHI [b3, E1-01CH] = 1, CMS [b2, E1-018H] = 0)

FE	DE		BOFF[2:0] (b2~0, E1-01CH)							
(b3, E1-018H)	(b4, E1-018H)	000	001	010	011	100	101	110	111	
0	0	4	6	8	10	12	14	16	18	
0	1	3	5	7	9	11	13	15	17	CER
1	0	3	5	7	9	11	13	15	17	
1	1	4	6	8	10	12	14	16	18	

Table - 32. Transmit System Interface Bit Offset (CHI [b3, E1-01CH] = 1, CMS [b2, E1-018H] = 1)

FE	DE		BOFF[2:0] (b2~0, E1-01CH)							
(b3, E1-018H)	(b4, E1-018H)	000	001	010	011	100	101	110	111	
0	0	6	10	14	18	22	26	30	34	
0	1	7	11	15	19	23	27	31	35	CER
1	0	7	11	15	19	23	27	31	35	
1	1	6	10	14	18	22	26	30	34	

For example: in Transmit Clock Slave mode, CMS (b2, E1-018H) = 0, DE (b4, E1-018H) = 0, FE (b3, E1-018H) = 0:

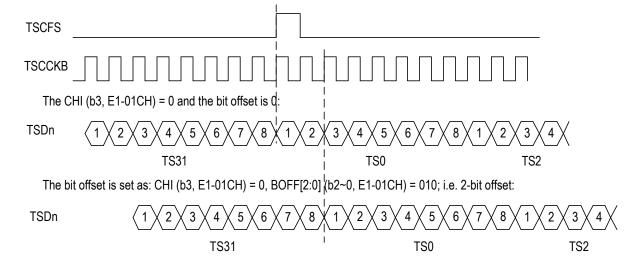


Figure - 46. Transmit Bit Offset in E1 Mode - 1

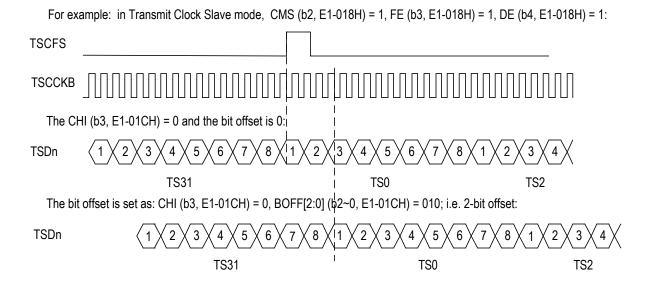


Figure - 47. Transmit Bit Offset in E1 Mode - 2

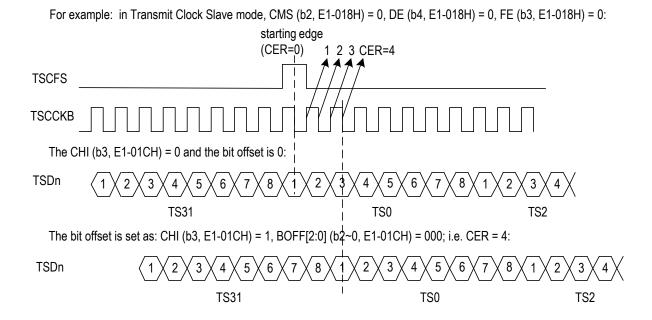


Figure - 48. Transmit Bit Offset in E1 Mode - 3

For example: in Transmit Clock Slave mode, CMS (b2, E1-018H) = 1, FE (b3, E1-018H) = 1, DE (b4, E1-018H) = 1:

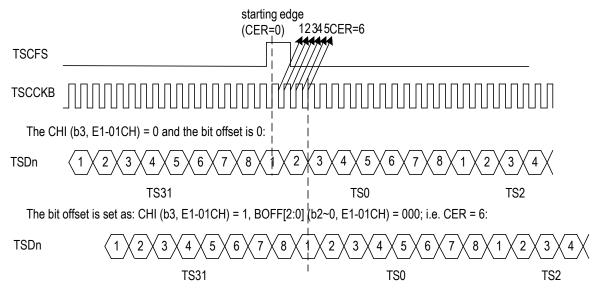


Figure - 49. Transmit Bit Offset in E1 Mode - 4

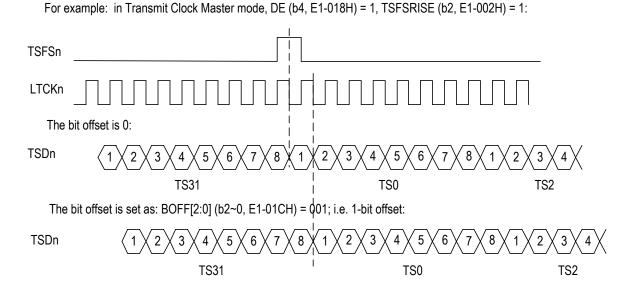


Figure - 50. Transmit Bit Offset in E1 Mode - 5

## 3.13.2 T1 / J1 MODE

In T1/J1 mode, the Transmit System Interface can be set in Non-multiplexed Mode or Multiplexed Mode. In Non-multiplexed Mode, the TSDn pin is used to input the data to each framer at the bit rate of 1.544 Mb/s or 2.048 Mb/s (T1/J1 mode E1 rate). While in the Multiplexed Mode, the data input to the eight framers are converted to 2.048 Mb/s format and byte interleaved from two high speed data streams and input on the MTSD1 and MTSD2 pins at the bit rate of 8.192 Mb/s.

In the Non-multiplexed Mode, if the timing signal for clocking data on the TSDn pin is provided by the system side and shared by all eight framers, the Transmit System Interface should be set in Transmit Clock Slave mode. If the timing signal for clocking data on each TSDn pin is provided from each line side (processed timing signal), the Transmit System Interface should be set in Transmit Clock Master mode.

In the Non-multiplexed Mode, if there is a common framing pulse provided by the system side for the eight framers, the Transmit System Interface should be set in Transmit Clock Slave mode. If there is not a common framing pulse, the Transmit System Interface should be set in Transmit Clock Master mode.

In the Transmit Clock Slave mode, if the multi-function pin TSFSn/TSSIGn is used to output the framing indication pulse, the Transmit System Interface is in Transmit Clock Slave TSFS Enable mode. If the TSFSn/TSSIGn is used to input the signaling bits to be inserted, the Transmit System Interface is in Transmit Clock Slave External Signaling mode.

The T1/J1 mode E1 rate, which means the system clock rate is 2.048 MHz in T1/J1 mode, can only be supported in the Transmit Clock Slave mode.

In the Transmit Clock Master mode, the multi-function pin TSFSn/TSSIGn is used as TSFSn to input the framing indication pulse.

Table - 33 summarizes the transmit system interface in different operation modes. To set the transmit system interface of each framer

into various operation modes, the registers must be configured as Table - 34

# 3.13.2.1 Transmit Clock Slave Mode

In the Transmit Clock Slave mode, the bit rate on the TSDn pin is 1.544 Mb/s. However, if the system clock rate is 2.048MHz, the data to be transmitted should be converted into the same rate as the line side, that is, to work in T1/J1 mode E1 rate. Thus the RATE[1:0] (b3~2, T1/J1-005H) should be set to '01'. The conversion complies as follows: The last bit of the Frame N of the system side is the F-bit of the Frame N in the device. Then one byte of the system side is discarded after the previous three bytes are converted into the device. This process repeats eight times and the conversion of one frame is completed. Then the process goes on (refer to Figure - 51).

In the Transmit Clock Slave mode, the Transmit Side System Common Clock B (TSCCKB) is provided by the system side. It is used as a common timing clock for all eight framers. The speed of the TSCCKB can be 1.544MHz or 2.048MHz. When it is 2.048MHz, the TSCCKB can be selected by the CMS (b5, T1/J1-015H) to be the same as the data (2.048Mb/S), or double the data (4.096Mb/s). The CMS (b5, T1/J1-015H) of the eight framers should be set to the same value. If the speed of the TSCCKB is double of the data, there will be two active edges in one bit duration. In this case, the COFF (b4, T1/J1-015H) determines the active edge to sample the signal on the TSDn and TSSIGn pins and the active edge to update the pulse on the TSFSn pin; however, the pulse on the TSCFS is always sampled on its first active edge.

In the Transmit Clock Slave mode, the Transmit Side System Common Clock A (TSCCKA) is provided by the system side. It is used as one of the reference clocks for the transmit jitter attenuator DPLL for all eight framers (refer to the Transmit Clock for details).

In the Transmit Clock Slave mode, the Transmit Side System Common Frame Pulse (TSCFS) is used as a common framing signal to align

Table - 33. T1/J1 Mode Transmit System Interface in Different Operation Modes

Operation Mode		Data Pin	Clock Pin	Framing Pin	Signaling Pin	Reference Clock	
Non-	Clock Slave TSFS Enable		TSDn	TSCCKB	TSCFS & TSFSn	No	TSCCKA
Multiplexed	Mode	External Signaling	TSDn	TSCCKB	TSCFS	TSSIGn	TSCCKA
Mode	Mode Clock Master Mode		TSDn	LTCKn	TSFSn	No	TSCCKA & TSCCKB
Multiplexed Mode		MTSD	MTSCCKB	MTSCFS	MTSSIG	TSCCKA	

Table - 34. Operation Mode Selection in T1/J1 Transmit Path

<b>RATE[1:0]</b> (b3~2, T1/J1-005H)	<b>EMODE[1:0]</b> (b7~6, T1/J1-005H)	Operation Mode
00 / 01 *	10	Transmit Clock Slave TSFS Enable
00 / 01	11	Transmit Clock Slave External Signaling
00	01	Transmit Clock Master
11 (in any of the eight framers)	11	Transmit Multiplexed

## Note:

<sup>\*</sup> When the RATE[1:0] are '00', the system clock rate is 1.544MHz.

When the RATE[1:0] are '01', the system clock rate is 2.048MHz, i.e., T1/J1 mode E1 rate.

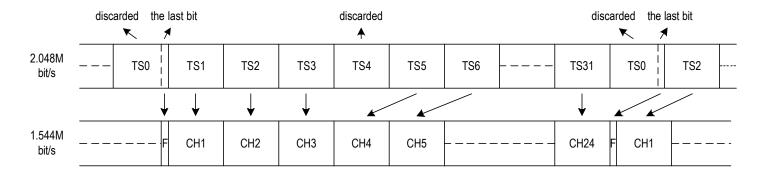


Figure - 51. E1 To T1/J1 Format Conversion

data streams for the eight framers. The TSCFS is asserted on the request of each F-bit or the first F-bit of every 12 SFs / every 24 ESFs, which is indicated by the TSCFSP (b1, T1/J1-005H). The valid polarity of the TSCFS is configured by the FPINV (b5, T1/J1-005H).

The Transmit Clock Slave Mode includes two sub-modes: Transmit Clock Slave TSFS Enable Mode and Transmit Clock Slave External Signaling Mode.

# 3.13.2.1.1 Transmit Clock Slave TSFS Enable Mode

In this mode (refer to Figure - 35), the data on the system interface are clocked by the TSCCKB. The active edge of the TSCCKB to sample the pulse on the TSCFS and the data on the TSDn and TSFSn is determined by the following bits in the registers (refer to Table - 35).

Figure - 52 to 54 show the functional timing examples. Bit 1 of each channel is the first bit to be transmitted.

Table - 35. Active Edge Selection of TSCCKB (in T1/J1 Transmit Clock Slave TSFS Enable Mode)

	the Bit Determining the Active Edge of the TSCCKB						
TSCFS	TCCCKDEALL (h2 T1/11 004H)						
TSD	TSCCKBFALL (b3, T1/J1-004H)						
TSFS	TSFSRISE (b5, T1/J1-004H)						

**Note:** The TSCCKBFALL (b3, T1/J1-004H) of the eight framers should be set to the same value to ensure the TSCFS for the eight framers is sampled on the same active edge.

Besides all the common functions described in the Transmit Clock Slave mode, the special feature in this mode is that the multi-functional pin TSFSn/TSSIGn is used as TSFSn to output a framing pulse to indicate every F-bit.

The CMS (b5, T1/J1-015H) is logic 0. The bankplane rate is 1.544Mbit/s. The TSCCKBFALL (b3, T1/J1-004H) is logic 1.

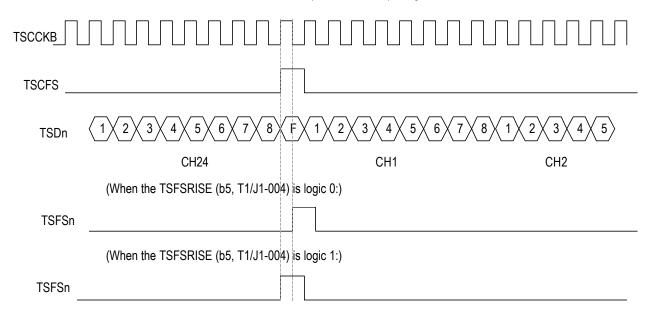


Figure - 52. T1/J1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 1

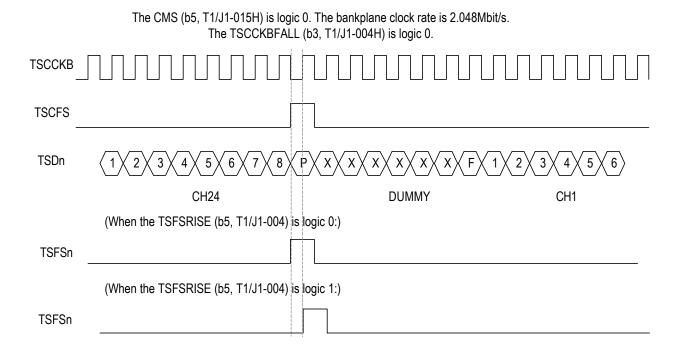
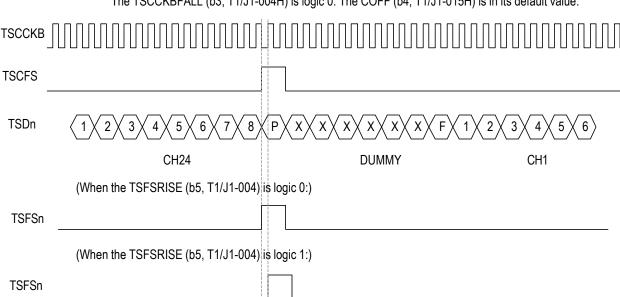


Figure - 53. T1/J1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 2



The CMS (b5, T1/J1-015H) is logic 1. The bankplane clock rate is 4.096Mbit/s. The TSCCKBFALL (b3, T1/J1-004H) is logic 0. The COFF (b4, T1/J1-015H) is in its default value.

Figure - 54. T1/J1 Transmit Clock Slave TSFS Enable Mode - Functional Timing Example 3

# 3.13.2.1.2 Transmit Clock Slave External Signaling Mode

In this mode (refer to Figure - 38), the data on the system interface are clocked by the TSCCKB. The active edge of the TSCCKB to sample the pulse on the TSCFS and the data on the TSDn and TSSIGn is determined by the TSCCKBFALL (b3, T1/J1-004H). The TSCCKBFALL (b3, T1/J1-004H) of the eight framers should be set to the same value to ensure the TSCFS for the eight framers is sampled on the same active edge.

Figure - 55 to 57 show the functional timing examples. Bit 1 of each channel is the first bit to be transmitted.

Besides all the common functions described in the Transmit Clock Slave mode, the special feature in this mode is that the multi-functional pin TSFSn/TSSIGn is used as TSSIGn to input the signaling. The signaling on the TSSIGn pin may be configured by the ABXXEN (b4, T1/J1-005H) to be valid only in the upper two-bit positions of the lower nibble of each channel (i.e. XXXXABXX) in T1 ESF mode.

The CMS (b5, T1/J1-015H) is logic 0. The bankplane rate is 1.544Mbit/s.

Figure - 55. T1/J1 Transmit Clock Slave External Signaling Mode - Functional Timing Example 1

The CMS (b5, T1/J1-015H) is logic 0. The bankplane rate is 2.048Mbit/s.

Figure - 56. T1/J1 Transmit Clock Slave External Signaling Mode - Functional Timing Example 2

The CMS (b5, T1/J1-015H) is logic 1. The bankplane clock rate is 4.096Mbit/s. The TSCCKBFALL (b3, T1/J1-004H) is logic 1.

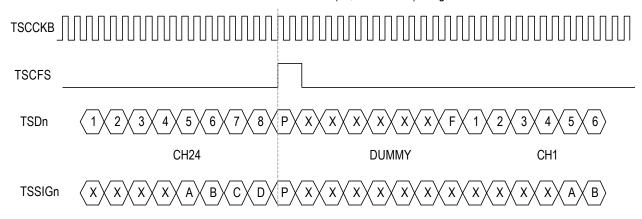


Figure - 57. T1/J1 Transmit Clock Slave External Signaling Mode - Functional Timing Example 3

# 3.13.2.2 Transmit Clock Master Mode

In the Transmit Clock Master mode (refer to Figure - 41), the Transmit Side System Common Clock A (TSCCKA) and Transmit Side System Common Clock B (TSCCKB) provided by the system side are used as one of the reference clocks for the transmit jitter attenuator DPLL for all eight framers (refer to the Transmit Clock for details).

In the Transmit Clock Master mode, the multi-functional pin TSFSn/TSSIGn is used as TSFSn to output a framing pulse to indicate every F-bit.

In the Transmit Clock Master mode, the bit rate on the TSDn pin is 1.544Mb/s.

In the Transmit Clock Master mode, each framer uses its own processed clock signal on LTCKn pin to sample/update the data on the system interface. The active edge of the LTCKn to sample the data on the TSDn pin is determinded by the TSDFALL (b1, T1/J1-004H). The active edge of the LTCKn to update the pulse on the TSFSn pin is determinded by the TSFSRISE (b5, T1/J1-004H).

Figure - 58 shows the functional timing examples. Bit 1 of each channel is the first bit to be transmitted.

# 3.13.2.3 Transmit Multiplexed Mode

In this mode (refer to Figure - 43), two multiplexed buses are used to input the data to all eight framers. Selected by the MTBS (b6, T1/J1-015H) in each framer, the data on one of the two multiplexed buses is byte-interleaved input to up to four framers. When each four framers are selected, the input sequence of the data on the multiplexed bus is arranged by setting the channel offset TSOFF[6:0] (b6~0, T1/J1-014H). The data for a different framer from one multiplexed bus must be shifted by a different channel offset to avoid data mixing. Then the data on the multiplexed bus will be input to each of the four selected framers with a byte-interleaved manner.

In the Transmit Multiplexed mode, the data on the system interface are clocked by the MTSCCKB. The active edge of the MTSCCKB to sample the data on the MTSCFS, MTSD and MTSSIG is determined by the TSCCKBFALL (b3, T1/J1-004H). The TSCCKBFALL (b3, T1/J1-004H) of the eight framers should be set to the same value.

In the Transmit Multiplexed mode, the Multiplexed Transmit Side Sys-

tem Common Clock B (MTSCCKB) is provided by the system side. It is used as a common timing clock for all eight framers. The speed of the MTSCCKB can be selected by the CMS (b5, T1/J1-015H) to be the same as the data to be transmitted (8.192MHz), or double the data (16.384MHz). If the speed of the MTSCCKB is double the data to be transmitted, there will be two active edges in one bit duration. In this case, the COFF (b4, T1/J1-015H) determines the active edge to sample the signal on the MTSD and MTSSIG pins and the active edge to update the pulse on the MTSFS pin; however, the pulse on the MTSCFS is always sampled on its first active edge. If the CMS (b5, T1/J1-015H) or the COFF (b4, T1/J1-015H) of any of the eight framers is configured as logic 1, all the others are taken as logic 1. That is, the CMS (b5, T1/J1-015H) and the COFF (b4, T1/J1-015H) of the eight framers should be configured to the same value in the Transmit Multiplexed mode.

In the Transmit Multiplexed mode, the Transmit Side System Common Clock A (TSCCKA) is provided by the system side. It is used as one of the reference clocks for the transmit jitter attenuator DPLL for all eight framers (refer to the Transmit Clock for details).

In the Transmit Multiplexed mode, the Multiplexed Transmit Side System Common Frame Pulse (MTSCFS) is used as a common framing signal to align data streams on the two multiplexed buses. The MTSCFS is asserted on the F-bit of the selected first framer. The valid polarity of the MTSCFS is congifured by the FPINV (b5, T1/J1-005H). The FPINV (b5, T1/J1-005H) of the eight framers should be the same value.

In the Transmit Multiplexed mode, the bit rate on the MTSD pin is 8.192Mb/s.

In the Transmit Multiplexed mode, the MTSSIG input the signaling bits to be inserted. The signaling bits are channel aligned with the data input from the MTSD. The signaling on the MTSSIG pin may be configured by the ABXXEN (b4, T1/J1-005H) to be valid only in the upper two-bit positions of the lower nibble of each channel (i.e. XXXXABXX) in T1 ESF mode.

Figure -  $59 \sim 60$  show the functional timing examples. Bit 1 of each channel is the first bit to be transmitted.

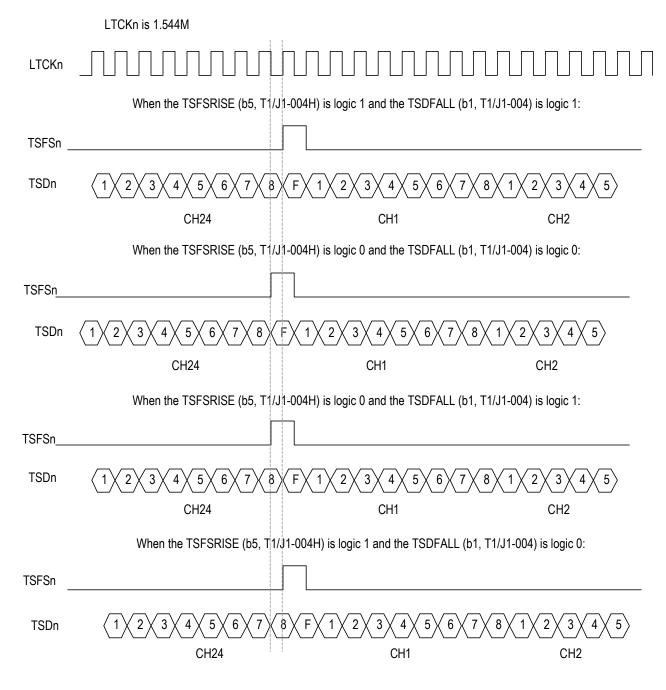


Figure - 58. T1/J1 Transmit Clock Master Mode - Functional Timing Example

The CMS (b5, T1/J1-015H) is logic 0, i.e., the bankplane rate is 8.192Mbit/s.

The TSCCKBFALL (b3, T1/J1-004H) is logic 1.

In this example, Framer1 to Framer4 are supposed to be demultiplexed from one multiplexed bus.

The TSOFF[6:0] of Framer1 are set to 7'b0000000, the TSOFF[6:0] of Framer2 are set to 7'b0000001, the TSOFF[6:0] of Framer4 are set to 7'b0000011, the BOFF[2:0] of the four Framers are set to logic 0:

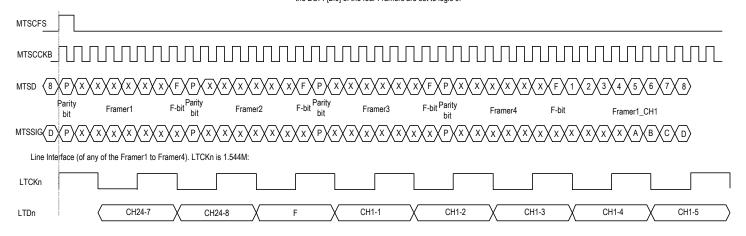


Figure - 59. T1/J1 Transmit Multiplexed Mode - Functional Timing Example 1

The CMS (b5, T1/J1-015H) is logic 1, i.e., the bankplane clock rate is 16.384Mbit/s.

The TSCCKBFALL (b3, T1/J1-004H) is logic 0.

In this example, Framer1 to Framer4 are supposed to be demultiplexed from one multiplexed bus.

The TSOFF[6:0] of Framer1 are set to 7'b0000000, the TSOFF[6:0] of Framer2 are set to 7'b0000001, the TSOFF[6:0] of Framer4 are set to 7'b00000011, the TSOFF[6:0] of Framer4 are set to 7'b0000011,

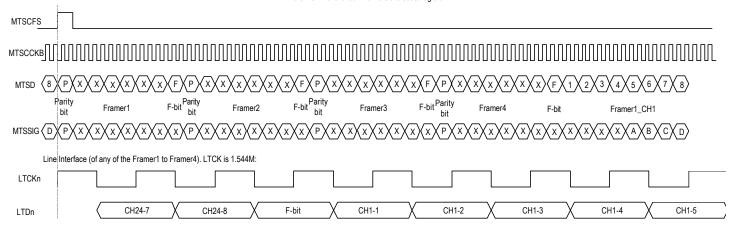


Figure - 60. T1/J1 Transmit Multiplexed Mode - Functional Timing Example 2

#### 3.13.2.4 Parity Check

In the above four modes, parity check is calculated over the bits in the previous frame and the result is input into the F-bit on the TSDn/MTSD and TSSIGn/MTSSIG pin. The even parity or odd parity is selected by the TPTYP (b7, T1/J1-002H) and whether the F-bit is calculated or not is determined by the PTY\_EXTD (b3, T1/J1-002H). The parity error event on the TSDn pin will be captured by the TSDI (b5, T1/J1-002H) and the parity error event on the TSSIGn pin will be captured by the TSSIGI (b4, T1/J1-002H). The TSDI (b5, T1/J1-002H) and TSSIGI (b4, T1/J1-002H) will be cleared after being read. The parity error will cause an interrupt on the  $\overline{\text{INT}}$  pin if the TPRTYE (b6, T1/J1-002H) is enabled.

## 3.13.2.5 Offset

When the system clock rate is 2.048MHz (in Transmit Clock Slave T1/J1 mode E1 rate mode) or 8.192MHz (in Transmit Multiplexed mode), the channel offset and/or bit offset between the TSCFS/MTSCFS and

the start of the corresponding frame on the TSDn/MTSD can be configured. The channel offset and bit offset can be set in both single clock mode (CMS [b5, T1/J1-015H] = 0) and double clock mode (CMS [b5, T1/J1-015H] = 1).

The channel offset is enabled by setting a non-zero value into the TSOFF[6:0] (b6~0, T1/J1-014H). The TSOFF[6:0] (b6~0, T1/J1-014H) give a binary representation.

The bit offset is enabled by setting a non-zero value into the BOFF[2:0] (b2~0, T1/J1-015H). The bit offset value equals the setting in the BOFF[2:0] (b2~0, T1/J1-015H). That is, '000' in the BOFF[2:0] (b2~0, T1/J1-015H) means no bit offset; '001' in the BOFF[2:0] (b2~0, T1/J1-015H) means one bit offset, and so on (refer to the examples in Figure - 61 and Figure - 62). When the bit offset is configured, the signal on the TSSIGn/MTSSIG or the pulse on the TSFSn is aligned to the RSDn/MRSD.

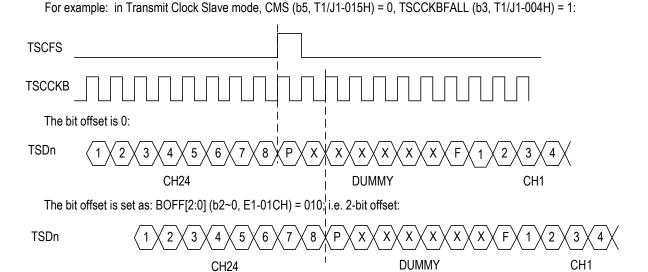


Figure - 61. Transmit Bit Offset in T1/J1 Mode - 1

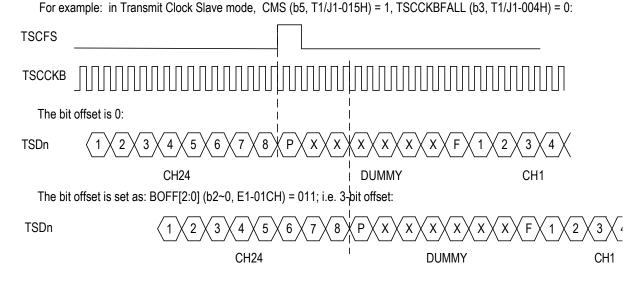


Figure - 62. Transmit Bit Offset in T1/J1 Mode - 2

# 3.14 TRANSMIT PAYLOAD CONTROL (TPLC)

Different test patterns can be inserted in the data to be transmitted or the data to be transmitted can be extracted to the PRBS Generator/Detector for test in this block. The Transmit Payload Control of each framer operates independently.

#### 3.14.1 E1 MODE

To enable the test for the data to be transmitted, the PCCE (b0, E1-060H) must be set to activate the setting in the indirect registers (from 20H to 7FH of TPLC indirect registers). The following methods can be used for test on a per-TS basis:

- Selected by the PRGDSEL[2:0] (b7~5, E1-00CH), the data to be transmitted on one of the eight framers will be extracted to the PRBS Generator/Detector when the RXPATGEN (b2, E1-00CH) is 1. The data can be extracted in framed or unframed mode. The selection is made by the UN\_DET (b0, E1-00CH). In unframed mode, all 32 timeslots are extracted and the per-timeslot configuration in the TEST (b3, E1-TPLC-indirect registers 20~3FH) is ignored. In framed mode, the data to be transmitted will only be extracted on the timeslot configured by the TEST (b3, E1-TPLC-indirect registers 20~3FH). Refer to the section of PRBS GENERATOR / DETECTOR (PRGD) for details.
- Enable the payload loopback by setting the LOOP (b2, E1-TPLC-indirect registers 20~3FH) (refer to Payload Loopback).
- Replace the data input from the TSDn/MTSD pin with the  $\mu$ -law or A-law milliwatt pattern (refer to Table 8 & Table 9) when the SUBS (b7, E1-TPLC-indirect registers 20~3FH), the DS0 (b4, E1-TPLC-indirect registers 20~3FH) and the DS1 (b5, E1-TPLC-indirect registers 20~3FH) are logic 1,1,1 or 1,1,0 respectively.
- Selected by the PRGDSEL[2:0] (b7~5, E1-00CH), the test pattern from the PRBS Generator/Detector will replace the data input from the TSDn/MTSD pin of one of the eight framers when the RXPATGEN (b2, E1-00CH) is 0. The test pattern can replace the data in framed or unframed mode. The selection is made by the UN\_GEN (b1, E1-00CH). In unframed mode, all the 32 timeslots are replaced and the per-timeslot configuration in the TEST (b3, E1-TPLC-indirect registers 20~3FH) is ignored. In framed mode, the received data will only be replaced on the timeslot configured by the TEST (b3, E1-TPLC-indirect registers 20~3FH). Refer to the section of PRBS GENERATOR / DETECTOR (PRGD) for details.
- Replace the data input from the TSDn/MTSD pin with the value in the IDLE[7:0] (b7~0, E1-TPLC-indirect registers 40~5FH) when the SUBS (b7, E1-TPLC-indirect registers 20~3FH) and the DS0 (b4, E1-TPLC-indirect registers 20~3FH) are logic 1,0.
- Invert the odd bits, even bits or all bits input from the TSDn/MTSD pin when the SUBS (b7, E1-TPLC-indirect registers 20~3FH), the DS0 (b4, E1-TPLC-indirect registers 20~3FH) and the DS1 (b5, E1-TPLC-indirect registers 20~3FH) are logic 0,0,1 or 0,1,0 or 0,1,1 respectively.

(The above methods are arranged from highest to lowest in priority.)

- Replace the signaling input from the TSSIGn pin with the value in the A, B, C, D (b3~0, E1-TPLC-indirect registers - 61~7FH) with the SIGSRC (b4, E1-TPLC indirect registers - 61~7FH) being logic 1 when the Channel Associated Signaling (CAS) is selected by the SIGEN (b6, E1-040H) & DLEN (b5, E1-040H).

Addressed by the A[6:0] (b6~0, E1-062H), the data read from or written into the indirect registers are in the D[7:0] (b7~0, E1-063H). The read or write operation is determined by the R/WB (b7, E1-062H). The indirect registers have a read/write cycle. Before the read/write operation is com-

pleted, the BUSY (b7, E1-061H) will be set. New operations on the indirect registers can only be implemented when the BUSY (b7, E1-061H) is cleared. The read/write cycle is 490ns.

#### 3.14.2 T1 / J1 MODE

To enable the test for the data to be transmitted, the PCCE (b0, T1/J1-030H) must be set to activate the setting in the indirect registers (from 01H to 48H of TPLC indirect registers). The following methods can be executed for test on a per-channel basis:

- Selected by the PRGDSEL[2:0] (b7~5, T1/J1-00FH), the data to be transmitted on one of the eight framers will be extracted to the PRBS Generator/Detector when the RXPATGEN (b2, T1/J1-00FH) is 1. The data can be extracted in framed or unframed mode. The selection is made by the UN\_DET (b0, T1/J1-00FH). In unframed mode, all 24 channels and the F-bit are extracted and the per-channel configuration in the TEST (b3, T1/J1-TPLC-indirect registers 01~18H) is ignored. In framed mode, the data to be transmitted will only be extracted on the channel specified by the TEST (b3, T1/J1-TPLC-indirect registers 01~18H). Fractional T1/J1 data can also be extracted in the specified channel when the Nx56k\_DET (b3, T1/J1-00FH) is set. Refer to the section of the PRBS GENERATOR / DETECTOR (PRGD) for details.
- Enable three types of Zero Code Suppression when the ZCS[1:0] (b1~0, T1/J1-TPLC-indirect registers 01~18H) is configured.
- Enable the payload loopback by setting the LOOP (b2, T1/J1-TPLC-indirect registers 01~18H) (refer to Payload Loopback).
- Replace the data input from the TSDn/MTSD pin with the milliwatt pattern when the DMW (b5, T1/J1-TPLC-indirect registers 01~18H) is logic 1. (The milliwatt is  $\mu$ -law. Refer to Table 9.)
- Selected by the PRGDSEL[2:0] (b7~5, T1/J1-00FH), the test pattern from the PRBS Generator/Detector will replace the data input from the TSDn pin of one of the eight framers when the RXPATGEN (b2, T1/J1-00FH) is 0. The test pattern can replace the data in framed of unframed mode. The selection is made by the UN\_GEN (b1, T1/J1-00FH). In unframed mode, all the 24 channels and the F-bit are replaced and the per-channel configuration in the TEST (b3, T1/J1-TPLC-indirect registers 01~18H) is ignored. In framed mode, the received data will only be replaced on the channel specified by the TEST (b3, T1/J1-TPLC-indirect registers 01~18H). Fractional T1/J1 signal can also be replaced in the specified channel when the Nx56k\_GEN (b4, T1/J1-00FH) is set. Refer to the section of PRBS GENERATOR / DETECTOR (PRGD) for details.
- Replace the data input from the TSDn/MTSD pin with the value in the IDLE[7:0] (b7~0, T1/J1-TPLC-indirect registers 19~30H) when the IDLE\_DS0 (b6, T1/J1-TPLC-indirect registers 01~18H) is set.
- Invert the most significant bit and/or the other bits in a channel input from the TSDn pin when the SIGNINV and the INVERT (b4 & b7, T1/J1-TPLC-indirect registers 01~18H) is set.

(The above methods are arranged from highest to lowest in priority.)

- Replace the signaling input from the TSSIGn pin with the value in the A, B, C, D (b3~0, T1/J1-TPLC-indirect registers - 31~48H) when the SIGC[1:0] (b7~6, T1/J1-TPLC-indirect registers - 31~48H) is configured.

The data of all channels can be selected by the GZCS[1:0] (b1 $\sim$ 0, T1/J1-044H) to be in GTE and Bell Zero Code Suppression when the bits in a channel are all zeros. The setting in the GZCS[1:0] (b1 $\sim$ 0, T1/J1-044H) are logically ORed with the setting in the ZCS[1:0] (b1 $\sim$ 0, T1/J1-TPLC-indirect registers - 01 $\sim$ 18H).

Addressed by the A[6:0] (b6~0, T1/J1-032H), the data read from or written into the indirect registers are in the D[7:0] (b7~0, T1/J1-033H).

The read or write operation is determined by the R/WB (b7, T1/J1-032H). Before the read/write operation is completed, the BUSY (b7, T1/J1-031H) will be set. New operations on the indirect registers can only be implemented when the BUSY (b7, T1/J1-031H) is cleared The read/write cycle is 650ns.

# 3.15 FRAME GENERATOR (FRMG)

The Frame Generator of each framer operates independently.

#### 3.15.1 E1 MODE

In E1 mode, the Frame Generator can generate Basic Frame, CRC-4 Multi-Frame and Channel Associated Signaling (CAS) Multi-Frame. The Frame Generator can also transmit alarm indication signal when special conditions occurs in the received data stream. International bits, National bits and Extra bits replacement and data invertion are all supported in the Frame Generator.

#### Generation

In E1 mode, the data to be transmitted can be formed to be Basic Frame, CRC-4 Multi-Frame and Signaling Multi-Frame.

The Basic Frame is generated when the FDIS (b3, E1-040H) is logic 0. The Basic Frame alignment sequence (FAS) - X0011011 will replace the data on the TS0 of each even frame and a logic 1 should be fixed in the 2nd bit of each odd frame.

The CRC-4 Multi-Frame is generated by setting the GENCRC (b4, E1-040H) when the INDIS (b1, E1-040H) is logic 0. The CRC-4 Multi-Frame alignment pattern - 001011 will replace the data on the international bits of the odd basic frames 1~11, and the calculated CRC bits will replace the data on the international bits of the even Basic Frames. The CRC bits are calculated every Sub Multi-Frame (SMF) and located in the next SMF. If the data input from the TSDn pin have already been in CRC Multi-Frame format, the CRC bits can be modified by setting the PATHCRC (b4, E1-002H) to transmit the CRC-4 transparently or modify the CRC-4 bits.

The Signaling Multi-Frame is generated by setting the SIGEN (b6, E1-040H) & the DLEN (b5, E1-040H) to logic 1 (CAS enable). The Signaling Multi-Frame alignment pattern - 0000 will replace the higher nibble (b1  $\sim$  b4) of the TS16 of Basic Frame 0, and the signaling source selected by the SIGSRC (b4, E1-TPLC-indirect registers - 61 $\sim$ 7FH) will replace the data on TS16 of Basic Frame 1 $\sim$ 15 (refer to Transmit Payload Control). When the Signaling Multi-Frame is not generated, setting a logic one in the MTRK (b7, E1-041H) will substitute the IDLE code set in the IDLE[7:0] (b7 $\sim$ 0, E1-TPLC-indirect registers - 40 $\sim$ 5FH) for all the data on the TS1 $\sim$ 31. When the Signaling Multi-Frame is generated, the

setting will only substitute the IDLE code for the TS1~15 and TS17~31. The TS16 is occupied by signaling. However, the MTRK (b7, E1-041H) takes effect only when the PCCE (b0, E1-060H) in the Transmit Payload Control is logic 1.

### **Alarm Indication**

When special conditions occurs in the received data stream, alarm indication will be transmitted automatically. The alarm indication can also be transmitted manually.

A logic 1 in the 3rd bit of NFAS (A bit) is the Remote Alarm Indication (RAI) signal. It is controlled by the REMAIS (b3, E1-041H), the AUTOYELLOW (b3, E1-000H) and the G706RAI (b0, E1-00EH) as illustrated in Table - 36.

When CRC-4 Multi-Frame is generated, the international bits of frame 13 & 15 (E1 & E2 bits) are used for FEBE indication only if the FEBEDIS (b2, E1-040H) is logic 0. When there are CRC calculated errors in SMF I or SMF II in the received data stream, a logic 0 will be automatically replaced in the E1 or the E2 bit for indication respectively. When the received data are out of CRC-4 Multi-Frame synchronization, the E1 and E2 bits can be forced to be logic 0 or logic 1, which is selected by the OOCMFE0 (b1, E1-00EH).

When Signaling Multi-Frame is generated, the 6th bit of TS16 of frame 0 (Y bit) is for Signaling Multi-frame Alarm Indication. A logic 1 in the Y bit means the Signaling Multi-frame Alarm. However, the value of the Y bit can be forced to be logic 0 or logic 1 by the MFAIS (b2, E1-041H).

## Control Over International / National / Extra Bits

After the Basic Frame is generated, the international bits (the first bit in TS0) can be replaced with the INDIS (b1, E1-040H) being logic 0.

The setting in the Si[1:0] (b7~6, E1-042H), the CRC-4 Multi-Frame and FEBE signal can all replace the international bits. Their priorities are controlled by the GENCRC (b4, E1-040H) and the FEBEDIS (b2, E1-040H) and illustrated in Table - 37.

When the setting in the SaX[1:4] (b3~0, E1-047H) is activated by the corresponding SaX\_EN[1:4] (b7~4, E1-047H), it will replace the data on the national bits whose position is selected by the SaSEL[2:0] (b7~5, E1-046H)

When Signaling Multi-Frame is generated, the extra bits (bits 4, 6 & 7 in TS16 of frame 0 of the Signaling Multi-Frame) can be replaced with the setting in the X[2:0] (b0~1 & b3, E1-043H) if the XDIS (b0, E1-040H) is logic 0.

Table - 36. Remote Alarm Indication

<b>REMAIS</b> (b3, E1-041H)	AUTOYELLOW (b3, E1-000H)	<b>G706RAI</b> (b0, E1-00EH)	Remote Alarm Indication Signal
1		· -	Manually force the remote alarm indication signal to be logic 1.
0	1	0	(per ETSI) The RAI is transmitted in any of the four conditions occurred in the received data stream: 1. out of Basic Frame; 2. during AISD; 3. in CRC-4 to non-CRC-4 interworking; 4. the offline searching is out of Basic Frame sync.
		1	(per Annex B of G.706) The RAI is transmitted in any of the two conditions occurred in the received data stream: 1. out of Basic Frame; 2. during AISD.
0	0	-	The RAI is not transmitted, that is, logic 0 is forced to transmit in its position.

Table - 37. Content in International Bits (when the INDIS [b1, E1-040H] is logic 0)

<b>GENCRC</b> (b4, E1-040H)	<b>FEBEDIS</b> (b2, E1-040H)	The Data on the International Bits
0	-	The international bits of the FAS frame represent the setting in the Si[1] (b7, E1-042H), while the international bits of the NFAS frame represent the setting in the Si[0] (b6, E1-042H).
1	0	The international bits of the FAS frame represent the calculated CRC-4 bits; the international bits of the former six NFAS frames represent the CRC-4 alignment sequence (001011). The other two international bits in frame 13 & 15 represent whether there are CRC-4 calculated errors in the received data stream (FEBE).
1	1	The international bits of the FAS frame represent the calculated CRC-4 bits; the international bits of the former six NFAS frames represent the CRC-4 alignment sequence (001011). The other two international bits in frame 13 & 15 represent the setting in the Si[1:0] (b7~6, E1-042H) respectively.

## **Diagnostics**

For diagnostic purposes, three kinds of data invertion can be executed:

- 1. When Basic Frame is generated, the FAS can be inverted from '0011011' to '1100100' by setting the FPATINV (b6, E1-041H);
- 2. When Basic Frame is generated, the 2nd bit of the NFAS can be inverted from '1' to '0' by setting the SPLRINV (b5, E1-041H);
- 3. When Signaling Multi-Frame is generated, the Signaling Multi-Frame alignment pattern can be inverted from '0000' to '1111' by setting the SPATINV (b4, E1-041H).

Of all the operations, transmitting all ones take the highest priority. All ones can be transmitted only in TS16 when the TS16AlS (b1, E1-041H) is set. All ones can also be transmitted on all the timeslots when the AlS (b0, E1-041H) is set.

A FIFO is employed in the Frame Generator to store the data stream to be transmitted. The FIFO can be initiated by setting the FRESH (b7, E1-040H).

## **Interrupt Summary**

The interrupt sources are summaried in Table - 38. When the conditions are met, the corresponding Interrupt Status bit will be logic 1. Then the interrupt will occur on the  $\overline{\text{INT}}$  pin if the Interrupt Enable bit is logic 1.

#### 3.15.2 T1 / J1 MODE

In T1/J1 mode, the data to be transmitted can be either the Super Frame (SF) or the Extended Super Frame (ESF) format. The selection is made by the ESF (b4, T1/J1-044H).

The SF/ESF is generated on the base of the UF (b6, T1/J1-046H)

and the FDIS (b3, T1/J1-006H) are logic 0, that is, the F-bit can be replaced with the Frame Alignment Pattern, DL and CRC-6 (the DL and CRC-6 bits only exist in the ESF format). Thus, the FAS can be replaced in its position when the FBITBYP (b2, T1/J1-006H) is logic 0. In SF format, the Frame Alignment Pattern is '10001101110X' and replaces the F-bit of each frame input from the TSDn pin (refer to Table - 3). In ESF format, the Frame Alignment Pattern is '001011' and replaces the F-bit in every 4th frame starting with Frame 4. The CRC-6 can replace the F-bit in every 4th frame starting with Frame 2 if the CRCBYP (b1, T1/J1-006H) is logic 0. The CRC-6 algorithm is selected between the T1 standard and the J1 standard by the J1\_CRC (b6, T1/J1-044H). The DL bits will replace the F-bit in every other frame starting with Frame 1 when the FDLBYP (b0, T1/J1-006H) is logic 0 (refer to Table - 4).

Before the data coming into the Frame Generator, if the SIGC[1:0] (b7~6, T1/J1-TPLC-indirect registers - 31~48H) select the signaling bit input from the TSDn pin to be replaced with the signaling input from the TSSIGn pin, the signaling bit of all channels can be replaced with the signaling of the 1st frame when the SIGAEN (b5, T1/J1-006H) is set. This configuration is to avoid the signaling change in the middle of a SF/ESF.

The data input from the TSDn pin will be replaced by the code set in the IDLE[7:0] (b7~0, T1/J1-TPLC-indirect registers - 19~30H) when the MTRK (b7, T1/J1-044H) is set. When the MTRK (b7, T1/J1-044H) is set, the signaling bits of all channels may also be replaced by the signaling input from the TSSIGn pin or the data set in the A, B, C, D (b3~0, T1/J1-TPLC-indirect registers - 31~48H) according to the setting in the SIGC[1:0] (b7~6, T1/J1-TPLC-indirect registers - 31~48H). The MTRK (b7, T1/J1-044H) takes effect only when the PCCE (b0, T1/J1-030H) in

Table - 38. Interrupt Summary

No.	Interrupt Sources	Indication Bits	Interrupt Mask Bits
1	The end of the first frame of a Signaling Multi-Frame is input to the Frame Generator when	SIGMFI	SIGMFE
	Signaling Multi-Frame is generated and coincides with the CRC Multi-Frame.	(b4, E1-045H)	(b4, E1-044H)
2	The end of the first frame of a CRC-4 Multi-Frame is input to the Frame Generator when	MFI	MFE
	CRC Multi-Frame is generated.	(b2, E1-045H)	(b4, E1-044H)
3	The end of the first frame of a CRC-4 Sub Multi-Frame is input to the Frame Generator	SMFI	SMFE
	when CRC Multi-Frame is generated.	(b1, E1-045H)	(b4, E1-044H)
4	The boundary of a FAS is input to the Frame Generator when Basic Frame is generated.	FASI	FASE
		(b3, E1-045H)	(b4, E1-044H)

the TPLC is logic 1.

Configured by the TXMFP (b1, T1/J1-00AH), the mimic pattern can be inserted in the 1st bit of each channel. The content of the mimic pattern is the same as the F-bit. The mimic pattern insertion is for diagnostic purposes.

The Yellow alarm signal can be inserted in the data stream to be transmitted when the XYEL (b2, T1/J1-045H) is enabled. The alarm signal pattern is selected between the T1 and J1 mode by the J1\_YEL (b5, T1/J1-044H). The pattern is:

- In T1 SF format: Transmit the logic 0 on the 2nd bit of each channel.
- In J1 SF format: Transmit the logic 1 on the 12th F-bit.
- In T1 ESF format: Transmit the 'FF00' on each DL of F-bit.
- In J1 ESF format: Transmit the 'FFFF' on each DL of F-bit.

The Yellow alarm signal can also be inserted automatically by setting the AUTOYELLOW (b3, T1/J1-000H) when Red alarm is declared in the received data stream.

In ESF format, if the Yellow alarm signal is stopped by setting the XYEL (b2, T1/J1-045H) to be logic 0, a Yellow alarm disabled pattern will be transmitted automatically. In T1 mode, the pattern is 'FFFF'. In J1 mode, the pattern is 'FF7E'. The disable pattern should be repeated 16 times before the BOC (refer to Bit-Oriented Message Transmitter) or the HDLC bits (refer to HDLC Transmitter) are inserted in the DL bit. The Yellow alarm takes the highest priority in these three kinds of insertion.

If there are no Yellow alarm signal, no BOC, no HDLC bits or noTPLC insertion in the DL of the F-bit, the DL position will be forced to transmit 'FFFF' in T1 mode or '7E7E' in J1 mode continuously.

A FIFO is employed in the Frame Generator to store the data stream to be transmitted. The FIFO can be initiated by setting the FRESH (b7, T1/J1-006H).

# 3.16 HDLC TRANSMITTER (THDLC)

The HDLC data insertion is performed in this block. The HDLC Transmitters #1, #2 and #3 in E1 mode or the HDLC Transmitter #1 and #2 in T1/J1 mode ESF format of each framer operate independently.

#### 3.16.1 E1 MODE

Three HDLC Transmitter blocks are provided for each framer to transmit a HDLC link. Before selecting the HDLC link, the TXCISEL (b3, E1-0AH) should be set to 1. Thus, the congifuration of Link Control and Bits Select registers (addressed from 028H to 02DH) is for THDLC. The THDLCSEL[1:0] (b5~4, E1-00AH) select one of the three HDLC controllers to be accessed by the microcontroller. The #2 and #3 blocks can also be disabled by setting the V52DIS (b3, E1-007H). The functionality of the HDLC link can be defined as the follows:

- 1. Set the DL\_EVEN (b7, E1-028H or b7, E1-02AH or b7, E1-02CH) and/or the DL\_ODD (b6, E1-028H or b6, E1-02AH or b6, E1-02CH) to select the even and/or odd frames (the even frames are FAS frames while the odd frames are NFAS frames);
- 2. Set the DL\_TS[4:0] (b4~0, E1-028H or b4~0, E1-02AH or b4~0, E1-02CH) to select the timeslot of the assigned frame or to select the TS16\_EN (b5, E1-028H) to define the TS16 of the assigned frame (this HDLC link can only be set in the #1 block and is enabled when the CCS is selected by the SIGEN [b6, E1-040H] and the DLEN [b5, E1-040H]);
- 3. Set the DL\_BIT[7:0] (b7~0, E1-029H or b7~0, E1-02BH or b7~0, E1-02DH) to select the bit of the assigned timeslot.

Thereafter, the HDLC packet will replace the data on the assigned data link. All the functions of the selected HDLC Transmitter block can

be realized only if the EN (b0, E1-050H) is set to logic 1; otherwise, all ones will be transmitted on the assigned data link.

A normal HDLC packet (refer to Figure - 3) is started with a 7E (Hex) flag, then the HDLC data are transmitted. Before closing, two bytes of CRC-CCITT frame check sequences (FCS) are added if the CRC (b1, E1-050H) is enabled. The HDLC packet is closed with another 7E flag. However, if the FLGSHARE (b7, E1-050H) is set, the closing flag of the current HDLC packet and the opening flag of the next HDLC packet are shared.

A FIFO buffer is used to store the HDLC data written to the TD[7:0] (b7~0, E1-055H). The UTHR[6:0] (b6~0, E1-051H) sets the upper threshold of the FIFO. When the data exceed the fill level, the data will be transmitted. The opening flag will be prepended before the data automatically. The transmission won't stop until the entire HDLC data are transmitted and the data in the FIFO are below the upper threshold. The end of the current entire HDLC frame is set by the EOM (b3, E1-050H). When it is set, the HDLC data should be transmitted even if they don't exceed the upper threshold of the FIFO. The FCS, if enabled, will be added before the closing flag automatically. Zero stuffing is automatically performed to the serial output data when there are five consecutive ones in the HDLC data or in the FCS. A 7F (Hex) abort sequence which deactivates the current HDLC packet can be inserted anytime the ABT (b2, E1-050H) is set. When the ABT (b2, E1-050H) is set, the current byte in the TD[7:0] (b7~0, E1-055H) is still transmitted, and then the FIFO is cleared and the 7F abort sequence is transmitted continuously. The low threshold of the FIFO can be set in the LINT[6:0] (b6~0, E1-052H), which should always be less than the value of the UTHR[6:0] (b6~0, E1-051H). The FIFO can be cleared anytime the FIFOCLR (b6, E1-050H) is set. Flags (7E) will consecutively be transmitted when there is no HDLC data to be transmitted if the data link is activated.

Four interrupt sources can be derived from this block.

- 1. When the data in the FIFO is empty or less than the setting in the LINT[6:0] (b6~0, E1-052H), the BLFILL (b5, E1-054H) will indicate. A transition from logic 0 to 1 on the BLFILL (b5, E1-054H) will cause a logic 1 in the LFILLI (b0, E1-054H). The interrupt on the INT pin will occur when the LFILLE (b0, E1-053H) is enabled;
- 2. When the data in the FIFO reach its maximum capacity 128 bytes, the FULL (b6, E1-054H) will be set for indication. A transition from logic 0 to 1 on the FULL (b6, E1-054H) will cause a logic 1 in the FULLI (b3, E1-054H). The interrupt on the INT pin will occur when the FULLE (b3, E1-053H) is enabled;
- 3. When the FIFO has already been filled with 128 bytes and new data are still written to it, the FIFO will overflow and the OVRI (b2, E1-054H) will be set for indication. The interrupt on the  $\overline{\text{INT}}$  pin will occur when the OVRE (b2, E1-053H) is enabled.
- 4. When the transmission is in process and it is out of data to be transmitted in the FIFO, the FIFO is underrun and the UDRI (b1, E1-054H) will be set for indication. The interrupt on the  $\overline{\text{INT}}$  pin will occur when the UDRE (b1, E1-053H) is enabled.

# 3.16.2 T1 / J1 MODE

In the SF format, there is no HDLC link.

In the ESF format, two HDLC Transmitter blocks (#1 and #2) are employed for each framer to transmit the HDLC link. Before selecting the HDLC link, the TXCISEL (b3, T1/J1-00DH) should be set to 1. Thus, the configuration of the Link Control and Bits Select registers (addressed from 070H to 071H) is for THDLC. Selected by the THDLCSEL[1:0]

(b5~4, T1/J1-00DH), registers in one of the two HDLC Transmitter blocks are accessable to the microprocessor. The #1 block transmits the HDLC link in the DL of F-bit (its position is shown in the Table - 4). The #2 block transmits the HDLC link in a channel and its position is selected as follows:

- 1. Set the DL2\_EVEN (b7, T1/J1-070H) and/or the DL2\_ODD (b6, T1/J1-070H) to select the even and/or odd frames;
- 2. Set the DL2\_TS[4:0] (b4~0, T1/J1-070H) to select the channel of the assigned frame;
- 3. Set the DL2\_BIT[7:0] (b7~0, T1/J1-071H) to select the bit of the assigned channel.

All the functions of the selected HDLC Transmitter block can be realized only if the EN (b0, T1/J1-034H) is enabled; otherwise, all ones will be transmitted on the assigned data link.

The structure of the HDLC packet (refer to Figure - 3) is the same as it is described in the E1 mode. When the FLGSHARE (b7, T1/J1-034H) is set, the closing flag of the current HDLC and the opening flag of the next HDLC is shared.

A FIFO buffer is used to store the HDLC data written to the TD[7:0] (b7~0, T1/J1-039H). The UTHR[6:0] (b6~0, T1/J1-035H) limit the upper threshold of the FIFO. When the data exceed the fill level, the data will be transmitted. The opening flag will be added before the data automatically. The transmission won't stop until an entire HDLC frame is transmitted and the data in the FIFO is below the upper threshold. The end of the current entire HDLC frame is indicated by the EOM (b3, T1/J1-034H). When it is set, the HDLC data should be transmitted even if it does not exceed the upper threshold of the FIFO. The FCS, if enabled by the CRC (b1, T1/J1-034H), will be added before the closing flag automatically. Zero stuffing is automatically performed to the serial output data when there are five consecutive ones in the HDLC data or in the FCS. A 7F abort sequence which deactivates the current HDLC packet can be inserted anytime the ABT (b2, T1/J1-034H) is set. When the ABT (b2, T1/J1-034H) is set, the current byte in the TD[7:0] (b7~0, T1/J1-039H) is still transmitted, and then the FIFO is cleared and the 7F abort sequence is transmitted continuously. The low threshold of the FIFO can be set in the LINT[6:0] (b6~0, T1/J1-036H), which should always be less than the value of the UTHR[6:0] (b6~0, T1/J1-035H). The FIFO can be cleared anytime the FIFOCLR (b6, T1/J1-034H) is set. Flags (7E) will consecutively be transmitted when there is no HDLC data to be transmitted during the data link activating.

Four interrupt sources can be derived from this block.

- 1. When the FIFO is empty or the data in the FIFO is less than the setting in the LINT[6:0] (b6~0, T1/J1-036H), the BLFILL (b5, T1/J1-038H) will be set for indication. A transition from logic 0 to 1 on the BLFILL (b5, T1/J1-038H) will cause a logic 1 in the LFILLI (b0, T1/J1-038H). The interrupt on the  $\overline{\text{INT}}$  pin will occur when the LFILLE (b0, T1/J1-037H) is enabled;
- 2. When the data in the FIFO reach its maximum capacity 128 bytes, the FULL (b6, T1/J1-038H) will be set for indication. A transition from logic 0 to 1 on the FULL (b6, T1/J1-038H) will cause a logic 1 in the FULLI (b3, T1/J1-038H). The interrupt on the  $\overline{\text{INT}}$  pin will occur when the FULLE (b3, T1/J1-037H) is enabled;
- 3. When the FIFO has been filled with 128 bytes already and new data are still written to it, the FIFO is will overflow and the OVRI (b2, T1/J1-038H) will be set for indication. The interrupt on the INT pin will occur when the OVRE (b2, T1/J1-037H) is enabled.
  - 4. When the transmission is in process and it is out of data to be

transmitted in the FIFO, the FIFO is underrun and the UDRI (b1, T1/J1-038H) will be set for indication. The interrupt on the  $\overline{\text{INT}}$  pin will occur when the UDRE (b1, T1/J1-037H) is enabled.

# 3.17 BIT-ORIENTED MESSAGE TRANSMITTER (TBOM) - T1 / J1 ONLY

The Bit Oriented Message (BOM) can only be transmitted in the ESF format in T1/J1 mode. The standard of the BOM is defined in the ANSI T1.403-1989. The Bit Oriented Message (BOM) of each framer operates independently.

The BOM pattern is '111111110XXXXXXX0' which occupies the DL of the F-bit in the ESF format (refer to Table - 4). The six 'X's represent the code that can be programmed in the BOC[5:0] (b5~0, T1/J1-05DH). When the BOC[5:0] (b5~0, T1/J1-05DH) are written with the bits other than the '111111', they will occupy the six 'X's' positions and the BOM will be transmitted.

If the BOM transmission is stopped by setting all ones in the BOC[5:0] (b5~0, T1/J1-05DH), a BOM disabled pattern will be transmitted automatically. In T1 mode, the pattern is 'FFFF'. In J1 mode, the pattern is 'FF7E'. The disable pattern should be repeated 16 times before the HDLC bits (refer to HDLC Transmitter) are inserted in the DL bit. The transmission of the BOM takes priority over any other substitutions of the DL bit except for the Yellow alarm signal.

# 3.18 INBAND LOOPBACK CODE GENERATOR (IBCG) - T1 / J1 ONLY

The Inband Loopback Code Generator can only transmit inband loopback code in a framed or unframed T1/J1 data stream. The Inband Loopback Code Generator of each framer operates independently.

The length and the content of the inband loopback code are programmed in the CL[1:0] (b1~0, T1/J1-046H) and the IBC[7:0] (b7~0, T1/J1-047H) respectively. The code can only be transmitted when the EN (b7, T1/J1-046H) is enabled. In framed mode, which is configured by the UF (b6, T1/J1-046H), the F-bit can be replaced by the Frame Alignment Pattern, DL and CRC-6 which are set in the Frame Generator block and the 24 channels are replaced with the inband loopback code. In unframed mode, which is configured by the UF (b6, T1/J1-046H), all 193 bits are replaced with the inband loopback code.

It is recommended that the setting of the EN (b7, T1/J1-046H) and the UF (b6, T1/J1-046H) should be the same.

# 3.19 JITTER ATTENUATOR (RJAT/TJAT)

The Jitter Attenuator of each framer operates independently

#### 3.19.1 E1 MODE

Two Jitter Attenuators are provided independently in the receive path and the transmit path.

The Jitter Attenuator integrates a FIFO and a DPLL. The smoothed clock output from the jitter attenuator is generated by adaptively dividing the 49.152MHz XCK according to the phase difference between the output smoothed clock and the input reference clock. The ratio between the frequency of the input reference clock and the frequency applied to the phase discriminator input is equal to the (N1 + 1) (the N1 is in b7~0, E1-021H for receive path and in b7~0, E1-025H for transmit path). The ratio between the frequency of the output smoothed clock and the frequency applied to the phase discriminator input is equal to the (N2 + 1) (the N2

is in b7~0, E1-022H for receive path and in b7~0, E1-026H for transmit path). The phase fluctuations of the input reference clock are attenuated by dividing the input reference clock and output smoothed clock by the (N1 + 1) and the (N2 + 1) respectively in the DPLL so that the frequency of the output smoothed clock is equal to the average frequency of the input reference clock. The phase fluctuations with a jitter frequency above 8.8Hz are attenuated by 6dB per octave when the N1 (b7~0, E1-021H for receive path and b7~0, E1-025H for transmit path) and the N2 (b7~0, E1-022H for receive path and b7~0, E1-026H for transmit path) are set to their default value. It will change when the N1 and the N2 are changed. Generally, when the N1 and the N2 increase, the curves of the Jitter Tolerance and Jitter Transfer in the graph will left-shift and when N1 and N2 decrease, they will right-shift. The phase fluctuations (wander) with frequency below 8.8Hz are tracked by the output smoothed clock. The output smoothed clock is used to clock the data out of the FIFO.

The FIFO is is 48 bits deep. If data is still written into the FIFO when the FIFO is already full, overflow will occur and the OVRI (b1, E1-020H for receive path and b1, E1-024H for transmit path) will indicate. If data is still read from the FIFO when the FIFO is already empty, under-run will occur and the UNDI (b0, E1-020H for receive path and b0, E1-024H for transmit path) will indicate. Thus, if the OVRE (b2, E1-023H for receive path and b2, E1-027H for transmit path) and the UNDE (b3, E1-023H for receive path and b3, E1-027H for transmit path) are set respectively, the interrupts on the  $\overline{\text{INT}}$  pin will occur. The jitter attenuation can be limited by setting the LIMIT (b0, E1-023H for receive path and b0, E1-027H for transmit path) to keep the FIFO 1UI away from being full or empty. Thus, the DPLL will track the jitter of the input reference clock by increasing or decreasing the frequency of the output smoothed clock to prevent the FIFO being empty or full. The FIFO can also selfcenter its read pointer by setting the CENT (b4, E1-023H for receive path and b4, E1-027H for transmit path). The FIFO can be set to be bypassed by the FIFOBYP (b7, E1-000H for receive path and b7, E1-002H for transmit path).

However, in Transmit Clock Master mode, the TJAT should be by-passed.

# **Jitter Characteristics**

Each Jitter Attenuator block provides excellent jitter tolerance and jitter attenuation while generating minimal residual jitter. It can accommodate up to 43Ulpp of input jitter at jitter frequencies above 9Hz. For jitter frequencies below 9Hz, which can be correctly called wander, the tolerance increases 20dB per decade. In most applications the each Jitter Attenuator block will limit jitter tolerance at lower jitter frequencies only. For high frequency jitter, above 10kHz for example, other factors such as clock and data recovery circuitry may limit jitter tolerance and must be considered. For low frequency wander, below 10Hz for example, other factors such as slip buffer hysteresis may limit wander tolerance and must be considered. The Jitter Attenuator blocks meet the low frequency jitter tolerance requirements ITU-T Recommendation G.823.

The Jitter Attenuator exhibits negligible jitter gain for jitter frequencies below 9Hz, and attenuates jitter at frequencies above 9Hz by 20dB per decade. In most applications the Jitter Attenuator blocks will determine jitter attenuation for higher jitter frequencies only. Wander, below 10Hz for example, will essentially be passed unattenuated through the Jitter Attenuator. Jitter, above 10Hz for example, will be attenuated as

specified, however, outgoing jitter may be dominated by the generated residual jitter in cases where incoming jitter is insignificant. This generated residual jitter is directly related to the use of 24X (49.152MHz) digital phase locked loop for transmit clock generation.

The Jitter Attenuator meets the jitter transfer requirements of ITU-T Recommendations G.737, G.738, G.739, and G.742.

#### Jitter Tolerance

Jitter tolerance is the maximum input phase jitter at a given jitter frequency that a device can accept without exceeding its linear operating range, or corrupting data. For the Jitter Attenuator, the input jitter tolerance is 43 Ulpp with no frequency offset. The frequency offset is the difference between the frequency of XCK divided by 24 and that of the input reference clock.

Refer to Figure - 63 for the Jitter Tolerance.

#### **Jitter Transfer**

The output jitter for jitter frequencies from 0 to 9 Hz is no more than 0.1dB greater than the input jitter. Jitter frequencies above 9 Hz are attenuated at a level of 6 dB per octave, as shown in Figure - 64.

## Frequency Range

In the non-attenuating mode, that is, when the FIFO is within one UI of overrunning or under running, the tracking range is 1.963 to 2.133 MHz. The guaranteed linear operating range is 2.048 MHz  $\pm$  1278 Hz with no jitter or XCK frequency offset.

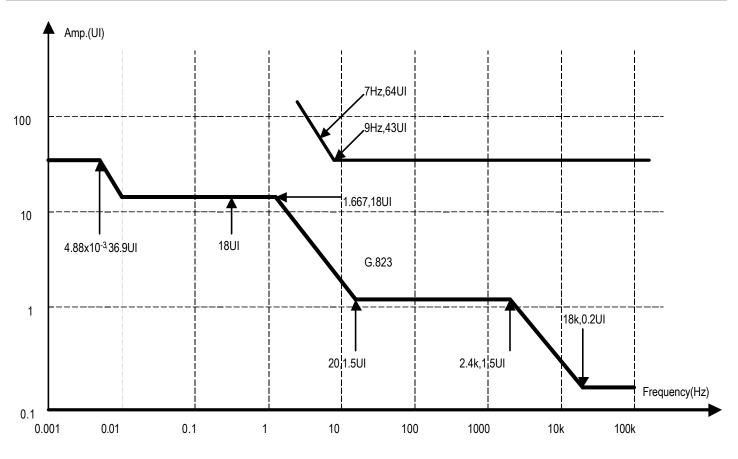


Figure - 63. E1 Mode Jitter Tolerance (N1 = N2 = 2fH)

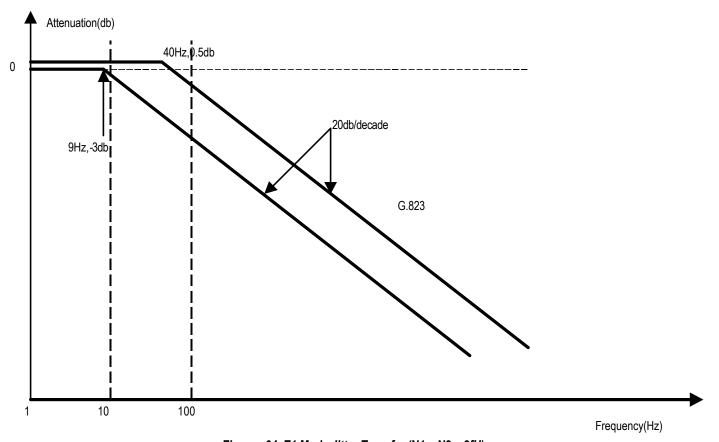


Figure - 64. E1 Mode Jitter Transfer (N1 = N2 = 2fH)

#### 3.19.2 T1 / J1 MODE

Two Jitter Attenuators are provided independly in the receive path and the transmit path.

The Jitter Attenuator integrates a FIFO and a DPLL. The smoothed clock output from the jitter attenuator is generated by adaptively dividing the 37.056MHz XCK according to the phase difference between the output smoothed clock and the input reference clock. The ratio between the frequency of the input reference clock and the frequency applied to the phase discriminator input is equal to the (N1 + 1) (the N1 is in b7~0, T1/ J1-011H for receive path and in b7~0, T1/J1-019H for transmit path). The ratio between the frequency of the output smoothed clock and the frequency applied to the phase discriminator input is equal to the (N2 + 1) (the N2 is in b7~0, T1/J1-012H for receive path and in b7~0, T1/J1-01AH for transmit path). The phase fluctuations of the input reference clock are attenuated by dividing the input reference clock and output smoothed clock by the (N1 + 1) and the (N2 + 1) respectively in the DPLL so that the frequency of the output smoothed clock is equal to the average frequency of the input reference clock. The phase fluctuations with a jitter frequency above 6.6Hz are attenuated by 6dB per octave when the N1 (b7~0, T1/J1-011H for receive path and b7~0, T1/J1-019H for transmit path) and the N2 (b7~0, T1/J1-012H for receive path and b7~0, T1/J1-01AH for transmit path) are in their default value. It will change when the N1 and the N2 are changed. Generally, when the N1 and the N2 increase, the curves of the Jitter Tolerance and Jitter Transfer in the graph will left-shift and When N1 and N2 decrease, they will right-shift. The phase fluctuations (wander) with frequency below 6.6Hz are tracked by the output smoothed clock. The output smoothed clock is used to clock the data out of the FIFO.

The FIFO is 48 bits deep. If data is still written into the FIFO when the FIFO is already full, overflow will occur and the OVRI (b1, T1/J1-010H for receive path and b1, T1/J1-018H for transmit path) will indicate. If data is still read from the FIFO when the FIFO is already empty, underrun will occur and the UNDI (b0, T1/J1-010H for receive path and b0, T1/J1-018H for transmit path) will indicate. Thus, if the OVRE (b2, T1/J1-013H for receive path and b2, T1/J1-01BH for transmit path) and the UNDE (b3, T1/J1-013H for receive path and b3, T1/J1-01BH for transmit path) are set respectively, the interrupts on the INT pin may occur. The jitter attenuation can be limited by setting the LIMIT (b0, T1/J1-013H for receive path and b0, T1/J1-01BH for transmit path) to keep the FIFO 1UI away from being full or empty,. Thus, the DPLL will track the jitter of the input reference clock by increasing or decreasing the frequency of the output smoothed clock to prevent the FIFO being empty or full. The FIFO can also self-center its read pointer by setting the CENT (b4, T1/J1-013H for receive path and b4, T1/J1-01BH for transmit path). The FIFO can be set to be bypassed by the FIFOBYP (b7, T1/J1-000H for receive path and b7, T1/J1-004H for transmit path).

However, in Transmit Clock Master mode, the TJAT should be by-passed.

## **Jitter Characteristics**

Each Jitter Attenuator block provides excellent jitter tolerance and jitter attenuation while generating minimal residual jitter. It can accommodate up to 45UI of input jitter at jitter frequencies above 12HZ. For jitter frequencies below 9 Hz, which can be correctly called wander, the tolerance increases 20dB per decade. In most applications the each Jitter Attenuator block will limit jitter tolerance at lower jitter frequencies

only. For high frequency jitter, above 10kHz for example, other factors such as clock and data recovery circuitry may limit jitter tolerance and must be considered. For low frequency wander, below 10Hz for example, other factors such as slip buffer hysteresis may limit wander tolerance and must be considered. The Jitter Attenuator blocks meet the low frequency jitter tolerance requirements AT&T TR 62411 for T1.

The Jitter Attenuator exhibits negligible jitter gain for jitter frequencies below 7Hz, and attenuates jitter at frequencies above 7Hz by 20 dB per decade. In most applications the Jitter Attenuator blocks will determine jitter attenuation for higher jitter frequencies only. Wander, below 10 Hz for example, will essentially be passed unattenuated through DJAT. Jitter, above 10 Hz for example, will be attenuated as specified, however, outgoing jitter may be dominated by the generated residual jitter in cases where incoming jitter is insignificant. This generated residual jitter is directly related to the use of 24X (37.056MHz) digital phase locked loop for transmit clock generation.

The Jitter Attenuator meets the jitter transfer requirements of AT&T TR 62411. The block allows to meet the implied jitter attenuation requirements for a TE or NT1 given in ANSI Standard T1.408, and the implied jitter attenuation requirements for a type II customer interface given in ANSI T1.403.

#### Jitter Tolerance

Jitter tolerance is the maximum input phase jitter at a given jitter frequency that a device can accept without exceeding its linear operating range, or corrupting data. For the Jitter Attenuator, the input jitter tolerance is 48 Ulpp with no frequency offset. The frequency offset is the difference between the frequency of XCK divided by 24 and that of the input reference clock.

Refer to Figure - 65 for the Jitter Tolerance.

### **Jitter Transfer**

The output jitter for jitter frequencies from 0 to 7Hz is no more than 0.1 dB greater than the input jitter. Jitter frequencies above 7Hz are attenuated at a level of 6 dB per octave, as shown in Figure - 66.

## Frequency Range

In the non-attenuating mode, that is, when the FIFO is within one UI of overrunning or under running, the tracking range is 1.48 to 1.608 MHz. The guaranteed linear operating range is 1.544 MHz  $\pm$  963 Hz (for T1) with no jitter or XCK frequency offset.

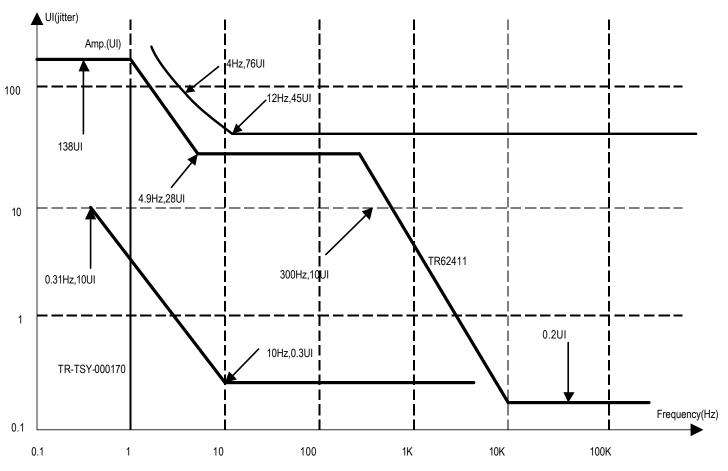


Figure - 65. T1/J1 Mode Jitter Tolerance (N1 = N2 = 2fH)

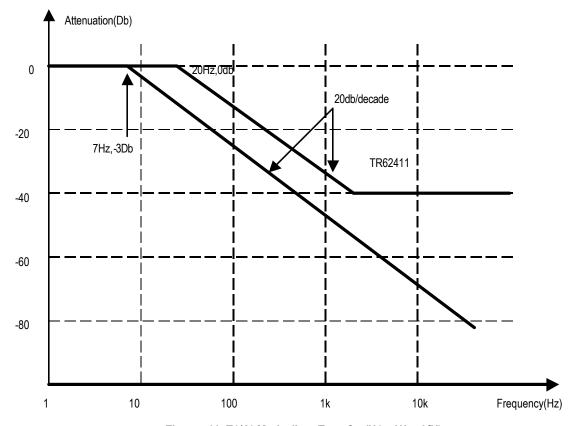


Figure - 66. T1/J1 Mode Jitter Transfer (N1 = N2 = 2fH)

### 3.20 TRANSMIT CLOCK

The Transmit Clock of each framer operates independently.

### 3.20.1 E1 MODE

The Transmit Clock helps the Transmit Jitter Attenuator to select the source of the input reference clock for the DPLL, and selects the clock source used to drive the clock to be output on the LTCKn pin. Refer to Figure - 67 for details.

#### 3.20.2 T1 / J1 MODE

The Transmit Clock helps the Transmit Jitter Attenuator to select the source of the input reference clock for the DPLL, and selects the clock source used to drive the clock to be output on the LTCKn pin. Refer to Figure - 67 for details.

### 3.21 LINE INTERFACE

#### 3.21.1 E1 MODE

On the receive line interface, the received data on the LRDn pin are sampled on the active edge of the LRCKn. The active edge of the

LRCKn is selected by the RCKFALL (b7, E1-001H).

On the transmit line interface, the data to be transmitted on the LTDn pin are updated on the active edge of the LTCKn. The active edge of the LTCKn is selected by the LTCKRISE (b0, E1-002H). All ones can be forced to transmit on the LTDn pin when the TAISEN (b6, E1-002H) is configured. All zeros can also be forced to transmitted when the TXDIS (b0, E1-007H) is configured.

### 3.21.2 T1 / J1 MODE

On the receive line interface, the received data on the LRDn pin are sampled on the active edge of the LRCKn. The active edge of the LRCKn is selected by the LRCKFALL (b2, T1/J1-003H).

On the transmit line interface, the data to be transmitted on the LTDn pin are updated on the active edge of the LTCKn. The active edge of the LTCKn is selected by the LTCKRISE (b0, T1/J1-004H). All ones can be forced to transmitted on the LTDn pin when the TAISEN (b6, T1/J1-004H) is configured. All zeros can also be forced to transmit when the TXDIS (b0, T1/J1-00AH) is configured.

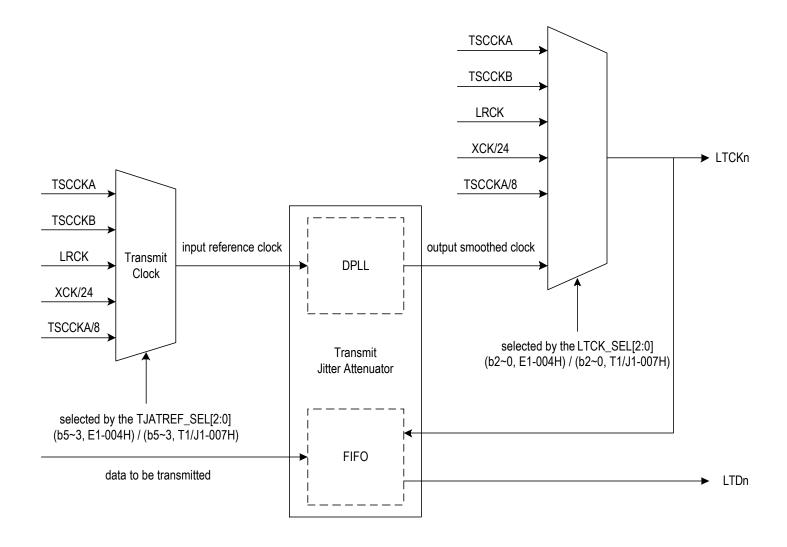


Figure - 67. Transmit Clock Select

#### 3.22 INTERRUPT SUMMARY

### 3.22.1 E1 MODE

When the  $\overline{\rm INT}$  pin asserts low, which means at least one interrupt has occurred in the device, reading the INT[8:1] (b7~0, E1-00BH) will find in which framer the interrupt occurs. After reading the INT regiser, the interrupt source bits from the interrupting framer are read. The Interrupt Source bits (PMON [b7, E1-005H], FRMG [b6, E1-005H], FRMP [b5, E1-005H], PRGD [b4, E1-005H], ELSB [b3, E1-005H], RHDLC#1 [b2, E1-005H], RHDLC#2 [b1, E1-005H], RHDLC#3 [b0, E1-005H], TRSI [b7, E1-006H], TJAT [b5, E1-006H], RJAT [b4, E1-006H], THDLC#1 [b3, E1-006H], THDLC#2 [b2, E1-006H], THDLC#3 [b1, E1-006H] and RCRB [b0, E1-006H]) will be logic 1 if there are interrupts in the corresponding block. To find the eventual interrupt sources, the interrupt Indication and Status bits in the block are polled if their Interrupt Enable bits are enabled. Then the sources are served after they are found.

### 3.22.2 T1 / J1 MODE

When the  $\overline{\text{INT}}$  pin asserts low, which means at least one interrupt has occurred in the device, reading the INT[8:1] (b7~0, T1/J1-00EH) will find that in which framer the interrupt occurs. After reading the INT regiser, the interrupt source bits from the interrupting framer are read. The Interrupt Source bits (PMON [b7, T1/J1-008H], IBCD [b6, T1/J1-008H], FRMP [b5, T1/J1-008H], PRGD [b4, T1/J1-008H], ELSB [b3, T1/J1-008H], RHDLC#1 [b2, T1/J1-008H], RBOM [b1, T1/J1-008H], ALMD [b0, T1/J1-008H], RHDLC#2 [b7, T1/J1-009H], TJAT [b5, T1/J1-009H], RJAT [b4, T1/J1-009H], THDLC#1 [b3, T1/J1-009H], THDLC#2 [b2, T1/J1-009H] and RCRB [b0, T1/J1-009H]) will be logic 1 if there are interrupts in the corresponding block. To find the eventual interrupt sources, the interrupt Indication and Status bits in the block are polled if their Interrupt Enable bits are enabled. Then the sources are served after they are found.

However, another Interrupt Source bit PRTY (b6, T1/J1-009H) is provided to route to the pending parity error.

# 3.23 LOOPBACK MODE

There are three diagnostic loopback modes: Line Loopback, Digital Loopback and Payload Loopback are provided in this device.

#### 3.23.1 LINE LOOPBACK

By programming the LINEB (b4, E1-007H / b4, T1/J1-00AH), each framer can be set in the Line Loopback mode. In this configuration, the jitter-attenuated clock and data from the Receive Jitter Attenuator are looped internally to the Line Transmit Clock and Data (LTDn and LTCKn). However, the Receive Jitter Attenuator can be bypassed if required. The received data stream is still output to the system side while the data stream input from the system side is ignored. Figure - 68 shows the process.

#### 3.23.2 DIGITAL LOOPBACK

By programming the DDLB (b2, E1-007H / b2, T1/J1-00AH), each framer can be set in the Digital Loopback mode. In this configuration, the data to be transmitted on the LTCKn and LTDn are looped internally to the Line Receive Clock and Data (LRDn and LRCKn). The data stream to be transmitted is still output to the line side while the data stream received from the line side is ignored. Figure - 69 shows the process.

#### 3.23.3 PAYLOAD LOOPBACK

By programming the LOOP (b2, E1-TPLC-indirect registers-20~3FH / b2, T1/J1-TPLC-indirect registers-10~18H) (the PCCE [b0, E1-060H / b0, T1/J1-030H] in the TPLC must be logic 1), each framer can be set in the Payload Loopback mode. When Receive Clock Master modes are enabled, the Elastic Store is used to align the line received data to the frame to be transmitted. When Receive Clock Slave modes are enabled, the Elastic Store is unavailable to implement the payload loopbacks, and loop-back functionality is provided only when the Transmit System Interfaces are also in a Transmit Clock Slave mode, and the received and transmitting clocks and frame alignment are identical (RSCCK = TSCCKB, RSCFS = TSCFS). Thus, the selected timeslot/channel in the transmit path will be overwritten by the corresponding received timeslot/channel. The remaining timeslots/channels in the transmit path are intact. Figure - 70 shows the process.

## 3.24 CLOCK MONITOR

The transition from low to high of the Crystal Clock (XCK), the Transmit Side System Common Clock #A (TSCCKA), the Transmit Side System Common Clock #B (TSCCKB), the Receive Side System Common Clock (RSCCK) and the Line Receive Clock (LRCK) are monitored and are reported by the XCK (b4, E1-00DH / b4, T1/J1-027H), the TSCCKB (b3, E1-00DH / b3, T1/J1-027H), the TSCCKA (b2, E1-00DH / b2, T1/J1-027H), the RSCCK (b1, E1-00DH / b1, T1/J1-027H) and the LRCK (b0, E1-00DH / b0, T1/J1-027H) respectively.

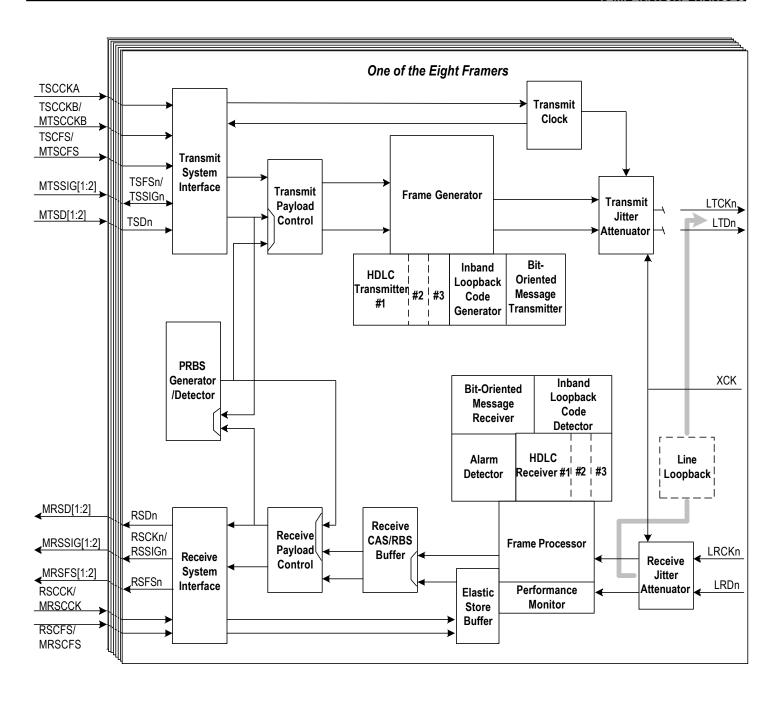


Figure - 68. Line Loopback

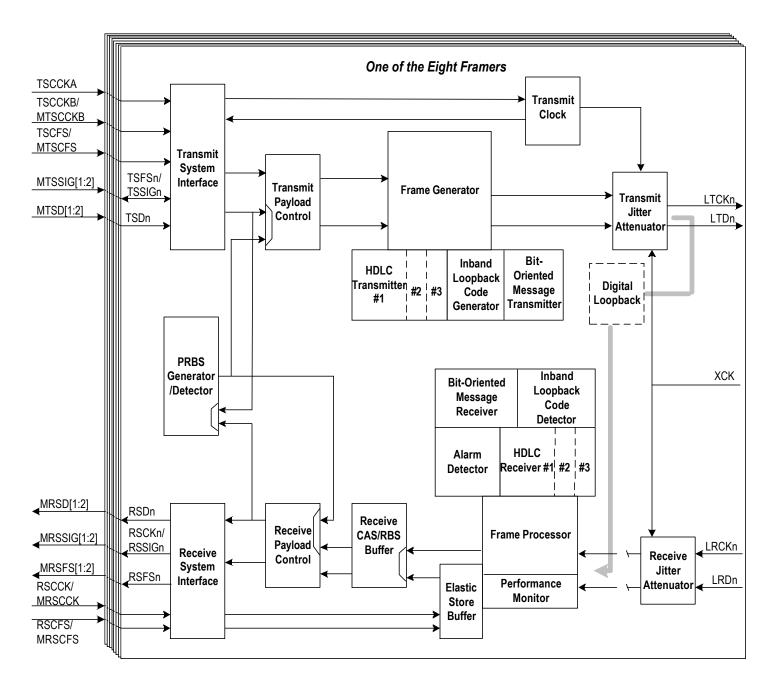


Figure - 69. Digital Loopback

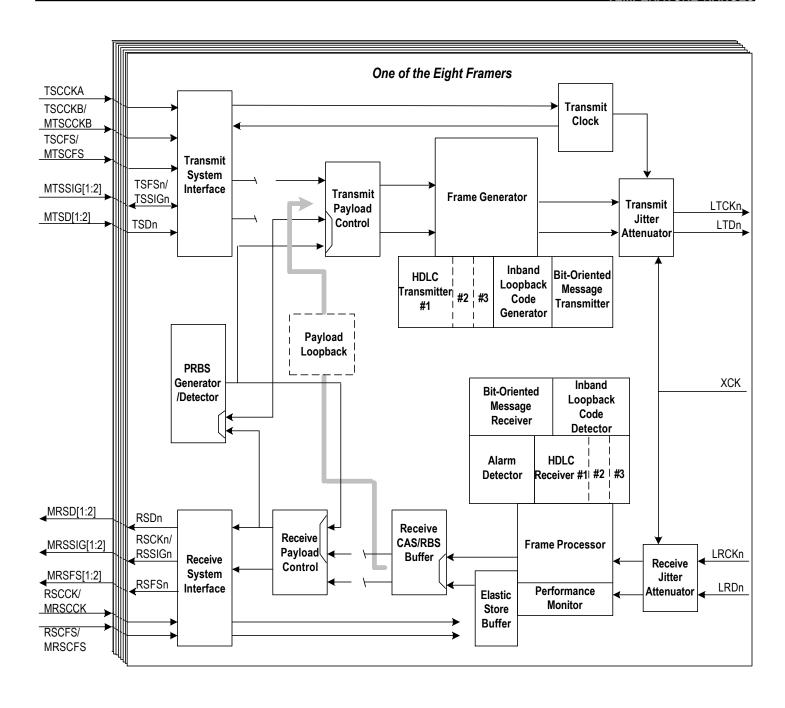


Figure - 70. Payload Loopback

# 4 OPERATION

## 4.1 E1 MODE

### 4.1.1 DEFAULT SETTING

When the device is powered-up, all the registers contain their default values.

Any of the eight framers can be reset anytime when the RESET (b0, E1-00AH / b0, T1/J1-00DH) in its framer is set. The device can also be reset anytime when the  $\overline{RST}$  pin is low for at least 100ns.

After the hardware reset, the IDT82V2108 will default to the following settings:

- Mode: the default operation mode of the device is T1 mode. When the E1 mode is desired, the TEMODE (b0, 400H) must be set to logic 0.
- Receive Path: the default setting of each block in the receive path is illustrated in Table 39.
- Transmit Path: the default setting of each block in the transmit path is illustrated in Table 40.

## 4.1.2 VARIOUS OPERATION MODES CONFIGURATION

Five operation modes can be set in the receive path and four operation modes can be set in the transmit path. In each operation mode, the configurations in Table - 41 and Table - 42 are illustrated for reference.

Table - 39. Default Setting in Receive Path

Function Block	Default Setting Description
Line Interface	• The LRDn inputs Non-Return to Zero (NRZ) data and are sampled on rising edge of the LRCKn.
	• The RJAT Clock Divisors (N1, N2) are set to '2F'.
Frame Processor	Basic Frame per G.704 with CRC Multi-Frame enabled.
	Channel Associated Signaling enabled.
HDLC Receiver #1, #2, #3	RHDLCs disabled.
Receive System Interface	In Receive Clock Slave External Signaling Mode.
	The data on the RSDn, RSSIGn pins are updated on rising edge of the RSCCK.
	RSCFS indicates Basic Frame Alignment.
	RSDn, RSSIGn, RSFSn pins are held in high-impedance state.
PRGD	The PRGD is configured to monitor the extracted data patterns in Frame One.

# Table - 40. Default Setting in Transmit Path

Function Block	Default Setting Description
PRGD	The PRGD is configured to insert test patterns to Frame One.
Transmit System Interface	In Transmit Clock Slave External Signaling Mode.
	The data on the TSDn and TSSIGn pins are sampled on rising edge of TSCCKB.
Frame Generator	CRC Multi-Frame is disabled.
	Channel Associated Signaling is enabled.
HDLC Transmitter #1, #2, #3	• THDLCs are disabled.
Line Interface	The LTDn outputs Non-Return to Zero (NRZ) data and is updated on falling edge of LTCKn.
	TJAT Clock Divisors (N1, N2) are set to '2F'
	Digital jitter attenuation is enabled. The PLL is synchronized to the TSCCKB clock. The smoothed clock
	output from the PLL is selected as the LTCKn.

Table - 41. Various Operation Modes in Receive Path for Reference

Mode	Register <sup>1</sup>	Value (from Bit7 to Bit0)	Description <sup>2</sup>				
	001H	00000000	In Receive Clock Slave RSCK Reference mode. The RSCK is 8KHZ.				
Receive Clock	00EH	00001000	The output on the RSFSn pin is determined by the ROHM, BRXSMFP, BRCMFP				
			and ALTIFP.				
Slave RSCK	010H	00100001	In Receive Clock Slave mode. The FE and DE are both 0. The receive				
			backplane rate is 2.048Mbit/s.				
Reference Mode	011H	00100000	The RSCFS is used.				
	012H	0000001	Enable the normal operation of the RSDn pin.				
	001H	01000000	In Receive Clock Slave External Signaling mode.				
Receive Clock	00EH	00001000	The output on the RSFSn pin is determined by the ROHM, BRXSMFP, BRCMFF and ALTIFP.  In Receive Clock Slave mode. The FE and DE are both 0. The receive backplane rate is 2.048Mbit/s.  The RSCFS is used.  Enable the normal operation of the RSDn pin.  In Receive Clock Slave External Signaling mode.				
Slave External	010H	00100001	The output on the RSFSn pin is determined by the ROHM, BRXSMFP, BRCMFF and ALTIFP.  In Receive Clock Slave mode. The FE and DE are both 0. The receive backplane rate is 2.048Mbit/s.  The RSCFS is used.  Enable the normal operation of the RSDn pin.  In Receive Clock Slave External Signaling mode.  The output on the RSFSn pin is determined by the ROHM, BRXSMFP, BRCMFF and ALTIFP.  In Receive Clock Slave mode. The FE and DE are both 0. The receive backplane rate is 2.048Mbit/s.  The RSCFS is used.  Enable the normal operation of the RSDn and RSSIGn pins.  In Receive Clock Master Full E1 Mode. The FE is logic 1 and the DE is logic 0.  The RSCFS is un-used.  Enable the normal operation of the RSDn pin.  In Receive Clock Master Nx64K Mode. The FE is logic 1 and the DE is logic 0.  The RSCFS is un-used.  Enable the normal operation of the RSDn pin.  Enable the Receive Payload Control.  The code in the DTRK[7:0] replaces the data output on the RSDn pin in the corresponding channel.  Multiplex the data stream of these four framers to the multiplexed bus 1.				
Signaling Mode	011H	00100000	•				
	012H	0000001	Enable the normal operation of the RSDn and RSSIGn pins.				
Receive Clock	010H	00001001	Enable the normal operation of the RSDn and RSSIGn pins.  In Receive Clock Master Full E1 Mode. The FE is logic 1 and the DE is logic 0.  The RSCFS is un-used.  Enable the normal operation of the RSDn pin.				
Master Full E1	011H	00000000	The RSCFS is un-used.				
	012H	0000001	Enable the normal operation of the RSDn pin.				
Receive	010H	10001001	In Receive Clock Master Nx64K Mode. The FE is logic 1 and the DE is logic 0.				
Clock	011H	00000000	The RSCFS is un-used.				
Master	012H	0000001	Enable the normal operation of the RSDn pin.				
Fractional	05CH	0000011	Enable the Receive Payload Control.				
E1	20H-3FH (RPLC	01000000	The code in the DTRK[7:0] replaces the data output on the RSDn pin in the				
Mode	indirect registers)		corresponding channel.				
	001H	01001000					
	081H	01001000	Multiplex the data stream of these four framers to the multiplexed bus 1.				
	101H	01001000					
Receive	181H	01001000					
	201H	01011000					
	281H	01011000	Multiplex the data stream of these four framers to the multiplexed bus 2.				
Multiplexed	301H	01011000					
	381H	01011000					
	010H	00111011					
Mode	090H	00111011					
	110H	00111011					
	190H	00111011	In Receive Multiplexed mode. The receive backplane rate is 8.192Mbit/s. The FE				
	210H	00111011	is logic 1 and the DE is logic 1.				
	290H	00111011					
	310H	00111011					
	390H	00111011					

Table - 41. Various Operation Modes in Receive Path for Reference (Continued)

Mode	Register 1	Value (from Bit7 to Bit0)	Description <sup>2</sup>				
	011H	00100000					
	091H	00100000					
	111H	00100000					
Receive	191H	00100000	The MRSCFS is used.				
	211H	00100000					
	291H	00100000					
Multiplexed	311H	00100000					
	391H	00100000					
	012H	0000001					
Mode	092H	0000001					
	112H	0000001	Enable the normal operation of the MRSD and MRSSIG pins.				
	192H	0000001					
(Continued)	212H	0000001					
	292H	0000001	1				
	312H	0000001					
	392H	0000001					
	013H	00000000	TSOFF[6:0] = 0. The timeslot offset is 0.				
	093H	0000001	TSOFF[6:0] = 1. The timeslot offset is 1.				
	113H	0000010	TSOFF[6:0] = 2. The timeslot offset is 2.				
	193H	00000011	TSOFF[6:0] = 3. The timeslot offset is 3.				
	213H	00000000	TSOFF[6:0] = 0. The timeslot offset is 0.				
	293H	0000001	TSOFF[6:0] = 1. The timeslot offset is 1.				
	313H	00000010	TSOFF[6:0] = 2. The timeslot offset is 2.				
	393H	00000011	TSOFF[6:0] = 3. The timeslot offset is 3.				

<sup>1.</sup> In the 'Register' column, except for the Receive Multiplexed mode, the register position of the Framer One is listed to represent the set of the registers of eight framers. The other registers position are tabulated in the 'Register Map'. However, in Receive Multiplexed mode, the register position of eight framers are all listed

<sup>2.</sup> The 'Description' illustrates the fundamental function of the operation mode. The others can be configured as desired.

Table - 42. Various Operation Modes in Transmit Path for Reference

Mode	Register 1	Value (from Bit7 to Bit0)	Description <sup>2</sup>
	018H	00100001	In Transmit Clock Slave mode. The FE is logic 0 and the DE is logic 0.
Transmit Clock	003H	01000000	In Transmit Clock Slave External Signaling mode.
•	040H	01110000	Channel Associated Signaling (CAS) is enabled. The CRC Multi-Frame is
Slave External			generated.
	004H	00001111	TSCCKB is selected as TJAT input reference clock. Smoothed clock is selected
Signaling Mode			as Line Transmit Clock (LTCK).
	027H	00010000	The FIFO is set to self-center its read pointer.
	018H	00100001	In Transmit Clock Slave mode. The FE is logic 0 and the DE is logic 0.
Transmit Clock	003H	00000000	In Transmit Clock Slave TSFS Enable mode.
	040H	01110000	Channel Associated Signaling (CAS) is enabled. The CRC Multi-Frame is
Slave TSFS			generated.
ľ	004H	00001111	TSCCKB is selected as TJAT input reference clock. Smoothed clock is selected
Enable Mode			as Line Transmit Clock (LTCK).
Ī	027H	00010000	The FIFO is set to self-center its read pointer.
	018H	00011001	In Transmit Clock Master Full E1 mode.
Transmit Clock	040H	01110000	Channel Associated Signaling (CAS) is enabled. The CRC Multi-Frame is
			generated.
Master Mode	004H	00100100	XCK/24 is selected as TJAT input reference clock. XCK/24 is selected as Line
			Transmit Clock (LTCK).
Transmit	003H	01000000	The data stream is taken from the multiplexed bus 1.
	083H	01010000	The data stream is taken from the multiplexed bus 2.
	103H	01000000	The data stream is taken from the multiplexed bus 1.
<u> </u>	183H	01010000	The data stream is taken from the multiplexed bus 2.
  -	203H	01000000	The data stream is taken from the multiplexed bus 1.
<u> </u>	283H	01010000	The data stream is taken from the multiplexed bus 2.
Multiplexed	303H	01000000	The data stream is taken from the multiplexed bus 1.
<u> </u>	383H	01010000	The data stream is taken from the multiplexed bus 2.
<u> </u>	018H	00110011	
<u> </u>	098H	00110011	
<u> </u>	118H	00110011	
Mode	198H	00110011	In Transmit Multiplexed mode. The FE is logic 0 and the DE is logic 1.
	218H	00110011	
	298H	00110011	
	318H	00110011	
	398H	00110011	
	01BH	0000000	TSOFF[6:0] = 0. The timeslot offset is 0.
	09BH	0000000	
	11BH	00000001	TSOFF[6:0] = 1. The timeslot offset is 1.
<u> </u>	19BH	00000001	TOOFFICE OF CITY OF THE STATE O
	21BH	00000010	TSOFF[6:0] = 2. The timeslot offset is 2.
	29BH	00000010	TOOFFICE OF CITY OF COMMENT
	31BH	00000011	TSOFF[6:0] = 3. The timeslot offset is 3.
	39BH	0000011	

Table - 42. Various Operation Modes in Transmit Path for Reference (Continued)

Mode	Register 1	Value (from Bit7 to Bit0)	Description <sup>2</sup>				
	040H	01110000					
	0C0H	01110000					
	140H	01110000					
	1C0H	01110000	Channel Associated Signaling (CAS) is enabled. The CRC Multi-Frame is				
Transmit	240H	01110000	generated.				
	2C0H	01110000					
	340H	01110000					
Multiplexed	3C0H	01110000					
	004H	00011101					
	084H	00011101					
Mode	104H	00011101					
	184H	00011101	TSCCKA is selected as TJAT input reference clock. Smoothed clock is selected as Line Transmit Clock (LTCK).				
	204H	00011101					
(Continued)	284H	00011101	, , ,				
	304H	00011101					
	384H	00011101					
	027H	00010000					
	0A7H	00010000					
	127H	00010000					
	1A7H	00010000	The FIFO is set to self-center its read pointer.				
	227H	00010000	'				
	2A7H	00010000					
	327H	00010000					
	3A7H	00010000					

<sup>1.</sup> In the 'Register' column, except for the Transmit Multiplexed mode, the register position of the Framer One is listed to represent the set of the registers of eight framers. The other registers position are tabulated in the 'Register Map'. However, in Transmit Multiplexed mode, the register position of eight framers are all listed.

<sup>2.</sup> The 'Description' illustrates the fundamental function of the operation mode. The others can be configured as desired.

#### 4.1.3 OPERATION EXAMPLE

In this chapter, some common operation examples are given for reference.

## 4.1.3.1 Using The HDLC Receiver

Before using the HDLC Receiver, the TXCISEL (b3, E1-00AH) must be set to 0 to enable the HDLC data link position for receive path.

Since three HDLC Receive data links are integrated in one framer, one of the three HDLC Receive data links must be selected in the RHDLCSEL[1:0] (b7~6, E1-00AH). Then the HDLC data link can be configured to extract from even and/or odd frames, from any timeslot, and from any bit. The following examples show how to select the HDLC Receiver data link positions:

- a. to extract the HDLC data link from all bits of TS16 in HDLC Receive #1:
  - set the TXCISEL (b3, E1-00AH) to 0;
  - set the RHDLCSEL[1:0] (b7~6, E1-00AH) to 00;
  - set the DL1\_EVEN (b7, E1-028H) to 0;
  - set the DL1\_ODD (b6, E1-028H) to 0;
  - set the TS16\_EN (b5, E1-028H) to 1.
- b. to extract the HDLC data link from the Sa8 National bit in HDLC Receive #1:
  - set the TXCISEL (b3, E1-00AH) to 0;
  - set the RHDLCSEL[1:0] (b7~6, E1-00AH) to 00;
  - set the DL1\_EVEN (b7, E1-028H) to 0;
  - set the DL1\_ODD (b6, E1-028H) to 1;
  - set the TS16\_EN (b5, E1-028H) to 0;
  - set the DL1\_TS[4:0] (b4~0, E1-028H) to 00000
  - set the DL1\_BIT[7:0] (b7~0, E1-029H) to 00000001.
- c. to extract the HDLC data link from all bits of TS20 of all frames in HDLC Receive #2:
  - set the TXCISEL (b3, E1-00AH) to 0;
  - set the RHDLCSEL[1:0] (b7~6, E1-00AH) to 01;
  - set the DL2 EVEN (b7, E1-02AH) to 1;
  - set the DL2\_ODD (b6, E1-02AH) to 1;
  - set the DL2\_TS [4:0] (b4~0, E1-02AH) to 10100;
  - set the DL2\_BIT [7:0] (b7~0, E1-02BH) to 11111111.

After setting the HDLC data link position properly, the selected HDLC Receiver should be enabled by setting the EN (b0, E1-048H)to logic 1. If needed, set the MEN (b3, E1-048H) and the MM (b2, E1-048H) to determine which Address Matching Mode to be selected (refer to Register Description for details). After setting these 3 bits, the RHDLC Primary Address Match register and the RHDLC Secondary Address Match register should be set to proper values. If the INTC[6:0] (b6~0, E1-049H) are set, whenever the number of bytes in the RHDLC FIFO exceeds the value set in the INTC[6:0] (b6~0, E1-049H), the INTR (b0, E1-04AH) will be set to logic 1. This interrupt will persist until the RHDLC FIFO becomes empty. Setting the INTE (b7, E1-049H) to logic 1 allows the internal interrupt status to be propagated to the INT output pin.

After setting these registers properly, the HDLC data can be received in a polled or interrupt driven mode.

## - Interrupt Driven Mode

When the INTE (b7, E1-049H) is set to logic 1, if the  $\overline{\text{INT}}$  pin is asserted, the source of the interrupt should be first identified by reading

the Interrupt ID register and Interrupt Source registers. If the source of the interrupt is HDLC Receive, the Interrupt Service procedure will be carried out as shown in Figure - 71.

## - Polling Mode

In polling mode, the operation procedure is the same as Figure - 71, except that the entry of the service is from a local timer rather than an interrupt.

To summarize the procedure of using HDLC Receive, a complete example is shown in Table - 43.

Table - 43. Example for Using HDLC Receiver

Register	Value	Description			
00AH	50H	RHDLC #2 is selected. The HDLC Receive is			
		accessable to the CPU interface.			
02AH	C4H	The TS4 of even frames and odd frames are			
		selected.			
02BH	FFH	All the 8 bits are selected.			
048H	0DH	The function of the RHDLC #2 is enabled. Set			
		the address match mode.			
049H	8FH	Set the INTE to 1. When the number of bytes in			
		the RHDLC FIFO exceeds 15, an interrupt is			
		generated.			
04CH	04CH 13H The primary address is set to 13H.				
04DH	FFH	The secondary address is set to FFH.			
Then read	the data	status in register 04AH. Until a complete packet is			

Then read the data status in register 04AH. Until a complete packet is received, read the data from register 04BH.

## 4.1.3.2 Using The HDLC Transmitter

Before using the HDLC Transmit, the TXCISEL (b3, E1-00AH) must be set to 1 to enable the HDLC data link position for transmit path.

Since three HDLC Transmit data links are integrated in one framer, one of the three HDLC Transmit data links must be selected in the THDLCSEL[1:0] (b5~4, E1-00AH). Then the HDLC data link can be configured to insert to even and/or odd frames, to any timeslot, and to any bit. The following examples show how to select the HDLC Transmit data link positions:

- a. to insert the HDLC data link to all bits of TS16 in HDLC Transmit
   #1.
  - set the TXCISEL (b3, E1-00AH) to 1;
  - set the THDLCSEL [1:0] (b5~4, E1-00AH) to 00;
  - set the DL1\_EVEN (b7, E1-028H) to 0;
  - set the DL1\_ODD (b6, E1-028H) to 0;
  - set the TS16\_EN (b5, E1-028H) to 1.
- b. to insert the HDLC data link to the Sa4-Sa8 National bits in HDLC Transmit #1:
  - set the TXCISEL (b3, E1-00AH) to 1;
  - set the THDLCSEL [1:0] (b5~4, E1-00AH) to 00;
  - set the DL1 EVEN (b7, E1-028H) to 0;
  - set the DL1\_ODD (b6, E1-028H) to 1;
  - set the TS16\_EN (b5, E1-028H) to 0;

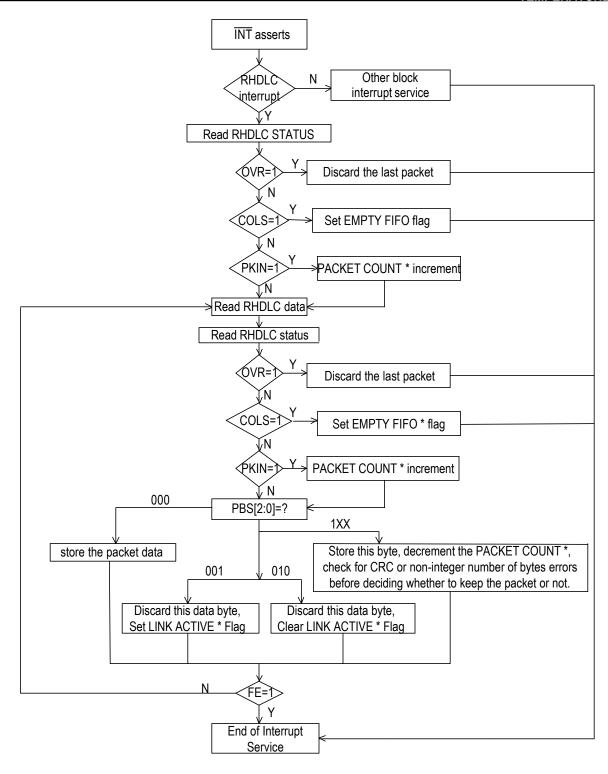


Figure - 71. Interrupt Service in E1 Mode HDLC Receiver

<sup>\*</sup> The PACKET COUNT, EMPTY FIFO and LINK ACTIVE are local software variable.

- set the DL1 TS[4:0] (b4~0, E1-028H) to 00000
- set the DL1\_BIT[7:0] (b7~0, E1-029H) to 00011111.
- c. to insert the HDLC data link to all bits of TS20 of all frames in HDLC Transmit #3:
  - set the TXCISEL (b3, E1-00AH) to 1;
  - set the THDLCSEL [1:0] (b5~4, E1-00AH) to 10;
  - set the DL3\_EVEN (b7, E1-02CH) to 1;
  - set the DL3\_ODD (b6, E1-02CH) to 1;
  - set the DL3\_TS [4:0] (b4~0, E1-02CH) to 10100;
  - set the DL3\_BIT [7:0] (b7~0, E1-02DH) to 11111111.

After setting the HDLC data link position properly, the selected HDLC Transmit should be enabled by setting the EN (b0, E1-050H) to logic 1. The FIFOCLR (b6, E1-050H) should be set and then cleared to initialize the THDLC FIFO.

Set the CRC (b1, E1-050H) to logic 1 if the Frame Check Sequences (FCS) generation is desired. Set the FULLE (b3, E1-053H), OVRE (b2, E1-053H), UDRE (b1, E1-053H) and LFILLE (b0, E1-053H) to logic 1 if interrupt driven mode is used. Set THDLC Upper Transmit Threshold and THDLC Lower Transmit Threshold registers to the desired values. If a complete packet has been written into THDLC FIFO, the EOM (b3, E1-050H) should be set.

After setting these registers properly, the HDLC data can be transmitted in a polled or interrupt driven mode.

## - Interrupt Driven Mode

Writing HDLC data to THDLC FIFO , the THDLC will transmit the HDLC data if the end of a packet was written or if the THDLC FIFO fill level reaches the Upper Transmit Threshold. The writing procedure is shown in Figure - 72.

When the FULLE (b3, E1-053H), OVRE (b2, E1-053H), UDRE (b1, E1-053H) and LFILLE (b0, E1-053H) are set to logic 1, if the  $\overline{\text{INT}}$  pin is asserted, the source of the interrupt should be identified firstly by reading the Interrupt ID register and Interrupt Source registers. If the source of the interrupt is HDLC Transmit, the Interrupt Service procedure will be carried out as shown in Figure - 73.

#### - Polling Mode

In packet transmission polling mode, the FULLE (b3, E1-053H), OVRE (b2, E1-053H), UDRE (b1, E1-053H) and LFILLE (b0, E1-053H) should be set to logic 0. The THDLC Lower Transmit Threshold should be set to such a value that sufficient warning of an underrun is given. The procedure shown in Figure - 74 should be followed.

To summarize the procedure of using HDLC Transmit, a complete example is shown in Table - 44.

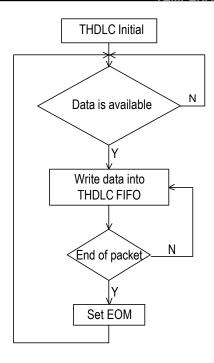
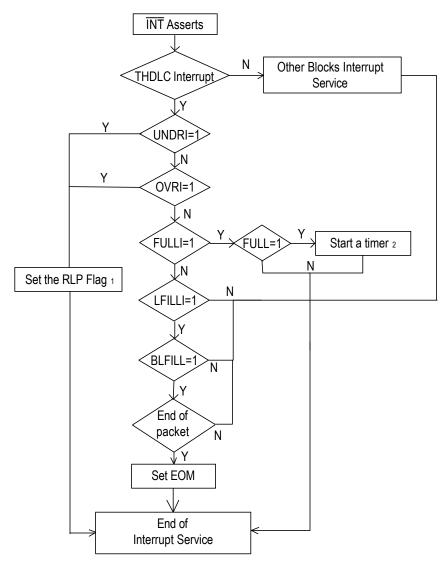


Figure - 72. Writing Data to E1 Mode THDLC FIFO

Table - 44. Example for Using HDLC Transmitter

Register	Value	Description
00AH	58H	THDLC #2 is selected. The HDLC Transmit is
		accessable to the CPU interface.
02AH	C4H	The TS4 of even frames and odd frames are
		selected.
02BH	FFH	All the 8 bits are selected.
050H	C3H	The function of the THDLC #2 is enabled. The
050H	83H	FCS is enables and the THDLC FIFO is reset.
053H	0FH	Enable the THDLC Interrupt Enable bits.
055H	12H	
055H	34H	
055H	56H	
055H	78H	Write data into THDLC FIFO.
055H	9AH	
055H	BCH	
055H	DEH	
055H	FFH	
050H	8BH	End of packet and set the EOM to 1.



- 1. RLP-Retransmit the last packet, a local software variable.
- 2. A local timer to wait for a certain time until the Full = 0 or the BLFILL = 1.

Figure - 73. Interrupt Service in E1 Mode HDLC Transmitter

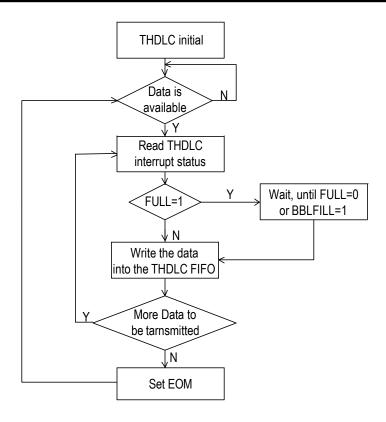


Figure - 74. Polling Mode in E1 Mode HDLC Transmitter

## 4.1.3.3 Using The PRBS Generator / Detector

The IDT82V2108 provides one PRBS generator/detector block to generate and detect an enormous variety of pseudo-random and repetitive patterns to diagnose E1 data stream of all eight framers. The common test patterns are shown in tabular form in Table - 45.

The PRBS generator/detector block can be used to test E1 line transmit-receive integrity and system backplane integrity.

- Example For Testing E1 Line Transmit-Receive Integrity

To monitor the errors in Framer 2 without taking the entire E1 span off line, the following procedure should be done:

- Use the PRGD block to test Framer 2;
- Configure the PRGD register;
- Chose a desired set of timeslots (for example TS2, TS4, TS5) for insertion/extraction of PRGD test data;
- Set the far end of the line to loop back at least the selected imeslots:
  - Monitor the E1 line transmit-receive integrity.

To realize the above function, the configuration in Table - 46 to Table - 48 should be set.

Table - 46 is the configuration for PRGD and loopback. Table - 47 shows the process to initialize the TPLC. Table - 48 shows the process to initialize the RPLC.

After the above settings, read the PRGD Interrupt Enable/Status (071H) register twice. If the SYNCV (b4, E1-071H) is logic 1 and the BEI (b2, E1-071H) is logic 0, the pattern detector is in synchronization state.

Then insert errors into this link. Here suppose to insert 3 errors, then the configuration is shown in Table - 49.

- Example For Testing E1 System Backplane Integrity

To test the E1 system backplane integrity, the RXPATGEN (b2, E1-00CH) should be set to logic 1 and the other registers are set as above. Then the PRGD can be used to test the system backplane integrity.

Table - 45. Test Pattern

Pseudo-Random Pattern Generation (the PS [b4, E1-070H] is logic 0)								
Pattern Type	TR <sup>1</sup>	LR <sup>2</sup>	IR#1 <sup>3</sup>	IR#2 ⁴	IR#3 <sup>5</sup>	IR#4 <sup>6</sup>	TINV 7	RINV 7
2 <sup>3</sup> -1	00	02	FF	FF	FF	FF	0	0
2 <sup>4</sup> -1	00	03	FF	FF	FF	FF	0	0
2 <sup>5</sup> -1	01	04	FF	FF	FF	FF	0	0
26_1	04	05	FF	FF	FF	FF	0	0
2 <sup>7</sup> –1 (Fractional T1 LB Activate)	00	06	FF	FF	FF	FF	0	0
2 <sup>7</sup> –1 (Fractional T1 LB Activate)	03	06	FF	FF	FF	FF	0	0
27-1	03	06	FF	FF	FF	FF	1	1
2 <sup>9</sup> –1 (O.153)	04	08	FF	FF	FF	FF	0	0
2 <sup>10</sup> -1	02	09	FF	FF	FF	FF	0	0
2 <sup>11</sup> -1 (0.152,0.153)	08	0A	FF	FF	FF	FF	0	0
2 13 -1 (0.151)	0D	0E	FF	FF	FF	FF	1	1
2 <sup>17</sup> -1	02	10	FF	FF	FF	FF	0	0
2 18 -1	06	11	FF	FF	FF	FF	0	0
2 <sup>20</sup> -1 (O.153)	02	13	FF	FF	FF	FF	0	0
2 <sup>21</sup> -1	01	14	FF	FF	FF	FF	0	0
2 <sup>22</sup> -1	00	15	FF	FF	FF	FF	0	0
2 <sup>23</sup> -1 (0.151)	11	16	FF	FF	FF	FF	1	1
2 <sup>25</sup> -1 2 <sup>28</sup> -1	02	18	FF	FF	FF	FF	0	0
2 <sup>28</sup> -1	02	1B	FF	FF	FF	FF	0	0
2 <sup>29</sup> -1	01	1C	FF	FF	FF	FF	0	0
2 31 -1	02	1E	FF	FF	FF	FF	0	0

# Repetitive Pattern Generation (the PS [b4, E1-070H] is logic 1)

· · · · · · · · · · · · · · · · · · ·								
Pattern Type	TR <sup>1</sup>	LR <sup>2</sup>	IR#1 <sup>3</sup>	IR#2 <sup>4</sup>	IR#3 <sup>5</sup>	IR#4 <sup>6</sup>	TINV 7	RINV 7
All ones	00	00	FF	FF	FF	FF	0	0
All zeros	00	00	FE	FF	FF	FF	0	0
Alternatingones/zeros	00	01	FE	FF	FF	FF	0	0
Double alternatingones/zeros	00	03	FC	FF	FF	FF	0	0
3 in 24	00	17	22	00	20	FF	0	0
1 in 16	00	0F	01	00	FF	FF	0	0
1 in 8	00	07	01	FF	FF	FF	0	0
1 in 4	00	03	F1	FF	FF	FF	0	0
DS1 Inbandloopback activate	00	04	F0	FF	FF	0F	0	0
DS1 InbandLoopback deactivate	00	02	FC	FF	FF	FF	0	0

- TR Tap Register
   LR Shift Register Length Register
   IR#1 PRGD Pattern Insertion #1 Register
   IR#2 PRGD Pattern Insertion #2 Register
- 5. IR#3 PRGD Pattern Insertion #3 Register
- 6. IR#4 PRGD Pattern Insertion #4 Register
- 7. TINV, RINV contained in the PRGD Control register

Table - 46. The Setting of PRGD

Table - 47. Initializtion of TPLC (Continued)

Register	Value	Description	
00CH	20H	Select Framer 2 to be tested by the PRGD	
		block. The PRGD pattern is inserted in the TPLC	
		and detected in the RPLC.	
070H	82H	Set Pattern Detector registers as error counter	
		register. Enable automatic resynchronization.	
072H	18H	Set the pattern length.	
073H	02H	Set the feedback tap position.	
078H	FFH	Set the Pattern Insertion registers.	
07BH	FFH	Load the data in the Pattern Insertion registers to	
		generate the pattern.	
087H	04H	Set diagnostic digital loopback mode.	
0E0H	01H	Enable the TPLC indirect registers to be	
		accessable.	
0DCH	01H	Enable the RPLC indirect registers to be	
		accessable.	

Table - 47. Initializtion of TPLC

Register	Value
0E3H	00H
0E2H	20H
0E3H	00H
0E2H	21H
0E3H	00H
0E2H	22H
0E3H	00H
0E2H	23H
0E3H	00H
0E2H	24H
0E3H	00H
0E2H	25H
0E3H	00H
0E2H	26H
0E3H	00H
0E2H	27H
0E3H	00H
0E2H	28H
0E3H	00H
0E2H	29H
0E3H	00H
0E2H	2AH
0E3H	00H
0E2H	2BH
0E3H	00H
0E2H	2CH
0E3H	00H
0E2H	2DH

Register	Value
0E3H	00H
0E2H	2EH
0E3H	00H
0E2H	2FH
0E3H	00H
0E2H	30H
0E3H	00H
0E2H	31H
0E3H	00H
0E2H	32H
0E3H	00H
0E2H	33H
0E3H	00H
0E2H	34H
0E3H	00H
0E2H	35H
0E3H	00H
0E2H	36H
0E3H	00H
0E2H	37H
0E3H	00H
0E2H	38H
0E3H	00H
0E2H	39H
0E3H	00H
0E2H	3AH
0E3H	00H
0E2H	3BH
0E3H	00H
0E2H	3CH
0E3H	00H
0E2H	3DH
0E3H	00H
0E2H	3EH
0E3H	00H
0E2H	3FH
0E3H	00H
0E2H	60H
0E3H	00H
0E2H	61H
0E3H	00H
0E2H	62H
0E3H	00H
0E2H	63H
0E3H	00H
0E2H	64H
0E3H	00H
0E2H	65H
UEZΠ	חטט

Table - 47. Initializtion of TPLC (Continued)

Register Value 0E3H 00H 0E2H 66H 0E3H 00H 0E2H 67H 0E3H 00H 0E2H 68H 0E3H 00H 0E2H 69H 0E3H 00H 0E2H 6AH 0E3H 00H 0E2H 6BH 0E3H 00H 0E2H 6CH 0E3H 00H 0E2H 6DH 0E3H 00H 0E2H 6EH 0E3H 00H 0E2H 6FH 00H 0E3H 70H 0E2H 00H 0E3H 0E2H 71H 00H 0E3H 0E2H 72H 0E3H 00H 0E2H 73H 0E3H 00H 74H 0E2H 0E3H 00H 0E2H 75H 0E3H 00H 0E2H 76H 0E3H 00H 77H 0E2H 0E3H 00H 0E2H 78H 0E3H 00H 79H 0E2H 0E3H 00H 0E2H 7AH 0E3H 00H 0E2H 7BH 0E3H 00H 0E2H 7CH 0E3H 00H 0E2H 7DH

Table - 47. Initializtion of TPLC (Continued)

Register	Value
0E3H	00H
0E2H	7EH
0E3H	00H
0E2H	7FH

Then set the TEST in TPLC Payload Control register for TS2, TS4 and TS5. The process is:

Register	Value	Description
0E3H	H80	Set the TEST in TPLC Payload Control register
0E2H	22H	for TS2.
0E3H	08H	Set the TEST in TPLC Payload Control register
0E2H	24H	for TS4.
0E3H	08H	Set the TEST in TPLC Payload Control register
0E2H	25H	forTS5.

Table - 48. Initializtion of RPLC

Value Register 0DFH 00H 0DEH 20H 0DFH 00H 0DEH 21H 0DFH 00H 0DEH 22H 0DFH 00H 0DEH 23H 0DFH 00H 0DEH 24H 0DFH 00H 0DEH 25H 0DFH 00H 0DEH 26H 0DFH 00H 0DEH 27H 0DFH 00H 28H 0DEH 0DFH 00H 0DEH 29H 0DFH 00H 0DEH 2AH 0DFH 00H 0DEH 2BH 0DFH 00H 2CH 0DEH 0DFH 00H 2DH 0DEH 0DFH 00H 0DEH 2EH 0DFH 00H 0DEH 2FH 0DFH 00H 0DEH 30H 0DFH 00H 31H 0DEH 0DFH 00H 0DEH 32H 0DFH 00H 0DEH 33H 0DFH 00H 0DEH 34H 0DFH 00H 35H 0DEH 0DFH 00H 0DEH 36H 0DFH 00H 0DEH 37H

Table - 48. Initializtion of RPLC (Continued)

Register	Value
0DFH	00H
0DEH	38H
0DFH	00H
0DEH	39H
0DFH	00H
0DEH	3AH
0DFH	00H
0DEH	3BH
0DFH	00H
0DEH	3CH
0DFH	00H
0DEH	3DH
0DFH	00H
0DEH	3EH
0DFH	00H
0DEH	3FH
	•

Then set the TEST in RPLC Payload Control register for TS2, TS4 and TS5. The process is:

Register	Value	Description	
0DFH	80H	Set the TEST in RPLC Payload Control	
0DEH	22H	register for TS2.	
0DFH	80H	Set the TEST in RPLC Payload Control	
0DEH	24H	register for TS4	
0DFH	80H	Set the TEST in RPLC Payload Control	
0DEH	25H	register for TS5	

Table - 49. Error Insertion

Value
08H
00H
08H
00H
08H
00H

Then write 00H into the 07CH register to update the error counter registers. Then read the registers from 07CH to 07FH to check the error numbers.

## 4.1.3.4 Using Payload Control and Receive CAS/RBS Buffer

Before using the Receive/Transmit Payload Control and Receive CAS/RBS Buffer, the indirect registers of these blocks must be initialized to eliminate erroneous control data. The PCCE (b0, E1-05CH & b0, E1-060H & b0, E1-064H) of these blocks must be set to logic 1 to enable these blocks.

Then the BUSY (b7, E1-05DH & b7, E1-061H & b7, E1-065H) must be checked before a new access request to the RPLC, TPLC and RCRB indirect registers. When the BUSY is logic 0, the new reading and writing access operations can be performed.

Figure - 75 shows the writing sequence of the RPLC, TPLC and RCRB indirect registers. Figure - 76 shows the reading sequence of the RPLC, TPLC and RCRB indirect registers.

## 4.1.3.5 Using TJAT / Timing Option

In different operation modes, the Timing Options and Clock Divisor Control registers can be set as the follows:

- Transmit Clock Slave Mode (System Backplane Rate: 2.048M bit/s) The TSCCKA or TSCCKB is selected as the TJAT DPLL input reference clock. The TSCCKA and TSCCKB are both equal to 2.048M. The N1 (b7~0, E1-025H) and N2 (b7~0, E1-026H) are set to their default value (2FH).

The smoothed clock output from the TJAT is selected as the LTCK.

- Transmit Clock Slave Mode (System Backplane Rate: 4.096M bit/s) The TSCCKA is selected as the TJAT DPLL input reference clock. The TSCCKA is equal to 2.048M. The N1 (b7~0, E1-025H) and N2 (b7~0, E1-026H) are set to their default value (2FH).

The smoothed clock output from the TJAT is selected as the LTCK.

- Transmit Clock Master Mode
   The XCK/24 is selected as the TJAT DPLL input reference clock.

   The XCK/24 is selected as the LTCK.
- Transmit Multiplexed Mode (System Backplane Rate: 8.192M bit/s)
  The TSCCKA is selected as the TJAT DPLL input reference clock.
  The TSCCKA is equal to 2.048M. The N1 (b7~0, E1-025H) and N2 (b7~0, E1-026H) are set to their default value (2FH).

The smoothed clock output from the TJAT is selected as the LTCK.

Transmit Multiplexed Mode (System Backplane Rate: 16.384M bit/s)
 The TSCCKA or TSCCKA/8 is selected as the TJAT DPLL input reference clock. The TSCCKA is equal to 2.048M or 16.384M. The N1 (b7~0, E1-025H) and N2 (b7~0, E1-026H) are set to their default value (2FH).

The smoothed clock output from the TJAT is selected as the LTCK.

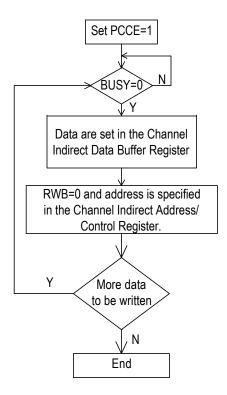


Figure - 75. Writing Sequence of Indirect Register in E1 Mode

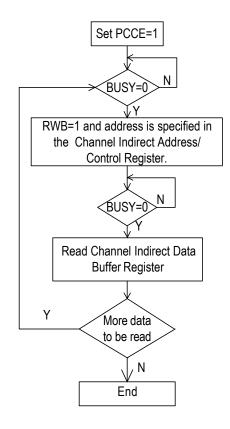


Figure - 76. Reading Sequence of Indirect Register in E1 Mode

## 4.2 T1/J1 MODE

### 4.2.1 DEFAULT SETTING

When the device is power up, all the registers are in their default values.

Any of the eight framers can be reset anytime when the RESET (b0, E1-00AH / b0, T1/J1-00DH) in its framer is set. The device can also be reset anytime when the  $\overline{RST}$  pin is low for at least 100ns.

After the hardware reset, the IDT82V2108 will default to the following settings:

- Mode: the default operation mode of the device is T1 mode.
- Receive Path: the default setting of each block in the receive path is illustrated in the Table 50.
- Transmit Path: the default setting of each block in the transmit path is illustrated in the Table 51.

### 4.2.2 OPERATION IN J1 MODE

IDT82V2108 can also be operated in J1 mode when the TEMODE (b0, 400H) is set to logic 1. Except for the setting of the JYEL in bit 3 of FRMP Configuration registers (020H, 0A0H, 120H, 1A0H, 220H, 2A0H, 320H, 3A0H), the J1\_YEL in bit 5 of ALMD Configuration registers (02CH, 0ACH, 12CH, 1ACH, 22CH, 2ACH, 32CH, 3ACH), the J1\_YEL

in bit 5 and the J1\_CRC in bit 6 of FRMG Configuration registers (044H, 0C4H, 144H, 1C4H, 244H, 2C4H, 344H, 3C4H), the setting of the other registers are the same as the setting in T1 mode.

The follows illustrate the setting in J1 mode difference from the setting in the T1 mode.

In receive path, set the JYEL in bit 3 of FRMP Configuration registers (020H, 0A0H, 120H, 1A0H, 220H, 2A0H, 320H, 3A0H) to logic 1, the Frame Processor will operate in J1 mode. Set the J1\_YEL in bit 5 of ALMD Configuration registers (02CH, 0ACH, 12CH, 1ACH, 22CH, 2ACH, 32CH, 3ACH) to logic 1, the Alarm Detector will operate in J1 mode.

In transmit path, set the J1\_CRC in bit 6 of FRMG Configuration registers (044H, 0C4H, 144H, 1C4H, 244H, 2C4H, 344H, 3C4H) to logic 1, the Frame Generator will generate J1 frame. Set the J1\_YEL in bit 5 of FRMG Configuration registers (044H, 0C4H, 144H, 1C4H, 244H, 2C4H, 344H, 3C4H) to logic 1, the IDT82V2108 will transmit the Yellow alarm in J1 format if Yellow alarm transmission is enabled.

## 4.2.3 VARIOUS OPERATION MODES CONFIGURATION

Five operation modes can be set in the receive path and four operation modes can be set in the transmit path. In each operation modes, the configurations in Table - 52 and Table - 53 are illustrated for reference.

Table - 50. Default Setting in Receive Path

Function Block	Default Setting Description	
Line Interface	• The LRDn inputs Non-Return to Zero (NRZ) data and is sampled on rising edge of the LRCKn.	
	<ul> <li>The RJAT Clock Divisors (N1, N2) are set to '2F'.</li> </ul>	
Frame Processor	Super Frame (SF) format is enabled.	
HDLC Receiver #1, #2	RHDLCs are disabled.	
Receive System Interface	In Recieve Clock Slave External Signaling Mode.	
	<ul> <li>The data on the RSDn, RSSIGn pins are updated on falling edge of the RSCCK.</li> </ul>	
	RSCFS indicates each F-bit.	
	<ul> <li>The data on the RSDn, RSSIGn, RSFSn pins are held in high-impedance state.</li> </ul>	
Receive Payload Control	• The RPLC is disabled.	
PRGD	The PRGD is configured to monitor the extracted data patterns in Frame One.	

Table - 51. Default Setting in Transmit Path

Function Block	Default Setting Description		
PRGD	The PRGD is configured to insert test patterns to Frame One.		
Transmit System Interface	In Transmit Clock Slave External Signaling Mode.		
	The data on the TSDn and TSSIGn pins are sampled on rising edge of TSCCKB.		
Transmit Payload Control	The TPLC is disabled.		
Frame Generator	Super Frame (SF) format is enabled.		
HDLC Transmitter #1, #2	THDLCs are disabled.		
Bit-Oriented Message Transmitter	• The BOMT is disabled.		
Inband Loop-back Code Generator	The Inband Loop-back Code Generator is disabled.		
Line Interface	• The LTDn outputs Non-Return to Zero (NRZ) data and is updated on falling edge of LTCKn.		
	TJAT Clock Divisors (N1, N2) are set to '2F'		
	• Digital jitter attenuation is enabled. The PLL is synchronized to the TSCCKB clock. The smoothed clock		
	output from the PLL is selected as the LTCKn.		

Table - 52. Various Operation Modes in Receive Path for Reference

Mod	е	Register 1	Value (from Bit7 to Bit0)	Description <sup>2</sup>
		001H	1000000	In Receive Clock Slave RSCK Reference mode.
Receive C	lock	003H	00010011	Enable the normal operation of the RSDn pin. The data on the RSDn are
				updated on the rising edge of the RSCCK. The data on the RSCFS are
Slave RS0	CK			sampled on the falling edge of the RSCCK.
		020H	00110000	The Frame Processor is set in ESF format. The CRC-6 calculation is
Reference	Mode			performed when mimic framing pattern is present.
		02CH	00010000	The Alarm Detector is set in ESF format.
		040H	00000100	The Receive CAS/RBS Buffer is set in ESF format.
		001H	11000000	In Receive Clock Slave External Signaling mode. The backplane rate is
Receive				1.544Mbit/s.
		003H	00010011	Enable the normal operation of the RSDn and RSSIGn pins. The data on the
	(1.544M			RSDn and RSSIGn are updated on the rising edge of the RSCCK. The data
Clock	`			on the RSCFS are sampled on the falling edge of the RSCCK.
	bit/s)	020H	00110000	The Frame Processor is set in ESF format. The CRC-6 calculation is
	,			performed when mimic framing pattern is present.
Slave		02CH	00010000	The Alarm Detector is set in ESF format.
		040H	00000100	The Receive CAS/RBS Buffer is set in ESF format.
		001H	11010000	In Receive Clock Slave External Signaling mode. The backplane rate is
External				2.048Mbit/s.
		003H	00010011	Enable the normal operation of the RSDn and RSSIGn pins. The data on the
	(2.048M			RSDn and RSSIGn are updated on the rising edge of the RSCCK. The data
Signaling				on the RSCFS are sampled on the falling edge of the RSCCK.
ŭ ŭ	bit/s)	020H	00110000	The Frame Processor is set in ESF format. The CRC-6 calculation is
	,			performed when mimic framing pattern is present.
Mode		02CH	00010000	The Alarm Detector is set in ESF format.
		040H	00000100	The Receive CAS/RBS Buffer is set in ESF format.
		001H	01000000	In Receive Clock Master Full T1/J1 mode.
Receive C	lock	003H	00010000	Enable the normal operation of the RSDn pin. The data on the RSDn and
Master Fu	II			RSFSn are updated on the rising edge of the RSCK.
T1/J1 Mod	de	020H	00000000	The Frame Processor is set in SF format.
		02CH	00000000	The Alarm Detector is set in SF format.
		040H	00000000	The Receive CAS/RBS Buffer is set in SF format.
		001H	00000000	In Receive Clock Master Nx64k mode.
Receive C	lock	003H	00010000	Enable the normal operation of the RSDn pin. The data on the RSDn and
				RSFSn are updated on the rising edge of the RSCK.
		020H	00110000	The Frame Processor is set in ESF format. The CRC-6 calculation is
Master Fra	actional			performed when mimic framing pattern is present.
		02CH	00010000	The Alarm Detector is set in ESF format.
		040H	00000100	The Receive CAS/RBS Buffer is set in ESF format.
T1/J1 Mod	de	050H	00000001	Enable the Receive Payload Control.
1 1/3 I Mode		01H-18H (RPLC	01000000	The code in the DTRK[7:0] replaces the data output on the RSDn pin in the
		Indirect Registers)		corresponding channel.

Table - 52. Various Operation Modes in Receive Path for Reference (Continued)

Mode	Register <sup>1</sup>	Value (from Bit7 to Bit0)	Description <sup>2</sup>	
	001H	11001000		
	081H	11001000		
	101H	11001000		
	181H	11001000	In Receive Multiplexed mode. The receive backplane rate is 8.192Mbit/s.	
	201H	11001000		
	281H	11001000		
	301H	11001000		
	381H	11001000		
	003H	01010011	Multiplex the data stream of these four framers to the multiplexed bus 1. Enable the	
'	083H	01010011	normal operation of the MRSD and MRSSIG pins. The data on the MRSD and MRSSIG	
	103H	01010011	are updated on the rising edge of the MRSCCK. The data on the MRSCFS are sampled	
Receive	183H	01010011	on the falling edge of the MRSCCK.	
	203H	11010011	Multiplex the data stream of these four framers to the multiplexed bus 2. Enable the	
	283H	11010011	normal operation of the MRSD and MRSSIG pins. The data on the MRSD and MRSSIG	
	303H	11010011	are updated on the rising edge of the MRSCCK. The data on the MRSCFS are sampled	
	383H	11010011	on the falling edge of the MRSCCK.	
	077H	00000000	TSOFF[6:0] = 0. The timeslot offset is 0.	
	0F7H	0000001	TSOFF[6:0] = 1. The timeslot offset is 1.	
	177H	0000010	TSOFF[6:0] = 2. The timeslot offset is 2.	
	1F7H	0000011	TSOFF[6:0] = 3. The timeslot offset is 3.	
·	277H	00000000	TSOFF[6:0] = 0. The timeslot offset is 0.	
	2F7H	0000001	TSOFF[6:0] = 1. The timeslot offset is 1.	
	377H	0000010	TSOFF[6:0] = 2. The timeslot offset is 2.	
Multiplexed	3F7H	0000011	TSOFF[6:0] = 3. The timeslot offset is 3.	
	020H	00110000		
	0A0H	00110000	The Frame Processor is set in ESF format. The CRC-6 calculation is performed when mimic framing pattern is present.	
	120H	00110000		
	1A0H	00110000		
	220H	00110000		
	2A0H	00110000		
	320H	00110000		
	3A0H	00110000		
	02CH	00010000		
	0ACH	00010000		
	12CH	00010000		
Mode	1ACH	00010000	The Alarm Detector is set in ESF format.	
	22CH	00010000		
	2ACH	00010000		
	32CH	00010000		
	3ACH	00010000		
	040H	00000100		
	0C0H	00000100	The Receive CAS/RBS Buffer is set in ESF format.	
	140H	00000100		
	1C0H	00000100		
	240H	00000100		
-	2C0H	00000100		
	340H	00000100		
	3C0H	00000100		

Table - 53. Various Operation Modes in Transmit Path for Reference

Mode		Register 1	Value (from Bit7 to Bit0)	Description <sup>2</sup>
Transmit Clock		004H	00001000	The data on the TSDn and TSCFS pins are sampled on the falling edge of the
				TSCCKB. The data on the TSFSn pin are updated on the falling edge of the TSCCKB
		005H	10000000	In Transmit Clock Slave TSFS Enabled mode. The backplane rate is 1.544Mbit/s.
Slave TSFS		044H	00010000	The Frame Generator is set in ESF format.
		007H	00001101	TSCCKB is selected as TJAT input reference clock. Smoothed clock is selected
Enable Mode				as Line Transmit Clock (LTCK).
		01BH	00010000	The FIFO is set to self-center its read pointer.
Transmit		004H	00001000	The data on the TSDn, TSSIGn and TSCFS pins are sampled on the falling edge
				of the TSCCKB.
		005H	11000000	In Transmit Clock Slave External Signaling mode. The backplane rate is
1	(1.544M			1.544Mbit/s.
Clock	bit/s)	044H	00010000	The Frame Generator is set in ESF format.
		007H	00001101	TSCCKB is selected as TJAT input reference clock. Smoothed clock is selected
l				as Line Transmit Clock (LTCK).
Slave		01BH	00010000	The FIFO is set to self-center its read pointer.
		004H	00001000	The data on the TSDn, TSSIGn and TSCFS pins are sampled on the falling edge
External		22-11	44000400	of the TSCCKB.
		005H	11000100	In Transmit Clock Slave External Signaling mode. The backplane rate is
	(0.04014	04411	00040000	2.048Mbit/s.
lo: 1:	(2.048M	044H	00010000	The Frame Generator is set in ESF format.
Signaling	bit/s)	007H	00001101	TSCCKB is selected as TJAT input reference clock. Smoothed clock is selected
		04011	444444	as Line Transmit Clock (LTCK).
Mode		019H	11111111	Set the Reference Clock Divisor(N1) to 255.
		01AH	11000000	Set the Output Clock Divisor(N2) to 192.
		01BH	00010000	The FIFO is set to self-center its read pointer.
Transmit Clock		004H	00000110	The data on the TSFSn pin are updated on the rising edge of the LTCK. The data
Transmit C	JIOCK	005H	01000000	on the TSDn pin are sampled on the falling edge of the LTCK.  In Transmit Clock Master Full T1/J1 mode. The backplane rate is 1.544Mbit/s.
MostorMe	ndo.	005H 044H	0000000	The Frame Generator is set in SF format.
Master Mode		007H	00100100	XCK/24 is selected as TJAT input reference clock and Line Transmit Clock
		00711	00100100	(LTCK).
		004H	00001000	(LTOR).
Transmit		084H	00001000	The data on the TSDn, TSSIGn and TSCFS pins are sampled on the falling edge of TSCCKB.
		104H	00001000	
		184H	00001000	
		204H	00001000	
ı		284H	00001000	of records.
1		304H	00001000	
Multiplexed		384H	00001000	
	-	005H	11001100	
		085H	11001100	In Transmit Multiplexed mode. The backplane rate is 8.192Mbit/s.
		105H	11001100	
Mode		185H	11001100	
		205H	11001100	
		285H	11001100	
		305H	11001100	

Table - 53. Various Operation Modes in Transmit Path for Reference (Continued)

Mode	Register 1	Value (from Bit7 to Bit0)	Description <sup>2</sup>	
	014H	00000000	TSOFF[6:0] = 0. The timeslot offset is 0.	
	094H	00000000		
	114H	0000001	TSOFF[6:0] = 1. The timeslot offset is 1.	
	194H	0000001		
	214H	0000010	TSOFF[6:0] = 2. The timeslot offset is 2.	
	294H	0000010		
	314H	0000011	TSOFF[6:0] = 3. The timeslot offset is 3.	
	394H	0000011		
Transmit	015H	00000000	The data stream is taken from the multiplexed bus 1.	
	095H	01000000	The data stream is taken from the multiplexed bus 2.	
	115H	00000000	The data stream is taken from the multiplexed bus 1.	
	195H	01000000	The data stream is taken from the multiplexed bus 2.	
	215H	00000000	The data stream is taken from the multiplexed bus 1.	
	295H	01000000	The data stream is taken from the multiplexed bus 2.	
	315H	00000000	The data stream is taken from the multiplexed bus 1.	
	395H	01000000	The data stream is taken from the multiplexed bus 2.	
Multiplexed	044H	00010000		
	0C4H	00010000		
	144H	00010000		
	1C4H	00010000	The Frame Generator is set in ESF format.	
	244H	00010000		
	2C4H	00010000		
	344H	00010000		
	3C4H	00010000		
Mode	007H	00011101		
	087H	00011101		
	107H	00011101		
	187H	00011101	TSCCKA is selected as TJAT input reference clock. Smoothed clock is selected as Line	
	207H	00011101	Transmit Clock (LTCK).	
	287H	00011101		
	307H	00011101		
	387H	00011101		
(Continued)	019H	11111111		
	099H	11111111		
	119H	11111111		
	199H	11111111	Set the Reference Clock Divisor(N1) to 255.	
	219H	11111111		
	299H	11111111		
	319H	11111111		
	399H	11111111		
	01AH	11000000		
	09AH	11000000		
	11AH	11000000		
	19AH	11000000	Set the Output Clock Divisor(N2) to 192.	
	21AH	11000000		
	29AH	11000000		
	31AH	11000000		
	39AH	11000000		

Table - 53. Various Operation Modes in Transmit Path for Reference (Continued)

Mode	Register 1	Value (from Bit7 to Bit0)	Description <sup>2</sup>
Transmit	01BH	00010000	
	09BH	00010000	
Multiplexed	11BH	00010000	
	19BH	00010000	The FIFO is set to self-center its read pointer.
Mode	21BH	00010000	
	29BH	00010000	
(Continued)	31BH	00010000	
	39BH	00010000	

#### Note:

- 1. In the 'Register' column, except for the Receive/Transmit Multiplexed mode, the register position of the Framer One is listed to represent the set of the registers of eight framers. The other registers position are tabulated in the 'Register Map'. However, in Receive/Transmit Multiplexed mode, the register position of eight framers are all listed.
- 2. The 'Description' illustrates the fundamental function of the operation mode. The others can be configured as desired.

# 4.2.4 OPERATION EXAMPLE

In this chapter, some common operation examples are given for reference.

# 4.2.4.1 Using The HDLC Receiver

In T1/J1 mode, the HDLC Receive can only be used in ESF format. Before using the HDLC#2 Receive, the TXCISEL (b3, T1/J1-00DH) must be set to 0 to enable the HDLC data link position for receive path.

Since two HDLC Receive data links are integrated in one framer, one of the two HDLC Receive data links must be selected in the RHDLCSEL[1:0] (b7~6, T1/J1-00DH). The RHDLC #1 can only extract from F-bit of each odd frame. The RHDLC #2 can be set to extract from even and/or odd frames, from any channel, and from any bit. The follow is an example for selecting the HDLC Receive data link positions in RHDLC #2:

a. to extract the HDLC data link from all bits of channel 20 of all framers in HDLC Receive #2:

- set the TXCISEL (b3, T1/J1-00DH) to 0;
- set the RHDLCSEL[1:0] (b7~6, T1/J1-00DH) to 01;
- set the DL2 EVEN (b7, T1/J1-070H) to 1;
- set the DL2\_ODD (b6, T1/J1-070H) to 1;
- set the DL2\_TS[4:0] (b4~0, T1/J1-070H) to 10100;
- set the DL2\_BIT[7:0] (b7~0, T1/J1-071H) to 11111111.

After setting the HDLC data link position properly, the selected HDLC Receive should be enabled by setting the EN (b0, T1/J1-054H)to logic 1. If needed, set the MEN (b3, T1/J1-054H) and the MM (b2, T1/J1-054H) to determine which Address Matching Mode to be selected (refer to Register Description for details). After setting these 3 bits, the RHDLC Primary Address Match register and the RHDLC Secondary Address Match register should be set to proper values. If the INTC[6:0] (b6~0, T1/J1-055H) are set, whenever the number of bytes in the RHDLC FIFO exceeds the value set in the INTC[6:0] (b6~0, T1/J1-055H), the INTR (b0, T1/J1-056H) will be set to logic 1. This interrupt will persist until the RHDLC FIFO becomes empty. Setting the INTE (b7, T1/J1-055H) to logic 1 allows the internal interrupt status to be propagated to the INT output pin.

After setting these registers properly, the HDLC data can be received in a polled or interrupt driven mode.

### - Interrupt Driven Mode

When the INTE (b7, T1/J1-055H) is set to logic 1, if the INT pin is asserted, the source of the interrupt should be identified firstly by reading the Interrupt ID register and Interrupt Source registers. If the source of the interrupt is HDLC Receive, the Interrupt Service procedure will be carried out as shown in Figure - 77.

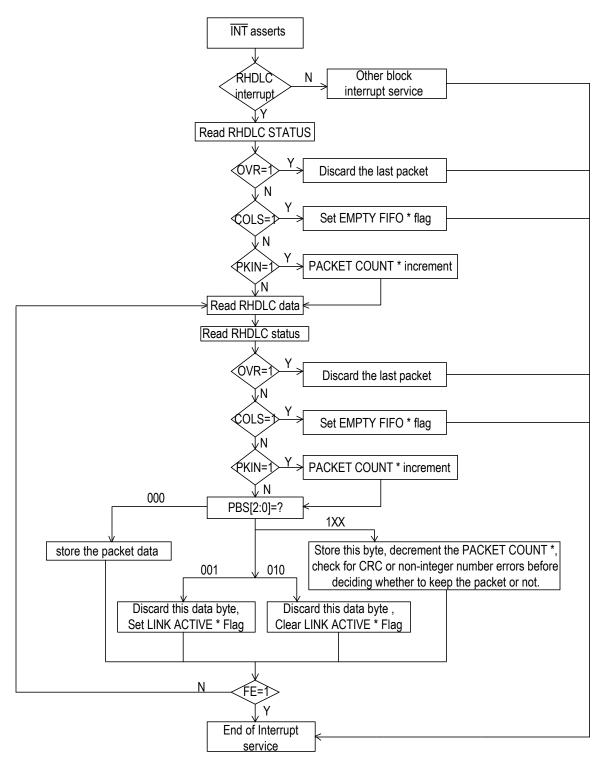
### - Polling Mode

In polling mode, the operation procedure is the same as Figure - 77, except that the entry of the service is from a local timer rather than an interrupt.

To summarize the procedure of using HDLC Receive, a complete example is shown in Table - 54.

Table - 54. Example for Using HDLC Receiver

Register	Value	Description	
00DH	50H	RHDLC #2 is selected. The HDLC Receive is	
		accessable to the CPU interface.	
070H	C4H	The TS4 of even frames and odd frames are	
		selected.	
071H	FFH	All the 8 bits are selected.	
054H	0DH	The function of the RHDLC #2 is enabled. Set	
		the address match mode.	
055H	8FH	Set the INTE to 1. When the number of bytes in	
		the RHDLC FIFO exceeds 15, an interrupt is	
		generated.	
058H	13H	The primary address is set to 13H.	
059H	FFH	The secondary address is set to FFH.	
Then read	Then read the data status in register 056H. Until a complete packet is		
received, read the data from register 057H.			



# Note:

Figure - 77. Interrupt Service in T1/J1 Mode HDLC Receiver

<sup>\*</sup> The PACKET COUNT ,EMPTY FIFO and LINK ACTIVE is a local software variable

#### 4.2.4.2 Using The HDLC Transmitter

In T1/J1 mode, the HDLC Transmit can only be used in ESF format. Before using the HDLC#2 Transmit, the TXCISEL (b3, T1/J1-00DH) must be set to 1 to enable the HDLC data link position for transmit path.

Since two HDLC Transmit data links are integrated in one framer, one of the two HDLC Transmit data links must be selected in the THDLCSEL[1:0] (b5~4, T1/J1-00DH). The THDLC #1 can only insert to F-bit of each odd frame. The THDLC #2 can be set to insert to even and/ or odd frames, to any channel, and to any bit. The follow is an example for selecting the HDLC Transmit data link positions in THDLC #2:

- a. to insert the HDLC data link to all bits of channel 20 of all framers in HDLC Transmit #2:
  - set the TXCISEL (b3, T1/J1-00DH) to 1;
  - set the THDLCSEL[1:0] (b5~4, T1/J1-00DH) to 01;
  - set the DL2\_EVEN (b7, T1/J1-070H) to 1;
  - set the DL2\_ODD (b6, T1/J1-070H) to 1;
  - set the DL2\_TS[4:0] (b4~0, T1/J1-070H) to 10100;
  - set the DL2\_BIT[7:0] (b7~0, T1/J1-071H) to 11111111.

After setting the HDLC data link position properly, the selected HDLC Transmit should be enabled by setting the EN (b0, T1/J1-034H) to logic 1. The FIFOCLR (b6, T1/J1-034H) should be set and then cleared to initialize the THDLC FIFO.

Set the CRC (b1, T1/J1-034H) to logic 1 if the Frame Check Sequences (FCS) generation is desired. Set the FULLE (b3, T1/J1-037H), OVRE (b2, T1/J1-037H), UDRE (b1, T1/J1-037H) and LFILLE (b0, T1/J1-037H) to logic 1 if interrupt driven mode is used. Set THDLC Upper Transmit Threshold and THDLC Lower Transmit Threshold registers to the desired values. If a complete packet has been written into THDLC FIFO, the EOM (b3, T1/J1-034H) should be set.

After setting these registers properly, the HDLC data can be transmitted in a polled or interrupt driven mode.

# - Interrupt Driven Mode

Writing HDLC data to THDLC FIFO, the THDLC will transmit the HDLC data if the end of a packet was written or if the THDLC FIFO fill level reaches the Upper Transmit Threshold. The writing procedure is shown in Figure - 78.

When the FULLE (b3, T1/J1-037H), OVRE (b2, T1/J1-037H), UDRE (b1, T1/J1-037H) and LFILLE (b0, T1/J1-037H) are set to logic 1, if the INT pin is asserted, the source of the interrupt should be identified firstly by reading the Interrupt ID register and Interrupt Source registers. If the source of the interrupt is HDLC Transmit, the Interrupt Service procedure will be carried out as shown in Figure - 79.

### - Polling Mode

In packet transmission polling mode, the FULLE (b3, T1/J1-037H), OVRE (b2, T1/J1-037H), UDRE (b1, T1/J1-037H) and LFILLE (b0, T1/J1-037H) should be set to logic 0. The THDLC Lower Transmit Threshold should be set to such a value that sufficient warning of an underrun is given. The procedure shown in Figure - 80 should be followed.

To summarize the procedure of using HDLC Transmit, a complete example is shown in Table - 55.

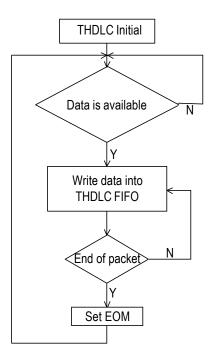
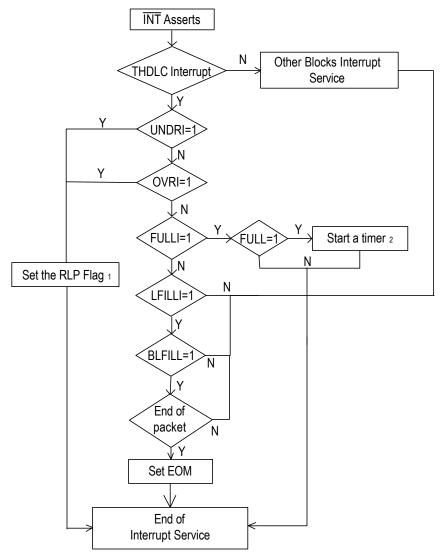


Figure - 78. Writing Data to T1/J1 Mode THDLC FIFO

Table - 55. Example for Using HDLC Transmitter

Register	Value	Description
_		,
00DH	58H	THDLC #2 is selected. The HDLC Transmit is
		accessable to the CPU interface.
070H	C4H	The TS4 of even frames and odd frames are
		selected.
071H	FFH	All the 8 bits are selected.
034H	C3H	The function of the THDLC #2 is enabled. The
034H	83H	FCS is enables and the THDLC FIFO is reset.
037H	0FH	Enable the THDLC Interrupt Enable bits.
039H	12H	
039H	34H	
039H	56H	
039H	78H	Write data into THDLC FIFO.
039H	9AH	
039H	BCH	
039H	DEH	
039H	FFH	
034H	8BH	End of packet and set the EOM to 1.



# Note:

- 1. RLP-Retransmit the last packet, a local software variable.
- 2. A local timer to wait for a certain time until the Full = 0 or the BLFILL = 1.

Figure - 79. Interrupt Service in T1/J1 Mode HDLC Transmitter

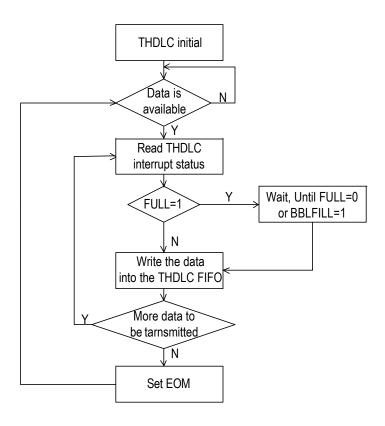


Figure - 80. Polling Mode in T1/J1 Mode HDLC Transmitter

# 4.2.4.3 Using The PRBS Generator / Detector

IDT82V2108 provide one PRBS generator/detector block to generate and detect an enormous variety of pseudo-random and repetitive patterns to diagnose T1/J1 data stream of eight framers. The common test patterns are tabulrized in Table - 56.

The PRBS generator/detector block can be used to test T1/J1 line transmit-receive integrity and system backplane integrity.

- Example For Testing T1/J1 Line Transmit-Receive Integrity

Suppose to monitor the errors in Framer Two without taking the entire T1/J1 offline. Following procedure should be done.

- Select Framer Two to be tested by the PRGD block;
- Configure the PRGD register;
- Chose a desired set of channels (for example CH2, CH4, CH5) for insert/extract PRGD test data;
- Set the far end of the line to loop back at least the selected channels;
  - Monitor the T1/J1 line transmit-receive integrity.

To realize the above function, the configuration in Table - 57 to Table - 59 should be set.

Table - 57 is the configuration for PRGD and loopback. Table - 58 shows the process to initialize the TPLC. Table - 59 shows the process to initialize the RPLC.

After the above setting, read the 061H register twice. If the SYNCV (b4, T1/J1-061H) is logic 1 and the BEI (b2, T1/J1-061H) is logic 0, the pattern detector is in synchronization state.

Then insert errors into this link. Here suppose to insert 3 errors, then the configuration is shown in Table - 60.

- Example For Testing T1/J1 System Backplane Integrity

To test the T1/J1 system backplane integrity, the RXPATGEN (b2, T1/J1-00FH) should be set to logic 1 and the other registers are set as above. Then the PRGD can be used to test the system backplane integrity.

Table - 56. Test Pattern

	Pseudo-Ran	dom Patterr	n Generation	(the PS [b4	, T1/J1-060H	] is logic 0)		
Pattern Type	TR <sup>1</sup>	LR <sup>2</sup>	IR#1 <sup>3</sup>	IR#2 <sup>4</sup>	IR#3 <sup>5</sup>	IR#4 <sup>6</sup>	TINV 7	RINV 7
2 <sup>3</sup> -1	00	02	FF	FF	FF	FF	0	0
2 <sup>4</sup> -1	00	03	FF	FF	FF	FF	0	0
2 <sup>5</sup> -1	01	04	FF	FF	FF	FF	0	0
2 <sup>6</sup> -1	04	05	FF	FF	FF	FF	0	0
2 <sup>7</sup> –1 (Fractional T1 LB Activate)	00	06	FF	FF	FF	FF	0	0
2 <sup>7</sup> –1 (Fractional T1 LB Activate)	03	06	FF	FF	FF	FF	0	0
27-1	03	06	FF	FF	FF	FF	1	1
2 <sup>9</sup> –1 (O.153)	04	08	FF	FF	FF	FF	0	0
2 <sup>10</sup> -1	02	09	FF	FF	FF	FF	0	0
2 <sup>11</sup> -1 (O.152,O.153)	08	0A	FF	FF	FF	FF	0	0
2 <sup>15</sup> -1 (O.151)	0D	0E	FF	FF	FF	FF	1	1
2 <sup>17</sup> -1	02	10	FF	FF	FF	FF	0	0
2 <sup>18</sup> -1	06	11	FF	FF	FF	FF	0	0
2 <sup>20</sup> -1 (O.153)	02	13	FF	FF	FF	FF	0	0
2 <sup>21</sup> -1	01	14	FF	FF	FF	FF	0	0
2 <sup>22</sup> -1	00	15	FF	FF	FF	FF	0	0
2 <sup>23</sup> -1 (0.151)	11	16	FF	FF	FF	FF	1	1
2 25 -1	02	18	FF	FF	FF	FF	0	0
2 <sup>28</sup> -1	02	1B	FF	FF	FF	FF	0	0
2 <sup>29</sup> -1	01	1C	FF	FF	FF	FF	0	0
2 <sup>31</sup> -1	02	1E	FF	FF	FF	FF	0	0

# Repetitive Pattern Generation (the PS [b4, T1/J1-060H] is logic 1)

Pattern Type	TR <sup>1</sup>	LR <sup>2</sup>	IR#1 <sup>3</sup>	IR#2 <sup>4</sup>	IR#3 <sup>5</sup>	IR#4 <sup>6</sup>	TINV 7	RINV 7
All ones	00	00	FF	FF	FF	FF	0	0
All zeros	00	00	FE	FF	FF	FF	0	0
Alternatingones/zeros	00	01	FE	FF	FF	FF	0	0
Double alternatingones/zeros	00	03	FC	FF	FF	FF	0	0
3 in 24	00	17	22	00	20	FF	0	0
1 in 16	00	0F	01	00	FF	FF	0	0
1 in 8	00	07	01	FF	FF	FF	0	0
1 in 4	00	03	F1	FF	FF	FF	0	0
DS1 Inbandloopback activate	00	04	F0	FF	FF	0F	0	0
DS1 InbandLoopback deactivate	00	02	FC	FF	FF	FF	0	0

# Note:

- 1. TR Tap Register
- 2. LR Shift Register Length Register
- 3. IR#1 PRGD Pattern Insertion #1 Register
- 4. IR#2 PRGD Pattern Insertion #2 Register
- 5. IR#3 PRGD Pattern Insertion #3 Register
- 6. IR#4 PRGD Pattern Insertion #4 Register
- 7. TINV, RINV contained in the PRGD Control register

Table - 57. The Setting of PRGD

Description Register Value 00FH 20H Select Framer 2 to be tested by the PRGD block. The PRGD pattern is inserted in the TPLC and detected in the RPLC. 060H 82H Set Pattern Detector registers as error counter register. Enable automatic resynchronization. 18H Set the pattern length. 062H 02H Set the feedback tap position. 063H 068H FFH Set the Pattern Insertion registers. Load the data in the Pattern Insertion registers to 06BH FFH generate the pattern. HA80 04H Set diagnostic digital loopback mode. 0B0H 01H Enable the TPLC indirect registers to be

Enable the RPLC indirect registers to be

Table - 58. Initializtion of TPLC

01H

0D0H

accessable.

Register	Value
0B3H	00H
0B2H	01H
0B3H	00H
0B2H	02H
0B3H	00H
0B2H	03H
0B3H	00H
0B2H	04H
0B3H	00H
0B2H	05H
0B3H	00H
0B2H	06H
0B3H	00H
0B2H	07H
0B3H	00H
0B2H	08H
0B3H	00H
0B2H	09H
0B3H	00H
0B2H	0AH
0B3H	00H
0B2H	0BH
0B3H	00H
0B2H	0CH
0B3H	00H
0B2H	0DH
0B3H	00H
0B2H	0EH
0B3H	00H

Table - 58. Initializtion of TPLC (Continued)

Register	Value
0B2H	0FH
0B3H	00H
0B2H	10H
0B3H	00H
0B2H	11H
0B3H	00H
0B2H	12H
0B3H	00H
0B2H	13H
0B3H	00H
0B2H	14H
0B3H	00H
0B2H	15H
0B3H	00H
0B2H	16H
0B3H	00H
0B2H	17H
0B3H	00H
0B2H	18H
0B3H	00H
0B2H	31H
0B3H	00H
0B2H	32H
OB3H	00H
0B2H	33H
OB3H	00H 34H
0B2H	
OB3H	00H
0B2H	35H
0B3H	00H
0B2H	36H
0B3H	00H
0B2H	37H
0B3H	00H
0B2H	38H
0B3H	00H
0B2H	39H
0B3H	00H
0B2H	3AH
0B3H	00H
0B2H	3BH
0B3H	00H
0B2H	3CH
0B3H	00H
0B2H	3DH
0B3H	00H
0B2H	3EH
0B3H	00H

Table - 58. Initializtion of TPLC (Continued)

Table - 59. Initializtion of RPLC (Continued)

Register	Value
0B2H	3FH
0B3H	00H
0B2H	40H
0B3H	00H
0B2H	41H
0B3H	00H
0B2H	42H
0B3H	00H
0B2H	43H
0B3H	00H
0B2H	44H
0B3H	00H
0B2H	45H
0B3H	00H
0B2H	46H
0B3H	00H
0B2H	47H
0B3H	00H
0B2H	48H

Then set the TEST in TPLC Payload Control register for CH2, CH4 and CH5. The process is:

Register	Value	Description
0B3H	08H	Set the TEST in TPLC Payload Control register
0B2H	02H	for CH2.
0B3H	08H	Set the TEST in TPLC Payload Control register
0B2H	04H	for CH4.
0B3H	08H	Set the TEST in TPLC Payload Control register
0B2H	05H	for CH5.

Table - 59. Initializtion of RPLC

Register	Value
0D3H	00H
0D2H	01H
0D3H	00H
0D2H	02H
0D3H	00H
0D2H	03H
0D3H	00H
0D2H	04H
0D3H	00H
0D2H	05H
0D3H	00H
0D2H	06H
0D3H	00H
0D2H	07H

Register	Value
0D3H	00H
0D2H	08H
0D3H	00H
0D2H	09H
0D3H	00H
0D2H	0AH
0D3H	00H
0D2H	0BH
0D3H	00H
0D2H	0CH
0D3H	00H
0D2H	0DH
0D3H	00H
0D2H	0EH
0D3H	00H
0D2H	0FH
0D3H	00H
0D2H	10H
0D3H	00H
0D2H	11H
0D3H	00H
0D2H	12H
0D3H	00H
0D2H	13H
0D3H	00H
0D2H	14H
0D3H	00H
0D2H	15H
0D3H	00H
0D2H	16H
0D3H	00H
0D2H	17H
0D3H	00H
0D2H	18H

Then set the TEST in RPLC Payload Control register for CH2, CH4 and CH5. The process is:

Register	Value	Description
0D3H	08H	Set the TEST in RPLC Payload Control
0D2H	02H	register for CH2.
0D3H	08H	Set the TEST in RPLC Payload Control
0D2H	04H	register for CH4
0D3H	08H	Set the TEST in RPLC Payload Control
0D2H	05H	register for CH5

Table - 60. Error Insertion

08H	
00H	
08H	
00H	
08H	
00H	
	00H 08H 00H 08H

Then write 00H into the 06CH register to update the error counter registers. Then read the registers from 06CH to 06FH to check the error numbers.

# 4.2.4.4 Using Payload Control and Receive CAS/RBS Buffer

Before using the Receive/Transmit Payload Control and Receive CAS/RBS Buffer, the indirect registers of these blocks must be initialized to eliminate erroneous control data. The the PCCE (b0, T1/J1-050H & b0, T1/J1-030H & b0, T1/J1-040H) of these blocks must be set to logic 1 to enable these blocks.

Then the BUSY (b7, T1/J1-051H & b7, T1/J1-031H & b7, T1/J1-041H) must be checked before a new access request to the RPLC, TPLC and RCRB indirect registers. When the BUSY is logic 0, the new reading and writing access operations can be performed.

Figure - 81 shows the writing sequence of the RPLC, TPLC and RCRB indirect registers. Figure - 82 shows the reading sequence of the RPLC, TPLC and RCRB indirect registers.

### 4.2.4.5 Using TJAT / Timing Option

In different operation modes, the Timing Options and Clock Divisor Control registers can be set as the follows:

- Transmit Clock Slave Mode (System Backplane Rate: 1.544M bit/s)

The TSCCKA or TSCCKB is selected as the TJAT DPLL input reference clock. The TSCCKA and TSCCKB are both equal to 1.544M. The N1 ( $b7\sim0$ , T1/J1-019H) and N2 ( $b7\sim0$ , T1/J1-01AH) are set to their default value (2FH).

The smoothed clock output from the TJAT is selected as the LTCK.

- Transmit Clock Slave Mode (System Backplane Rate: 2.048M bit/s) The TSCCKA or TSCCKB is selected as the TJAT DPLL input reference clock. The TSCCKA and TSCCKB are both equal to 2.048M. The N1 (b7~0, T1/J1-019H) is set to 'b111111111 and the N2 (b7~0, T1/J1-01AH) is set to 'b11000000.

The smoothed clock output from the TJAT is selected as the LTCK.

- Transmit Clock Slave Mode (System Backplane Rate: 4.096M bit/s)
The TSCCKA is selected as the TJAT DPLL input reference clock.
The TSCCKA is equal to 2.048M. The N1 (b7~0, T1/J1-019H) is set to 'b11111111 and the N2 (b7~0, T1/J1-01AH) is set to 'b11000000.

The smoothed clock output from the TJAT is selected as the LTCK.

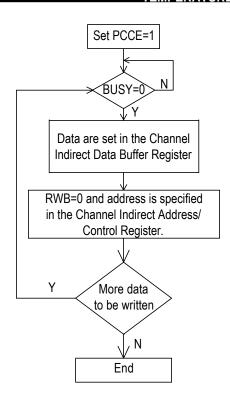


Figure - 81. Writing Sequence of Indirect Register in T1/J1 Mode

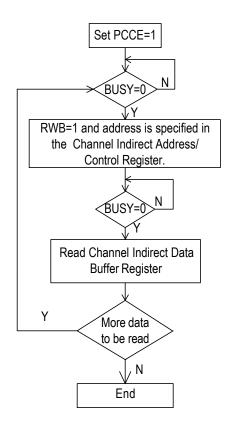


Figure - 82. Reading Sequence of Indirect Register in T1/J1 Mode

- Transmit Clock Master Mode
   The XCK/24 is selected as the TJAT DPLL input reference clock.

   The XCK/24 is selected as the LTCK.
- Transmit Multiplexed Mode (System Backplane Rate: 8.192M bit/s)
  The TSCCKA is selected as the TJAT DPLL input reference clock.
  The TSCCKA is equal to 2.048M. The N1 (b7~0, T1/J1-019H) is set to 'b11111111 and the N2 (b7~0, T1/J1-01AH) is set to 'b11000000.

The smoothed clock output from the TJAT is selected as the LTCK.

- Transmit Multiplexed Mode (System Backplane Rate: 16.384M bit/s)
The TSCCKA or TSCCKA/8 is selected as the TJAT DPLL input reference clock. The TSCCKA is equal to 2.048M or 16.384M. The N1 (b7~0, T1/J1-019H) is set to 'b111111111 and the N2 (b7~0, T1/J1-01AH) is set to 'b11000000.

The smoothed clock output from the TJAT is selected as the LTCK.

# 5 PROGRAMMING INFORMATION

The Micro-Processor Interface provides the logic to connect the microprocessor interface. For all accesses,  $\overline{\text{CS}}$  must be low. The data bus and address bus of the interface can work in multiplexed on non-multiplexed mode. In non-multiplexed mode, ALE pin should be connected to high. In multiplexed mode, data bus and address bus should be externally connected.

### 5.1 REGISTER MAP

The registers are devided into two parts: E1 part and T1/J1 part. Before operation, the TEMODE (b0, 400H) must be set to specify which part to be accessed by the microprocessor.

Table - 61. T1/E1 Mode Selection Register

Address	Register
400	T1 / E1 Mode Selection

# 5.1.1 E1 MODE REGISTER MAP

When the TEMODE (b0, 400H) is logic 0, the E1 mode registers are accessed.

Table - 62a. E1 Mode Register Map - Direct Register

			E1 A	ddress				Register		
Framer 1	Framer 2	Framer 3	Framer4	Framer 5	Framer 6	Framer 7	Framer8	•		
000	080	100	180	200	280	300	380	Receive Path Line Options		
001	081	101	181	201	281	301	381	Receive Side System Interface Options		
002	082	102	182	202	282	302	382	Transmit Path Configuration		
003	083	103	183	203	283	303	383	Transmit Side System Interface Options		
004	084	104	184	204	284	304	384	Transmit Timing Options		
005	085	105	185	205	285	305	385	Interrupt Source #1		
006	086	106	186	206	286	306	386	Interrupt Source #2		
007	087	107	187	207	287	307	387	Diagnostic		
800	088	108	188	208	288	308	388	Reserved		
			00	)9				Chip ID/ Global PMON Update		
00A	A80	10A	18A	20A	28A	30A	38A	HDLC Micro Select/Framer Reset		
			0	0B				Framer Interrupt ID		
			00	)C				PRGD Positioning/control		
00D	08D	10D	18D	20D	28D	30D	38D	Clock Monitor		
00E	08E	10E	18E	20E	28E	30E	38E	Receive Path Frame Pulse Configuration		
00F	08F	10F	18F	20F	28F	30F	38F	Reserved		
010	090	110	190	210	290	310	390	RESI Configuration		
011	091	111	191	211	291	311	391	RESI Frame Pulse Configuration		
012	092	112	192	212	292	312	392	RESI Parity Configuration		
013	093	113	193	213	293	313	393	RESI Timeslot Offset		
014	094	114	194	214	294	314	394	RESI Bit Offset		
015~017	095~097	115~117	195~197	215~217	295~297	315~317	395~397	Reserved		
018	098	118	198	218	298	318	398	TRSI Configuration		
019	099	119	199	219	299	319	399	TRSI Frame Pulse Configuration		
01A	09A	11A	19A	21A	29A	31A	39A	TRSI Parity Configuration and Status		
01B	09B	11B	19B	21B	29B	31B	39B	TRSI Timeslot Offset		
01C	09C	11C	19C	21C	29C	31C	39C	TRSI Bit Offset		
01D~01F	09D~09F	11D~11F	19D~19F	21D~21F	29D~29F	31D~31F	39D~39F	Reserved		
020	0A0	120	1A0	220	2A0	320	3A0	RJAT Interrupt Status		
021	0A1	121	1A1	221	2A1	321	3A1	RJAT Reference Clock Divisor(N1)		
022	0A2	122	1A2	222	2A2	322	3A2	RJAT Reference Clock Divisor(N2)		
023	0A3	123	1A3	223	2A3	323	3A3	RJAT Configuration		
024	0A4	124	1A4	224	2A4	324	3A4	TJAT Interrupt Status		
025	0A5	125	1A5	225	2A5	325	3A5	TJAT Reference Clock Divisor(N1)		
026	0A6	126	1A6	226	2A6	326	3A6	TJAT Reference Clock Divisor(N2)		
027	0A7	127	1A7	227	2A7	327	3A7	TJAT Configuration		
028	0A8	128	1A8	228	2A8	328	3A8	RHDLC 1(TXCISEL=0) Link Control/		
								THDLC 1(TXCISEL=1) Link Control		
029	0A9	129	1A9	229	2A9	329	3A9	RHDLC 1(TXCISEL=0) bits select/		
								THDLC 1(TXCISEL=1) bits select		
02A	0AA	12A	1AA	22A	2AA	32A	3AA	RHDLC 2(TXCISEL=0) Link Control/		
								THDLC 2(TXCISEL=1) Link Control		

Table - 62a. E1 Mode Register Map - Direct Register (Continued)

			E1 Ad	dress	Register			
Framer 1	Framer 2	Framer 3	Framer4	Framer 5	Framer 6	Framer 7	Framer8	1
02B	0AB	12B	1AB	22B	2AB	32B	3AB	RHDLC 2(TXCISEL=0) bits select/
								THDLC 2(TXCISEL=1) bits select
02C	0AC	12C	1AC	22C	2AC	32C	3AC	RHDLC 3(TXCISEL=0) Link Control/
								THDLC 3(TXCISEL=1) Link Control
02D	0AD	12D	1AD	22D	2AD	32D	3AD	RHDLC 3(TXCISEL=0) bits select/
								THDLC 3(TXCISEL=1) bits select
02E~02F	0AE~0AF	12E~12F	1AE~1AF	22E~22F	2AE~2AF	32E~32F	3AE~3AF	Reserved
030	0B0	130	1B0	230	2B0	330	3B0	FRMP Frame Alignment Option
031	0B1	131	1B1	231	2B1	331	3B1	FRMP Maintenance Mode Options
032	0B2	132	1B2	232	2B2	332	3B2	FRMP Framing Status Interrupt Enable
033	0B3	133	1B3	233	2B3	333	3B3	FRMP Maintenance/Alarm Status Interrupt Enable
034	0B4	134	1B4	234	2B4	334	3B4	FRMP Framing Status Interrupt Indication
035	0B5	135	1B5	235	2B5	335	3B5	FRMP Maintenance/Alarm Status Interrupt Indication
036	0B6	136	1B6	236	2B6	336	3B6	FRMP Framing Status
037	0B7	137	1B7	237	2B7	337	3B7	FRMP Maintenance/Alarm Status
038	0B8	138	1B8	238	2B8	338	3B8	FRMP TS0 International/National Bits
039	0B9	139	1B9	239	2B9	339	3B9	FRMP CRC Error Counter-LSB
03A	0BA	13A	1BA	23A	2BA	33A	3BA	FRMP CRC Error Counter-MSB/TS16 Extra Bits
03B	0BB	13B	1BB	23B	2BB	33B	3BB	FRMP National Bit Code-word Interrupt Enable
03C	0BC	13C	1BC	23C	2BC	33C	3BC	FRMP National Bit Code-word Interrupts
03D	0BD	13D	1BD	23D	2BD	33D	3BD	FRMP National Bit Code-word
03E	0BE	13E	1BE	23E	2BE	33E	3BE	FRMP Frame pulse/Alarm/V5.2 Link ID Interrupt Enable
03F	0BF	13F	1BF	23F	2BF	33F	3BF	FRMP Frame Pulse/Alarm Interrupts
040	0C0	140	1C0	240	2C0	340	3C0	FRMG Configuration
041	0C1	141	1C1	241	2C1	341	3C1	FRMG Alarm/Diagnostic Control
042	0C2	142	1C2	242	2C2	342	3C2	FRMG International Bits
043	0C3	143	1C3	243	2C3	343	3C3	FRMG Extra Bits
044	0C4	144	1C4	244	2C4	344	3C4	FRMG Interrupt Enable
045	0C5	145	1C5	245	2C5	345	3C5	FRMG Interrupt Status
046	0C6	146	1C6	246	2C6	346	3C6	FRMG National Bit Code-word Select
047	0C7	147	1C7	247	2C7	347	3C7	FRMG National Bit Code-word
048	0C8	148	1C8	248	2C8	348	3C8	RHDLC #1, 2, 3 Configuration
049	0C9	149	1C9	249	2C9	349	3C9	RHDLC #1, 2, 3 Interrupt Control
04A	0CA	14A	1CA	24A	2CA	34A	3CA	RHDLC #1, 2, 3 Status
04B	0CB	14B	1CB	24B	2CB	34B	3CB	RHDLC #1, 2, 3 Data
04C	0CC	14C	1CC	24C	2CC	34C	3CC	RHDLC #1, 2, 3 Primary Address Match
04D	0CD	14D	1CD	24D	2CD	34D	3CD	RHDLC #1, 2, 3 Secondary Address Match
	0CE~0CF				2CE~2CF			Reserved
050	0D0	150	1D0	250	2D0	350	3D0	THDLC #1, 2, 3 Configuration
051	0D1	151	1D1	251	2D1	351	3D1	THDLC #1, 2, 3 Upper Transmit Threshold
052	0D2	152	1D2	252	2D2	352	3D2	THDLC #1, 2, 3 Lower Transmit Threshold
053	0D3	153	1D3	253	2D3	353	3D3	THDLC #1, 2, 3 Interrupt Enable
054	0D4	154	1D4	254	2D4	354	3D4	THDLC #1, 2, 3 interrupt Status
055	0D5	155	1D5	255	2D5	355	3D5	THDLC #1, 2, 3 Transmit Data
	0D6~0D8		1D6~1D8		2D6~2D8			Reserved
059	0D9	159	1D9	259	2D9	359	3D9	ELSB Interrupt Enable/ Status
05A	0DA	15A	1DA	25A	2DA	35A	3DA	ELSB Idle Code

Table - 62a. E1 Mode Register Map - Direct Register (Continued)

			E1 Ad	Idress	Register					
Framer 1	Framer 2	Framer 3	Framer4	Framer 5	Framer 6	Framer 7	Framer8	·		
05B	0DB	15B	1DB	25B	2DB	35B	3DB	Reserved		
05C	0DC	15C	1DC	25C	2DC	35C	3DC	RPLC Configuration		
05D	0DD	15D	1DD	25D	2DD	35D	3DD	RPLC μP Access Status		
05E	0DE	15E	1DE	25E	2DE	35E	3DE	RPLC Channel indirect Address/Control		
05F	0DF	15F	1DF	25F	2DF	35F	3DF	RPLC Channel Indirect Data Buffer		
060	0E0	160	1E0	260	2E0	360	3E0	TPLC Configuration		
061	0E1	161	1E1	261	2E1	361	3E1	TPLC μP Access Status		
062	0E2	162	1E2	262	2E2	362	3E2	TPLC Channel indirect Address/Control		
063	0E3	163	1E3	263	2E3	363	3E3	TPLC Channel Indirect Data Buffer		
064	0E4	164	1E4	264	2E4	364	3E4	RCRB Configuration (COSS=0) /		
								RCRB COSS[30:25] (COSS=1)		
065	0E5	165	1E5	265	2E5	365	3E5	RCRB μP Access Status (COSS=0) /		
								RCRB COSS[24:17] (COSS=1)		
066	0E6	166	1E6	266	2E6	366	3E6	RCRB CH IND Addr/Control (COSS=0) /		
								RCRB COSS[16:9] (COSS=1)		
067	0E7	167	1E7	267	2E7	367	3E7	RCRB CH Indirect Data Buffer (COSS=0) /		
								RCRB COSS[8:1] (COSS=1)		
068	0E8	168	1E8	268	2E8	368	3E8	PMON Interrupt Enable/Status		
069	0E9	169	1E9	269	2E9	369	3E9	PMON FER Count		
06A	0EA	16A	1EA	26A	2EA	36A	3EA	PMON FEBE Count (LSB)		
06B	0EB	16B	1EB	26B	2EB	36B	3EB	PMON FEBE Count (MSB)		
06C	0EC	16C	1EC	26C	2EC	36C	3EC	PMON CRC Count (LSB)		
06D	0ED	16D	1ED	26D	2ED	36D	3ED	PMON CRC Count (MSB)		
06E~06F	0EE~0EF	16E~16F	1EE~1EF	26E~26F	2EE~2EF	36E~36F	3EE~3EF	Reserved		
			07					PRGD Control		
			07	71				PRGD Interrupt Enable/Status		
			07					PRGD Shift Register Length		
			07	73				PRGD Tap		
			07					PRGD Error Insertion		
			075 ^	- 077				Reserved		
			07	78				PRGD Pattern Insertion #1		
			07					PRGD Pattern Insertion #2		
			07					PRGD Pattern Insertion #3		
			07					PRGD Pattern Insertion #4		
				C C				PRGD Pattern Detector #1		
				D .				PRGD Pattern Detector #2		
				7E				PRGD Pattern Detector #3		
			07	7F				PRGD Pattern Detector #4		

Table - 62b. E1 Mode Register Map - Indirect Register

	Address	Register
RPLC	20-3FH	Payload control byte for TS0 to TS 31
Indirect Registers	40-5FH	Data trunk conditioning code for TS0 to TS31
	61-7FH	Signaling trunk conditioning code for TS1 to TS31
TPLC	20-3FH	Payload control byte for TS0 to TS 31
Indirect Registers	40-5FH	Idle code for TS0 to TS31
	61-7FH	Signaling control byte for TS1 to TS31
RCRB	01-1FH / 21-3FH	Signaling data for TS1 to TS31
Indirect Registers	41-5FH	Signaling control for TS1 to TS31

# 5.1.2 T1 / J1 MODE REGISTER MAP

When the TEMODE (b0, 400H) is logic 1, the T1/J1 mode registers are accessed.

Table - 63a. T1/J1 Mode Register Map - Direct Register

			T1 / J1 /	Address	Register					
Framer 1	Framer 2	Framer 3	Framer4	Framer 5	Framer 6	Framer 7	Framer8			
000	080	100	180	200	280	300	380	Receive Line Options		
001	081	101	181	201	281	301	381	Receive Side System Interface Options		
002	082	102	182	202	282	302	382	Back-plane Parity configuration/Status		
003	083	103	183	203	283	303	383	Receive Interface Configuration		
004	084	104	184	204	284	304	384	Transmit Interface Configuration		
005	085	105	185	205	285	305	385	Transmit Side System Interface Options		
006	086	106	186	206	286	306	386	Transmit Framing and Bypass Options		
007	087	107	187	207	287	307	387	Transmit Timing Options		
800	088	108	188	208	288	308	388	Interrupt Source #1		
009	089	109	189	209	289	309	389	Interrupt Source #2		
00A	08A	10A	18A	20A	28A	30A	38A	Diagnostic		
			00	В				Reserved		
			00	С				Chip ID/ Global PMON Update		
00D	08D	10D	18D	20D	28D	30D	38D	HDLC Micro Select/Framer Reset		
			00	E				Framer Interrupt ID		
			00	F				PRGD Positioning/control		
010	090	110	190	210	29F	310	390	RJAT Interrupt Status		
011	091	111	191	211	291	311	391	RJAT Reference Clock Divisor(N1)		
012	092	112	192	212	292	312	392	RJAT Reference Clock Divisor(N2)		
013	093	113	193	213	293	313	393	RJAT Configuration		
014	094	114	194	214	294	314	394	TRSI Timeslot Offset		
015	095	115	195	215	295	315	395	TRSI Bit Offset		
	096~097	116~117	196~197	216~217	296~297	316~317	396~397	Reserved		
018	098	118	198	218	298	318	398	TJAT Interrupt Status		
019	099	119	199	219	299	319	399	TJAT Reference Clock Divisor(N1)		
01A	09A	11A	19A	21A	29A	31A	39A	TJAT Reference Clock Divisor(N2)		
01B	09B	11B	19B	21B	29B	31B	39B	TJAT Configuration		
01C	09C	11C	19C	21C	29C	31C	39C	Reserved		
01D	09D	11D	19D	21D	29D	31D	39D	ELSB Interrupt Enable/ Status		

Table - 63a. T1/J1 Mode Register Map - Direct Register (Continued)

			T1 / J1 A	ddress	Register			
Framer 1	Framer 2	Framer 3		Framer 5	Framer 6	Framer 7	Framer8	1.09.000
01E	09E	11E	19E	21E	29E	31E	39E	ELSB Idle Code
01F	09F	11F	19F	21F	29F	31F	39F	Reserved
020	0A0	120	1A0	220	2A0	320	3A0	T1 FRMP Configuration
021	0A1	121	1A1	221	2A1	321	3A1	T1 FRMP interrupt enable
022	0A2	122	1A2	222	2A2	322	3A2	T1 FRMP interrupt status
023~026		123~126	1A3~1A6	223~226	2A3~2A6	323~326	3A3~3A6	Reserved
027	0A7	127	1A7	227	2A7	327	3A7	Clock Monitor
028~029		128~129	1A8~1A9		2A8~2A9	328~329		Reserved
02A	0AA	12A	1AA	22A	2AA	32A	3AA	RBOM Configuration
02B	0AB	12B	1AB	22B	2AB	32B	3AB	RBOM Code Status
02C	0AC	12C	1AC	22C	2AC	32C	3AC	ALMD Configuration
02D	0AD	12D	1AD	22D	2AD	32D	3AD	ALMD Interrupt Enable
02E	0AE	121	1A1	221	2A1	321	3A1	ALMD Interrupt Status
02F	0AF	12F	1AF	22F	2AF	32F	3AF	ALMD Alarm Detection Status
030	0B0	130	1B0	230	2B0	330	3B0	TPLC Configuration
031	0B1	131	1B1	231	2B1	331	3B1	TPLC μP Access Status
032	0B2	132	1B2	232	2B2	332	3B2	TPLC Channel indirect Address/Control
033	0B3	133	1B3	233	2B3	333	3B3	TPLC Channel Indirect Data Buffer
034	0B4	134	1B4	234	2B4	334	3B4	THDLC #1, 2, 3 Configuration
035	0B5	135	1B5	235	2B5	335	3B5	THDLC #1, 2, 3 Upper Transmit Threshold
036	0B6	136	1B6	236	2B6	336	3B6	THDLC #1, 2, 3 Lower Transmit Threshold
037	0B7	137	1B7	237	2B7	337	3B7	THDLC #1, 2, 3 Interrupt Enable
038	0B8	138	1B8	238	2B8	338	3B8	THDLC #1, 2, 3 interrupt Status
039	0B9	139	1B9	239	2B9	339	3B9	THDLC #1, 2, 3 Transmit Data
03A~03B	0BA~0BB	13A~13B	1BA~1BB	23A~23B	2BA~2BB	33A~33B	3BA~3BB	Reserved
03C	0BC	13C	1BC	23C	2BC	33C	3BC	IBCD Configuration
03D	0BD	13D	1BD	23D	2BD	33D	3BD	IBCD Interrupt Enable/Status
03E	0BE	13E	1BE	23E	2BE	33E	3BE	IBCD Active Code
03F	0BF	13F	1BF	23F	2BF	33F	3BF	IBCD Deactivate Code
040	0C0	140	1C0	240	2C0	340	3C0	RCRB Configuration (COSS=0) /
								RCRB COSS[30:25] (COSS=1)
041	0C1	141	1C1	241	2C1	341	3C1	RCRB μP Access Status (COSS=0) /
								RCRB COSS[24:17] (COSS=1)
042	0C2	142	1C2	242	2C2	342	3C2	RCRB CH IND Addr/Control (COSS=0) /
								RCRB COSS[16:9] (COSS=1)
043	0C3	143	1C3	243	2C3	343	3C3	RCRB CH Indirect Data Buffer (COSS=0) /
								RCRB COSS[8:1] (COSS=1)
044	0C4	144	1C4	244	2C4	344	3C4	FRMG Configuration
045	0C5	145	1C5	245	2C5	345	3C5	FRMG Alarm Transmit
046	0C6	146	1C6	246	2C6	346	3C6	IBCG Configuration
047	0C7	147	1C7	247	2C7	347	3C7	IBCG Loop-back Code
048	0C8	148	1C8	248	2C8	348	3C8	Reserved
049	0C9	149	1C9	249	2C9	349	3C9	PMON Interrupt Enable/Status
04A	0CA	14A	1CA	24A	2CA	34A	3CA	PMON BEE Count (LSB)
04B	0CB	14B	1CB	24B	2CB	34B	3CB	PMON BEE Count (MSB)
04C	0CC	14C	1CC	24C	2CC	34C	3CC	PMON FER Count (LSB)
04D	0CD	14D	1CD	24D	2CD	34D	3CD	PMON FER Count (MSB)

Table - 63a. T1/J1 Mode Register Map - Direct Register (Continued)

		7	Γ1 / J1 Add	Register							
Framer 1	Framer 2	Framer 3	Framer4	Framer 5	Framer 6	Framer 7	Framer8	-			
04E	0CE	14E	1CE	24E	2CE	34E	3CE	PMON OOF Count			
04F	0CF	14F	1CF	24F	2CF	34F	3CF	PMON COFA Count			
050	0D0	150	1D0	250	2D0	350	3D0	RPLC Configuration			
051	0D1	151	1D1	251	2D1	351	3D1	RPLC μP Access Status			
052	0D2	152	1D2	252	2D2	352	3D2	RPLC Channel indirect Address/Control			
053	0D3	153	1D3	253	2D3	353	3D3	RPLC Channel Indirect Data Buffer			
054	0D4	154	1D4	254	2D4	354	3D4	RHDLC #1, 2, 3 Configuration			
055	0D5	155	1D5	255	2D5	355	3D5	RHDLC #1, 2, 3 Interrupt Control			
056	0D6	156	1D6	256	2D6	356	3D6	RHDLC #1, 2, 3 Status			
057	0D7	157	1D7	257	2D7	357	3D7	RHDLC #1, 2, 3 Data			
058	0D8	158	1D8	258	2D8	358	3D8	RHDLC #1, 2, 3 Primary Address Match			
059	0D9	159	1D9	259	2D9	359	3D9	RHDLC #1, 2, 3 Secondary Address Match			
05A~05C	0DA~0DC	15A~15C	1DA~1DC	25A~25C	2DA~2DC	35A~35C	3DA~3DC	Reserved			
05D	0DD	15D	1DD	25D	2DD	35D	3DD	TBOM Code			
05E~05F	0DE~0DF	15E~15F	1DE~1DF	25E~25F	3DE~3DF	Reserved					
			06	0				PRGD Control			
			06	61				PRGD Interrupt Enable/Status			
			06	62				PRGD Shift Register Length			
			06	3				PRGD Tap			
			06	64				PRGD Error Insertion			
			065 ~	067				Reserved			
			06	88				PRGD Pattern Insertion #1			
			06	69				PRGD Pattern Insertion #2			
			06	SA .				PRGD Pattern Insertion #3			
			06	SB .				PRGD Pattern Insertion #4			
			06	SC				PRGD Pattern Detector #1			
			06	SD.				PRGD Pattern Detector #2			
			06	6E				PRGD Pattern Detector #3			
			06	6F				PRGD Pattern Detector #4			
070	0F0	170	1F0	270	2F0	370	3F0	RHDLC 2(TXCISEL=0) Link Control /			
								THDLC 2(TXCISEL=1) Link Control			
071	0F1	171	1F1	271	2F1	371	3F1	RHDLC 2(TXCISEL=0) bits select /			
								THDLC 2(TXCISEL=1) bits select			
072~076	0F2~0F6	172~176	1F2~1F6	272~276	2F2~0F6	372~376	3F2~3F6	Reserved			
077	0F7	177	1F7	277	2F7	377	3F7	RESI Timeslot Offset			
078	0F8	178	1F8	278	2F8	378	3F8	RESI Bit Offset			

Table - 63b. T1/J1 Mode Register Map - Indirect Register

	Address	Register
RPLC	01-18H	Payload control byte for CH1 to CH24
Indirect Register	19-30H	Data trunk conditioning code for CH1 to CH24
	31-48H	Signaling trunk conditioning code for CH1 to CH24
TPLC	01-18H	Payload control byte for CH1 to CH24
Indirect Register	19-30H	Idle code for CH1 to CH24
	31-48H	Signaling control byte for CH1 to CH24
RCRB	01-18H / 21-38H	Signaling data for CH1 to CH24
Indirect Register	41-58H	Signaling control for CH1 to CH24

### 5.2 REGISTER DESCRIPTION

# E1 Or T1 / J1 Mode Selection (400H)

Bit No.	7	6	5	4	3	2	1	0					
Bit Name													
Туре		Reserved											
Default								1					

#### TEMODE:

This bit selects the operation mode globally for the chip.

- = 0: The chip operates in the E1 mode.
- = 1: The chip operates in the T1/J1 mode.

#### 5.2.1 E1 MODE

E1 Receive Path Line Options (000H, 080H, 100H, 180H, 200H, 280H, 300H, 380H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FIFOBYP	UNF	WORDERR	CNTNFAS	AUTOYELLOW	AUTORED	AUTOOOF	AUTOUPDATE
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

#### FIFOBYP:

This bit decides whether the received data should pass through or bypass the Receive Jitter Attenuation FIFO.

- = 0: The received data pass through the RJAT FIFO.
- = 1: The RJAT FIFO is bypassed. The delay is reduced by typically 24 bits.

### UNF:

- = 0: The Frame Processor operates normally.
- = 1: Frame searching is disabled, the Receive CAS/RBS Buffer holds its signaling frozen, and Auto\_OOF function, if enabled, will consider OOF to be declared.

# WORDERR, CNTNFAS:

WORDERR	CNTNFAS	Framing Bit Error
0	0	Each bit error in a 7-bit FAS pattern is counted as a single framing bit error.
1	0	One or more than one bit errors in a 7-bit FAS pattern is counted as a single framing bit error.
0	1	Each bit error in a 7-bit FAS pattern is counted as a single framing bit error, and
		a logic 0 in the second bit of TS0 of NFAS is counted as a single bit error too.
1	1	An 8-bit Error Word is consisted of a 7-bit FAS pattern and the second bit of timeslot 0 in the next NFAS
		frame. One or more than one bit errors in this 8-bit Error Word is counted as a single framing bit error.

# **AUTOYELLOW:**

This bit decides whether to send Yellow Alarm signal automatically.

- = 0: The automatic Yellow Alarm Transmission is disabled. It means that the RAI bit, the 3rd bit of NFAS frame, can only be transmitted when the REMAIS (b3, E1-041H) is set to 1.
- = 1: The automatic Yellow Alarm Transmission is enabled. It means that the RAI bit (the 3rd bit of NFAS frame) in the transmit data stream will be set to 1 automatically during loss of frame alignment or receiving AIS. The G706RAI (b0, E1-00EH) is used to select the conditions, under which the Yellow Alarm signal will be transmitted automatically.

### AUTORED:

This bit decides whether to start trunk conditioning (replacing data on RSDn with the data stored in the data trunk conditioning registers in RPLC) automatically when Red Alarm is declared.

- = 0: The trunk conditioning is not activated automatically when RED (b3, E1-037 H) becomes 1.
- = 1: The trunk conditioning will be initiated automatically when the RED (b3, E1-037H) becomes 1.

#### AUTOOOF:

This bit decides whether to start trunk conditioning (replacing data on RSDn with the data stored in the data trunk conditioning registers in RPLC) automatically in the duration of loss of basic frame.

- = 0: The trunk conditioning is not activated automatically when the OOFV (b6, E1-036 H) becomes 1.
- = 1: The trunk conditioning will be activated automatically when the OOFV (b6, E1-036 H) becomes 1.

#### AUTOUPDATE:

This bit decides whether the PMON and PRGD registers are automatically updated once every second.

- = 0: The PMON and PRGD registers are not automatically updated. They can only be updated by MCU operation.
- = 1: The PMON and PRGD registers will be automatically updated once every second.

### E1 Receive Side System Interface Options (001H, 081H, 101H, 181H, 201H, 281H, 301H, 381H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	LRCKFALL	RSSIG_EN	RSCKSEL	MRBS	MRBC	OOSMFAIS	TRKEN	RXMTKC
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	1	0	0	0	0	0	0

### LRCKFALL:

This bit selects the active edge of LRCKn to sample the data on the corresponding LRDn.

- = 0: the rising edge is selected.
- = 1: the falling edge is selected.

#### RSSIG EN:

When Receive Clock Slave Mode is enabled (RSCKSLV = 1, b5, E1-010H), this bit configures the receive side system interface.

- = 0: the Receive Clock Slave RSCK Reference Mode is selected. The RSCKn/RSSIGn pin will be used as RSCKn to output a 2.048 MHz jitter attenuated version of LRCKn or an 8KHz clock.
- = 1: the Receive Clock Slave External Signaling mode is selected. The RSCKn/RSSIGn pin is used as RSSIGn to output the extracted signaling data. Each time-slot's signaling bits are timeslot aligned with the RSDn data stream and located in lower nibble (b5b6b7b8).

### RSCKSEL:

When Receive Clock Slave RSCK Reference Mode is selected, this bit selects the frequency of the RSCKn.

- = 0: the RSCKn outputs an 8 KHz timing reference that is generated by dividing the jitter attenuated version of LRCKn.
- = 1: the RSCKn outputs a jitter attenuated version of the 2.048 MHz Line Receive Clock (LRCKn).

#### MRBS:

In Receive Multiplexed mode, this bit decides which bus the corresponding framer will use to output the received data.

- = 0: The first multiplexed bus (MRSD[1], MRSFS[1], MRSSIG[1]) is selected.
- = 1: The second multiplexed bus (MRSD[2], MRSFS[2], MRSSIG[2]) is selected.

### MRBC:

This bit turns on or off the transmission of received data from the corresponding framer to the selected multiplexed receive bus. Users should complete the setting in the MRBS (b4, E1-001H) before enabling this bit.

- = 0: The corresponding framer will not output its data stream on the multiplexed bus.
- = 1: The corresponding framer will output its data stream on the multiplexed bus.

### OOSMFAIS:

This bit decides whether to send Alarm Indication Signals (All Ones Signals) on RSSIGn to the system side in the condition of out of signaling multi-frame. This bit affects the corresponding timeslot of the MRSSIGn data stream if the multiplexed bus is enabled.

- = 0: The output on RSSIGn/MRSSIG pin will not be affected by the indication of out of Signaling Multi-Frame.
- = 1: The output on RSSIGn/MRSSIG pin will be set to all "1" in the condition of out of Signaling Multi-Frame.

#### TRKEN:

This bit decides whether to substitute the data on RSDn with the contents in the ELSB Idle Code Register during out of Basic frame. After substitution, the ELSB Idle Code can still be overwritten by the contents in RPLC Data Trunk Conditioning Registers and Signaling Trunk Conditioning Registers on per timeslot basis. This bit only has effect in Receive Clock Slave mode, and it affects the corresponding timeslot of multiplexed bus MRSD when multiplexed bus operation is enabled.

- = 0: ELSB Idle Code Substitution is disabled.
- = 1: Data in all timeslots on the RSDn will be replaced by the contents in ELSB Idle Code Register during out of Basic frame.

#### RXMTKC:

This bit decides how to substitute the received data stream on RSDn and RSSIGn with the contents in the RPLC Data Trunk Conditioning Registers and the RPLC Signaling Trunk Conditioning Registers. This bit affects the corresponding timeslot of the MRSD and MRSSIG if the multiplexed backplane is enabled.

- = 0: The data and signaling are substituted on a per-timeslot basis in accordance with the control bits contained in the per-timeslot Payload Control Byte registers in the RPLC.
- = 1: the data on RSDn of all timeslots are replaced with the data contained in the Data Trunk Conditioning registers in RPLC, and the data on RSSIGn of all timeslots are replaced with the data contained in the Signaling Trunk Conditioning registers. To enable this function, the PCCE (b0, E1-05CH) of the RPLC must be set to logic 1.

# E1 Transmit Path Configuration (002H, 082H, 102H, 182H, 202H, 282H, 302H, 382H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FIFOBYP	TAISEN		PATHCRC		TSFSRISE		LTCKRISE
Туре	R/W	R/W	Reserved	R/W	Reserved	R/W	Reserved	R/W
Default	0	0		0		0		0

#### FIFOBYP:

This bit decides whether the transmit data should pass through or bypass the Transmit Jitter Attenuation FIFO.

- = 0: The transmit data pass through the TJAT FIFO.
- = 1: The TJAT FIFO is bypassed. The delay is reduced by typically 24 bits.

# TAISEN:

This bit enables the line interface to generate an un-framed all-ones Alarm Indication Signal on the LTDn pin.

- = 0: normal operation.
- = 1: LTDn transmits all ones.

#### PATHCRC:

This bit allows upstream bit errors to be transmitted to the downstream transparently. When the data stream on TSDn is already in the CRC Multi-Frame format, and the IDT82V2108 is going to change some bits in the data stream, this bit decides whether to replace the original CRC-4 bits with re-calculated CRC-4 bits or just modify the original CRC-4 bits according to the contribution caused by changing bits in the data stream. This bit only takes effect when the FPTYP (b1, E1-019H) is set to 1 and one of the INDIS (b1, E1-040H) or FDIS (b3, E1-045H) is set to 1.

- = 0: A new re-calculated CRC-4 value will overwrite the incoming CRC-4 word. As the new CRC-4 value is transmitted to downstream, the bit errors in upstream can not be detected by the downstream.
- = 1: The incoming CRC-4 value is modified to just reflect the bit changes made by IDT82V2108. If there is any bit error in the upstream, it will be transmitted to the downstream transparently, and the downstream machine can detect it.

# TSFSRISE:

This bit selects the active edge of TSCCKB to update the Transmit Frame Pulse on TSFSn pin.

- = 0: the signal on TSFSn is updated on the falling edge of TSCCKB.
- = 1: the signal on TSFSn is updated on the rising edge of TSCCKB.

#### LTCKRISE:

This bit selects the active edge of LTCKn to update the data on LTDn.

- = 0: the data on LTDn pin is updated on the falling edge of LTCKn.
- = 1: the data on LTDn pin is updated on the rising edge of LTCKn.

# E1 Transmit Side System Interface Options (003H, 083H, 103H, 183H, 203H, 283H, 303H, 383H)

Bit No.	7	6	5	4	3	2		1		0
Bit Name		TSSIG_EN		MTBS						
Туре	Reserved	R/W	Reserved	R/W		Reserved				
Default		1		0						

# TSSIG EN:

- In Transmit Clock Slave mode (TSCKSLV=1, b5, E1-018H), this bit configures the transmit side system interface.
- = 0: the Transmit Clock Slave TSFS Enable mode is selected. The TSFSn/TSSIGn pin is used as TSFSn output.
- = 1: the Transmit Clock Slave External Signaling mode is selected. The TSFSn/TSSIGn pin is used as TSSIGn input.
- In Transmit Multiplexed mode, this bit must be set to 1.

# MTBS:

- In Transmit Multiplexed mode, this bit selects which multiplexed bus will interface with the corresponding framer.
- = 0: The incoming data is taken from the first multiplexed bus (MTSD1, MTSSIG1).
- = 1: The incoming data is taken from the second multiplexed bus (MTSD2, MTSSIG2).

# **E1 Transmit Timing Options** (004H, 084H, 104H, 184H, 204H, 284H, 304H, 384H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			TJATREF_SEL[2] TJATREF_SEL[1]		TJATREF_SEL[0]	LTCK_SEL[2]	LTCK_SEL[1]	LTCK_SEL[0]
Туре	Reserved		R/W	R/W	R/W	R/W	R/W	R/W
Default			1	0	0	1	0	1

# TJATREF\_SEL[2:0] - Transmit Jitter Attenuation DPLL Input Reference Clock Selection

The TJATREF\_SEL[2:0] select the input reference clock for the TJAT DPLL.

TJATREF_SEL[2:0]	Input Reference Clock
000	TSCCKA/8
001	TSCCKB
010	LRCK
011	TSCCKA
100	XCK / 24
Others	TSCCKB

# LTCK\_SEL[2:0] - Line Transmit Clock (LTCKn) Selection

LTCK_SEL[2:0]	Line Transmit Clock
000	TSCCKA/8
001	TSCCKB
010	LRCK
011	TSCCKA
100	XCK / 24
Others	A smoothed clock output from the TJAT DPLL

E1 Interrupt Source #1 (005H, 085H, 105H, 185H, 205H, 285H, 305H, 385H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PMON	FRMG	FRMP	PRGD	ELSB	RHDLC#1	RHDLC#2	RHDLC#3
Туре	R	R	R	R	R	R	R	R
Default	X	X	X	Χ	Χ	Χ	Χ	Χ

Bits in this register indicate which function block caused an interrupt signal on  $\overline{\text{INT}}$  pin. Reading this register does not remove the interrupt indication. To remove the interrupt indication on the  $\overline{\text{INT}}$  pin, the corresponding interrupt status register must be read.

**E1 Interrupt Source #2** (006H, 086H, 106H, 186H, 206H, 286H, 306H, 386H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TRSI		TJAT	RJAT	THDLC#1	THDLC#2	THDLC#3	RCRB
Туре	R	Reserved	R	R	R	R	R	R
Default	Х		Χ	Х	X	X	Х	Χ

Bits in this register indicate which function block caused an interrupt signal on  $\overline{\text{INT}}$  pin. Reading this register does not remove the interrupt indication. To remove the interrupt indication on the  $\overline{\text{INT}}$  pin, the corresponding interrupt status register must be read.

E1 Diagnostics (007H, 087H, 107H, 187H, 207H, 287H, 307H, 387H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				LINELB	V52DIS	DDLB	RAIS	TXDIS
Туре	Reserved			R/W	R/W	R/W	R/W	R/W
Default				0	0	0	0	0

#### LINELB:

Line Loop back means that the transmit line interface data and clock (LTDn and LTCKn) are internal directly comes from the received line data and clock (LRDn and LRCKn). The loop back data stream can pass through the Receive Jitter Attenuator or bypass the Receive Jitter Attenuator (if the Receive Jitter Attenuator is configured to be bypassed)

- = 0: Line loop back is disabled.
- = 1: Line loop back is enabled.

# V52DIS:

- = 0: All HDLC controllers of the corresponding framer are available to use.
- = 1: Only the first HDLC controller in receive direction (RHDLC#1) and transmit direction (THDLC#1) are available to use, the remaining HDLC controllers are disabled.

Note that this bit can not be reset by software reset. It can only reset by hardware reset.

#### DDLB:

Digital Loop back means that the received line data and clock (LRDn and LRCKn) are internal directly comes from the transmit line data and clock (LTDn and LTCKn) without the Receive Jitter Attenuator.

- = 0: Digital loop back is disabled.
- = 1: Digital loop back is enabled

### RAIS:

- = 0: normal operation.
- = 1: force the data output on RSDn to be all ones, and freeze the signal on RSSIGn at the current valid signaling in Receive Clock Slave External Signaling mode. In Receive Multiplexed mode, the data of the corresponding framer output on MRSD is forced to be all ones, and the signal of the corresponding framer output on MRSSIG is frozen at the current valid signaling.

# TXDIS:

- = 0: normal transmission.
- = 1: force the data to be transmitted on the TLDn pin to be all zeros.

# E1 Revision / Chip ID / Global PMON Update (009H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TYPE[2]	TYPE[1]	TYPE[0]	ID[4]	ID[3]	ID[2]	ID[1]	ID[0]
Туре	R	R	R	R	R	R	R	R
Default	0	0	0	0	0	0	0	1

Writing to this register causes all Performance Monitor and PRGD Generator/Detector counters to be updated simultaneously.

# TYPE[2:0]:

TYPE[2:0] are fixed to 000, representing the IDT82V2108 chip.

### ID[4:0]:

ID[4:0] are fixed to 00011, representing the current version number of the IDT82V2108.

# E1 Data Link Micro Select / Framer Reset (00AH, 08AH, 10AH, 18AH, 20AH, 28AH, 30AH, 38AH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RHDLCSEL[1]	RHDLCSEL[0]	THDLCSEL[1]	THDLCSEL[0]	TXCISEL			RESET
Туре	R/W	R/W	R/W	R/W	R/W	Res	erved	R/W
Default	Х	X	X	X	X			0

### RHDLCSEL[1:0]:

The RHDLCSEL[1:0] select one of the three HDLC Receivers to be accessed by the microprocessor. At one time, the microprocessor can only access one HDLC controller. These bits must be set before using the HDLC controller.

RHDLCSEL[1:0]	the HDLC Receiver
00	RHDLC #1
01	RHDLC #2
10	RHDLC #3
11	Reserved

### THDLCSEL[1:0]:

The THDLCSEL[1:0] select one of the three HDLC Transmitters to be accessed by the microprocessor. At one time, the microprocessor can only access one HDLC controller. These bits must be set before using the HDLC controller.

THDLCSEL[1:0]	the HDLC Transmitter
00	THDLC #1
01	THDLC#2
10	THDLC#3
11	Reserved

#### TXCISEL:

The registers addressed from E1-028H to E1-02DH are shared by the HDLC Receiver and HDLC Transmitter to decide the position of the extracted bit in the received data stream and the inserted bit in the transmitting data stream respectively. So this bit is used to decide whether the Read/Write operation on the registers addressed from E1-028H to E1-02DH is for the HDLC receiver or for the HDLC transmitter.

- = 0: the Read/Write operation on registers addressed from 028 H to 02D H is for HDLC receiver.
- = 1: the Read/Write operation on registers addressed from 028H to 02D H is for the HDLC transmitter.

### RESET:

This bit implements a software reset for individual framer.

- = 0: normal operation.
- = 1: The corresponding framer is held in reset. However, this bit, the bits in this register and the V52DIS (b3, E1-007H) can not be reset. Therefor, a logic 0 must be written to bring the framer out of reset. Holding the framer in a reset state effectively puts it into a low power standby mode. A hardware reset clears the RESET bit, the bits in this register and the V52DIS (b3, E1-007H).

### E1 Interrupt ID (00BH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	INT[8]	INT[7]	INT[6]	INT[5]	INT[4]	INT[3]	INT[2]	INT[1]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Χ	Χ	Χ	Χ	0	0	0

This register indicates which one of the eight framers caused the interrupt  $\overline{\text{INT}}$  pin to be logic low. When any one of the eight framers caused the interrupt, the corresponding bit in the INT[8:1] will be high.

# E1 Pattern Generator / Detector Positioning / Control (00CH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PRGDSEL[2]	PRGDSEL[1]	PRGDSEL[0]			RXPATGEN	UNF_GEN	UNF_DET
Туре	R/W	R/W	R/W	Rese	rved	R/W	R/W	R/W
Default	0	0	0			0	0	0

The IDT82V2108 has only one PRBS Generator/Detector (PRGD) shared by all the eight framers. At one time, only one framer can use this PRGD. This register selects which framer will use the PRGD and how the PRGD will be used.

# PRGDSEL[2:0]:

PRGDSEL[2:0] select one of the eight framers to be tested by the PRGD block.

, <u> </u>	
PRGDSEL[2:0]	Selected Framer
000	Framer 1
001	Framer 2
010	Framer 3
011	Framer 4
100	Framer 5
101	Framer 6
110	Framer 7
111	Framer 8

#### RXPATGEN:

- = 0: the pattern in PRGD is generated in the transmit path and is detected in the receive path.
- = 1: the pattern in PRGD is generated in the receive path and is detected in the transmit path.

# UNF GEN:

- = 0: which timeslots of the selected path will be replaced by the PRGD pattern is specified in TPLC or RPLC.
- = 1: all the 32 timeslots of the selected path will be replaced by the PRGD pattern.

# UNF DET:

- = 0: which timeslots of the selected path will be detected by PRGD pattern is specified in TPLC or RPLC.
- = 1: all the 32 timeslots of the selected path will be detected by PRGD pattern.

E1 Clock Monitor (00DH, 08DH, 10DH, 18DH, 20DH, 28DH, 30DH, 38DH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				XCK	TSCCKB	TSCCKA	RSCCK	LRCK
Туре		Reserved	'	R	R	R	R	R
Default				Χ	Х	X	Χ	Х

This register provides activity monitoring on the IDT82V2108 clocks. When a monitored clock signal makes a low to high transition, the corresponding bit in this register is set to 1, and this bit remains to be 1 until this register is read. After a read operation on this register, all the bits in this register will be cleared to 0. A lack of transitions of the monitored clock will be indicated by 0 in the corresponding bit, which means that the clock fails. This register should be read periodically to detect clock failures.

#### XCK:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the XCK.

#### TSCCKB:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the TSCCKB.

#### TSCCKA:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the TSCCKA.

### RSCCK:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the RSCCK.

#### LRCK:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the LRCK.

# E1 Receive Path Frame Pulse Configuration (00EH, 08EH, 10EH, 18EH, 20EH, 28EH, 30EH, 38EH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					PERTS_RSFS	REF_MRSFS	OOCMFE0	G706RAI
Туре	Reserved			R/W	R/W	R/W	R/W	
Default					0	0	0	0

#### PERTS\_RSFS, REF\_MRSFS:

PERTS_RSFS	REF_MRSFS	the Pulse on the RSFSn/MRSFS
0	0	The pulse output on the RSFS/MRSFS pin is forced to be logic 0.
1	0	The signal on the RSFS/MRSFS pin is determined by the ROHM,
		BRXSMFP, BRCMFP and ALTIFP (b3b2b1b0, E1-011 H).
Х	1	RSFSn/MRSFS contains a reference frame pulse identical to the receive
		system side common frame pulse on the RSCFS/MRSCFS pin.

In Receive Multiplexed mode, these two bits in the eight framers should be set in the same value.

# OOCMFE0:

This bit selects one of two operation modes concerning the transmission of E-bits when the framer is out of CRC-4 multiframe.

- = 0: transmit ones for the E-bits while out of CRC-4 Multi-Frame.
- = 1: transmit zeros for the E-bits while out of CRC-4 Multi-Frame. (This setting is compliant with the CRC-4 to non-CRC-4 interworking procedure in Annex B of G.706)

#### G706RAI:

When the AUTOYELLOW (b3, E1-00H) is set as 1, which means the RAI bit will be transmitted automatically in certain conditions, this bit selects one of two criteria defining the conditions. If the AUTOYELLOW (b3, E1-00H) is 0, G706RAI does not have any effect.

- = 0: The RAI bit will be transmitted when out of Basic Frame, when AISD is declared, when CRC-4 to non-CRC-4 interworking is declared or when the off-line searching indicates out of Basic frame. This definition follows the ETSI standards.
- = 1: The RAI bit will be transmitted when out of Basic frame or when AISD is declared, but not when CRC-4 to non-CRC-4 interworking is declared nor when offline out-of frame is declared. This definition follows the Annex B of G.706.

# E1 Receive Backplane Configuration (010H, 090H, 110H, 190H, 210H, 290H, 310H, 390H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FRACTN[1]	FRACTN[0]	RSCKSLV	DE	FE	CMS	RATE[1]	RATE[0]
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	1	1	1	0	0	0

### FRACTN[1:0]:

When Receive Clock Master mode is selected, (RSCKSLV=0, b5, E1-010H), these two bits selects one of the operation modes shown in the following table. The two bits will be ignored if the Receive Clock Slave mode is selected. (RSCKSLV=1, b5, E1-010H).

FRACTN[1:0]	Operation Mode
0 0	Receive Clock Master Full E1 mode
01	Reserved
10	Receive Clock Master Fractional E1 mode
11	Receive Clock Master Fractional E1 with F-bit mode

<sup>&</sup>quot;Full E1" mode means that the received entire frame (256 bits) is clocked out from RSDn pin, and there are no gaps in the RSCKn clock pulse.

"Fractional E1" mode means that the RSCKn only clocks out on the selected time slots, and RSCKn does not pulse during those un-selected time slots. The time slots selection is decided by DTRKC/NxTS (b6, E1-RPLC-Indirect Register-20-3F H).

"Fractional E1 with F-bit" mode is to support ITU recommendation G.802 where 1.544 Mbit/s data is carried within a 2.048 Mbit/s data stream. In this configuration, bits from the second bit of TS 26 to the last bit of the Basic Frame are suppressed, and the remaining bits can be selectively gapped by the DTRKC/NxTS (b6, E1-RPLC-Indirect Register-20-3F H).

#### RSCKSLV:

- = 0: Receive Clock Master mode is selected.
- = 1: Received Clock Slave mode is selected.

RSCKSLV must be set to 1 to support multiplexed backplane.

### DE:

- = 0: the signal on the RSDn and RSSIGn pins are updated on the falling edge of the RSCCK or the RSCK.
- = 1: the signal on the RSDn and RSSIG pins are updated on the rising edge of the RSCK or the RSCK.

In Receive Multiplexed mode, the DE in all eight framers should be set at the same value.

#### FE:

If FE is not equal to DE, the frame pulse will be sampled or updated one clock edge after the corresponding data pulse.

- = 0: the signal on the RSCFS pin is sampled or the signal on the RSFSn pin is updated on the falling edge of the RSCCK or the RSCKn.
- = 1: the signal on the RSCFS pin is sampled or the signal on the RSFSn pin is updated on the rising edge of the RSCCK or the RSCKn.
- In Receive Multiplexed mode, the FE in all eight framers should be set at the same value.

#### CMS:

- = 0: the clock frequency of RSCCK/MRSCCK is the same as the bit rate of the backplane.
- = 1: the clock frequency of RSCCK/MRSCCK is the double of the bit rate of the backplane.

The CMS of all eight framers should be set at the same value.

### RATE[1:0]:

These bits determine the bit rate of the received data stream on the backplane. Note that to operate in the Receive Multiplexed mode, the RATE[1:0] in all eight framers should be configured to select the 8.192 Mbit/s backplane bit rate. When the RATE[1:0] selects the 8.192 Mbit/s, the RSCKSLV (b5, E1-010H) must be set to 1.

RATE[1:0]	Backplane Rate
0 0	Reserved
0 1	2.048M bit/s
1 0	Reserved
11	8.192M bit/s

# E1 Receive Backplane Frame Pulse Configuration (011H, 091H, 111H, 191H, 211H, 291H, 311H, 391H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		FPINV	FPMODE		ROHM	BRXSMFP	BRXCMFP	ALTIFP
Туре	Reserved	R/W	R/W	Reserved	R/W	R/W	R/W	R/W
Default		0	1		0	0	0	0

#### FPINV:

- = 0: The framing pulse RSCFS and RSFSn/MRSFS are active high.
- = 1: The framing pulse RSCFS and RSFSn/MRSFS are active low.

When this bit is used to indicate the active pulse for RSCFS or MRSFS, then it should be set to the same value for all eight framers.

### FPMODE:

This bit decides whether to use RSCFS as the framing pulse or not. In Receive Clock Master mode (RSCKSLV=0, b5, E1-010H), the FPMODE must be 0.

- = 0: RSCFS/MRSCFS is unused.
- = 1: RSCFS/MRSCFS is used.

In Receive Multiplexed mode, the FPMODE in all eight framers should be set to the same value.

#### ROHM:

When the PERTS\_RSFS and the REF\_MRSFS (b3~2, E1-00EH) are 1 and 0 respectively, this bit decides whether to use RSFSn pin to indicate TS0 and TS16. Details are tabulated in the following table.

# BRXSMFP, BRXCMFP:

When the PERTS\_RSFS and the REF\_MRSFS (b3~2, E1-00EH) are 1 and 0 respectively, these two bits, together with the ALTIFP bit, select the output signal seen on the RSFSn pin. Details are tabulated in the following table.

### ALTIFP:

When the RSFSn pin is configured to output the framing pulse for Basic Frame, Signaling Multiframe or CRC Multiframe, this bit permits suppression of every other framing pulse. The following table shows the details for the different configurations of RSFSn pin.

ROHM	BRXSMFP	BRXCMFP	ALTIFP	RSFSn / MRSFS Indication
0	0	0	0	The RSFSn asserts for 1 bit cycle on the first bit of each Basic Frame output on the RSDn.
0	0	0	1	The RSFSn asserts for 1 bit cycle on the first bit of every second Basic Frame output on the RSDn.
0	0	1	0	The RSFSn asserts for 1 bit cycle on the first bit of the first frame of each CRC Multi-Frame output on the RSDn (in case CRC Multi-Frame is disabled, the RSFSn asserts every 16 frames).
0	0	1	1	The RSFSn asserts for 1 bit cycle on the first bit of the first frame of every second CRC Multi-Frame output on the RSDn (in case CRC Multi-Frame is disabled, the RSFSn asserts every 32 frames).
0	1	0	0	The RSFSn asserts for 1 bit cycle on the first bit of the first frame of each Signaling Multi-Frame output on the RSDn (in case Signaling Multi-Frame is disabled, the RSFSn asserts every 16 frames).
0	1	0	1	The RSFSn asserts for 1 bit cycle on the first bit of the first frame of every second Signaling Multi-Frame output on the RSDn (in case Signaling Multi-Frame is disabled, the RSFSn asserts every 32 frames).
0	1	1	0	The RSFSn goes high/low at the start of the first bit of the first frame of each Signaling Multi- Frame, and does the opposite at the end of the first bit of the first frame of each CRC Multi- Frame.
0	1	1	1	The RSFSn goes high/low at the start of the first bit of the first frame of every second Signaling Multi-Frame, and does the opposite at the end of the first bit of the first frame of every second CRC Multi-Frame.
1	Х	Χ	Χ	The RSFSn pin pulses during the entire TS0 period and the entire TS16.

In Receive Multiplexed mode, when the PERTS\_RSFS and the REF\_MRSFS (b3~2, E1-00EH) are 1 and 0 respectively, the MRSFS can only indicate the first bit of a Basic Frame of the selected first framer no matter what is set in the ROHM, BRXSMFP, BRXCMFP and ALTIFP.

# E1 Receive Backplane Parity / F-bit Configuration (012H, 092H, 112H, 192H, 212H, 292H, 312H, 392H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RPTYP	RPTYE	FIXF	FIXPOL	PTY_EXTD		TRI[1]	TRI[0]
Туре	R/W	R/W	R/W	R/W	R/W	Reserved	R/W	R/W
Default	0	0	0	0	0		0	0

#### RPTYP:

This bit selects the parity type for the receive side system data.

- = 0: even parity is employed, which means a logic one should be inserted in the first bit of TS0 of each basic frame when the number of ones in the previous basic frame is odd.
- = 1: odd parity is employed, which means a logic one should be inserted in the first bit of TS0 of each basic frame when the number of ones in the previous basic frame is even.

#### RPTYE:

This bit enables the parity for the receive side system data. The bit is invalid in Receive Clock Master Fractional E1 (with F-bit) mode.

- = 0: disable the parity on the RSDn/MRSD pin.
- = 1: enable the parity on the RSDn/MRSD pin.

# FIXF:

This bit controls whether the parity bit position is fixed at the level defined by the FIXPOL. It is invalid in Receive Clock Master Fractional E1 (with F-bit) mode and valid when RPTYE = 0.

- = 0: no action.
- = 1: the setting in the FIXPOL is valid. The first bit of TS0 of each basic frame output on the RSDn/MRSD pin is fixed with the value of FIXPOL.

#### FIXPOL:

This bit is invalid in Receive Clock Master Fractional E1 (with F-bit) mode and valid when the RPTYE = 0 and the FIXF = 1.

- = 0: force the first bit of TS0 of each basic frame output on the RSDn/MRSD pin to be logic 0.
- = 1: force the first bit of TS0 of each basic frame output on the RSDn/MRSD pin to be logic 1.

### PTY EXTD:

When the parity is calculated over the previous basic frame, the first bit of TS0 on the RSDn pin can be included or not. The decision is made by this bit.

- = 0: the first bit of TS0 on the RSDn/MRSD pin is not calculated.
- = 1: the first bit of TS0 on the RSDn/MRSD pin is calculated.

# TRI[1:0]:

TRI[1:0]	Output Status on the RSDn/MRSD and RSSIGn/MRSSIG pin
0 0	in high impedance
10	Reserved
0 1	normal output
11	Reserved

### E1 Receive Backplane Time Slot Offset (013H, 093H, 113H, 193H, 213H, 293H, 313H, 393H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		TSOFF[6]	TSOFF[5]	TSOFF[4]	TSOFF[3]	TSOFF[2]	TSOFF[1]	TSOFF[0]
Туре	Reserved	R/W						
Default		0	0	0	0	0	0	0

These bits determine the timeslot offset between the signal on the RSCFS pin and the start of the Basic Frame output on the RSDn & RSSIGn pin. If the RSCFS does not exist, the timeslot offset is between the RSFSn and the start of the Basic Frame output on the RSDn & RSSIGn. In Receive Multiplexed mode, each framer contributes every fourth timeslot on MRSD[1:2] and MRSSIG[1:2].

They define a binary number. The offset can be set from 0 to 127 timeslots.

# E1 Receive Backplane Bit Offset (014H, 094H, 114H, 194H, 214H, 294H, 314H, 394H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	Reserved		RSD_RSCFS_EDGE		BOFF_EN	BOFF[2]	BOFF[1]	BOFF[0]
Туре			R/W	Reserved	R/W	R/W	R/W	R/W
Default			0		0	0	0	0

### RSD RSCFS EDGE:

Valid when the CMS (b2, E1-010H) is logic 1 and the DE (b4, E1-010H) is not equal to the FE (b3, E1-010H).

- = 0: select the second active edge of the RSCCK to update the signal on the RSDn, RSSIGn and RSFSn pins, or select the first active edge of the MRSCCK to update the signal on the MRSD, MRSSIG and MRSFS pins.
- = 1: select the first active edge of the RSCCK to update the signal on the RSDn, RSSIGn and RSFSn pins, or select the second active edge of the MRSCCK to update the signal on the MRSD, MRSSIG and MRSFS pins.

(The signal on the RSCFS/MRSCFS pin is always sampled on the first active edge.)

In Receive Multiplexed mode, the RSD\_RSCFS\_EDGE in all eight framers should be set at the same value.

When the CMS (b2, E1-010H) is logic 1 and the DE (b4, E1-010H) is equal to FE (b3, E1-010H), the signals on the RSDn/MRSD, RSSIGn/MRSSIG and RSFSn/MRSFS pins are updated on the first active edge of the RSCCK/MRSCCK.

# BOFF EN:

Valid when the CMS (b2, E1-010H) is logic 0.

- = 0: disable the bit offset.
- = 1: enable the bit offset.

# BOFF[2:0]:

Valid when the CMS (b2, E1-010H) is logic 0 and the BOFF\_EN is logic 1.

These bits define a binary number. The content in the BOFF[2:0] determines the bit offset between the signal on the RSCFS pin and the start of the Basic Frame output on the RSDn & RSSIGn pin. If the RSCFS does not exist, the timeslot offset is between the RSFSn and the start of the Basic Frame output on the RSDn & RSSIGn. It is also available in Receive Multiplexed mode.

Programming of the Bit Offsets is consistent with the convention established by the Concentration Highway Interface (CHI) specification. Refer to the Functional Description for details.

E1 Transmit Backplane Configuration (018H, 098H, 118H, 198H, 218H, 298H, 318H, 398H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			TSCKSLV	DE	FE	CMS	RATE[1]	RATE[0]
Туре	Reser	Reserved		R/W	R/W	R/W	R/W	R/W
Default			1	1	1	0	0	0

#### TSCKSLV:

- = 0: Transmit Clock Master mode is selected.
- = 1: Transmit Clock Slave mode or Transmit Multiplexed mode is selected.

### DE:

- = 0: the data on the TSDn/MTSD and TSSIGn/MTSSIG pins are sampled on the falling edge of the TSCCKB/MTSCCKB or the LTCKn.
- = 1: the data on the TSDn/MTSD and TSSIGn/MTSSIG pins are sampled on the rising edge of the TSCCKB/MTSCCKB or the LTCKn.
- In Transmit Multiplexed mode, the DE of the eight framers should be set to the same value.

#### FE:

Valid in Transmit Clock Slave mode and Transmit Multiplexed mode.

- = 0: the data on the TSCFS/MTSCFS pin is sampled on the falling edge of TSCCKB/MTSCCKB.
- = 1: the data on the TSCFS/MTSCFS pin is sampled on the rising edge of TSCCKB/MTSCCKB.
- In Transmit Multiplexed mode, the FE of the eight framers should be set to the same value.

### CMS:

- = 0: the clock rate of TSCCKB/MTSCCKB is the same as that of the backplane.
- = 1: the clock rate of TSCCKB/MTSCCKB is twice that of the backplane.

The CMS of the eight framers should be set to the same value.

### RATE[1:0]:

These bits determine the bit rate of the transmit data stream on the backplane. Note that if any of the eight framers selects the 8.192 Mbit/s backplane bit rate, the multiplied bus will be enabled for the chip. When the RATE[1:0] selects the 8.192 Mbit/s, the TSCKSLV (b5, E1-018H) must be set to 1.

RATE[1:0]	Backplane Rate
0 0	Reserved
0 1	2.048M bit/s
10	Reserved
11	8.192M bit/s (valid to eight frames)

# E1 Transmit Backplane Frame Pulse Configuration (019H, 099H, 119H, 199H, 219H, 299H, 319H, 399H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					FPINV		FPTYP	
Туре		Reserved				Reserved	R/W	Reserved
Default					0		0	

### FPINV:

- = 0: the positive pulse on the TSCFS pin is valid.
- = 1: the negative pulse on the TSCFS pin is valid.

The FPINV of the eight framers should be the same value.

### FPTYP:

- = 0: indicate that the signal on the TSCFS pin pulses during the first bit of each Basic Frame.
- = 1: indicate that the signal on the TSCFS pin asserts on the first bit of each Signaling Multi-Frame and asserts oppositely following the first bit of each CRC Multi-Frame.

The FPTYP of the eight framers should be the same value.

### E1 Transmit Backplane Parity Configuration and Status (01AH, 09AH, 11AH, 19AH, 21AH, 29AH, 31AH, 39AH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TPTYP	TPTYE	TDI		PTY_EXTD	Reserved		
Туре	R/W	R/W	R	Reserved	R/W			
Default	0	0	Х		0			

#### TPTYP:

- = 0: even parity is employed in the first bit of TS0 of each Basic Frame input from the TSDn/MTSD pin, which means a logic one is expected in the position when the number of ones in the previous Basic Frame is odd.
- = 1: odd parity is employed in the first bit of TS0 of each Basic Frame input from the TSDn/MTSD pin, which means a logic one is expected in the position when the number of ones in the previous Basic Frame is even.

#### TPTYE:

This bit decides whether to generate an interrupt when a parity error is detected on the TSDn/MTSD pin.

- = 0: No interrupt is generated when a parity error is detected on the TSDn/MTSD pin.
- = 1: An interrupt on the  $\overline{\rm INT}$  pin is generated when a parity error is detected on the TSDn/MTSD pin.

### TDI:

This bit indicates the parity error detected on the TSDn/MTSD pin.

- = 0: no parity error is detected on the TSDn/MTSD pin.
- = 1: a parity error is detected on the TSDn/MTSD pin.

This bit is cleared to 0 when it is read.

# PTY\_EXTD:

- = 0: the parity checking is calculated over the previous basic frame, excluding the first bit of TS0 on the TSDn/MTSD pin.
- = 1: the parity checking is calculated over the previous basic frame, including the first bit of TS0 on the TSDn/MTSD pin.

#### E1 Transmit Backplane Time Slot Offset (01BH, 09BH, 11BH, 19BH, 21BH, 29BH, 31BH, 39BH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		TSOFF[6]	TSOFF[5]	TSOFF[4]	TSOFF[3]	TSOFF[2]	TSOFF[1]	TSOFF[0]
Туре	Reserved	R/W						
Default		0	0	0	0	0	0	0

In Transmit Clock Slave mode, the content in the TSOFF[6:0] determines the timeslot offset between the TSCFS and the start of the Basic Frame transmitted on the TSDn & TSSIGn. In Transmit Multiplexed mode, the content in the TSOFF[6:0] determines the timeslot offset between the MTSCFS and the start of the Basic Frame transmitted on the MTSD & MTSSIG for the corresponding framer.

In Transmit Clock Master mode, the timeslot offset is disabled, that is, the TSOFF[6:0] must be logic 0.

The TSOFF[6:0] define a binary number. The offset can be set from 0 to 127 timeslots.

### E1 Transmit Backplane Bit Offset (01CH, 09CH, 11CH, 19CH, 21CH, 29CH, 31CH, 39CH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				COFF	CHI	BOFF[2]	BOFF[1]	BOFF[0]
Туре		Reserved			R/W	R/W	R/W	R/W
Default				0	0	0	0	0

#### COFF:

Valid when the CMS (b2, E1-018H) is logic 1.

- = 0: select the first active edge of the TSCCKB/MTSCCKB to sample the data on the TSDn/MTSD, TSSIGn/MTSSIG and to update the data on the TSFSn.
- = 1: select the second active edge of the TSCCKB/MTSCCKB to sample the data on the TSDn/MTSD, TSSIGn/MTSSIG and to update the data on the TSFSn.

(The signal on the TSCFS/MTSCFS pin is always sampled on the first active edge.)

#### CHI:

This bit controls if the value in the BOFF[2:0] is the actual value or meets the Concentration Highway Interface (CHI) specification.

- = 0: disable the CHI specification.
- = 1: enable the CHI specification.

### BOFF[2:0]:

In Transmit Clock Master mode, the content in the BOFF[2:0] determines the bit offset between the signal on the TSFSn and the start of the Basic Frame transmitted on the TSDn.

In Transmit Clock Slave mode, the content in the BOFF[2:0] determines the bit offset between the TSCFS and the start of the Basic Frame transmitted on the TSDn & TSSIGn.

In Transmit Multiplexed mode, the content in the BOFF[2:0] determines the bit offset between the MTSCFS and the start of the Basic Frame transmitted on the MTSD & MTSSIG.

These bits define a binary number. When the CHI = 0, the setting in the BOFF[2:0] is their actual value (0 stands for 0 bit offset, 1 stands for 1 bit offset). When the CHI = 1, programming of the BOFF[2:0] is consistent with the convention established by the Concentration Highway Interface (CHI) specification. Refer to the Functional Description for details.

# E1 RJAT Interrupt Status (020H, 0A0H, 120H, 1A0H, 220H, 2A0H, 320H, 3A0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			OVRI	UNDI				
Туре			R	R				
Default							Х	X

### OVRI:

If data are still attempted to write into the FIFO when the FIFO is already full, the overwritten event will occur.

- = 0: the RJAT FIFO is not overwritten.
- = 1: the RJAT FIFO is overwritten.

This bit is cleared to 0 when it is read.

# UNDI:

If data are still attempted to read from the FIFO when the FIFO is already empty, the under-run event will occur.

- = 0: the RJAT FIFO is not under-run.
- = 1: the RJAT FIFO is under-run.

This bit is cleared to 0 when it is read.

# E1 RJAT Reference Clock Divisor (N1) Control (021H, 0A1H, 121H, 1A1H, 221H, 2A1H, 3A1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N1[7]	N1[6]	N1[5]	N1[4]	N1[3]	N1[2]	N1[1]	N1[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N1[7:0] + 1) is the divisor of the input reference clock, which is the ratio between the frequency of the input reference clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the RJAT.

# E1 RJAT Output Clock Divisor (N2) Control (022H, 0A2H, 122H, 1A2H, 222H, 2A2H, 322H, 3A2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N2[7]	N2[6]	N2[5]	N2[4]	N2[3]	N2[2]	N2[1]	N2[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N2[7:0] + 1) is the divisor of the output smoothed clock, which is the ratio between the frequency of the output smoothed clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the RJAT.

# E1 RJAT Configuration (023H, 0A3H, 123H, 1A3H, 223H, 2A3H, 323H, 3A3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				CENT	UNDE	OVRE		LIMIT
Туре		Reserved			R/W	R/W	Reserved	R/W
Default				0	0	0		1

#### CENT:

The CENT allows the RJAT FIFO to self-center its read pointer, maintaining the pointer at least 4 UI away from the FIFO being empty or full.

- = 0: disable the self-center. Data are pass through uncorrupted.
- = 1: enable the FIFO to self-center its read pointer when the FIFO is 4 UI away from being empty or full.

A positive transition in this bit will execute a self-center action immediately.

# UNDE:

This bit decides whether to generate an interrupt when the RJAT FIFO is under-run.

- = 0: No interrupt is generated when the RJAT FIFO is under-run.
- = 1: An interrupt on the INT pin is generated when the RJAT FIFO is under-run.

### OVRE:

This bit decides whether to generate an interrupt when the RJAT FIFO is overwritten.

- = 0: No interrupt is generated when the RJAT FIFO is overwritten.
- = 1: An interrupt on the INT pin is generated when the RJAT FIFO is overwritten.

#### LIMIT:

- = 0: disable the limitation of the jitter attenuation.
- = 1: enable the DPLL to limit the jitter attenuation by enabling the FIFO to increase or decrease the frequency of the output smoothed clock when the read pointer is 1 UI away from the FIFO being empty or full. This limitation of jitter attenuation ensures that no data is lost during high phase shift conditions.

# **E1 TJAT Interrupt Status** (024H, 0A4H, 124H, 1A4H, 224H, 2A4H, 324H, 3A4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			OVRI	UNDI				
Туре			R	R				
Default							Χ	Χ

#### OVRI:

If data are still attempted to write into the FIFO when the FIFO is already full, the overwritten event will occur.

- = 0: the TJAT FIFO is not overwritten.
- = 1: the TJAT FIFO is overwritten.

This bit is cleared to 0 when it is read.

# UNDI:

If data are still attempted to read from the FIFO when the FIFO is already empty, the under-run event will occur.

- = 0: the TJAT FIFO is not under-run.
- = 1: the TJAT FIFO is under-run.

This bit is cleared to 0 when it is read.

# E1 TJAT Reference Clock Divisor (N1) Control (025H, 0A5H, 125H, 1A5H, 225H, 2A5H, 325H, 3A5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N1[7]	N1[6]	N1[5]	N1[4]	N1[3]	N1[2]	N1[1]	N1[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N1[7:0] + 1) is the divisor of the input reference clock, which is the ratio between the frequency of the input reference clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the TJAT.

# E1 TJAT Output Clock Divisor (N2) Control (026H, 0A6H, 126H, 1A6H, 226H, 2A6H, 326H, 3A6H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N2[7]	N2[6]	N2[5]	N2[4]	N2[3]	N2[2]	N2[1]	N2[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N2[7:0] + 1) is the divisor of the output smoothed clock, which is the ratio between the frequency of the output smoothed clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the TJAT.

### **E1 TJAT Configuration** (027H, 0A7H, 127H, 1A7H, 227H, 2A7H, 327H, 3A7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				CENT	UNDE	OVRE		LIMIT
Туре		Reserved			R/W	R/W	Reserved	R/W
Default				0	0	0		1

### CENT:

The CENT allows the TJAT FIFO to self-center its read pointer, maintaining the pointer at least 4 UI away from the FIFO being empty or full.

- = 0: disable the self-center. Data are pass through uncorrupted.
- = 1: enable the FIFO to self-center its read pointer when the FIFO is 4 UI away from being empty or full.

A positive transition in this bit will execute a self-center action immediately.

# UNDE:

This bit decides whether to generate an interrupt when the TJAT FIFO is under-run.

- = 0: No interrupt is generated when the TJAT FIFO is under-run.
- = 1: An interrupt on the INT pin is generated when the TJAT FIFO is under-run.

#### OVRE:

This bit decides whether to generate an interrupt when the TJAT FIFO is overwritten.

- = 0: No interrupt is generated when the TJAT FIFO is overwritten.
- = 1: An interrupt on the INT pin is generated when the TJAT FIFO is overwritten.

### LIMIT:

- = 0: disable the limitation of the jitter attenuation.
- = 1: enable the DPLL to limit the jitter attenuation by enabling the FIFO to increase or decrease the frequency of the output smoothed clock when the read pointer is 1 UI away from the FIFO being empty or full. This limitation of jitter attenuation ensures that no data is lost during high phase shift conditions.

## E1 RHDLC Receive Data Link 1 Control (TXCISEL = 0) (028H, 0A8H, 128H, 1A8H, 228H, 2A8H, 328H, 3A8H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL1_EVEN	DL1_ODD	TS16_EN	DL1_TS[4]	DL1_TS[3]	DL1_TS[2]	DL1_TS[1]	DL1_TS[0]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	1	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 0, this register is used for the Receive HDLC #1.

### DL1 EVEN:

- = 0: the data is not extracted from the even frames.
- = 1: the data is extracted from the even frames.

The even frames are FAS frames.

# DL1 ODD:

- = 0: the data is not extracted from the odd frames.
- = 1: the data is extracted from the odd frames.

The odd frames are NFAS frames.

#### TS16 EN:

This bit is valid when the DL1\_EVEN and DL1\_ODD are both 0.

- = 0: the data is not extracted from the TS16.
- = 1: the data is extracted from the TS16.

### DL1 TS[4:0]:

These bits represent the binary value of the timeslot to extract the data from. They are invalid when the DL1\_EVEN and the DL1\_ODD are both logic 0.

# E1 RHDLC Data Link 1 Bit Select (TXCISEL = 0) (029H, 0A9H, 129H, 1A9H, 229H, 2A9H, 3A9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL1_BIT[7]	DL1_BIT[6]	DL1_BIT[5]	DL1_BIT[4]	DL1_BIT[3]	DL1_BIT[2]	DL1_BIT[1]	DL1_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 0, this register is used for the Receive HDLC #1.

# DL1\_BITn:

- = 0: the data is not extracted from the corresponding bit.
- = 1: the data is extracted from the corresponding bit of the assigned timeslot.

These bits are invalid when the DL1\_EVEN and the DL1\_ODD are both logic 0.

The DL1\_BIT[7] corresponds to the first bit (MSB) of the selected timeslot.

## E1 RHDLC Receive Data Link 2 Control (TXCISEL = 0) (02AH, 0AAH, 12AH, 1AAH, 22AH, 2AAH, 3AAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_EVEN	DL2_ODD		DL2_TS[4]	DL2_TS[3]	DL2_TS[2]	DL2_TS[1]	DL2_TS[0]
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 0, this register is used for the Receive HDLC #2.

# DL2 EVEN:

- = 0: the data is not extracted from the even frames.
- = 1: the data is extracted from the even frames.

The even frames are FAS frames.

### DL2 ODD:

- = 0: the data is not extracted from the odd frames.
- = 1: the data is extracted from the odd frames.

The odd frames are NFAS frames.

## DL2 TS[4:0]:

These bits represent the binary value of the timeslot to extract the data from. They are invalid when the DL2\_EVEN and the DL2\_ODD are both logic 0.

# E1 RHDLC Data Link 2 Bit Select (TXCISEL = 0) (02BH, 0ABH, 12BH, 1ABH, 22BH, 2ABH, 32BH, 3ABH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_BIT[7]	DL2_BIT[6]	DL2_BIT[5]	DL2_BIT[4]	DL2_BIT[3]	DL2_BIT[2]	DL2_BIT[1]	DL2_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 0, this register is used for the Receive HDLC #2.

# DL2\_BITn:

- = 0: the data is not extracted from the corresponding bit.
- = 1: the data is extracted from the corresponding bit of the assigned timeslot.

These bits are invalid when the DL2\_EVEN and the DL2\_ODD are both logic 0.

The DL2\_BIT[7] corresponds to the first bit (MSB) of the selected timeslot.

# E1 RHDLC Receive Data Link 3 Control (TXCISEL = 0) (02CH, 0ACH, 12CH, 1ACH, 22CH, 2ACH, 32CH, 3ACH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL3_EVEN	DL3_ODD		DL3_TS[4]	DL3_TS[3]	DL3_TS[2]	DL3_TS[1]	DL3_TS[0]
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 0, this register is used for the Receive HDLC #3.

### DL3 EVEN:

- = 0: the data is not extracted from the even frames.
- = 1: the data is extracted from the even frames.

The even frames are FAS frames.

# DL3 ODD:

- = 0: the data is not extracted from the odd frames.
- = 1: the data is extracted from the odd frames.

The odd frames are NFAS frames.

# DL3\_TS[4:0]:

These bits represent the binary value of the timeslot to extract the data from. They are invalid when the DL3\_EVEN and the DL3\_ODD are both logic 0.

# E1 RHDLC Data Link 3 Bit Select (TXCISEL = 0) (02DH, 0ADH, 12DH, 1ADH, 22DH, 2ADH, 32DH, 3ADH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL3_BIT[7]	DL3_BIT[6]	DL3_BIT[5]	DL3_BIT[4]	DL3_BIT[3]	DL3_BIT[2]	DL3_BIT[1]	DL3_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 0, this register is used for the Receive HDLC #3.

# DL3 BITn:

- = 0: the data is not extracted from the corresponding bit.
- = 1: the data is extracted from the corresponding bit of the assigned timeslot.

These bits are invalid when the DL3\_EVEN and the DL3\_ODD are both logic 0.

The DL3\_BIT[7] corresponds to the first bit (MSB) of the selected timeslot.

## E1 THDLC Transmit Data Link 1 Control (TXCISEL = 1) (028H, 0A8H, 128H, 1A8H, 228H, 2A8H, 3A8H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL1_EVEN	DL1_ODD	TS16_EN	DL1_TS[4]	DL1_TS[3]	DL1_TS[2]	DL1_TS[1]	DL1_TS[0]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	1	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 1, this register is used for the Transmit HDLC #1.

### DL1 EVEN:

- = 0: the data is not inserted to the even frames.
- = 1: the data is inserted to the even frames.

The even frames are FAS frames.

#### DL1 ODD:

- = 0: the data is not inserted to the odd frames.
- = 1: the data is inserted to the odd frames.

The odd frames are NFAS frames.

#### TS16 EN:

This bit is valid when the DL1\_EVEN and DL1\_ODD are both 0 and the CCS is selected (the SIGEN [b6, E1-040H] and the DLEN [b5, E1-040H] are logic 1).

- = 0: the data is not inserted to the TS16.
- = 1: the data is inserted to the TS16.

# DL1\_TS[4:0]:

The data is inserted into the timeslot defined by the binary number in these bits. They are invalid when the DL1\_EVEN and the DL1\_ODD are both logic 0.

# E1 THDLC Data Link 1 Bit Select (TXCISEL = 1) (029H, 0A9H, 129H, 1A9H, 229H, 2A9H, 3A9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL1_BIT[7]	DL1_BIT[6]	DL1_BIT[5]	DL1_BIT[4]	DL1_BIT[3]	DL1_BIT[2]	DL1_BIT[1]	DL1_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 1, this register is used for the Transmit HDLC #1.

# DL1\_BITn:

- = 0: the data is not inserted to the corresponding bit.
- = 1: the data is inserted to the corresponding bit of the assigned timeslot.

These bits are invalid when the DL1\_EVEN and the DL1\_ODD are both logic 0.

The DL1\_BIT[7] corresponds to the first bit (MSB) of the selected timeslot.

## E1 THDLC Transmit Data Link 2 Control (TXCISEL = 1) (02AH, 0AAH, 12AH, 1AAH, 22AH, 2AAH, 3AAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_EVEN	DL2_ODD		DL2_TS[4]	DL2_TS[3]	DL2_TS[2]	DL2_TS[1]	DL2_TS[0]
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 1, this register is used for the Transmit HDLC #2.

### DL2 EVEN:

- = 0: the data is not inserted to the even frames.
- = 1: the data is inserted to the even frames.

The even frames are FAS frames.

# DL2 ODD:

- = 0: the data is not inserted to the odd frames.
- = 1: the data is inserted to the odd frames.

The odd frames are NFAS frames.

# DL2\_TS[4:0]:

The data is inserted into the timeslot defined by the binary number in these bits. They are invalid when the DL2\_EVEN and the DL2\_ODD are both logic 0.

# E1 THDLC Data Link 2 Bit Select (TXCISEL = 1) (02BH, 0ABH, 12BH, 1ABH, 22BH, 2ABH, 32BH, 3ABH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_BIT[7]	DL2_BIT[6]	DL2_BIT[5]	DL2_BIT[4]	DL2_BIT[3]	DL2_BIT[2]	DL2_BIT[1]	DL2_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 1, this register is used for the Transmit HDLC #2.

# DL2 BITn:

- = 0: the data is not inserted to the corresponding bit.
- = 1: the data is inserted to the corresponding bit of the assigned timeslot.

These bits are invalid when the DL2\_EVEN and the DL2\_ODD are both logic 0.

The DL2 BIT[7] corresponds to the first bit (MSB) of the selected timeslot.

# E1 THDLC Transmit Data Link 3 Control (TXCISEL = 1) (02CH, 0ACH, 12CH, 1ACH, 22CH, 2ACH, 32CH, 3ACH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL3_EVEN	DL3_ODD		DL3_TS[4]	DL3_TS[3]	DL3_TS[2]	DL3_TS[1]	DL3_TS[0]
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 1, this register is used for the Transmit HDLC #3.

### DL3 EVEN:

- = 0: the data is not inserted to the even frames.
- = 1: the data is inserted to the even frames.

The even frames are FAS frames.

# DL3 ODD:

- = 0: the data is not inserted to the odd frames.
- = 1: the data is inserted to the odd frames.

The odd frames are NFAS frames.

### DL3 TS[4:0]:

The data is inserted into the timeslot defined by the binary number in these bits. They are invalid when the DL3\_EVEN and the DL3\_ODD are both logic 0.

# E1 THDLC Data Link 3 Bit Select (TXCISEL = 1) (02DH, 0ADH, 12DH, 1ADH, 22DH, 2ADH, 32DH, 3ADH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL3_BIT[7]	DL3_BIT[6]	DL3_BIT[5]	DL3_BIT[4]	DL3_BIT[3]	DL3_BIT[2]	DL3_BIT[1]	DL3_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, E1-00AH) is 1, this register is used for the Transmit HDLC #3.

# DL3\_BITn:

- = 0: the data is not inserted to the corresponding bit.
- = 1: the data is inserted to the corresponding bit of the assigned timeslot.

These bits are invalid when the DL3\_EVEN and the DL3\_ODD are both logic 0.

The DL3\_BIT[7] corresponds to the first bit (MSB) of the selected timeslot.

# E1 FRMP Frame Alignment Options (030H, 0B0H, 130H, 1B0H, 230H, 2B0H, 330H, 3B0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	CRCEN	CASDIS	C2NCIWCK	Reserved		REFR	REFCRCE	REFRDIS
Туре	R/W	R/W	R/W			R/W	R/W	R/W
Default	1	0	0			0	1	0

#### CRCEN:

- = 0: disable searching for the CRC Multi-Frame.
- = 1: enable searching for the CRC Multi-Frame alignment signal and monitor the errors in the CRC Multi-Frame.

### CASDIS:

- = 0: enable searching for the Channel Associated Signaling (CAS) Multi-Frame alignment signal and monitor the errors in the Signaling Multi-Frame.
  - = 1: disable searching for the Channel Associated Signaling Multi-Frame.

#### C2NCIWCK:

- = 0: stop searching for the CRC Multi-Frame alignment signal in CRC to non-CRC inter-working mode.
- = 1: continue searching for the CRC Multi-Frame alignment signal even if CRC to non-CRC inter-working has been declared.

#### REFR:

A transition from logic 0 to logic 1 forces to re-search for a new Basic Frame.

#### REFCRCE:

This bit decides if the Frame Processor re-searches for the Basic Frame when there are excessive CRC errors. The excessive CRC errors is defined as more than 914 CRC errors in one second. One CRC error is counted when the local calculated CRC-4 is not equal to the received CRC-4

- = 0: disable re-searching for the Basic Frame when there are excessive CRC errors.
- = 1: enable re-searching for the Basic Frame when there are excessive CRC errors.

#### REFRDIS:

- 0 = enable re-searching for the Basic Frame when it is out of basic frame sync or there are excessive CRC errors.
- 1 = "locked in frame" once initial frame alignment has been found. Disable re-searching for the Basic Frame under any error conditions once the initial Basic Frame sync is acquired.

While the FRMP remains locked in frame due to REFRDIS=1, a received AIS will not be detected since the Frame Processor must be out-of-frame to detect AIS.

# E1 FRMP Maintenance Mode Options (031H, 0B1H, 131H, 1B1H, 231H, 2B1H, 331H, 3B1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		BIT2C	SMFASC	TS16C	RAIC		AISC	EXCRCERR
Туре	Reserved	R/W	R/W	R/W	R/W	Reserved	R/W	R
Default		1	0	0	0		Х	Х

## BIT2C:

- = 0: out of basic frame sync is declared on 3 consecutive FAS errors.
- = 1: enable the additional criteria to declare out of basic frame sync. Thus, out of basic frame sync is declared when 3 consecutive logic 0 are received in bit 2 of TS0 of NFAS or 3 consecutive FAS are in errors.

#### SMFASC:

- = 0: enable the declaration of out of signaling multi-frame sync when 2 consecutive Signaling Multi-Frame alignment patterns have been received in error.
- = 1: enable the declaration of out of signaling multi-frame sync when 2 consecutive Signaling Multi-Frame alignment patterns have been received in error or when all the content in TS16 of Frame 0 are logic 0 for one or two consecutive multi-frames which is defined in TS16C (b4, E1-031H).

#### TS16C:

Valid when the SMFASC (b5, E1-031H) is logic 1.

- = 0: enable the declaration of out of signaling multi-frame sync when all the content in TS16 are logic 0 for one multi-frame.
- = 1: enable the declaration of out of signaling multi-frame sync when all the content in TS16 are logic 0 for two consecutive multi-frames.

### RAIC:

- = 0: set the RAIV (b7, E1-037H) to be logic 1 on the reception of any A bit being logic one, and set the RAIV (b7, E1-037H) to be logic 0 on the reception of any A bit being logic zero.
- = 1: set the RAIV (b7, E1-037H) to be logic 1 on the reception of the A bit being logic one for 4 or more consecutive occasions, and set the RAIV (b7, E1-037H) to be logic 0 on the reception of any A bit being logic zero.

### AISC:

- = 0: set the AISD (b5, E1-037H) to logic 1 when it is out of basic frame sync and less than 3 zeros are detected in a 512-bit stream, and set the AISD (b5, E1-037H) to logic 0 when 3 or more zeros are detected in a 512-bit stream.
- = 1: set the AISD (b5, E1-037H) to logic 1 when it is out of basic frame sync and less than 3 zeros are detected in each of 2 consecutive 512-bit stream, and set the AISD (b5, E1-037H) to logic 0 when 3 or more zeros are detected in each of 2 consecutive 512-bit stream.

#### EXCRCERR:

The excessive CRC errors is defined as more than 914 CRC errors in one second. One CRC error is counted when the local calculated CRC-4 is not equal to the received CRC-4.

- = 0: normal operation.
- = 1: indicate that there are excessive CRC errors in the received data stream.

This bit is cleared to 0 after it is read

# E1 FRMP Framing Status Interrupt Enable (032H, 0B2H, 132H, 1B2H, 232H, 2B2H, 332H, 3B2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	C2NCIWE	OOFE	OOSMFE	OOCMFE	COFAE	FERE	SMFERE	CMFERE
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	1	0

#### C2NCIWE:

- = 0: disable the interrupt on the INT pin when the C2NCIWI (b7, E1-034H) is logic one.
- = 1: enable the interrupt on the INT pin when the C2NCIWI is logic one.

#### OOFE:

- = 0: disable the interrupt on the INT pin when the OOFI (b6, E1-034H) is logic one.
- = 1: enable the interrupt on the INT pin when the OOFI is logic one.

#### OOSMFE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the OOSMFI (b5, E1-034H) is logic one.
- = 1: enable the interrupt on the INT pin when the OOSMFI is logic one.

#### OOCMFE:

- = 0: disable the interrupt on the INT pin when the OOCMFI (b4, E1-034H) is logic one.
- = 1: enable the interrupt on the INT pin when the OOCMFI is logic one.

#### COFAE:

- = 0: disable the interrupt on the INT pin when the position of the basic frame alignment signal changes.
- = 1: enable the interrupt on the  $\overline{\rm INT}$  pin when the position of the basic frame alignment signal changes.

#### FERE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when there is error in the basic frame alignment pattern.
- = 1: enable the interrupt on the INT pin when there is error in the basic frame alignment pattern.

#### SMFERE:

- = 0: disable the interrupt on the INT pin when there is an error in the signaling multi-frame alignment pattern.
- = 1: enable the interrupt on the INT pin when there is an error in the signaling multi-frame alignment pattern.

### CMFERE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when there is error in the CRC multi-frame alignment pattern.
- = 1: enable the interrupt on the  $\overline{\text{INT}}$  pin when there is error in the CRC multi-frame alignment pattern.

## E1 FRMP Maintenance / Alarm Status Interrupt Enable (033H, 0B3H, 133H, 1B3H, 233H, 2B3H, 333H, 3B3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RAIE	RMAIE	AISDE		REDE	AISE	FEBEE	CRCEE
Туре	R/W	R/W	R/W	Reserved	R/W	R/W	R/W	R/W
Default	0	0	0		0	0	0	0

#### RAIE:

- = 0: disable the interrupt on the INT pin when the RAII (b7, E1-035H) is logic one.
- = 1: enable the interrupt on the  $\overline{\text{INT}}$  pin when the RAII is logic one.

# RMAIE:

- = 0: disable the interrupt on the INT pin when the RMAII (b6, E1-035H) is logic one.
- = 1: enable the interrupt on the INT pin when the RMAII is logic one.

### AISDE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the AISDI (b5, E1-035H) is logic one.
- = 1: enable the interrupt on the INT pin when the AISDI is logic one.

#### REDE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the REDI (b3, E1-035H) is logic one.
- = 1: enable the interrupt on the INT pin when the REDI is logic one.

#### AISE:

- = 0: disable the interrupt on the INT pin when the AISI (b2, E1-035H) is logic one.
- = 1: enable the interrupt on the  $\overline{INT}$  pin when the AISI is logic one.

### FEBEE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when a logic 0 is received in the E1 (the first bit in TS0 in the 13th Frame of CRC-4 Multi-Frame) or E2 (the first bit in TS0 in the 15th Frame of CRC-4 Multi-Frame) bit.
  - = 1: enable the interrupt on the  $\overline{\text{INT}}$  pin when a logic 0 is received in the E1 or E2 bit.

#### CRCEE:

- = 0: disable the interrupt on the INT pin when there is difference between the calculated CRC-4 remainder and the received CRC-4.
- = 1: enable the interrupt on the INT pin when there is difference between the calculated CRC-4 remainder and the received CRC-4.

## E1 FRMP Framing Status Interrupt Indication (034H, 0B4H, 134H, 1B4H, 234H, 2B4H, 334H, 3B4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	C2NCIWI	OOFI	OOSMFI	OOCMFI	COFAI	FERI	SMFERI	CMFERI
Туре	R	R	R	R	R	R	R	R
Default	X	Х	Х	Χ	X	Χ	X	Х

All the bits in this register are clear to 0 after the register is read.

#### C2NCIWI:

- = 0: no status change on the C2NCIWV (b7, E1-036H).
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the C2NCIWV (b7, E1-036H).

### OOFI:

- = 0: no status change on the OOFV (b6, E1-036H)
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the OOFV (b6, E1-036H).

### OOSMFI:

- = 0: no status change on the OOSMFV (b5, E1-036H)
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the OOSMFV (b5, E1-036H).

### OOCMFI:

- = 0: no status change on the OOCMFV (b4, E1-036H)
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the OOCMFV (b4, E1-036H).

### COFAI:

- = 0: the position of the basic frame alignment signal does not change
- = 1: the position of the basic frame alignment signal changes.

# FERI:

- = 0: there is no error in the basic frame alignment pattern.
- = 1: there is an error in the basic frame alignment pattern.

### SMFERI:

- = 0: there is no error in the signaling multi-frame alignment pattern.
- = 1: there is an error in the signaling multi-frame alignment pattern.

#### CMFERI:

- = 0: there is no error in the CRC multi-frame alignment pattern.
- = 1: there is an error in the CRC multi-frame alignment pattern.

## E1 FRMP Maintenance / Alarm Status Interrupt Indication (035H, 0B5H, 135H, 1B5H, 235H, 2B5H, 335H, 3B5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RAII	RMAII	AISDI		REDI	AISI	FEBEI	CRCEI
Туре	R	R	R	Reserved	R	R	R	R
Default	Х	Х	Х		X	Х	X	Х

All the bits in this register are clear to 0 after the register is read.

#### RAII:

- = 0: no status change on the RAIV (b7, E1-037H).
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the RAIV (b7, E1-037H).

# RMAII:

- = 0: no status change on the RMAIV (b6, E1-037H).
- = 1: there is a transition from 0 to 1 or from 1 to 0 on the RMAIV (b6, E1-037H).

# AISDI:

- = 0: no status change on the AISD (b5, E1-037H).
- = 1: there is a transition from 0 to 1 or from 1 to 0 on the AISD (b5, E1-037H).

#### REDI:

- = 0: no status change on the RED (b3, E1-037H).
- = 1: there is a transition from 0 to 1 or from 1 to 0 on the RED (b3, E1-037H).

### AISI:

- = 0: no status change on the AIS (b2, E1-037H).
- = 1: there is a transition from 0 to 1 or from 1 to 0 on the AIS (b2, E1-037H).

# FEBEI:

- = 0: No logic 0 is received in the E1 or E2 bit.
- = 1: A logic 0 is received in the E1 or E2 bit.

### CRCEI:

- = 0: No difference between the calculated CRC-4 remainder and the received CRC-4.
- = 1: There is difference between the calculated CRC-4 remainder and the received CRC-4 remainder.

## E1 FRMP Framing Status (036H, 0B6H, 136H, 1B6H, 236H, 2B6H, 336H, 3B6H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	C2NCIWV	OOFV	OOSMFV	OOCMFV	OOOFV	RAICCRCV	CFEBEV	V52LINKV
Туре	R	R	R	R	R	R	R	R
Default	X	X	Х	Х	Χ	Х	Χ	Х

# C2NCIWV:

- = 0: the Frame Processor does not operate in CRC to non-CRC inter-working mode.
- = 1: the Frame Processor operates in CRC to non-CRC inter-working mode.

# OOFV:

- = 0: the Basic Frame is in sync.
- = 1: the Basic Frame is out of sync.

## OOSMFV:

- = 0: the Signaling Multi-Frame is in sync.
- = 1: the Signaling Multi-Frame is out of sync.

# OOCMFV:

- = 0: the CRC Multi-Frame is in sync.
- = 1: the CRC Multi-Frame is out of sync

### OOOFV:

- = 0: the offline frame is in sync.
- = 1: the offline frame is out of sync.

#### RAICCRCV:

- = 0: normal operation.
- = 1: the remote alarm (logic 1 in A bit) and the FEBE (logic 0 in bit E1 or E2) have existed for a period of 10ms.

### CFEBEV:

- = 0: normal operation.
- = 1: FEBE (logic 0 in bit E1 and E2) has existed for more than or equal to 990 occasions in each second for 5 consecutive seconds.

# V52LINKV:

- = 0: V5.2 link ID signal is not received.
- = 1: V5.2 link ID signal is received, i.e., 2 out of 3 Sa7 bits are logic zeros

E1 FRMP Maintenance / Alarm Status (037H, 0B7H, 137H, 1B7H, 237H, 2B7H, 337H, 3B7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RAIV	RMAIV	AISD		RED	AIS		
Туре	R	R	R	Reserved	R	R	Reserved	
Default	X	X	X		X	Х		

## RAIV:

This bit indicates the value of the Remote Alarm Indication (A) bit.

- = 0: the A bit is logic 0.
- = 1: RAI is detected according to the criterion set in the RAIC (b3, E1-031H). When the RAIC is 0, RAI is detected when the A bit is received as logic 1. When the RAIC is 1, RAI is detected when the A bit is received as logic 1 for 4 or more consecutive occasions.

The RAIV is updated every two frames.

#### RMAIV:

This bit indicates the value of the Remote Signaling Multi-Frame Alarm Indication (Y) bit.

- = 0: the Y bit is logic 0.
- = 1: logic 1 has been received in Y bit for 3 consecutive signaling multi-frames.

The RMAIV is updated every 16 frames.

### AISD:

This bit indicates the Alarm Indication Signal (AIS) detect value. The detection of AIS is disabled in unframed mode.

- = 0: AIS is clear according to the criterion set in the AISC (b1, E1-031H). When the AISC is 0, AIS is clear when 3 or more zeros are detected in a 512-bit stream. When the AISC is 1, AIS is clear when 3 or more zeros are detected in each of 2 consecutive 512-bit stream.
- = 1: AIS is detected according to the criterion set in the AISC (b1, E1-031H). When the AISC is 0, AIS is detected when it is out of basic frame sync and less than 3 zeros are detected in a 512-bit stream. When the AISC is 1, AIS is detected when it is out of basic frame sync and less than 3 zeros are detected in each of 2 consecutive 512-bit stream.

The AISD bit is updated once every 512 bit periods.

#### RED:

- = 0: out of basic frame sync has been absent for 100ms.
- = 1: out of basic frame sync has persisted for 100ms.

#### AIS:

- = 0: the condition of AIS has been absent for 100ms.
- = 1: the condition of AIS has persisted for 100ms.

## E1 FRMP Timeslot0 International / National Bits (038H, 088H, 138H, 188H, 238H, 288H, 338H, 388H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	Si[1]	Si[0]	Α	Sa[4]	Sa[5]	Sa[6]	Sa[7]	Sa[8]
Туре	R	R	R	R	R	R	R	R
Default	X	Χ	Χ	Χ	X	Х	X	Х

The content in this register reflects the international bits, Remote Alarm Indication bit and national bits. The Si[1:0] bits are the international bits. The A bit is the Remote Alarm Indication bit. The Sa[4:8] bits are the national bits. Their position is shown in the following table:

Frame		the Eight Bits in TS0								
Туре	0	0 1 2 3 4 5 6 7								
FAS	Si[1]	0	0	1	1	0	1	1		
NFAS	Si[0]	1	Α	Sa[4]	Sa[5]	Sa[6]	Sa[7]	Sa[8]		

Note that the contents of this register are not updated while the received data stream is out of Basic Frame.

# Si[1]:

Directly reflect the content in the International bit in the latest received FAS frame and is updated on the generation of the IFPI interrupt on FAS frames.

## Si[0]:

Directly reflect the content in the International bit in the latest received NFAS frame and is updated on the generation of the IFPI interrupt on NFAS frames.

### A:

Directly reflect the content in the Remote Alarm Indication (A) bit in the latest received NFAS frame and is updated on the generation of the IFPI interrupt on NFAS frames.

### Sa[4:8]:

Directly reflect the content in the National bit in the latest received NFAS frame and is updated on the generation of the IFPI interrupt on NFAS frames.

# **E1 FRMP CRC Error Counter-LSB** (039H, 0B9H, 139H, 1B9H, 239H, 2B9H, 339H, 3B9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	CRCERR[7]	CRCERR[6]	CRCERR[5]	CRCERR[4]	CRCERR[3]	CRCERR[2]	CRCERR[1]	CRCERR[0]
Туре	R	R	R	R	R	R	R	R
Default	Х	Χ	Х	Х	X	X	Χ	Х

The CRCERR[7:0], together with the CRCERR[9:8], represent the number of the CRC errors and update every second. The CRCERR[0] is the LSB.

E1 FRMP CRC Error Counter-MSB / Timeslot16 Extra Bits (03AH, 0BAH, 13AH, 1BAH, 23AH, 2BAH, 33AH, 3BAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	OVR	NEWDATA	X[0]	Υ	X[1]	X[2]	CRCERR[9]	CRCERR[8]
Туре	R	R	R	R	R	R	R	R
Default	0	0	X	Χ	X	X	X	X

## OVR:

The overwritten means that the data is still written into the CRCERR[9:0] (b1~0, E1-03AH & b7~0, E1-039H) without the data being read in the latest one second interval.

- = 0: the CRCERR[9:0] (b1~0, E1-03AH & b7~0, E1-039H) are not overwritten.
- = 1: the CRCERR[9:0] (b1~0, E1-03AH & b7~0, E1-039H) are overwritten.

This bit is clear to 0 after it is read.

#### NEWDATA:

- = 0: the value in the CRCERR[9:0] (b1~0, E1-03AH & b7~0, E1-039H) has not been updated with new value.
- = 1: the value in the CRCERR[9:0] (b1~0, E1-03AH & b7~0, E1-039H) has been updated with new value.

This bit is clear to 0 after it is read. This bit can be polled to determine the 1 second timing boundary used by the Frame Processor.

# X[0:2], Y:

Directly reflect the content in the Extra bits (X[0:2]) and the Remote Signaling Multi-frame Alarm bit (Y) in Frame0 of TS16 of the latest received Signaling Multi-Frame. They are updated on the generation of the IFPI interrupt on NFAS frames. Note that these bits are not updated when the received data stream is out of Basic Frame. The position of the X[2:0] and Y bit is shown in the following table:

the Eight Bits in TS16								
Frame 0	0	0 1 2 3 4 5 6 7						
	0	0	0	0	X[0]	Υ	X[1]	X[2]

## CRCERR[9:8]:

The CRCERR[9:8], together with the CRCERR[7:0], represent the number of the CRC errors and update every second. The CRCERR[9] is the MSB.

# E1 FRMP National Bit Codeword Interrupt Enables (03BH, 0BBH, 13BH, 1BBH, 23BH, 2BBH, 33BH, 3BBH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	SaSEL[2]	SaSEL[1]	SaSEL[0]	Sa4E	Sa5E	Sa6E	Sa7E	Sa8E
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

# SaSEL[2:0]:

The SaSEL[2:0] select the National Bit Codeword (SaX) to appear in the SaX[1:4] (b3~0, E1-03DH) of the National Bit Codeword register.

SaSEL[2:0]	National Bit Codeword
001	
010	Reserved
011	
100	Sa4
101	Sa5
110	Sa6
111	Sa7
000	Sa8

# Sa4E, Sa5E, Sa6E, Sa7E, Sa8E:

- = 0 (in any of the 5 bits): disable the interrupt on the INT pin when the value is changed in its corresponding SaX[1:4] (b3~0, E1-03DH).
- = 1 (in any of the 5 bits): enable the interrupt on the INT pin when the value is changed in its corresponding SaX[1:4] (b3~0, E1-03DH) (X is 4 through 8).

The interrupt enable should be logic 0 for any bit receiving a HDLC data link.

# E1 FRMP National Bit Codeword Interrupts (03CH, 0BCH, 13CH, 1BCH, 23CH, 2BCH, 33CH, 3BCH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				Sa4I	Sa5I	Sa6I	Sa7l	Sa8I
Туре		Reserved			R	R	R	R
Default				Х	Х	Х	X	Х

#### Sa4I, Sa5I, Sa6I, Sa7I, Sa8I:

- = 0 (in any of the 5 bits): the value is not changed in its corresponding SaX[1:4] (b3~0, E1-03DH) bits (X is 4 through 8).
- = 1 (in any of the 5 bits): the value is changed in its corresponding SaX[1:4] (b3~0, E1-03DH) bits (X is 4 through 8).

This bit is clear to 0 after the register is read.

## E1 FRMP National Bit Codeword (03DH, 0BDH, 13DH, 1BDH, 23DH, 2BDH, 33DH, 3BDH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		-			SaX[1]	SaX[2]	SaX[3]	SaX[4]
Туре		Reserved				R	R	R
Default					X	X	Х	X

These bits directly reflect the content in the SaX nibble codeword of the CRC Sub Multi-Frame. "X" is determined by the SaSEL[2:0] (b7~5, E1-03BH). SaX[1] is the first SaX bit of the Sub Multi-Frame and analogically. The SaX[1:4] are debounced. They are updated only when two consecutive codewords are the same.

### E1 FRMP Frame Pulse/Alarm/V5.2 Link ID Interrupt Enables (03EH, 0BEH, 13EH, 1BEH, 23EH, 2BEH, 33EH, 3BEH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	OOOFE	RAICCRCE	CFEBEE	V52LINKE	IFPE	ICSMFPE	ICMFPE	ISMFPE
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

#### OOOFE:

- = 0: disable the interrupt on the INT pin when the OOOFI (b7, E1-03FH) is logic one.
- = 1: enable the interrupt on the INT pin when the OOOFI (b7, E1-03FH) is logic one.

## RAICCRCE:

- = 0: disable the interrupt on the INT pin when the remote alarm (logic 1 in A bit) and the FEBE (logic 0 in bit E1 or E2) being existed for a period of 10ms.
- = 1: enable the interrupt on the INT pin when the remote alarm (logic 1 in A bit) and the FEBE (logic 0 in bit E1 or E2) being existed for a period of 10ms.

#### CFEBEE:

- = 0: disable the interrupt on the INT pin when the FEBE (logic 0 in bit E1 or E2) has existed for more than 990 occasions in each second for 5 consecutive seconds.
- = 1: enable the interrupt on the INT pin when the FEBE (logic 0 in bit E1 or E2) has existed for more than 990 occasions in each second for 5 consecutive seconds.

#### V52LINKE:

- = 0: disable the interrupt on the INT pin when the V52LINKI is logic one.
- = 1: enable the interrupt on the INT pin when the V52LINKI (b4, E1-03FH) is logic one.

### IFPE:

- = 0: disable the interrupt on the INT pin when the first bit of each basic frame is received.
- = 1: enable the interrupt on the INT pin when the first bit of each basic frame is received.

## ICSMFPE:

- = 0: disable the interrupt on the INT pin when the first bit of each CRC sub-multi-frame is received.
- = 1: enable the interrupt on the INT pin when the first bit of each CRC sub-multi-frame is received.

### ICMFPE:

- = 0: disable the interrupt on the INT pin when the first bit of each CRC multi-frame is received.
- = 1: enable the interrupt on the INT pin when the first bit of each CRC multi-frame is received.

# ISMFPE:

- = 0: disable the interrupt on the INT pin when the first bit of each signaling multi-frame is received.
- = 1: enable the interrupt on the INT pin when the first bit of each signaling multi-frame is received.

## E1 FRMP Frame Pulse / Alarm Interrupts (03FH, 0BFH, 13FH, 1BFH, 23FH, 2BFH, 33FH, 3BFH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	OOOFI	RAICCRCI	CFEBEI	V52LINKI	IFPI	ICSMFPI	ICMFPI	ISMFPI
Туре	R	R	R	R	R	R	R	R
Default	X	Χ	Х	Χ	Χ	Χ	X	Х

The bits of this register are clear to 0 after the register is read.

#### 000FI:

- = 0: there is no transition (from 0 to 1 or from I to 0) on the OOOFV (b3, E1-036H).
- = 1: there is a transition (from 0 to 1 or from I to 0) on the OOOFV (b3, E1-036H).

## RAICCRCI:

- = 0: there is no transition from normal operation to the remote alarm (logic 1 in A bit) or the FEBE (logic 0 in bit E1 or E2) has being absent for a period of 10ms.
- = 1: there is a transition from normal operation to the remote alarm (logic 1 in A bit) and the FEBE (logic 0 in bit E1 or E2) has being existed for a period of 10ms.

#### CFEBEI:

- = 0: there is no transition from normal operation to FEBE (logic 0 in bit E1 or E2) existed for more than 990 occasions in each second for 5 consecutive seconds.
- = 1: there is a transition from normal operation to FEBE (logic 0 in bit E1 or E2) existed for more than 990 occasions in each second for 5 consecutive seconds.

#### V52LINKI:

- = 0: there is no transition (from 0 to 1 or from 1 to 0) on the V52LINKV (b0, E1-036H).
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the V52LINKV (b0, E1-036H).

# IFPI:

- = 0: the received bit is not the first bit of each Basic Frame.
- = 1: the first bit of each Basic Frame is received.

### ICSMFPI:

- = 0: the received bit is not the first bit of each CRC sub-Multi-Frame.
- = 1: the first bit of each CRC sub-Multi-Frame is received.

### ICMFPI:

- = 0: the received bit is not the first bit of each CRC multi-frame.
- = 1: the first bit of each CRC multi-frame is received.

#### ISMFPI:

- = 0: the received bit is not the first bit of each Signaling Multi-Frame.
- = 1: the first bit of each Signaling Multi-Frame is received.

E1 FRMG Configuration (040H, 0C0H, 140H, 1C0H, 240H, 2C0H, 340H, 3C0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FRESH	SIGEN	DLEN	GENCRC	FDIS	FEBEDIS	INDIS	XDIS
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	1	1	0	0	0	0	0

## FRESH:

- = 0: normal operation.
- = 1: initiate the FIFO in the Frame Generator block.

After initialization of the backplane interface, the user should write 1 into this bit and then clear it.

## SIGEN, DLEN:

These two bits select the signaling sources for TS16. They are valid when the AIS (b0, E1-041H) is logic 0:

SIGEN, DLEN	Signaling Source
0 0	Signaling insertion disabled or CCS enabled. TS16 data is taken directly from the input TSDn TS16 or from the THDLC if the THDLC selects this inserted position. The XDIS (b0, E1-040H) must also be set to logic 1 to disable the insertion of the extra bits in TS16 of frame 0.
0 1	Reserved
0 1	Reserved
11	CAS enabled. TS16 data is taken from either TSSIGn stream or from the TPLC Signaling/PCM Control byte as selected on a per-timeslot basis via the SIGSRC (b4, E1-TPLC-indirect registers - 61~7FH). However, the TS16 of Frame0 of Signaling Multi-Frame is overwritten by '0000X[0]YX[1]X[2]'.

#### GENCRC:

- = 0: CRC Multi-Frame generation is disabled. Then the International Bits are replaced with the value contained in the Si[1:0] (b7~6, E1-042H) if the INDIS (b1, E1-040H) is enabled (logic 0), or, if the INDIS (b1, E1-040H) is not enabled, the international bits are taken directly from TSDn/MTSD.
- = 1: CRC Multi-Frame generation is enabled. When CRC Multi-Frame is generated, the international bits on the TSDn pin are replaced with CRC Multi-Frame alignment pattern and calculated CRC-4 bits. The CRC bits calculated during the transmission of the SMFn are transmitted in the following SMF (SMF n+1). If the FEBEDIS (b2, E1-040H) is enabled (logic 0), the FEBE indication is inserted in the E1 and E2 bit positions. The setting to 1 is valid when the FDIS (b3, E1-040H) and the INDIS (b1, E1-040H) are logic 0.

### FDIS:

- = 0: replace the data on the TS0 of FAS on the TSDn/MTSD pin with Basic Frame alignment sequence (FAS).
- = 1: keep the data on the TSDn/MTSD pin to pass through the Frame Generation transparently. The values in the control bits GENCRC (b4, E1-040H), FEBEDIS (b2, E1-040H) and INDIS (b1, E1-040H) are ignored.

#### FEBEDIS:

Valid when the FDIS (b3, E1-040H) and the INDIS (b1, E1-040H) are logic 0 and the GENCRC (b4, E1-040H) is logic 1.

- = 0: the international bit of frame 13 & 15 are for FEBE indication.
- = 1: FEBE indication is disabled.

# INDIS:

- = 0: enabled to replace the international bit.
- = 1: disable to replace the international bit. The value of the international bit is directly taken from the TSDn/MTSD or from the THDLC if the THDLC selects this inserted position.

### XDIS:

Valid when FDIS (b3, E1-040H) is logic 0, and the SIGEN (b6, E1-040H) and the DLEN (b5, E1-040H) are logic 1.

- = 0: replace the extra bits with the setting in the X[2:0].
- = 1: ignore the setting in the X[2:0] bits. The value in the extra bits is taken from the TSDn/MTSD.

# E1 FRMG Transmit Alarm / Diagnostic Control (041H, 0C1H, 141H, 1C1H, 241H, 2C1H, 341H, 3C1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	MTRK	FPATINV	SPLRINV	SPATINV	REMAIS	MFAIS	TS16AIS	AIS
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

#### MTRK:

Valid when the FDIS (b3, E1-040H) is logic 0 and the PCCE (b0, E1-060H) is 1.

- = 0: ignore the setting in the IDLE Code Byte register.
- = 1: replace the data on the TS1~15 & TS17~31 with the IDLE code. And when the SIGEN (b6, E1-040H) is logic 1, replace the data on the TS16 with signaling; when the SIGEN (b6, E1-040H) is logic 0, replace the data on the TS16 with IDLE code.

## FPATINV:

Valid when the FDIS (b3, E1-040H) is logic 0.

- = 0: disable the inversion of the FAS.
- = 1: enable the inversion of the FAS (from '0011011' to '1100100').

### SPLRINV:

Valid when the FDIS (b3, E1-040H) is logic 0.

- = 0: disable the inversion of the 2nd bit of NFAS.
- = 1: enable the inversion of the 2nd bit of NFAS (from 1 to 0).

#### SPATINV:

Valid when the FDIS (b3, E1-040H) is logic 0 and the SIGEN (b6, E1-040H) & the DLEN (b5, E1-040H) are logic 1.

- = 0: disable the inversion of the Signaling Multi-Frame alignment signal.
- = 1: enable the inversion of the Signaling Multi-Frame alignment signal (from '0000' to '1111').

### REMAIS:

Valid when the FDIS (b3, E1-040H) is logic 0.

- = 0: normal operation.
- = 1: force the 3rd bit of NFAS to be logic 1.

### MFAIS:

Valid when the FDIS (b3, E1-040H) is logic 0.

- = 0: normal operation.
- = 1: force to transmit the Y bit as logic 1.

### TS16AIS:

Valid when the FDIS (b3, E1-040H) is logic 0 and Signaling Multi-Frame generator is enabled.

- = 0: normal transmission.
- = 1: force to transmit all ones in TS16 unconditionally.

### AIS:

- = 0: normal transmission.
- = 1: force to transmit all ones on all timeslots unconditionally.

# **E1 FRMG International Bits Control** (042H, 0C2H, 142H, 1C2H, 242H, 2C2H, 342H, 3C2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	Si[1]	Si[0]						
Туре	R/W	R/W			Rese	rved		
Default	1	1						

# Si[1:0]:

Valid when the FDIS (b3, E1-040H) and the INDIS (b1, E1-040H) are logic 0.

When CRC Multi-Frame generation is disabled (GENCRC is logic 0), the Si[1] and Si[0] bits can be programmed to any value and will be inserted into the first of each FAS frame and NFAS frame, respectively. When CRC Multi-Frame generation is enabled (GENCRC is logic 1), and FEBE indication is disabled (FEBEDIS is logic 1), the values programmed in the Si[1] and Si[0] bit positions are inserted into the E1 & E2 bit positions respectively. When GENCRC is logic 1 and FEBEDIS is logic 0, both Si[1] and Si[0] are ignored.

# **E1 FRMG Extra Bits Control** (043H, 0C3H, 143H, 1C3H, 243H, 2C3H, 343H, 3C3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					X[0]		X[1]	X[2]
Туре		Rese	erved		R/W	Reserved	R/W	R/W
Default					1		1	1

# X[2:0]:

Valid when the FDIS (b3, E1-040H), the XDIS (b0, E1-040H), the SIGEN (b6, E1-040H) and the DLEN (b5, E1-040H) are all logic 0. Replace the extra bits located in bits 5, 7 & 8 in TS16 of frame 0 of the Signaling Multi-Frame.

TEMPERATURE RANGES

# E1 FRMG Interrupt Enable (044H, 0C4H, 144H, 1C4H, 244H, 2C4H, 344H, 3C4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				SIGMFE	FASE	MFE	SMFE	
Туре		Reserved		R/W	R/W	R/W	R/W	Reserved
Default				0	0	0	0	(must be 0)

# SIGMFE:

- = 0: disable the interrupt on the INT pin when the SIGMFI (b4, E1-045H) is logic one.
- = 1: enable the interrupt on the INT pin when the SIGMFI (b4, E1-045H) is logic one.

# FASE:

- = 0: disable the interrupt on the  $\overline{INT}$  pin when the FASI (b3, E1-045H) is logic one.
- = 1: enable the interrupt on the INT pin when the FASI (b3, E1-045H) is logic one.

### MFE:

- = 0: disable the interrupt on the  $\overline{\rm INT}$  pin when the MFI (b2, E1-045H) is logic one.
- = 1: enable the interrupt on the  $\overline{INT}$  pin when the MFI (b2, E1-045H) is logic one.

### SMFE:

- = 0: disable the interrupt on the INT pin when the SMFI (b1, E1-045H) is logic one.
- = 1: enable the interrupt on the  $\overline{\text{INT}}$  pin when the SMFI (b1, E1-045H) is logic one.

E1 FRMG Interrupt Status (045H, 0C5H, 145H, 1C5H, 245H, 2C5H, 345H, 3C5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				SIGMFI	FASI	MFI	SMFI	
Туре	Reserved			R	R	R	R	Reserved
Default				Х	X	X	Χ	

The bits in this register are clear to 0 after the register is read.

## SIGMFI:

Valid when the Signaling Multi-Frame is generated and coincides with the CRC Multi-Frame.

- = 0: not at the end of the first frame of a Signaling Multi-Frame.
- = 1: indicate the end of the first frame of a Signaling Multi-Frame.

#### FASI:

Valid when the Basic Frame is generated.

- = 0: not on the boundary of a FAS frame.
- = 1: indicate the boundary of a FAS frame.

#### MFI:

Valid when the CRC-4 Multi-Frame is generated.

- = 0: not at the end of the first frame of a CRC-4 Multi-Frame.
- = 1: indicate the end of the first frame of a CRC-4 Multi-Frame.

# SMFI:

Valid when the CRC-4 Multi-Frame is generated.

- = 0: not at the end of the first frame of a CRC-4 Sub Multi-Frame.
- = 1: indicate the end of the first frame of a CRC-4 Sub Multi-Frame.

# E1 FRMG National Bit Codeword Select (046H, 0C6H, 146H, 1C6H, 246H, 2C6H, 346H, 3C6H)

Bit No.	7	6	5	4	3	2	1	0	
Bit Name	SaSEL[2]	SaSEL[1]	SaSEL[0]	Reserved					
Туре	R/W	R/W	R/W						
Default	0	0	0						

# SaSEL[2:0]:

The SaSEL[2:0] select which National Bit Codeword (SaX) will be replaced by the SaX[1:4] (b3~0, E1-047H).

SaSEL[2:0]	National Bit Codeword
000	
001	Reserved
010	7
011	Sa4
100	Sa5
101	Sa6
110	Sa7
111	Sa8

# E1 FRMG National Bit Codeword (047H, 0C7H, 147H, 1C7H, 247H, 2C7H, 347H, 3C7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	SaX_EN[1]	SaX_EN[2]	SaX_EN[3]	SaX_EN[4]	SaX[1]	SaX[2]	SaX[3]	SaX[4]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	1	1	1	1

### SaX ENn:

Valid when the FDIS (b3, E1-040H) is logic 0, and the INDIS (b1, E1-040H) is logic 0.

- = 0: disable the corresponding bit in the SaX[1:4] to replace the national bit codeword selected by the SaSEL[2:0].
- = 1: enable the corresponding bit in the SaX[1:4] to replace the national bit codeword selected by the SaSEL[2:0].

# SaX[1:4]:

These bits are the codeword to be inserted into a CRC-4 sub-multiframe.

The setting in the SaX[1:4] will replace the national bits which are assigned by the SaSEL[2:0].

If the code word is written during SMF I of a CRC-4 Multi-Frame, it will appear in the SaX[1:4] bits of SMF II of the same Multi-frame. If the code word is written during SMF II of a Multi-Frame, its contents will be latched internally and will appear in SMF I of the next Multi-Frame.

# E1 RHDLC #1, #2, #3 Configuration (048H, 0C8H, 148H, 1C8H, 248H, 2C8H, 348H, 3C8H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					MEN	MM	TR	EN
Туре		Resei	rved		R/W	R/W	R/W	R/W
Default					0	0	0	0

Selection of the RHDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, E1-00AH).

### MEN, MM:

The MEN & MM define the address matching mode:

MEN	MM	Address Matching Mode
0	Χ	No address matching is needed. All the HDLC data are stored in the FIFO.
1	0	The HDLC data are stored in the FIFO when the first byte is all ones or the same as the setting in the PA[7:0] (b7~0,
		E1-04CH) or the SA[7:0] (b7~0, E1-04DH).
1	1	The HDLC data are stored in the FIFO when the most significant 6 bits in the first byte are all ones or the same as the
		setting in the PA[7:2] (b7~2, E1-04CH)or the SA[7:2] (b7~2, E1-04DH).

# TR:

- = 0: Normal operation.
- = 1: force the RHDLC to immediately terminate the reception of the current data frame, empty the FIFO buffer, clear the interrupts and initiate a new HDLC search.

This bit is clear to 0 after a rising and falling edge occur on the internal clock or after the register is read.

### EN:

- = 0: disable the operation of the RHDLC block and all the FIFO buffer and interrupts are cleared.
- = 1: enable the operation of the RHDLC block and the HDLC opening flag will be searched immediately.

If the EN is set from logic 1 to logic 0 and back to logic 1, the RHDLC will immediately terminate the reception of the current data frame, empty the FIFO buffer, clear the interrupts and initiate a new HDLC search.

# E1 RHDLC #1, #2, #3 Interrupt Control (049H, 0C9H, 149H, 1C9H, 249H, 2C9H, 349H, 3C9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	INTE	INTC[6]	INTC[5]	INTC[4]	INTC[3]	INTC[2]	INTC[1]	INTC[0]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

Selection of the RHDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, E1-00AH).

### INTE:

- = 0: disable the interrupt on the INT pin when there is a transition from 0 to 1 on the INTR (b0, E1-04AH).
- = 1: enable the interrupt on the INT pin when there is a transition from 0 to 1 on the INTR (b0, E1-04AH).

# INTC[6:0]:

These bits set the interrupt threshold point of the FIFO buffer. Exceeding the set point will cause an interrupt, and the interrupt will persist until the FIFO is empty. The set point is decimal 128 when the INTC[6:0] is all zeros.

The contents of this register should only be changed when the EN (b0, E1-048H) is logic 0. This prevents any erroneous interrupt generation.

E1 RHDLC #1, #2, #3 Status (04AH, 0CAH, 14AH, 1CAH, 24AH, 2CAH, 34AH, 3CAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FE	OVR	COLS	PKIN	PBS[2]	PBS[1]	PBS[0]	INTR
Туре	R	R	R	R	R	R	R	R
Default	Х	X	Χ	Χ	Χ	Х	X	Х

Selection of the RHDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, E1-00AH).

### FE:

- = 0: the FIFO is loaded with data.
- = 1: the FIFO is empty.

#### OVR:

The overwritten condition occurs when data are written over unread data in the FIFO buffer. This bit is cleared to 0 after the register is read.

- = 0: no overwriting occurs.
- = 1: the FIFO is overwritten, and then the FIFO is reset, which cause the COLS and PKIN to be reset to logic 0.

#### COLS:

This bit reflects the HDLC link status change.

- = 0: normal operation.
- = 1: the first HDLC opening flag sequence (7E) activated the HDLC or the HDLC abort sequence (7F) deactivated the HDLC is detected. This bit is cleared to 0 after the bit is read, or after the OVR transits to be logic 1, or after the EN is clear.

#### PKIN:

- = 0: a HDLC packet has not been written into the FIFO.
- = 1: a HDLC packet has been written into the FIFO.

This bit is cleared to 0 after the bit is read, or after the OVR transitions to logic 1.

#### PBS[2:01:

The PBS[2:0] indicate the status of the last byte read from the FIFO.

PBS[2:0]	Status of the Data
000	Normal data
001	A dummy byte to indicate the first HDLC opening flag sequence (7E) was detected, which means the HDLC link became active.
010	A dummy byte to indicate the HDLC abort sequence (7F) was detected, which means the HDLC link became inactive.
011	Reserved.
100	The last byte of a non-aborted HDLC packet was received. The HDLC packet is in an integer number of bytes and has no FCS
	error
101	The last byte of a non-aborted HDLC packet was received and a non-integer number of bytes are in the packet.
110	The last byte of a non-aborted HDLC packet was received. The HDLC packet is in an integer number of bytes and has FCS errors.
111	The last byte of a non-aborted HDLC packet was received. The HDLC packet is in a non-integer number of bytes and has FCS
	errors

### INTR:

- = 0: no interrupt sources in the HDLC Receiver block occurs
- = 1: any one of the interrupt sources in the HDLC Receiver block occurs. The interrupt sources in the HDLC Receiver are: 1. Receiving the first 7E opening flag sequence which activates the HDLC link; 2. A packet was received; 3. Change of link status; 4. Exceeding the set point of the FIFO which is defined in the INTC[6:0] (b6~0, E1-049H); 5. Over-writting the FIFO.

This bit is cleared to 0 after the bit is read.

TEMPERATURE RANGES

## E1 RHDLC #1, #2, #3 Data (04BH, 0CBH, 14BH, 1CBH, 24BH, 2CBH, 34BH, 3CBH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RD[7]	RD[6]	RD[5]	RD[4]	RD[3]	RD[2]	RD[1]	RD[0]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Х	Х	X	Х	X	Χ	X

Selection of the RHDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, E1-00AH).

## RD[7:0]:

This register represents the bytes read from the FIFO. This register should not be accessed at a rate greater than 1/15 of the XCK rate. The RD[0] corresponds to the first bit of the serial received data from the FIFO.

# E1 RHDLC #1, #2, #3 Primary Address Match (04CH, 0CCH, 14CH, 1CCH, 24CH, 2CCH, 34CH, 3CCH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PA[7]	PA[6]	PA[5]	PA[4]	PA[3]	PA[2]	PA[1]	PA[0]
Туре	R/W							
Default	1	1	1	1	1	1	1	1

Selection of the RHDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, E1-00AH).

# PA[7:0]:

These bits stipulate the primary address pattern.

PA[0] stores the first bit of the serial data.

## E1 RHDLC #1, #2, #3 Secondary Address Match (04DH, 0CDH, 14DH, 1CDH, 24DH, 2CDH, 34DH, 3CDH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	SA[7]	SA[6]	SA[5]	SA[4]	SA[3]	SA[2]	SA[1]	SA[0]
Туре	R/W							
Default	1	1	1	1	1	1	1	1

Selection of the RHDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, E1-00AH).

### SA[7:0]:

These bits stipulate the secondary address pattern.

SA[0] stores the first bit of the serial data.

## E1 THDLC #1, #2, #3 Configuration (050H, 0D0H, 150H, 1D0H, 250H, 2D0H, 350H, 3D0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FLGSHARE	FIFOCLR			EOM	ABT	CRC	EN
Туре	R/W	R/W	Rese	Reserved		R/W	R/W	R/W
Default	1	0			0	0	1	0

Selection of the THDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, E1-00AH).

### FLGSHARE:

- = 0: the opening flag of the next HDLC frame and closing flag of the current HDLC frame are separate.
- = 1: the opening flag of the next HDLC frame and closing flag of the current HDLC frame are shared

#### FIFOCLR:

- = 0: normal operation.
- = 1: clear the FIFO.

#### EOM:

- = 0: normal operation.
- = 1: a positive transition of this bit starts a packet transmission. Then if the CRC(b1, E1-050H) is set, the 16-bit FCS word is appended to the last data byte transmitted.

#### ABT:

- = 0: normal operation.
- = 1: transmit the 7F abort sequence after the current setting in the Transmit Data register is transmitted, so that the FIFO is cleared and all data in the FIFO will be lost.

Aborts are continuously sent and the FIFO is held in reset until this bit is reset to a logic 0. At least one Abort sequence will be sent when the ABT transitions from logic 0 to logic 1.

#### CRC:

- = 0: do not append the CRC-16 frame check sequences (FCS) to the end of the HDLC data.
- = 1: append the FCS to the end of the HDLC data.

#### EN:

- = 0: disable the operation of the THDLC block.
- = 1: enable the operation of the THDLC block and flag sequences are sent until data is written into the THDLC Transmit Data register and the EOM is set to logic 1.

# E1 THDLC #1, #2, #3 Upper Transmit Threshold (051H, 0D1H, 151H, 1D1H, 251H, 2D1H, 351H, 3D1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		UTHR[6]	UTHR[5]	UTHR[4]	UTHR[3]	UTHR[2]	UTHR[1]	UTHR[0]
Туре	Reserved	R/W						
Default		1	0	0	0	0	0	0

Selection of the THDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, E1-00AH).

## UTHR[6:0]:

These bits define the upper fill level of the FIFO. Once the fill level exceeds the UTHR[6:0] value, the data stored in the FIFO will start to transmit. The transmission will not stop until the complete packet is transmitted and the THDLC FIFO fill level is below UTHR[6:0] + 1.

It should be greater than the value of the LINT[6:0] unless both are equal to 00H.

# E1 THDLC #1, #2, #3 Lower Interrupt Threshold (052H, 0D2H, 152H, 1D2H, 252H, 2D2H, 352H, 3D2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		LINT[6]	LINT[5]	LINT[4]	LINT[3]	LINT[2]	LINT[1]	LINT[0]
Туре	Reserved	R/W						
Default		0	0	0	0	1	1	1

Selection of the THDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, E1-00AH).

## LINT[6:0]:

These bits define the fill level of the FIFO that can cause an interrupt. That is, when the fill level of the FIFO is below the LINT[6:0], an interrupt will be generated. The LINT[6:0] should be less than the value of the UTHR[6:0] unless both are equal to 00H.

# E1 THDLC #1, #2, #3 Interrupt Enable (053H, 0D3H, 153H, 1D3H, 253H, 2D3H, 353H, 3D3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					FULLE	OVRE	UDRE	LFILLE
Туре		Rese	rved		R/W	R/W	R/W	R/W
Default					0	0	0	0

Selection of the THDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, E1-00AH).

### FULLE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the FULLI (b3, E1-054H) is logic 1.
- = 1: enable the interrupt on the INT pin when the FULLI (b3, E1-054H) is logic 1.

### OVRE:

- = 0: disable the interrupt on the INT pin when the OVRI (b2, E1-054H) is logic 1.
- = 1: enable the interrupt on the INT pin when the OVRI (b2, E1-054H) is logic 1.

# UDRE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the UDRI (b1, E1-054H) is logic 1.
- = 1: enable the interrupt on the  $\overline{\text{INT}}$  pin when the UDRI (b1, E1-054H) is logic 1.

# LFILLE:

- = 0: disable the interrupt on the INT pin when the LFILLI (b0, E1-054H) is logic 1.
- = 1: enable the interrupt on the INT pin when the LFILLI (b0, E1-054H) is logic 1.

## E1 THDLC #1, #2, #3 Interrupt Status / UDR Clear (054H, 0D4H, 154H, 1D4H, 254H, 2D4H, 354H, 3D4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		FULL	BLFILL		FULLI	OVRI	UDRI	LFILLI
Туре	Reserved	R	R	Reserved	R	R	R	R
Default		Х	Χ		Χ	Χ	Χ	X

Selection of the THDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, E1-00AH).

#### FULL:

- = 0: the THDLC FIFO is not full.
- = 1: the THDLC FIFO is full (128 bytes).

#### BLFILL:

- = 0: the fill level in the THDLC FIFO is not below the value of the LINT[6:0] (b6~0, E1-052H).
- = 1: the fill level in the THDLC FIFO is empty or below the value of the LINT[6:0] (b6~0, E1-052H).

# FULLI:

- = 0: there is no transition (from 0 to 1) on the FULL.
- = 1: there is a transition (from 0 to 1) on the FULL.

This bit is cleared to 0 after the bit is read.

### OVRI:

The Over-Written is that the THDLC FIFO was already full when another data byte was written to the THDLC Transmit Data register.

- = 0: the THDLC FIFO is not overwritten.
- = 1: the THDLC FIFO is overwritten.

This bit is cleared to 0 after the bit is read.

# UDRI:

The Under-Run is that the THDLC was in the process of transmitting a packet when it ran out of data to be transmitted.

- = 0: the THDLC FIFO is not under-run.
- = 1: the THDLC FIFO is under-run.

This bit is cleared to 0 after the bit is read.

### LFILLI:

- = 0: there is no transition (from 0 to 1) on the BLFILL.
- = 1: there is a transition (from 0 to 1) on the BLFILL.

This bit is cleared to 0 after the bit is read.

# E1 THDLC #1, #2, #3 Transmit Data (055H, 0D5H, 155H, 1D5H, 255H, 2D5H, 355H, 3D5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TD[7]	TD[6]	TD[5]	TD[4]	TD[3]	TD[2]	TD[1]	TD[0]
Туре	R/W							
Default	Χ	Χ	Х	Х	Х	X	X	Х

Selection of the THDLC block (#1, #2, or #3) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, E1-00AH).

The content is the data to be transmitted. It is serially transmitted (TD[0] is the first).

# E1 ELSB Interrupt Enable / Status (059H, 0D9H, 159H, 1D9H, 259H, 2D9H, 359H, 3D9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				SLIPE	SLIPD	SLIPI		
Туре		Reserved					R	R
Default						0	Х	Х

## SLIPE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when a slip occurs.
- = 1: enable the interrupt on the INT pin when a slip occurs.

### SLIPD:

This bit makes sense only when the SLIPI is logic 1.

- = 0: the latest slip is due to the Elastic Store Buffer being empty; a frame was duplicated.
- = 1: the latest slip is due to the Elastic Store Buffer being full; a frame was deleted.

### SLIPI:

- = 0: no slip occurs.
- = 1: a slip occurs.

This bit is cleared to 0 after the bit is read.

# E1 ELSB Idle Code (05AH, 0DAH, 15AH, 1DAH, 25AH, 2DAH, 35AH, 3DAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Туре	R/W							
Default	1	1	1	1	1	1	1	1

These bits set the idle code that will replace the data on the RSDn/MRSD when it is out of Basic Frame and the TRKEN (b1, E1-001H) is logic 1. D7 is the first bit to be inserted.

The writing of the idle code pattern is asynchronous with respect to the output data clock. One timeslot of idle code data will be corrupted if the register is written to when the framer is out of frame.

# E1 RPLC Configuration (05CH, 0DCH, 15CH, 1DCH, 25CH, 2DCH, 35CH, 3DCH)

Bit No.	7	6	5	4	3	2	1	0			
Bit Name											
Туре		Reserved									
Default											

# PCCE:

- = 0: the per-TS functions in RPLC are disabled.
- = 1: the per-TS functions in RPLC are enabled.

# E1 RPLC mP Access Status (05DH, 0DDH, 15DH, 1DDH, 25DH, 2DDH, 35DH, 3DDH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	BUSY							
Туре	R				Reserved			
Default	0							

#### BUSY:

- = 0: no reading or writing operation on the indirect registers.
- = 1: an internal indirect register is being accessed, any new operation on the internal indirect register is not allowed.

This bit goes low timed to an internal high-speed clock rising edge after the operation has been completed. The operation cycle is 490ns. No operations to the indirect registers are possible until this bit is logic 0.

# E1 RPLC Channel Indirect Address / Control (05EH, 0DEH, 15EH, 1DEH, 25EH, 2DEH, 35EH, 3DEH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	R/WB	A6	A5	A4	A3	A2	A1	A0
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

Writing to this register with a valid address and R/WB bit initiates an internal operation cycle to the indirect registers.

### R/WB:

- = 0: write the data to the specified indirect register.
- = 1: read the data from the specified indirect register.

# A[6:0]:

Specify the address of the indirect registers (from 20H to 7FH) for the microprocessor access.

### E1 RPLC Channel Indirect Data Buffer (05FH, 0DFH, 15FH, 1DFH, 25FH, 2DFH, 35FH, 3DFH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Туре	R/W							
Default	0	0	0	0	0	0	0	0

This register hold the value which will be read from or write into the indirect registers (from 20H to 7FH). If data is to be written to the indirect registers, the byte to be written must be written into this register before the target indirect register's address and R/WB=0 is written into the Address/Control register, initiating the access. If data is to be read from the indirect registers, only the target indirect register's address and R/WB=1 is written into the Address/Control register, initiating the request. After 490 ns, this register will contain the requested data byte.

RPLC Indirect Registers Map						
20H~3FH	Per-TS Configuration Byte for TS0 ~ TS31					
40H~5FH	Data Trunk Conditioning Code Byte for TS0 ~ TS31					
61H~7FH	Signaling Trunk Conditioning Byte for TS1 ~ TS31					

# E1 RPLC Per-TS Configuration Registers (RPLC Indirect Registers 20H – 3FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TEST	DTRKC/NxTS	STRKC	DMW	DMWALAW	SIGNINV	RINV[1]	RINV[0]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	X	X	X	Х	X	X	X	X

#### TEST:

- = 0: disable the data in the corresponding timeslot to be tested by PRGD.
- = 1: enable the data in the corresponding timeslot to be extracted to PRGD for test (when the RXPATGEN [b2, E1-00CH] is logic 0), or enable the test pattern from PRGD to replace the data in the corresponding timeslot for test (when the RXPATGEN [b2, E1-00CH] is logic 1).

All the timeslots that are extracted to the PRGD are concatenated and treated as a continuous stream in which pseudo random are searched for. Similarly, all timeslots set to be replaced with PRGD test pattern data are concatenated replaced by the PRBS.

# DTRKC/NxTS:

- = 0: disable the data in the corresponding timeslot to be replaced by the data set in the DTRK[7:0] (b7~0, E1-RPLC-indirect registers-40~5FH) when output on the RSDn/MRSD pin.
- = 1: enable the data in the corresponding timeslot to be replaced by the data set in the DTRK[7:0] (b7~0, E1-RPLC-indirect registers-40~5FH) when output on the RSDn/MRSD pin.

In addition, it controls the RSCKn of the corresponding timeslot in Receive Clock Slave Fractional E1 mode:

- = 0: RSCKn is clocked for the corresponding timeslot.
- = 1: RSCKn is held in its inactive state.

#### STRKC:

- = 0: disable the signaling of the corresponding timeslot to be replaced by the data set in the A, B, C, D (b3~0, E1-RPLC-indirect registers-61~7FH) when output on the RSSIGn/MRSSIG pin.
- = 1: enable the signaling of the corresponding timeslot to be replaced by the data set in the A, B, C, D (b3~0, E1-RPLC-indirect registers-61~7FH) when output on the RSSIGn/MRSSIG pin.

## DMW:

- = 0: disallow the data in the corresponding timeslot to be replaced by a digital milliwatt pattern when output on the RSDn/MRSD pin.
- = 1: enable the data in the corresponding timeslot to be replaced by a digital milliwatt pattern when output on the RSDn/MRSD pin.

## DMWALAW:

- = 0: the milliwatt pattern is the  $\mu$ -Law pattern.
- = 1: the milliwatt pattern is the A-Law pattern.

#### SIGNINV, RINV[1:0]:

The SIGNINV and the RINV[1:0] bits select the bits in the corresponding timeslot to be inverted when output on the RSDn/MRSD pin:

SIGNINV	RINV[1:0]	Bits Inverted
0	0 0	No inversion
0	01	Invert the even bits (2, 4, 6, 8) of the timeslot (bit 1 is the MSB)
0	10	Invert the odd bits (1, 3, 5, 7) of the timeslot (bit 1 is the MSB)
0	11	Invert all the bits of the timeslot
1	00	Invert the bit 1 (MSB) of the timeslot
1	01	Invert the bits 1, 2, 4, 6 and 8 of the timeslot
1	10	Invert the bits 3, 5 and 7 of the timeslot
1	11	Invert all the bits of the timeslot except the MSB (bit1)

The priority of the RPLC operation on the RSDn pin from high to low is:

Extract data to PRGD for test; Replace the data with the value in the DTRK[7:0]; Replace the data with the milliwatt pattern; Replace the data with the pattern generated in the PRGD; Invert the bit.

# E1 RPLC Data Trunk Conditioning Code Byte Registers (RPLC Indirect Registers 40H – 5FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DTRK7	DTRK6	DTRK5	DTRK4	DTRK3	DTRK2	DTRK1	DTRK0
Туре	R/W							
Default	X	Χ	Χ	Х	X	X	Χ	X

They contain the data that will replace the data output on the RSDn/MRSD pin when the corresponding bit DTRKC/NxTS (b6, E1-RPLC-indirect registers-20~3FH) is logic 1. The DTRK7 is the MSB.

# **E1 RPLC Signaling Trunk Conditioning Byte Registers** (RPLC Indirect Registers 61H – 7FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					Α	В	С	D
Туре	Reserved				R/W	R/W	R/W	R/W
Default					Χ	Χ	Х	Χ

These bits contain the data that will replace the data output on the RSSIGn/MRSSIG pin when the corresponding STRKC (b5, E1-RPLC-indirect registers-20~3FH) is logic 1. They are in the least significant nibble.

# E1 TPLC Configuration (060H, 0E0H, 160H, 1E0H, 260H, 2E0H, 360H, 3E0H)

Bit No.	7	6	5	4	3	2	1	0		
Bit Name										
Туре	Reserved									
Default								0		

# PCCE:

- = 0: the per-TS functions in TPLC are disabled.
- = 1: the per-TS functions in TPLC are enabled.

# E1 TPLC INP Access Status (061H, 0E1H, 161H, 1E1H, 261H, 2E1H, 361H, 3E1H)

Bit No.	7	6	5	4	3	2	1	0				
Bit Name	BUSY											
Туре	R		Reserved									
Default	0											

## BUSY:

- = 0: no reading or writing operation on the indirect registers.
- = 1: an internal indirect register is being accessed, any new operation on the internal indirect register is not allowed.

This bit goes low timed to an internal high-speed clock rising edge after the operation has been completed. The operation cycle is 490ns. No more operations to the indirect registers could be done until this bit is cleared.

# E1 TPLC Channel Indirect Address / Control (062H, 0E2H, 162H, 1E2H, 262H, 2E2H, 362H, 3E2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	R/WB	A6	A5	A4	A3	A2	A1	A0
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

Writing to this register with a valid address and R/WB bit initiates an internal operation cycle to the indirect registers.

# R/WB:

- = 0: write the data to the specified indirect register.
- = 1: read the data from the specified indirect register.

# A[6:0]:

Specify the address of the indirect registers (from 20H to 7FH) for the microprocessor access.

# E1 TPLC Channel Indirect Data Buffer (063H, 0E3H, 163H, 1E3H, 263H, 2E3H, 363H, 3E3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Туре	R/W							
Default	0	0	0	0	0	0	0	0

This register hold the value which will be read from or write into the indirect registers (from 20H to 7FH). If data are to be written to the indirect registers, the byte to be written must be written into this register before the target indirect register's address and R/WB=0 is written into the Address/Control register, initiating the access. If data are to be read from the indirect registers, only the target indirect register's address and R/WB=1 is written into the Address/Control register, initiating the request. After 490 ns, this register will contain the requested data byte.

TP	TPLC Indirect Registers Map							
20H~3FH	Per-TS Control Byte for TS0 ~ TS31							
40H~5FH	IDLE Code Byte for TS0 ~ TS31							
61H~7FH	Signaling /PCM Control Byte for TS1 ~ TS31							

## E1 TPLC Per-TS Control Registers (TPLC Indirect Registers 20H – 3FH)

Bit No.	7	6	5	4	3	2	1	0	
Bit Name	SUBS		DS1	DS0	TEST	LOOP			
Туре	R/W	Reserved	R/W	R/W	R/W	R/W	Reserved		
Default	Х		Χ	Х	Χ	Х			

## SUBS, DS1, DS0:

The SUBS, DS[1:0] bits select one of the operation to the corresponding timeslot:

SUBS	DS[1]	DS[0]	OPERATION
0	0	0	No change to the timeslot
0	1	0	Invert the odd bits (1, 3, 5, 7) of the timeslot (bit 1 is the LSB)
0	0	1	Invert the even bits (2, 4, 6, 8) of the timeslot (bit 8 is the MSB)
0	1	1	Invert all the bits of the timeslot
1	-	0	Replace the timeslot with the IDLE code
1	0	1	Replace the timeslot with the A-law digital milliwatt pattern (per G.711)
1	1	1	Replace the timeslot with the $\mu$ -law digital milliwatt pattern (per G.711)

## TEST:

- = 0: disable the data in the corresponding timeslot to be tested by PRGD.
- = 1: enable the data in the corresponding timeslot to be extracted to PRGD for test (when the RXPATGEN [b2, E1-00CH] is logic 1), or enable the test pattern from PRGD to replace the data in the corresponding timeslot for test (when the RXPATGEN [b2, E1-00CH] is logic 0).

All the timeslots that are extracted to the PRGD are concatenated and treated as a continuous stream in which pseudo random are searched for. Similarly, all timeslots set to be replaced with PRGD test pattern data are concatenated replaced by the PRBS.

#### LOOP:

- = 0: disable the payload loopback.
- = 1: enable the payload loopback. When the Receive Clock Master modes are enabled, the Elastic Store is used to align the receive line data to the data to be transmitted. When Receive Clock Slave modes are enabled, the Elastic Store is unavailable to facilitate the payload loopbacks, and loopback functionality is provided only when the transmit path is also in Transmit Clock Slave mode, and the received clock and the clock to be transmitted and Common Frame Pulse are identical (RSCCK = TSCCKB, RSCFS = TSCFS).

The priority of the TPLC operation on the TSDn pin from high to low is:

Extract data to PRGD for test; Payload loopback; Replace the data with the milliwatt pattern; Replace the data with the pattern generated in the PRGD; Replace the data with the value in the IDLE[7:0]; Invert the even bits or/and odd bits.

# E1 TPLC IDLE Code Byte Registers (TPLC Indirect Registers 40H – 5FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	IDLE7	IDLE6	IDLE5	IDLE4	IDLE3	IDLE2	IDLE1	IDLE0
Туре	R/W							
Default	Χ	Χ	Х	Χ	Χ	Х	Х	Χ

They contain the data that will replace the data input from the TSDn pin when the corresponding SUBS and DS[1:0] are allowed. IDLE7 is the MSB.

# E1 TPLC Signaling / PCM Control Byte Registers (TPLC Indirect Registers 61H – 7FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				SIGSRC	Α	В	С	D
Туре		Reserved			R/W	R/W	R/W	R/W
Default				Х	Χ	Χ	Χ	Х

# SIGSRC:

This bit is valid only if the Channel Associated Signaling (CAS) is selected in the E1 FRMG Configuration Register.

- = 0: use the data on the TSSIGn pin as the signaling.
- = 1: use the data in the A, B, C, D bits as the signaling.

# A, B, C, D:

They contain the data that can be used as signaling when the corresponding SIGSRC is logic 1. They are in the least significant nibble.

# E1 RCRB Configuration (COSS = 0) (064H, 0E4H, 164H, 1E4H, 264H, 2E4H, 364H, 3E4H)

Bit No.	7	6	5	4	3	2	1	0	
Bit Name		COSS	SIGE					PCCE	
Туре	Reserved	R/W	R/W		Reserved				
Default		0	0		0				

# COSS:

- = 0: allow the RCRB registers to access the indirect registers.
- = 1: allow the RCRB registers to reflect the change of the signaling of its corresponding timeslot.

## SIGE:

- = 0: disable generation of an interrupt on the  $\overline{\text{INT}}$  pin when there is a signaling change in any one of the 30 timeslots.
- = 1: enable generation of an interrupt on the INT pin when there is a signaling change in any one of the 30 timeslots.

## PCCE:

- = 0: the per-TS functions in RCRB are disabled.
- = 1: the per-TS functions in RCRB are enabled.

# E1 RCRB Timeslot Indirect Status (COSS = 0) (065H, 0E5H, 165H, 1E5H, 265H, 2E5H, 365H, 3E5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	BUSY							
Туре	R				Reserved			
Default	0							

#### BUSY:

- = 0: no reading or writing operation on the indirect registers.
- = 1: an internal indirect register is being accessed, any new operation on the internal indirect register is not allowed.

This bit goes low timed to an internal high-speed clock rising edge after the operation has been completed. The operation cycle is 490ns. No operations to the indirect registers can be done until this bit is cleared.

# E1 RCRB Timeslot Indirect Address / Control (COSS = 0) (066H, 0E6H, 166H, 1E6H, 266H, 2E6H, 366H, 3E6H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	R/WB	A6	A5	A4	A3	A2	A1	A0
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

## R/WB:

- = 0: write the data to the specified indirect register.
- = 1: read the data from the specified indirect register.

# A[6:0]:

Specified the address of the indirect registers (from 20H to 5FH) for the microprocessor access.

# E1 RCRB Timeslot Indirect Data Buffer (COSS = 0) (067H, 0E7H, 167H, 1E7H, 267H, 2E7H, 367H, 3E7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Type	R/W							
Default	X	X	X	X	X	X	X	X

This register holds the value which will be read from or written to the indirect registers (from 20H to 7FH). If data are to be written to the indirect registers, the byte to be written must be written into this register before the target indirect register's address and R/WB=0 is written into the Address/Control register, initiating the access. If data are to be read from the indirect registers, only the target indirect register's address and R/WB=1 is written into the Address/Control register, initiating the request. After 490 ns, this register will contain the requested data byte.

# E1 RCRB Change of Signaling State (COSS = 1) (064H, 0E4H, 164H, 1E4H, 264H, 2E4H, 364H, 3E4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		COSS	COSS[30]	COSS[29]	COSS[28]	COSS[27]	COSS[26]	COSS[25]
Туре	Reserved	R/W	R	R	R	R	R	R
Default		0	Х	Χ	Χ	Χ	Χ	Х

## COSS:

- = 0: allow the RCRB registers to access the indirect registers.
- = 1: allow the RCRB registers to reflect the change of the signaling of its corresponding timeslot.

#### COSSn:

- = 0: the signaling in its corresponding timeslot is not changed.
- = 1: the signaling in its corresponding timeslot is changed.

These bits are cleared to 0 after the register is read. COSS[30:25] correspond to timeslots 31 to 26.

## E1 RCRB Change of Signaling State (COSS = 1) (065H, 0E5H, 165H, 1E5H, 265H, 2E5H, 365H, 3E5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	COSS[24]	COSS[23]	COSS[22]	COSS[21]	COSS[20]	COSS[19]	COSS[18]	COSS[17]
Туре	R	R	R	R	R	R	R	R
Default	Х	X	Х	Х	Х	Х	Х	X

## COSSn:

- = 0: the signaling in its corresponding timeslot is not changed.
- = 1: the signaling in its corresponding timeslot is changed.

These bits are cleared to 0 after the register is read. COSS[24:17] correspond to timeslots 25 to 18.

# E1 RCRB Change of Signaling State (COSS = 1) (066H, 0E6H, 166H, 1E6H, 266H, 2E6H, 366H, 3E6H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	COSS[16]	COSS[15]	COSS[14]	COSS[13]	COSS[12]	COSS[11]	COSS[10]	COSS[9]
Туре	R	R	R	R	R	R	R	R
Default	Х	Х	Х	Х	Х	Х	Х	Х

## COSSn:

- = 0: the signaling in its corresponding timeslot is not changed.
- = 1: the signaling in its corresponding timeslot is changed.

These bits are cleared to 0 after the register is read. COSS[16] corresponds to timeslots 17. COSS[15:9] correspond to timeslot 15 to 9.

## E1 RCRB Change of Signaling State (COSS = 1) (067H, 0E7H, 167H, 1E7H, 267H, 2E7H, 367H, 3E7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	COSS[8]	COSS[7]	COSS[6]	COSS[5]	COSS[4]	COSS[3]	COSS[2]	COSS[1]
Туре	R	R	R	R	R	R	R	R
Default	X	X	X	Х	X	X	X	X

# COSSn:

- = 0: the signaling in its corresponding timeslot is not changed.
- = 1: the signaling in its corresponding timeslot is changed.

These bits are cleared to 0 after the register is read. COSS[8:1] correspond to timeslots 8 to 1.

RCRB Indire	ect Registers Map
20H	-
01H ~ 0FH / 21H ~ 2FH	Signaling Data Register for TS1 ~ 15
10H, 30H	-
11H ~ 1FH / 31H ~ 3FH	Signaling Data Register for TS17 ~31
40H ~ 5FH	TS0 ~ 31 Configuration Data

# E1 RCRB Timeslot / Channel Signaling Data Registers (COSS = 0) (RCRB Indirect Registers 01H - 1FH / 21H – 3FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					А	В	С	D
Туре		Reserved				R	R	R
Default					X	Х	Х	Х

# A, B, C, D:

They contain the signaling of the corresponding timeslot. The value for TS0 and TS16 are not valid.

There is a maximum 2 ms delay between the transition of the COSS[n] bit (E1-064H & E1-065H & E1-066H & E1-067H) and the updating of the A, B, C, D code in the corresponding indirect registers  $21H \sim 3FH$ . To avoid this 2ms delay, users can read the corresponding  $b3\sim0$  in the indirect registers  $01H \sim 1FH$  first. If the value of these four bits are different from the previous A, B, C, D code, then the content of  $b3\sim0$  in the  $01H \sim 1FH$  is the updated A, B, C, D code. If the conternt of the four bits is the same as the previous A, B, C, D code, then users should read the  $b3\sim0$  in the  $21H \sim 3FH$  to get the updated A, B, C, D code.

# E1 RCRB Per-Timeslot Configuration Registers (COSS = 0) (RCRB Indirect Registers 40H – 5FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name								DEB
Туре				Reserved				R/W
Default								Х

#### DEB:

- = 0: disable signaling debounce.
- = 1: enable signaling debounce (valid only if the PCCE is logic 1). That is, the signaling is acknowledged only when 2 consecutive signaling of a timeslot are the same.

**E1 PMON Interrupt Enable / Status** (068H, 0E8H, 168H, 1E8H, 268H, 2E8H, 368H, 3E8H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name						INTE	XFER	OVR
Туре			Reserved			R/W	R	R
Default						0	0	0

#### INTE:

- = 0: disable the interrupt on the INT pin when the counter data has been transferred into the Error Count registers.
- = 1: enable the interrupt on the INT pin when the counter data has been transferred into the Error Count registers.

# XFER:

- = 0: indicate that the counter data has not been transferred to the Error Count registers.
- = 1: indicate that the counter data has been transferred to the Error Count registers.

This bit is clear to 0 after the bit is read.

#### OVR:

- = 0: indicate that no overwritten on the Error Count registers has occurred.
- = 1: indicate that one of the Error Count registers is overwritten.

This bit is clear to 0 after the bit is read.

#### Registers 069H-06DH, 0E9H-0EDH, 16H-16DH, 1E9H-1EDH, 269H-26DH, 2E9H-2EDH, 369H-36DH, 3E9H-3EDH:

The PMON Error Count registers for a single framer are updated as a group by writing to any of the PMON count registers or updated every 1 second when the AUTOUPDATE (b0, E1-000H) is set. The PMON Error Count registers for eight framers are updated by writing to the Chip ID/ Global PMON Update register (E1-009H).

When the chip is reset, the contents of the PMON Error Count registers are unknown until the first latching of performance data is performed.

# **E1 PMON Framing Bit Error Count** (069H, 0E9H, 169H, 1E9H, 269H, 2E9H, 369H, 3E9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		FER[6]	FER[5]	FER[4]	FER[3]	FER[2]	FER[1]	FER[0]
Туре	Reserved	R	R	R	R	R	R	R
Default		X	Х	X	Х	Х	Х	Х

These bits are valid when it is in the Basic Frame Sync. They represent the number of the basic frame alignment signal errors and update on the defined intervals. The basic frame alignment signal errors are defined in the WORDERR (b5, E1-000H) and the CNTNFAS (b4, E1-000H).

E1 PMON Far End Block Error Count LSB (06AH, 0EAH, 16AH, 1EAH, 26AH, 2EAH, 36AH, 3EAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FEBE[7]	FEBE[6]	FEBE[5]	FEBE[4]	FEBE[3]	FEBE[2]	FEBE[1]	FEBE[0]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Χ	Χ	Χ	Χ	Χ	Х	Х

# E1 PMON Far End Block Error Count MSB (06BH, 0EBH, 16BH, 1EBH, 26BH, 2EBH, 36BH, 3EBH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			FEBE[9]	FEBE[8]				
Туре			R	R				
Default			Χ	Х				

The FEBE[9:0] are valid when it is in the CRC Multi-Frame Sync. They represent the number of the Far End Block Errors (FEBE) and update on the defined intervals.

# **E1 PMON CRC Error Count LSB** (06CH, 0ECH, 16CH, 1ECH, 26CH, 2ECH, 36CH, 3ECH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	CRCE[7]	CRCE[6]	CRCE[5]	CRCE[4]	CRCE[3]	CRCE[2]	CRCE[1]	CRCE[0]
Туре	R	R	R	R	R	R	R	R
Default	X	X	X	X	X	X	X	Х

# **E1 PMON CRC Error Count MSB** (06DH, 0EDH, 16DH, 1EDH, 26DH, 2EDH, 36DH, 3EDH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			CRCE[9]	CRCE[8]				
Туре			Rese	erved			R	R
Default							X	X

The CRCE[9:0] are valid when it is in the CRC Multi-Frame Sync. They represent the number of the CRC errors and update on the defined intervals.

## E1 PRGD Control (070H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PDR[1]	PDR[0]		PS	TINV	RINV	AUTOSYNC	MANSYNC
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	1	0

## PDR[1:0]:

The PDR[1:0] define the function of the four PRGD Pattern Detector registers:

PDR[1:0]	PRGD Pattern Detector Registers (#1 ~ #4)
00, 01	Pattern Receive
10	Error Count
11	Bit Count
(The #1 is the LSB,	while the #4 is the MSB.)

# PS:

- = 0: a pseudo-random pattern is generated/detected by the PRGD.
- = 1: a repetitive pattern is generated/detected by the PRGD.

This bit should be set first of all the PRGD registers.

## TINV:

- = 0: disable the inversion of the generated pattern before being transmitted.
- = 1: enable the inversion of the generated pattern before being transmitted.

## RINV:

- = 0: disable the inversion of the received pattern before being processed.
- = 1: enable the inversion of the received pattern before being processed.

## **AUTOSYNC:**

- = 0: disable automatically re-searching for the sync of the pattern when the PRGD pattern is out of synchronization.
- = 1: enable automatically re-searching for the sync of the pattern when the PRGD pattern is out of synchronization.

## MANSYNC:

Trigger on the rising edge. A transition from logic 0 to logic 1 on this bit manually initiates a re-search for the sync of a pattern.

Every time when the setting of the PRGD registers is changed or the detector data source changes, a manual sync operation is recommended to ensure that the detector works correctly.

## E1 PRGD Interrupt Enable / Status (071H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	SYNCE	BEE	XFERE	SYNCV	SYNCI	BEI	XFERI	OVR
Туре	R/W	R/W	R/W	R	R	R	R	R
Default	0	0	0	Χ	Χ	Х	Х	Х

## SYNCE:

- = 0: disable the interrupt on the INT pin when the SYNCI is logic one.
- = 1: enable the interrupt on the INT pin when the SYNCI is logic one.

#### BEE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when bit error has been detected in the received pattern.
- = 1: enable the interrupt on the  $\overline{\text{INT}}$  pin when each bit error is detected in the received pattern.

#### XFERE:

- = 0: disable the interrupt on the INT pin when the the data in the PRGD pattern detector register is updated.
- = 1: enable the interrupt on the INT pin when the the data in the PRGD pattern detector register is updated.

## SYNCV:

- = 0: the pattern is out of sync (the pattern detector has detected 10 or more bit errors in a hopping 48-bit window).
- = 1: the pattern is in sync (the pattern detector has observed at least 48 consecutive error-free bit-periods).

## SYNCI:

- = 0: there is no transition on the SYNCV.
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the SYNCV.

This bit is cleared to 0 after the bit is read.

# BEI:

- = 0: no bit error is detected in the received pattern.
- = 1: at least one bit error has been detected in the received pattern.

This bit is cleared to 0 after the bit is read.

# XFERI:

- = 0: the data in the PRGD pattern detector register is not updated.
- = 1: the data in the PRGD pattern detector register is updated.

This bit is cleared to 0 after the bit is read.

# OVR:

- = 0: the PRGD pattern detector register is not overwritten.
- = 1: the PRGD pattern detector register is overwritten.

This bit is cleared to 0 after the bit is read.

# E1 PRGD Shift Register Length (072H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				PL[4]	PL[3]	PL[2]	PL[1]	PL[0]
Туре		Reserved			R/W	R/W	R/W	R/W
Default				0	0	0	0	0

These bits determine the length of the valid data in the PRGD pattern insertion register. The length is equal to the value of PL[4:0] + 1.

# E1 PRGD Tap Bit Type Function Default (073H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				PT[4]	PT[3]	PT[2]	PT[1]	PT[0]
Туре		Reserved			R/W	R/W	R/W	R/W
Default					0	0	0	0

These bits determine the feedback tap position of the generated pseudo random pattern before it is transmitted. The feedback tap position is equal to the value of PT[4:0] + 1. In application, the PT is always less than the PL.

# E1 PRGD Error Insertion (074H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					EVENT	EIR[2]	EIR[1]	EIR[0]
Туре		Rese	rved		R/W	R/W	R/W	R/W
Default					0	0	0	0

#### EVENT:

A single bit error is generated when the state of this bit is changed from 0 to 1. To insert another bit error, this bit must be cleared to 0, and then set from 0 to 1 again.

# EIR[2:0]:

The EIR[2:0] bits determine the bit error rate that will be inserted in the PRGD test pattern. If the bit error rate is changed from one non-zero value to another non-zero value, it is recommended to set the EIR[2:0] to '000' first, then set the EIR[2:0] to the desired value.

EIR[2:0]	Bit error rate
000	No error inserted
0 0 1	No error inserted
010	10 <sup>-2</sup>
011	10 <sup>-3</sup>
100	10 <sup>-4</sup>
101	10 <sup>-5</sup>
110	10 <sup>-6</sup>
111	10 <sup>-7</sup>

# E1 PRGD Pattern Insertion #1 (078H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[7]	PI[6]	PI[5]	PI[4]	PI[3]	PI[2]	PI[1]	PI[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

# E1 PRGD Pattern Insertion #2 (079H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[15]	PI[14]	PI[13]	PI[12]	PI[11]	PI[10]	PI[9]	PI[8]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

# E1 PRGD Pattern Insertion #3 (07AH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[23]	PI[22]	PI[21]	PI[20]	PI[19]	PI[18]	PI[17]	PI[16]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

# E1 PRGD Pattern Insertion #4 (07BH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[31]	PI[30]	PI[29]	PI[28]	PI[27]	PI[26]	PI[25]	PI[24]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When a repetitive pattern is selected to transmit, the data in these registers are the repetitive pattern.

When a pseudo random pattern is selected to transmit, the data in these registers should be set to FFFFFFFH. They are the initial value for the pseudo random pattern.

Writing to PI[31:24] updates the PRGD configuration.

When a repetitive pattern is transmitted, the PI[31] bit is transmitted first, followed by the remaining bits in sequence down to PI[0]. The length of the valid data in these four registers is determined by the PL[4:0]. When the length is less than 31, the bits in higher PI are not used.

## E1 PRGD Pattern Detector #1 (07CH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[7]	PD[6]	PD[5]	PD[4]	PD[3]	PD[2]	PD[1]	PD[0]
Туре	R	R	R	R	R	R	R	R
Default	X	Х	X	X	X	Χ	Χ	Χ

# E1 PRGD Pattern Detector #2 (07DH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[15]	PD[14]	PD[13]	PD[12]	PD[11]	PD[10]	PD[9]	PD[8]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Х	Х	Х	Χ	Χ	Х	Х

## E1 PRGD Pattern Detector #3 (07EH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[23]	PD[22]	PD[21]	PD[20]	PD[19]	PD[18]	PD[17]	PD[16]
Туре	R	R	R	R	R	R	R	R
Default	X	X	Х	Х	X	X	Х	X

# E1 PRGD Pattern Detector #4 (07FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[31]	PD[30]	PD[29]	PD[28]	PD[27]	PD[26]	PD[25]	PD[24]
Туре	R	R	R	R	R	R	R	R
Default	X	X	X	X	X	X	X	Х

When the PDR[1:0] (b7~6, E1-070H) are set to 00 or 01, the four PRGD pattern detector registers are configured as Pattern Receive registers. They reflect the content of the received pattern.

When the PDR[1:0] (b7~6, E1-070H) are set to 10, the four PRGD pattern detector registers are configured as Error Counter registers. The value in these registers represents the number of the bit errors. The bit errors are not accumulated when the pattern is out of sync.

When the PDR[1:0] (b7~6, E1-070H) are set to 11, the four PRGD pattern detector registers are configured as Bit Counter registers. The value in these registers represent the total received bit number.

These registers can be configured to update every second automatically when the AUTOUPDATE (b0, E1-000H) is set to 1, or by writing to any of these four registers, or to the Revision / Chip ID / Global PMON register.

#### 5.2.2 T1 / J1 MODE

# T1 / J1 Receive Line Options (000H, 080H, 100H, 180H, 200H, 280H, 300H, 380H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FIFOBYP	UNF	IBCD_IDLE		AUTOYELLOW	AUTORED	AUTOOOF	AUTOUPDATE
Туре	R/W	R/W	R/W	Reserved	R/W	R/W	R/W	R/W
Default	0	0	0		0	0	0	0

#### FIFOBYP:

This bit decides whether the received data should pass through or bypass the Receive Jitter Attenuation FIFO.

- = 0: The received data pass through the RJAT FIFO.
- = 1: The RJAT FIFO is bypassed. The delay is reduced by typically 24 bits.

#### UNF:

- = 0: The Frame Processor operates normally.
- = 1: Frame searching is disabled, the RCRB holds its signaling frozen, and Auto\_OOF function, if enabled, will consider OOF to be declared.

#### IBCD IDLE:

This bit is valid in framed mode.

- = 0: compare the inband loopback activate/deactivate code with all received data stream, including the F-bit. However, the result of F-bit comparison is discarded.
  - = 1: compare the inband loopback activate/deactivate code with the received data stream, excluding the F-bit.

## **AUTOYELLOW:**

This bit decides whether to send the Yellow Alarm signal automatically.

- = 0: The automatic Yellow Alarm Transmission is disabled. It means that the Yellow Alarm can only be transmitted when the XYEL (b1, T1/J1-045H) is set to 1.
- = 1: The automatic Yellow Alarm Transmission is enabled. It means that the Yellow Alarm will be transmitted automatically when Red alarm is declared in the received data stream.

## AUTORED:

This bit decides whether to start trunk conditioning (replacing data on RSDn/MRSD with the data stored in the data trunk conditioning registers in RPLC) automatically when Red Alarm is declared.

- = 0: The trunk conditioning is not activated automatically when Red Alarm is declared in the Alarm Detector block.
- = 1: The trunk conditioning will be initiated automatically when Red Alarm is declared in the Alarm Detector block.

#### AUTOOOF:

This bit decides whether to start trunk conditioning (replacing data on RSDn/MRSD with the data stored in the data trunk conditioning registers in RPLC) automatically in the duration of loss of SF/ESF frame.

- = 0: The trunk conditioning is not activated automatically when the INFR (b0, T1/J1-022H) is declared in the Frame Processor block.
- = 1: The trunk conditioning will be activated automatically when the INFR (b0, T1/J1-022H) is declared in the Frame Processor block.

## AUTOUPDATE:

This bit decides whether the PMON and PRGD registers are automatically updated once every second.

- = 0: The PMON and PRGD registers are not automatically updated. They can only be updated by MCU operation.
- = 1: The PMON and PRGD registers will be automatically updated once every second.

# T1 / J1 Receive Side System Interface Options (001H, 081H, 101H, 181H, 201H, 281H, 301H, 381H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	IMODE[1]	IMODE[0]	RSCKSEL	RSCCK2M	RSCCK8M	RSFSP	ALTIFP	IMTKC
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	1	1	0	0	0	0	0	0

#### IMODE[1:0]:

These bits select the operation mode in the receive path.

IMODE[1:0]	Operation Mode In Receive Path
0 0	Receive Clock Master Fractional T1/J1 mode
01	Receive Clock Master Full T1/J1 mode
10	Receive Clock Slave RSCK Reference mode
11	Receive Clock Slave External Signaling mode

"Receive Clock Master Full T1/J1" mode means that the received entire frame (193 bits) is clocked out from RSDn pin, and there are no gaps in the RSCKn clock pulse.

"Receive Clock Master Fractional T1/J1" mode means that the RSCKn only clocks out on the selected channels, and RSCKn does not pulse during those un-selected channels. The selection of the channel is decided by the EXTRACT (b2, T1/J1-RPLC-Indirect Register-01-18 H).

When Receive Clock Slave RSCK Reference Mode is selected, the RSCKn/RSSIGn pin will be used as RSCKn to output a 1.544 MHz jitter attenuated version of LRCKn or an 8KHz clock.

When Receive Clock Slave External Signaling mode is selected, the RSCKn/RSSIGn pin is used as RSSIGn to output the extracted signaling data. Each channel's signaling data is channel aligned with the RSDn data stream and located in lower nibble (b5b6b7b8) of the timeslot.

#### RSCKSEL:

When Receive Clock Slave RSCK Reference Mode is selected, this bit selects the frequency of the RSCKn.

- = 0: the RSCKn outputs an 8 KHz timing reference that is generated by dividing the jitter attenuated version of LRCKn.
- = 1: the RSCKn outputs a jitter attenuated version of the 1.544MHz receive line clock (LRCK).

# RSCCK2M, RSCCK8M:

These bits determine the bit rate of the received data stream on the backplane. The 2.048 Mbit/s rate can only be supported when the backplane is configured in Receive Clock Slave mode. If the Receive Multiplexed mode is desired, all the RSCCK2M & RSCCK8M in eight framers must be set the same to select the 8.192 Mbit/s backplane bit rate. When the RSCCK2M, RSCCK8M selects the 8.192 Mbit/s, the IMODE[1:0] (b7~6, T1/J1-001H) must be set to 11.

RSCCK2M, RSCCK8M	Backplane Rate
0 0	1.544M bit/s
10	2.048M bit/s
01	8.192M bit/s
11	Reserved

#### RSFSP, ALTIFP:

RSFSP	ALTIFP	RSFSn Indication
0	0	the RSFSn pin asserts on each F-bit.
0	1	the RSFSn pin asserts on every second F-bit (i.e., the F-bit of even frame if there is no channel offset).
1	0	the signal on the RSFSn pin asserts on the first F-bit of every 12 frames (in SF format) or every 24 frames (in ESF format).
1	1	-

In Receive Multiplexed mode, regardless of the setting in the RSFSP and the ALTRSFS, the MRSFS can only indicate each F-bit of SF/ESF of the selected first framer.

#### IMTKC:

This bit decides how to substitute the received data stream on RSDn and RSSIGn with contents in RPLC Data Trunk Conditioning Registers and RPLC Signaling Trunk Conditioning Registers. This bit affects the corresponding timeslot of the MRSD and MRSSIG even if the multiplexed backplane is enabled.

- = 0: The data and signaling signals are substituted on a per-timeslot basis in accordance with the control bits contained in the per-timeslot Payload Control Byte registers in the RPLC.
- = 1: the data on RSDn of all timeslots are replaced with the data contained in the Data Trunk Conditioning registers in RPLC, and the data on RSSIGn of all timeslots are replaced with the data contained in the Signaling Trunk Conditioning registers. To enable this function, the PCCE (b0, T1/J1-050H) of the RPLC must be set to logic 1.

## T1/J1 Backplane Parity Configuration / Status (002H, 082H, 102H, 182H, 202H, 282H, 302H, 382H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TPTYP	TPRTYE	TSDI	TSSIGI	PTY_EXTD		RPTYP	RPRTYE
Туре	R/W	R/W	R	R	R/W	Reserved	R/W	R/W
Default	0	0	X	X	0		0	0

# TPTYP:

- = 0: even parity check is employed in the F-bit input from the TSDn/MTSD and TSSIGn/MTSSIG pin, which means a logic one is expected in the F-bit position when the number of ones in the previous SF/ESF is odd.
- = 1: odd parity check is employed in the F-bit input from the TSDn/MTSD and TSSIGn/MTSSIG pin, which means a logic one is expected in the F-bit position when the number of ones in the previous SF/ESF is even.

#### TPRTYE:

This bit is invalid in Receive Clock Master Fractional T1/J1 mode.

- = 0: disable the interrupt on the INT pin when a parity error is detected on the TSDn/MTSD pin or a parity error is detected on the TSSIGn/MTSSIG pin.
- = 1: enable the interrupt on the INT pin when a parity error is detected on the TSDn/MTSD pin or a parity error is detected on the TSSIGn/MTSSIG pin.

#### TSDI:

- = 0: indicate that no parity error is detected on the TSDn/MTSD pin.
- = 1: indicate that a parity error is detected on the TSDn/MTSD pin.

This bit is cleared to 0 after the bit is read.

# TSSIGI:

- = 0: indicate that no parity error is detected on the TSSIGn/MTSSIG pin.
- = 1: indicate that a parity error is detected on the TSSIGn/MTSSIG pin.

This bit is cleared to 0 after the bit is read.

## PTY EXTD:

- = 0: the parity is calculated over the previous SF/ESF, excluding the F-bit on the TSDn/MTSD, TSSIGn/MTSSIG, RSDn/MRSD and RSSIGn/MRSSIG pin.
- = 1: the parity is calculated over the previous SF/ESF, including the F-bit on the TSDn/MTSD, TSSIGn/MTSSIG, RSDn/MRSD and RSSIGn/MRSSIG pin.

#### RPTYP:

This bit is invalid in Receive Clock Master Fractional T1/J1 mode and valid when the RPRTYE = 1.

- = 0: even parity check is employed in the F-bit output on the RSDn/MRSD and RSSIGn/MRSSIG pin, which means a logic one should be replaced in the F-bit when the number of ones in the previous SF/ESF is odd.
- = 1: odd parity check is employed in the F-bit output on the RSDn/MRSD and RSSIGn/MRSSIG pin, which means a logic one should be replaced in the F-bit when the number of ones in the previous SF/ESF is even.

#### RPRTYE:

This bit is invalid in Receive Clock Master Fractional T1/J1 mode.

- = 0: disable the parity on the RSDn/MRSD and RSSIGn/MRSSIG pin.
- = 1: enable the parity on the RSDn/MRSD and RSSIGn/MRSSIG pin.

## T1/J1 Receive Interface Configuration (003H, 083H, 103H, 183H, 203H, 283H, 303H, 383H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	MRBS	MRBC	TRI[1]	TRI[0]	RSCKRISE	LRCKFALL	RSCFSFALL	RSCCKRISE
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

#### MRBS:

In Receive Multiplexed mode, this bit decides which bus the corresponding framer will use to output the received data.

- = 0: The first multiplexed bus (MRSD[1], MRSFS[1], MRSSIG[1]) is selected.
- = 1: The second multiplexed bus (MRSD[2], MRSFS[2], MRSSIG[2]) is selected.

#### MRBC:

This bit turns on or turn off the transmission of received data from the corresponding framer to the selected multiplexed receive bus. Users should complete the setting in the MRBS (b7, T1/J1-003H) before turn on this bit.

- = 0: The corresponding framer will not output its data stream on the multiplexed bus.
- = 1: The corresponding framer will output its data stream on the multiplexed bus.

## TRI[1:0]:

These bits decide the output status of the RSDn/MRSD and RSSIGn/MRSSIG pins.

TRI[1:0]	Output Status on the RSDn/MRSD and RSSIGn/MRSSIG pin			
0 0	in high impedance			
10	Reserved			
0 1	Normal operation			
11	Reserved			

#### RSCKRISE:

This bit selects the active edge of RSCKn to update the data on the corresponding RSDn and RSFSn. This bit is valid in Receive Clock Master mode.

- = 0: the falling edge is selected.
- = 1: the rising edge is selected.

#### LRCKFALL:

This bit selects the active edge of LRCKn to sample the data on the corresponding LRDn.

- = 0: the rising edge is selected.
- = 1: the falling edge is selected.

## RSCFSFALL:

This bit selects the active edge of RSCCK/MRSCCK to sample the data on the corresponding RSCFS/MRSCFS. This bit is valid in Receive Clock Slave mode and Receive Multiplexed mode.

- = 0: the rising edge is selected.
- = 1: the falling edge is selected

This bit in all eight framers must be set to the same value.

# RSCCKRISE:

This bit selects the active edge of RSCCK/MRSCCK to update the data on the corresponding RSDn/MRSD, RSSIGn/MRSSIG and RSFSn/MRSFS. This bit is valid in Receive Clock Slave mode and Receive Multiplexed mode.

- = 0: the falling edge is selected.
- = 1: the rising edge is selected.

In Receive Multiplexed mode, the RSCCKRISE of the eight framers must be set to the same value.

## T1/J1 Transmit Interface Configuration (004H, 084H, 104H, 184H, 204H, 284H, 304H, 384H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FIFOBYP	TAISEN			TSCCKBFALL	TSFSRISE	TSDFALL	LTCKRISE
Туре	R/W	R/W	Rese	Reserved		R/W	R/W	R/W
Default	0	0			0	0	0	0

#### FIFOBYP:

This bit decides whether the transmit data should pass through or bypass the Transmit Jitter Attenuation FIFO. The bit is valid in Clock Slave mode.

- = 0: the data to be transmitted pass through the TJAT FIFO.
- = 1: the TJAT FIFO is bypassed. The delay is reduced by typically 24 bits.

#### TAISEN:

This bit enables the line interface to generate an un-framed all-ones Alarm Indication Signal on the TLDn pin or the corresponding framer on the MTLD.

- = 0: normal operation.
- = 1: TLDn or the corresponding framer on the MTLD transmits all ones.

# TSCCKBFALL:

This bit selects the active edge of TSCCKB/MTSCCKB to sample the data on the corresponding TSDn/MTSD, TSSIGn/MTSSIG and TSCFS/MTSCFS. This bit is valid in Transmit Clock Slave mode and Transmit Multiplexed mode.

- = 0: the rising edge is selected.
- = 1: the falling edge is selected.

The TSCCKBFALL of the eight framers should be set to the same value.

#### TSFSRISE:

This bit is valid in Transmit Clock Slave TSFS Enabled mode and Transmit Clock Master mode.

- = 0: In Transmit Clock Slave TSFS Enabled mode, the signal on the TSFSn pin is updated on the falling edge of the TSCCKB. In Transmit Clock Master mode, the signal on the TSFSn pin is updated on the falling edge of the LTCKn.
- = 1: In Transmit Clock Slave TSFS Enabled mode, the signal on the TSFSn pin is updated on the rising edge of the TSCCKB. In Transmit Clock Master mode, the signal on the TSFSn pin is updated on the rising edge of the LTCKn.

## TSDFALL:

This bit selects the active edge of LTCKn to sample the data on the corresponding TSDn in Transmit Clock Master mode.

- = 0: the TSDn is sampled on the rising edge of the LTCKn.
- = 1: the TSDn is sampled on the falling edge of the LTCKn.

## LTCKRISE:

This bit selects the active edge of LTCKn to update the data on the corresponding LTDn.

- = 0: the falling edge is selected.
- = 1: the rising edge is selected.

## T1/J1 Transmit Side System Interface Options (005H, 085H, 105H, 185H, 205H, 285H, 305H, 385H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	EMODE[1]	EMODE[0]	FPINV	ABXXEN	RATE[1]	RATE[0]	TSCFSP	
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reserved
Default	1	1	0	0	0	0	0	

# EMODE[1:0]:

EMODE[1:0]	Operation Mode In Transmitter Path
0 0	Reserved
01	Transmit Clock Master mode
1 0 Transmit Clock Slave TSFS Enable mo	
11	Transmit Clock Slave External Signaling mode

In Transmit Multiplexed mode, these bits must be set to 11.

#### FPINV:

- = 0: the positive pulse on the TSCFS/MTSCFS pin is valid.
- = 1: the negative pulse on the TSCFS/MTSCFS pin is valid.

This bit of the eight framers should be set to the same value.

#### ABXXEN:

This bit is valid only in T1 ESF mode.

- = 0: the valid signaling on the TSSIGn/MTSSIG pin is in the lower four nibble of each channel (i.e. XXXXABCD).
- = 1: the valid signaling on the TSSIGn/MTSSIG pin is in the upper two-bit positions of the lower nibble of each channel (i.e. XXXXABXX). Thus, the 'A' bit will be inserted to the signaling bit of Frame 6 and 18, and the 'B' bit will be inserted to the signaling bit of Frame 12 and 24.

# RATE[1:0]:

These bits determine the bit rate of the transmit data stream on the backplane. Note that if any of the eight framers selects the 8.192 Mbit/s backplane bit rate, the multiplied bus will be enabled for the chip. When the RATE[1:0] selects the 2.048 Mbit/s or 8.192 Mbit/s, the EMODE[1] (b7, T1/J1-005H) must be set to 1 (i.e., in Transmit Clock Slave mode).

RATE[1:0]	Backplane Rate
0 0	1.544M bit/s
01	2.048M bit/s
10	Reserved
11	8.192M bit/s

## TSCFSP:

- = 0: indicate that the signal on the TSCFS pin asserts on each F-bit.
- = 1: indicate that the signal on the TSCFS pin asserts on the first F-bit of every 12 SFs or every 24 ESFs.

This bit of the eight framers should be set to the same value.

# T1 / J1 Transmit Framing and Bypass Options (006H, 086H, 106H, 186H, 206H, 286H, 306H, 386H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FRESH		SIGAEN		FDIS	FBITBYP	CRCBYP	FDLBYP
Туре	R/W	Reserved	R/W	Reserved	R/W	R/W	R/W	R/W
Default	0		0		0	0	0	0

#### FRESH:

- = 0: normal operation.
- = 1: initiate the FIFO in the Frame Generator block.

After initialization of the backplane interface, the user should write 1 into this bit and then clear it.

## SIGAEN:

- = 0: track the signaling input from the TSSIGn/MTSSIG pin for the signaling bit.
- = 1: take a snapshot of the 1st frame input from the TSSIGn/MTSSIG pin and lock it for the signaling bit of the whole SF/ESF.

## FDIS:

This bit is valid when the UF (b6, T1/J1-046H) is logic 0.

- = 0: the Frame Generator is enabled to generate and insert the framing bits into the transmit data.
- = 1: the Frame Generator is bypassed. Data on TSDn/MTSD pin is transmitted transparently.

#### FBITBYP:

This bit is valid when the UF (b6, T1/J1-046H) and the FDIS (b3, T1/J1-006H) are logic 0.

- = 0: the frame synchronization bits in the output data stream are generated by the Frame Generator.
- = 1: the frame synchronization bits in the input data stream on the TSDn/MTSD pin will be output transparently.

## CRCBYP:

This bit is valid when the UF (b6, T1/J1-046H) and the FDIS (b3, T1/J1-006H) are logic 0.

- = 0: the framing bit corresponding to the CRC-6 bit position in the output data stream are generated by the Frame Generator.
- = 1: the framing bit corresponding to the CRC-6 bit position in the input data stream on the TSDn/MTSD pin will be output transparently.

# FDLBYP:

This bit is valid when the UF (b6, T1/J1-046H) and the FDIS (b3, T1/J1-006H) are logic 0.

- = 0: the framing bit corresponding to the data link bit position in the output data stream are generated by the Frame Generator.
- = 1: the framing bit corresponding to the data link bit position in the input data stream on the TSDn/MTSD pin will be output transparently.

# **T1/J1 Transmit Timing Options** (007H, 087H, 107H, 187H, 207H, 287H, 307H, 387H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			TJATREF_SEL[2]	TJATREF_SEL[1]	TJATREF_SEL[0]	LTCK_SEL[2]	LTCK_SEL[1]	LTCK_SEL[0]
Туре	Rese	erved	R/W	R/W	R/W	R/W	R/W	R/W
Default			1	0	0	1	0	1

# TJATREF\_SEL[2:0]:

The TJATREF\_SEL[2:0] select the input reference clock for the TJAT DPLL.

TJATREF_SEL[2:0]	Input Reference Clock
000	TSCCKA / 8
001	TSCCKB
010	LRCK
011	TSCCKA
100	XCK / 24
the others	TSCCKB

# LTCK\_SEL[2:0]:

LTCK_SEL[2:0]	Line Transmit Clock				
000	TSCCKA/8				
001	TSCCKB				
010	LRCK				
011	TSCCKA				
100	XCK / 24				
the others	A smoothed clock output from the TJAT DPLL				

# T1/J1 Interrupt Source #1 (008H, 088H, 108H, 188H, 208H, 288H, 308H, 388H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PMON	IBCD	FRMP	PRGD	ELSB	RHDLC#1	RBOM	ALMD
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

Bits in this register indicate which function block caused an interrupt signal on  $\overline{\text{INT}}$  pin. Reading this register does not clear the interrupt indication. To clear the interrupt indication on  $\overline{\text{INT}}$  pin, the corresponding interrupt status register must be read.

# T1 / J1 Interrupt Source #2 (009H, 089H, 109H, 189H, 209H, 289H, 309H, 389H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RHDLC#2	PRTY	TJAT	RJAT	THDLC#1	THDLC#2		RCRB
Туре	R/W	R/W	R/W	R/W	R/W	R/W	Reserved	R/W
Default	0	0	0	0	0	0		0

Bits in this register indicate which function block caused an interrupt signal on  $\overline{\text{INT}}$  pin.

The PRTY bit indicates a pending parity error indication needs serving in the Backplane Parity Configuration and Status registers.

Reading this register does not clear the interrupt indication. To clear the interrupt indication on INT pin, the corresponding interrupt status register must be read.

## T1 / J1 Diagnostics (00AH, 08AH, 10AH, 18AH, 20AH, 28AH, 30AH, 38AH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				LINELB		DDLB	TXMFP	TXDIS
Туре		Reserved			Reserved	R/W	R/W	R/W
Default				0		0	0	0

## LINELB:

Line Loop back means that the transmit line interface data and clock (LTDn and LTCKn) are internal directly comes from the received line data and clock (LRDn and LRCKn). The loop back data stream can pass through the Receive Jitter Attenuator or bypass the Receive Jitter Attenuator (if the Receive Jitter Attenuator is configured to be bypassed)

- = 0: Line loop back is disabled.
- = 1: Line loop back is enabled.

#### DDLB:

Digital Loop back means that the received line data and clock (LRDn and LRCKn) are internal directly comes from the transmit line data and clock (LTDn and LTCKn) without the Receive Jitter Attenuator.

- = 0: Digital loop back is disabled.
- = 1: Digital loop back is enabled.

## TXMFP:

This bit controls whether the mimic pattern is generated. The mimic pattern is a copy of the F-bit. The mimic pattern is generated in the 1st bit of each channel.

- = 0: the mimic pattern is not generated.
- = 1: the mimic pattern is generated.

#### TXDIS:

- = 0: normal transmission.
- = 1: force the data to be transmitted on the TLDn pin to be all zeros.

## T1 / J1 Revision / Chip ID / Global PMON Update (00CH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TYPE[2]	TYPE[1]	TYPE[0]	ID[4]	ID[3]	ID[2]	ID[1]	ID[0]
Туре	R	R	R	R	R	R	R	R
Default	0	0	0	0	0	0	0	1

Writing to this register causes all Performance Monitor and PRGD Generator/Detector counters to be updated simultaneously.

#### TYPE[2:0]:

TYPE[2:0] are fixed to 000, representing the IDT82V2108 chip.

# ID[4:0]:

ID[4:0] are fixed to 00011, representing the current version number of the IDT82V2108.

## T1/J1 Data Link Micro Select / Framer Reset (00DH, 08DH, 10DH, 18DH, 20DH, 28DH, 30DH, 38DH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RHDLCSEL[1]	RHDLCSEL[0]	THDLCSEL[1]	THDLCSEL[0]	TXCISEL	Reserved		RESET
Туре	R/W	R/W	R/W	R/W	R/W			R/W
Default	X	Х	Χ	Χ	Х			0

# RHDLCSEL[1:0]:

The RHDLCSEL[1:0] select one of the two HDLC Receivers to be accessed by the microprocessor in ESF format. When RHDLC #1 is selected, the HDLC link position is fixed in the DL of F-bit. When RHDLC #2 is selected, the microprocessor can access the HDLC #2 controller to assign the link to any one of 24 channels. These bits must be set before using the HDLC controller.

RHDLCSEL[1:0]	the HDLC Receiver
00	RHDLC#1
01	RHDLC #2
10	Reserved
11	

## THDLCSEL[1:0]:

The THDLCSEL[1:0] select which one of the two HDLC Transmitters to be accessed by the microprocessor in ESF format. When THDLC #1 is selected, the HDLC link position is fixed in the DL of F-bit. When THDLC #2 is selected, the microprocessor can access the HDLC #2 controller to assign the link to any one of 24 channels. These bits must be set before using the HDLC controller.

	<u> </u>
THDLCSEL[1:0]	the HDLC Transmitter
00	THDLC #1
01	THDLC #2
10	Reserved
11	

## TXCISEL:

The registers addressed from T1/J1-070H to T1/J1-071H are shared by the HDLC Receiver and HDLC Transmitter to decide the position of the extracted bit in the received data stream and the inserted bit in the transmitting data stream respectively. This bit is used to decide whether the Read/Write operation on the registers addressed from T1/J1-070H to T1/J1-071H is for the HDLC receiver or for the HDLC transmitter.

- = 0: the Read/Write operation on registers addressed from T1/J1-070H to T1/J1-071H is for HDLC receiver.
- = 1: Read/Write operation on registers addressed from T1/J1-070H to T1/J1-071H is for the HDLC transmitter.

## RESET:

This bit implements a software reset for individual framer.

- = 0: normal operation.
- = 1: The corresponding framer is held in reset. However, this bit and the bits in this register can not be reset. Therefore, a logic 0 must be written to bring the framer out of reset. Holding the framer in a reset state effectively puts it into a low power standby mode. A hardware reset clears the RESET bit and the bits in this register.

## T1 / J1 Interrupt ID (00EH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	INT[8]	INT[7]	INT[6]	INT[5]	INT[4]	INT[3]	INT[2]	INT[1]
Туре	R	R	R	R	R	R	R	R
Default	0	0	0	0	0	0	0	0

This register indicates which one of the eight framers caused the interrupt  $\overline{\text{INT}}$  pin to logic low. When any one of the eight framers caused the interrupt, the corresponding bit in the INT[8:1] will be high.

# T1 / J1 Pattern Generator / Detector Positioning / Control (00FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PRGDSEL[2]	PRGDSEL[1]	PRGDSEL[0]	Nx56k_GEN	Nx56k_DET	RXPATGEN	UNF_GEN	UNF_DET
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

The IDT82V2108 has only one Pattern Generator/Detector (PRGD) shared by all eight framers. At one time, only one framer can use this PRGD. This register selects which framer will use the PRGD and how the PRGD will be used.

# PRGDSEL[2:0]:

PRGDSEL[2:0] select one of the eight framers to be tested by the PRGD block.

PRGDSEL[2:0]	Tested Framer
000	Framer 1
001	Framer 2
010	Framer 3
011	Framer 4
100	Framer 5
101	Framer 6
110	Framer 7
111	Framer 8

## Nx56k\_GEN:

This bit is invalid when the UNF\_GEN (b1, T1/J1-00FH) is logic 1.

- = 0: eight bits are all replaced with the PRGD pattern when one channel is selected in the TPLC or RPLC.
- = 1: the 7 most significant bits are replaced with the PRGD pattern when one channel is selected in the TPLC or RPLC.

## Nx56k DET:

This bit is invalid when the UNF\_DEL (b0, T1/J1-00FH) is logic 1.

- = 0: eight bits are all detected by the PRGD when one channel is selected in the TPLC or RPLC.
- = 1: the 7 most significant bits are detected by the PRGD when one channel is selected in the TPLC or RPLC.

## RXPATGEN:

- = 0: the pattern in PRGD is generated in the transmit path and is detected in the receive path.
- = 1: the pattern in PRGD is generated in the receive path and is detected in the transmit path.

## UNF GEN:

- = 0: which channels of the selected path will be replaced by the PRGD pattern is specified in TPLC or RPLC.
- = 1: all 24 channels and the F-bit of the selected path will be replaced by the PRGD pattern.

# UNF\_DET:

- = 0: which channels of the selected path will be detected by PRGD pattern is specified in TPLC or RPLC.
- = 1: all 24 channels and the F-bit of the selected path will be detected by PRGD pattern.

# T1 / J1 RJAT Interrupt Status (010H, 090H, 110H, 190H, 210H, 290H, 310H, 390H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			OVRI	UNDI				
Туре			R	R				
Default							Χ	Х

# OVRI:

If data are still attempted to write into the FIFO when the FIFO is already full, the overwritten event will occur.

- = 0: the RJAT FIFO is not overwritten.
- = 1: the RJAT FIFO is overwritten.

This bit is cleared to 0 when it is read.

## UNDI:

If data are still attempted to read from the FIFO when the FIFO is already empty, the under-run event will occur.

- = 0: the RJAT FIFO is not under-run.
- = 1: the RJAT FIFO is under-run.

This bit is cleared to 0 when it is read.

# T1 / J1 RJAT Reference Clock Divisor (N1) Control (011H, 091H, 111H, 191H, 211H, 291H, 311H, 391H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N1[7]	N1[6]	N1[5]	N1[4]	N1[3]	N1[2]	N1[1]	N1[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N1[7:0] + 1) is the divisor of the input reference clock, which is the ratio between the frequency of the input reference clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the RJAT.

# T1 / J1 RJAT Output Clock Divisor (N2) Control (012H, 092H, 112H, 192H, 212H, 292H, 312H, 392H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N2[7]	N2[6]	N2[5]	N2[4]	N2[3]	N2[2]	N2[1]	N2[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N2[7:0] + 1) is the divisor of the output smoothed clock, which is the ratio between the frequency of the output smoothed clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the RJAT.

## T1/J1 RJAT Configuration (013H, 093H, 113H, 193H, 213H, 293H, 313H, 393H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				CENT	UNDE	OVRE		LIMIT
Туре		Reserved			R/W	R/W	Reserved	R/W
Default				0	0	0	1	1

## CENT:

The CENT allows the RJAT FIFO to self-center its read pointer, maintaining the pointer at least 4 UI away from the FIFO being empty or full.

- = 0: disable the self-center. Data are pass through uncorrupted when the FIFO is empty or full.
- = 1: enable the FIFO to self-center its read pointer when the FIFO is 4 UI away from being empty or full.

A positive transition in this bit will execute a self-center action immediately.

## UNDE:

This bit decides whether to generate an interrupt when the RJAT FIFO is under-run.

- = 0: No interrupt is generated when the RJAT FIFO is under-run.
- = 1: An interrupt on the INT pin is generated when the RJAT FIFO is under-run.

## OVRE:

This bit decides whether to generate an interrupt when the RJAT FIFO is overwritten.

- = 0: No interrupt is generated when the RJAT FIFO is overwritten.
- = 1: An interrupt on the INT pin is generated when the RJAT FIFO is overwritten.

#### LIMIT:

- = 0: disable the limitation of the jitter attenuation.
- = 1: enable the DPLL to limit the jitter attenuation by enabling the FIFO to increase or decrease the frequency of the output smoothed clock when the read pointer is 1 UI away from the FIFO being empty or full. This limitation of jitter attenuation ensures that no data is lost during high phase shift conditions.

## T1/J1 TRSI Timeslot Offset (014H, 094H, 114H, 194H, 214H, 294H, 314H, 394H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		TSOFF[6]	TSOFF[5]	TSOFF[4]	TSOFF[3]	TSOFF[2]	TSOFF[1]	TSOFF[0]
Туре	Reserved	R/W						
Default		0	0	0	0	0	0	0

In T1/J1 Transmit Clock Slave External Signaling mode E1 rate, the content in the TSOFF[6:0] determines the channel offset between the signal on the TSCFS pin and the start of the corresponding frame transmitted on the TSDn & TSSIGn.

In T1/J1 Transmit Clock Slave TSFS Enabled mode E1 rate, the content in the TSOFF[6:0] determine the channel offset between the signal on the TSCFS pin and the start of the corresponding frame transmitted on the TSDn.

In Transmit Multiplexed mode, the content in the TSOFF[6:0] determine the channel offset between the signal on the MTSCFS pin and the start of the corresponding frame transmitted on the MTSD.

Except for the above three modes, the channel offset is disabled. Thus, the TSOFF[6:0] must be logic 0.

These bits define a binary number. The offset can be set from 0 to 127 channels.

T1 / J1 TRSI Bit Offset (015H, 095H, 115H, 195H, 215H, 295H, 315H, 395H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		MTBS	CMS	COFF		BOFF[2]	BOFF[1]	BOFF[0]
Туре	Reserved	R/W	R/W	R/W	Reserved	R/W	R/W	R/W
Default		0	0	0		0	0	0

#### MTBS:

Valid in Transmit Multiplexed mode.

- = 0: the data of the current channel are taken from the first multiplexed bus (MTSD[1], MTSSIG[1]).
- = 1: the data of the current channel are taken from the second multiplexed bus (MTSD[2], MTSSIG[2]).

## CMS:

- = 0: the clock rate of the TSCCKB/MTSCCKB is the same as that of the backplane.
- = 1: the clock rate of the TSCCKB/MTSCCKB is twice that of the backplane.

The CMS of the eight framers should be set to the same value.

#### COFF:

Valid when the CMS (b5, T1/J1-015H) is logic 1.

- = 0: select the first active edge of the TSCCKB/MTSCCKB to sample / update the data.
- = 1: select the second active edge of the TSCCKB/MTSCCKB to sample / update the data.
- In Transmit Clock Multiplexed mode, the COFF of the eight framers should be set to the same value.

# BOFF[2:0]:

In T1/J1 Transmit Clock Slave External Signaling mode E1 rate, the content in the BOFF[2:0] determines the bit offset between the signal on the TSCFS pin and the start of the SF/ESF transmitted on the TSDn & TSSIGn.

In T1/J1 Transmit Clock Slave TSFS Enabled mode E1 rate, the content in the BOFF[2:0] determines the bit offset between the signal on the TSCFS pin and the start of the SF/ESF transmitted on the TSDn.

In Transmit Multiplexed mode, the content in the BOFF[2:0] determines the bit offset between the signal on the MTSCFS pin and the start of the SF/ESF transmitted on the MTSD & MTSSIG.

Except for the above three modes, the bit offset is disabled. Thus, the BOFF[2:0] must be logic 0.

These bits define a binary number. Refer to the Functional Description for details.

# T1/J1 TJAT Interrupt Status (018H, 098H, 118H, 198H, 218H, 298H, 318H, 398H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			OVRI	UNDI				
Туре			R	R				
Default							Х	X

#### OVRI:

If data are still attempted to write into the FIFO when the FIFO is already full, the overwritten event will occur.

- = 0: the TJAT FIFO is not overwritten.
- = 1: the TJAT FIFO is overwritten.

This bit is cleared to 0 when it is read.

# UNDI:

If data are still attempted to read from the FIFO when the FIFO is already empty, the under-run event will occur.

- = 0: the TJAT FIFO is not under-run.
- = 1: the TJAT FIFO is under-run.

This bit is cleared to 0 when it is read.

## T1/J1 TJAT Reference Clock Divisor (N1) Control (019H, 099H, 119H, 199H, 219H, 299H, 319H, 399H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N1[7]	N1[6]	N1[5]	N1[4]	N1[3]	N1[2]	N1[1]	N1[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N1[7:0] + 1) is the divisor of the input reference clock, which is the ratio between the frequency of the input reference clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the TJAT.

# T1/J1 TJAT Output Clock Divisor (N2) Control (01AH, 09AH, 11AH, 19AH, 21AH, 29AH, 31AH, 39AH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	N2[7]	N2[6]	N2[5]	N2[4]	N2[3]	N2[2]	N2[1]	N2[0]
Туре	R/W							
Default	0	0	1	0	1	1	1	1

These bits define a binary number. The (N2[7:0] + 1) is the divisor of the output smoothed clock, which is the ratio between the frequency of the output smoothed clock and the frequency applied to the phase discriminator input.

Writing to this register will reset the DPLL in the TJAT.

## **T1 / J1 TJAT Configuration** (01BH, 09BH, 11BH, 19BH, 21BH, 29BH, 31BH, 39BH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				CENT	UNDE	OVRE		LIMIT
Туре		Reserved		R/W	R/W	R/W	Reserved	R/W
Default				0	0	0		1

#### CENT:

The CENT allows the TJAT FIFO to self-center its read pointer, maintaining the pointer at least 4 UI away from the FIFO being empty or full.

- = 0: disable the self-center. Data are pass through uncorrupted when the FIFO is empty or full.
- = 1: enable the FIFO to self-center its read pointer when the FIFO is 4 UI away from being empty or full.

A positive transition in this bit will execute a self-center action immediately.

# UNDE:

This bit decides whether to generate an interrupt when the TJAT FIFO is under-run.

- = 0: No interrupt is generated when the TJAT FIFO is under-run.
- = 1: An interrupt on the  $\overline{\text{INT}}$  pin is generated when the TJAT FIFO is under-run.

#### OVRE:

This bit decides whether to generate an interrupt when the TJAT FIFO is overwritten.

- = 0: No interrupt is generated when the TJAT FIFO is overwritten.
- = 1: An interrupt on the INT pin is generated when the TJAT FIFO is overwritten.

#### LIMIT:

- = 0: disable the limitation of the jitter attenuation.
- = 1: enable the DPLL to limit the jitter attenuation by enabling the FIFO to increase or decrease the frequency of the output smoothed clock when the read pointer is 1 UI away from the FIFO being empty or full. This limitation of jitter attenuation ensures that no data is lost during high phase shift conditions.

# T1 / J1 ELSB Interrupt Enable / Status (01DH, 09DH, 11DH, 19DH, 21DH, 29DH, 31DH, 39DH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				SLIPE	SLIPD	SLIPI		
Туре			Reserved			R/W	R	R
Default						0	Χ	Χ

#### SLIPE:

- = 0: disable the interrupt on the  $\overline{INT}$  pin when a slip occurs.
- = 1: enable the interrupt on the  $\overline{INT}$  pin when a slip occurs.

# SLIPD:

This bit is applicable when the SLIPI is logic 1.

- = 0: the latest slip is due to the Elastic Store Buffer being empty; a frame was duplicated.
- = 1: the latest slip is due to the Elastic Store Buffer being full; a frame was deleted.

#### SLIPI:

- = 0: no slip occurs.
- = 1: a slip occurs.

This bit is cleared to 0 after the bit is read.

## T1 / J1 ELSB Idle Code (01EH, 09EH, 11EH, 19EH, 21EH, 29EH, 31EH, 39EH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Туре	R/W							
Default	1	1	1	1	1	1	1	1

These bits set the idle code that will replace the data on the RSDn/MRSD automatically when it is out of SF/ESF sync. D7 is the first bit to be inserted.

The writing of the idle code pattern is asynchronous with respect to the output data clock. One channel of idle code data will be corrupted if the register is written to when the framer is out of frame.

# **T1 / J1 FRMP Configuration** (020H, 0A0H, 120H, 1A0H, 220H, 2A0H, 320H, 3A0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	M2O[1]	M20[0]	ESFFA	ESF	JYEL			
Туре	R/W	R/W	R/W	R/W	R/W		Reserved	
Default	0	0	0	0	0			

# M20[1:0]:

These bits select the SF/ESF frame loss criteria.

- = 00: 2 of 4 frame alignment bits in error.
- = 01: 2 of 5 frame alignment bits in error.
- = 10: 2 of 6 frame alignment bits in error.
- = 11: Reserved

#### ESFFA:

This bit selects the framing algorithm for ESF format.

- = 0: if four consecutive Frame Alignment Patterns are detected in the F-Bit in the received data stream without the mimic framing pattern, the ESF synchronization is acquired. However, if there are mimic framing patterns in the received data stream, the ESF In-frame is not declared.
- = 1: when 6 consecutive Frame Alignment Patterns are received error free and the CRC-6 checksum is also error free, the synchronization is acquired. In this condition, the existence of the mimic framing patterns is not considered.

## ESF:

This bit selects the SF or ESF format in the Frame Processor block.

- = 0: SF format is selected.
- = 1: ESF format is selected.

# JYEL:

This bit selects the T1 or J1 mode in the Frame Processor block.

- = 0: T1 mode is selected.
- = 1: J1 mode is selected.

## T1 / J1 FRMP Interrupt Enable (021H, 0A1H, 121H, 1A1H, 221H, 2A1H, 321H, 3A1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	Reserved		COFAE	FERE	BEEE	SFEE	MFPE	INFRE
Туре			R/W	R/W	R/W	R/W	R/W	R/W
Default			0	0	0	0	0	0

#### COFAE:

When the frame alignment pattern has been achieved and the position of the new frame alignment pattern differs from the previous one, this bit decides whether to generate an interrupt or not.

- = 0: Disables the interrupt when there is a shift on the framing signal position.
- =1: Enables the interrupt on INT pin when there is a shift on the framing signal position.

#### FERE:

- = 0: No interrupt is generated when there is a framing bit error.
- = 1: An interrupt on  $\overline{INT}$  pin is generated when a framing bit error is detected.

## BEEE:

- = 0: No interrupt is generated when there is a bit error event.
- = 1: An interrupt on INT pin is generated when a bit error event occurs. Here, the bit error event is defined as a framing bit error for SF formatted data and a CRC-6 error (the local calculated CRC-6 result is not the same as the received CRC-6 bits) for ESF formatted data.

(In SF mode, this bit has the same function as the FERE.)

#### SFEE:

The Severe Framing Error is defined as 2 or more framing bit errors during the current super-frame of SF or ESF data.

- = 0: No interrupt is generated when there is a Severe Framing Error.
- = 1: An interrupt on INT pin is generated when Severe Framing Error event occurs.

## MFPE:

Mimic Framing Pattern is defined as more than one framing alignment pattern existing simultaneously in the receiving data stream. This bit decides whether to generate an interrupt when Mimic Framing Pattern appears or disappears.

- = 0: No interrupt is generated when there is a transition of the status of Mimic Framing Pattern.
- = 1: An interrupt on  $\overline{\text{INT}}$  pin is generated when there is a transition (exist to non-exist, or non-exist to exist) of the status of Mimic Framing Pattern.

## INFRE:

This bit decides whether to generate an interrupt when the status of incoming data stream changes from in-frame to out-of-frame or from out-of-frame to in-frame.

- = 0: No interrupt is generated when there is a transition of Frame Synchronize Status.
- = 1: An interrupt on INT pin is generated when there is a transition of Framing Synchronize Status.

### T1 / J1 FRMP Interrupt Status (022H, 0A2H, 122H, 1A2H, 222H, 2A2H, 322H, 3A2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	COFAI	FERI	BEEI	SFEI	MFPI	INFRI	MFP	INFR
Туре	R	R	R	R	R	R	R	R
Default	0	0	0	0	0	0	0	0

#### COFAI:

- = 0: Indicates the framing signal position shift has not occurred.
- = 1: Indicates the occurrence of framing signal position shift, which means when the frame alignment pattern has been achieved, the position of the new alignment pattern differs from the previous one.

This bit is cleared to 0 after it is read.

### FERI:

- = 0: Indicates that there is no framing bit error.
- = 1: Indicates the occurrence of a framing bit error.

This bit is cleared to 0 after it is read.

### BEEI:

This bit indicates the occurrence of a bit error event. The bit error event is defined as a framing bit error for SF format or a CRC-6 error (the local calculated CRC-6 result is not the same as the received CRC-6 bits) for ESF format.

- = 0: Indicates there is no bit error.
- = 1: Indicates the occurrence of a bit error.

(For SF formatted data, this bit has the same function as FERI bit.)

This bit is cleared to 0 after it is read

#### SFEI:

The Severe Framing Error is defined as 2 or more framing bit errors during the current super-frame of SF or ESF data.

- = 0: Indicates there is no severe framing error.
- = 1: Indicates the occurrence of severe framing error.

This bit is cleared to 0 after it is read.

#### MFPI:

This bit indicates the transition of the status of the current mimic framing pattern.

- = 0: When the status of current mimic framing pattern is not changed.
- = 1: When there is a transition (exist to non-exist, or non-exist to exist) of the status of mimic framing pattern.

This bit is cleared to 0 after it is read.

### INFRI:

This bit indicates the transition of frame synchronization status.

- = 0: When the frame synchronization status is not changed.
- = 1: When the frame synchronization status of the receiving data stream changes from in-frame to out-of-frame or from out-of-frame to in-frame.

#### MFP:

This bit reflects the current status of mimic framing pattern.

- = 0: Indicates that the mimic framing pattern does not exist.
- = 1: Indicates the presence of more than one framing alignment patterns in the receiving data stream.

Read operation will not change the status of this bit.

### INFR:

This bit reflects the current status of frame synchronization.

- = 0: The received data stream is out-of-frame.
- = 1: The received data stream is in-frame.

Read operation will not change the status of this bit.

# T1 / J1 Clock Monitor (027H, 0A7H, 127H, 1A7H, 227H, 2A7H, 327H, 3A7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				XCK	TSCCKB	TSCCKA	RSCCK	LRCK
Туре		Reserved		R	R	R	R	R
Default				X	X	X	X	X

This register provides monitoring on the IDT82V2108 clocks. When a monitored clock signal makes a low to high transition, the corresponding bit in this register is set to 1, and this bit remains to be 1 until this register is read. After a read operation on this register, all the bits in this register will be cleared to 0. A lack of transitions of the monitored clock will be indicated by 0 in the corresponding bit, which means that the clock fails. This register should be read periodically to detect clock failures.

# XCK:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the XCK.

# TSCCKB:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the TSCCKB.

# TSCCKA:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the TSCCKA.

### RSCCK:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the RSCCK.

### LRCK:

- = 0: after the bit is read.
- = 1: a low to high transition occurs on the LRCK.

### T1 / J1 RBOM Enable (02AH, 0AAH, 12AH, 1AAH, 22AH, 2AAH, 32AH, 3AAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name						IDLE	AVC	BOCE
Туре			Reserved	R/W	R/W	R/W		
Default						0	0	0

### IDLE:

- = 0: disable the interrupt on the INT pin when there is a transition from BOM to non-BOM in the received data stream.
- = 1: enable the interrupt on the INT pin when there is a transition from BOM to non-BOM in the received data stream.

### AVC:

This bit selects the validation criteria used to acknowledge the Bit Oriented Message (BOM) in the received data stream, or to acknowledge the Yellow signal in T1/J1 ESF format.

- = 0: the BOM or the Yellow signal is acknowledged when the pattern is matched and the received code is identical 8 out of 10 times.
- = 1: the BOM or the Yellow signal is acknowledged when the pattern is matched and the received code is identical 4 out of 5 times.

#### BOCE:

- = 0: disable the interrupt on the INT pin when a valid BOM code is detected in the received data stream.
- = 1: enable the interrupt on the INT pin when a valid BOM code is detected in the received data stream.

# T1/J1 RBOM Code Status (02BH, 0ABH, 12BH, 1ABH, 22BH, 2ABH, 32BH, 3ABH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	IDLEI	BOCI	BOC[5]	BOC[4]	BOC[3]	BOC[2]	BOC[1]	BOC[0]
Туре	R	R	R	R	R	R	R	R
Default	0	0	1	1	1	1	1	1

#### IDLEI:

- = 0: no transition from Bit Oriented Message (BOM) to non-BOM in the received data stream.
- = 1: a transition from BOM to non-BOM in the received data stream.

This bit is cleared to 0 after the register is read.

# BOCI:

- = 0: no Bit Oriented Message (BOM) is detected.
- = 1: BOM is detected in the received data stream.

This bit is cleared to 0 after the register is read.

### BOC[5:0]:

These bits directly reflect the content of the Bit Oriented Message (BOM) in the received data stream.

All ones in the BOC[5:0] mean there is no BOM received.

The BOC[5] corresponds to the MSB of the code while the BOC[0] corresponds to the LSB.

## T1/J1 ALMD Configuration (02CH, 0ACH, 12CH, 1ACH, 22CH, 2ACH, 32CH, 3ACH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			J1_YEL	ESF				
Туре	Rese	rved	R/W	R/W	Reserved			
Default			0	0	7			

# J1\_YEL:

This bit selects the T1 or J1 mode in the Alarm Detector block.

- = 0: T1 mode is selected in the ALMD block.
- = 1: J1 mode is selected in the ALMD block.

#### ESF:

This bit selects the SF or ESF format in the Alarm Detector block.

- = 0: SF format is selected.
- = 1: ESF format is selected.

# **T1/J1 ALMD Interrupt Enable** (02DH, 0ADH, 12DH, 1ADH, 22DH, 2ADH, 32DH, 3ADH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				FASTD		YELE	REDE	AISE
Туре		Reserved		R/W	Reserved	R/W	R/W	R/W
Default				0		0	0	0

#### FASTD:

- = 0: the RED Alarm is cleared when the out of SF/ESF sync event has been absent for 16.6sec (±500ms); the AIS Alarm is cleared when the AIS signal has been absent for 16.8sec (±500ms).
- = 1: the RED Alarm is cleared when the out of SF/ESF sync event has been absent for 120ms; the AIS Alarm is cleared when the AIS signal has been absent for 180ms.

### YELE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the YELI (b5, T1/J1-02EH) is logic one.
- = 1: enable the interrupt on the INT pin when the YELI (b5, T1/J1-02EH) is logic one.

### REDE:

- = 0: disable the interrupt on the INT pin when the REDI (b4, T1/J1-02EH) is logic one.
- = 1: enable the interrupt on the INT pin when the REDI (b4, T1/J1-02EH) is logic one.

### AISE:

- = 0: disable the interrupt on the INT pin when the AISI (b3, T1/J1-02EH) is logic one.
- = 1: enable the interrupt on the  $\overline{\text{INT}}$  pin when the AISI (b3, T1/J1-02EH) is logic one.

# T1/J1 ALMD Interrupt Status (02EH, 0AEH, 12EH, 1AEH, 22EH, 2AEH, 32EH, 3AEH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		Reserved		REDI	AISI	YEL	RED	AIS
Туре	Rese			R	R	R	R	R
Default			0	0	0	0	0	0

### YELI:

- = 0: there is no transition (from 1 to 0 or from 0 to 1) on the YEL (b2, T1/J1-02EH).
- = 1: there is a transition (from 1 to 0 or from 0 to 1) on the YEL (b2, T1/J1-02EH).

This bit is clear to 0 after the register is read.

# REDI:

- = 0: there is no transition (from 1 to 0 or from 0 to 1) on the RED (b1, T1/J1-02EH).
- = 1: there is a transition (from 1 to 0 or from 0 to 1) on the RED (b1, T1/J1-02EH).

This bit is clear to 0 after the register is read.

# AISI:

- = 0: there is no transition (from 1 to 0 or from 0 to 1) on the AIS (b0, T1/J1-02EH).
- = 1: there is a transition (from 1 to 0 or from 0 to 1) on the AIS (b0, T1/J1-02EH).

This bit is clear to 0 after the register is read.

# YEL:

- = 0: the Yellow signal has been absent for 425ms ( $\pm$ 50ms).
- = 1: the Yellow signal has been present for 425ms ( $\pm$ 50ms).

#### RED:

- = 0: the REDD (b2, T1/J1-02FH) has been logic 0 for 16.6sec (±500ms), or for 120ms if the FASTD (b4, T1/J1-02DH) is set.
- = 1: the REDD (b2, T1/J1-02FH) has been logic 1 for 2.55sec ( $\pm$ 40ms).

### AIS:

- = 0: the AIS signal has been absent for 16.8sec (±500ms), or for 180ms if the FASTD (b4, T1/J1-02DH) is set.
- = 1: the AIS signal has been present for 1.5sec ( $\pm 100$ ms).

### T1/J1 ALMD Alarm Detection Status (02FH, 0AFH, 12FH, 1AFH, 22FH, 2AFH, 32FH, 3AFH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name							YELD	AISD
Туре			Reserved			R	R	R
Default						Χ	Χ	Χ

### REDD:

- = 0: no out of SF/ESF sync event has occurred in the latest 40ms period.
- = 1: one or more out of SF/ESF sync events have occurred in the latest 40ms period.

### YELD:

- = 0: In SF format, the Yellow signal is absent during the latest 40ms period; in ESF format, the Yellow signal is absent during the latest 4ms period.
- = 1: In SF format, the Yellow signal is present during the latest 40ms period; in ESF format, when the AVC (b1, T1/J1-02AH) is 0, the Yellow signal is present during the latest 40ms period, when the AVC (b1, T1/J1-02AH) is 1, the Yellow signal is present during the latest 20ms period.

The Yellow signal is acknowledged differently in each format:

- In T1 SF format: The Yellow signal occupies the 2nd bit of each channel. When the bit is logic 1 for 16 or fewer times during the 40ms period, the Yellow signal is present.
- In J1 SF format: The Yellow signal occupies the F-bit of the 12th frame. However, when the bit is logic 0 for 2 or fewer times during the 40ms period, the Yellow signal is present.
- In T1/J1 ESF format: The Yellow signal occupies the DL of the F-bit (refer to Table-4). The pattern is 'FF00' in T1 mode and 'FFFF' in J1 mode. When the AVC (b1, T1/J1-02AH) is logic 0, the Yellow signal is acknowledged if the pattern is matched in 8 out of 10 successive DL. When the AVC (b1, T1/J1-02AH) is logic 1, the Yellow signal is acknowledged if the pattern is matched in 4 out of 5 successive DL.

### AISD:

- = 0: the AIS signal is absent during the latest 60ms period.
- = 1: the AIS signal is present during the latest 60ms period.

The AIS signal is acknowledged when the received data are out of SF/ESF synchronization for 60ms and the received logic 0 is less than 127 times in the same period.

## T1/J1 TPLC Configuration (030H, 0B0H, 130H, 1B0H, 230H, 2B0H, 330H, 3B0H)

Bit No.	7	6	5	4	3	2	1	0			
Bit Name		_									
Туре		Reserved									
Default											

### PCCE:

- = 0: the per-channel functions in TPLC are disabled.
- = 1: the per-channel functions in TPLC are enabled.

# T1 / J1 TPLC **TIP** Access Status (031H, 0B1H, 131H, 1B1H, 231H, 2B1H, 331H, 3B1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	BUSY							
Туре	R				Reserved			
Default	0							

#### BUSY:

- = 0: no reading or writing operation on the indirect registers.
- = 1: an internal indirect register is being accessed, any new operation on the internal indirect register is not allowed.

This bit goes low timed to an internal high-speed clock rising edge after the operation has been completed. The operation cycle is 650ns. No more operations to the indirect registers could be done until this bit is cleared.

# T1 / J1 TPLC Channel Indirect Address / Control (032H, 0B2H, 132H, 1B2H, 232H, 2B2H, 332H, 3B2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	R/WB	A6	A5	A4	A3	A2	A1	A0
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

Writing to this register with a valid address and R/WB bit initiates an internal operation cycle to the indirect registers.

# R/WB:

- = 0: write the data to the specified indirect register.
- = 1: read the data from the specified indirect register.

# A[6:0]:

Specify the address of the indirect registers (from 01H to 48H) for the microprocessor access.

T1 / J1 TPLC Channel Indirect Data Buffer (033H, 0B3H, 133H, 1B3H, 233H, 2B3H, 333H, 3B3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Туре	R/W							
Default	0	0	0	0	0	0	0	0

This register hold the value which will be read from or write into the indirect registers (from 01H to 48H). If data are to be written to the indirect registers, the byte to be written must be written into this register before the target indirect register's address and R/WB=0 is written into the Address/Control register, initiating the access. If data are to be read from the indirect registers, only the target indirect register's address and R/WB=1 is written into the Address/Control register, initiating the request. After 490 ns, this register will contain the requested data byte.

TPLC Indirect Registers Map								
01H ~ 18H	Per-Channel Control for Channel 1 ~ 24							
19H ~ 30H	IDLE Code Byte for Channel 1 ~ 24							
31H~48H	Signaling Control Byte for Channel 1 ~ 24							

# T1/J1 TPLC Per-Channel Control Registers (TPLC Indirect Registers 01H - 18H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	INVERT	IDLE_DS0	DMW	SIGNINV	TEST	LOOP	ZCS[1]	ZCS[0]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	Х	Χ	Χ	Χ	Х	Χ	Χ	Χ

### INVERT:

This bit, together with the SIGNINV (b4, T1/J1-TPLC-indirect register - 01~18H), determines the bit inversion of the corresponding channel when input from the TSDn/MTSD pin.

INVERT	SIGNINV	Bit Inversion					
0	0	No bit inversion					
0	1	Invert the MSB of the corresponding channel					
1	0	Invert all the bits of the corresponding channel					
1	1	Invert all the bits except the MSB of the corresponding channel					

### IDLE DS0:

- = 0: disable the data in the corresponding channel to be replaced by the data set in the IDLE[7:0] when input from the TSDn/MTSD pin.
- = 1: enable the data in the corresponding channel to be replaced by the data set in the IDLE[7:0] when input from the TSDn/MTSD pin.

### DMW:

- = 0: disable the data in the corresponding channel to be replaced with a digital milliwatt pattern when input from the TSDn/MTSD pin.
- = 1: enable the data in the corresponding channel to be replaced with a digital milliwatt pattern when input from the TSDn/MTSD pin.

#### SIGNINV

Refer to the INVERT (b7, T1/J1-TPLC-indirect register - 01~18H).

#### TEST:

- = 0: disable the data in the corresponding channel to be tested by PRGD.
- = 1: enable the data in the corresponding channel to be extracted to PRGD for test (when the RXPATGEN [b2, T1/J1-00FH] is logic 1), or enable the test pattern from PRGD to replace the data in the corresponding channel for test (when the RXPATGEN [b2, T1/J1-00FH] is logic 0).

All the timeslots that are extracted to the PRGD are concatenated and treated as a continuous stream in which pseudo random are searched for. Similarly, all timeslots set to be replaced with PRGD test pattern data are concatenated replaced by the PRBS.

### LOOP:

- = 0: disable the payload loopback.
- = 1: enable the payload loopback. When the Receive Clock Master modes are enabled, the Elastic Store is used to align the receive line data to the data to be transmitted. When Receive Clock Slave modes are enabled, the Elastic Store is unavailable to facilitate the payload loopbacks, and loopback functionality is provided only when the transmit path is also in Transmit Clock Slave mode, and the received clock and the clock to be transmitted and Common Frame Pulse are identical (RSCCK = TSCCKB, RSCFS = TSCFS).

# ZCS[1:0]:

ZCS[1:0]	Per-Channel Zero Code Suppression
0 0	No zero code suppression.
0 1	Every bit 8 in the corresponding channel is forced to be logic one.
10	GTE Zero Code Suppression – Every bit 8 (or bit 7 in signaling frames) is forced to be logic one when the bits in the
	corresponding channel are all zeros.
11	Bell Zero Code Suppression – Every bit 7 is forced to be logic one when the bits in the corresponding channel are all zeros.

The priority of the TPLC operation on the TSDn/MTSD pin from high to low is:

Extract data to PRGD for test; Zero Code Suppression; Payload loopback; Replace the data with the milliwatt pattern; Replace the data with the pattern generated in the PRGD; Replace the data with the value in the IDLE[7:0]; Invert the bit.

# T1/J1 TPLC IDLE Code Byte Registers (TPLC Indirect Registers 19H – 30H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	IDLE7	IDLE6	IDLE5	IDLE4	IDLE3	IDLE2	IDLE1	IDLE0
Туре	R/W							
Default	Χ	Х	X	Χ	Х	Х	Х	X

They contain the data that will replace the data input from the TSDn/MTSD pin when the corresponding IDLE\_DS0 is logic 1. IDLE7 is the MSB.

# **T1/J1 TPLC Signaling Control Byte Registers** (TPLC Indirect Registers 31H – 48H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	SIGC0	SIGC1			Α	В	С	D
Туре	R/W	R/W	Rese	Reserved		R/W	R/W	R/W
Default	Х	X			Х	Χ	Χ	X

# SIGC0:

This bit is valid when the corresponding SIGC1 is logic 1.

- = 0: use the data input from the TSSIGn/MTSSIG pin as the signaling.
- = 1: use the data in the A, B, C, D as the signaling.

# SIGC1:

- = 0: disable replacing the signaling bit with the data input from the TSSIGn/MTSSIG pin or the data in the A, B, C, D.
- = 1: enable replacing the signaling bit with the data input from the TSSIGn/MTSSIG pin or the data in the A, B, C, D.

### A, B, C, D:

They contain the data that can be used as signaling when the corresponding SIGC0 is logic 1. They are in the least significant nibble.

## T1/J1 THDLC #1, #2 Configuration (034H, 0B4H, 134H, 1B4H, 234H, 2B4H, 334H, 3B4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FLGSHARE	FIFOCLR			EOM	ABT	CRC	EN
Туре	R/W	R/W	Rese	Reserved		R/W	R/W	R/W
Default	1	0			0	0	1	0

Selection of the THDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, T1/J1-00DH).

#### FLGSHARE:

- = 0: the closing flag of the current HDLC and the opening flag of the next HDLC are separate.
- = 1: the closing flag of the current HDLC and the opening flag of the next HDLC are shared.

#### FIFOCLR:

- = 0: normal operation.
- = 1: clear the FIFO.

### EOM:

- = 0: normal operation.
- = 1: a positive transition of this bit starts a packet transmission. Then if the CRC(b1, E1-050H) is set, the 16-bit FCS word is appended to the last data byte transmitted.

### ABT:

- = 0: normal operation.
- = 1: transmit the 7F abort sequence after the current setting in the Transmit Data register is transmitted, so that the FIFO is cleared and all data in the FIFO will be lost.

Aborts are continuously sent and the FIFO is held in reset until this bit is reset to a logic 0. At least one Abort sequence will be sent when the ABT transitions from logic 0 to logic 1.

### CRC:

- = 0: do not append the CRC-16 frame check sequences (FCS) to the end of the HDLC data.
- = 1: append the FCS to the end of the HDLC data

#### EN:

- = 0: disable the operation of the THDLC block and transmit all ones on the assigned data link.
- = 1: enable the operation of the THDLC block and flag sequences are sent until data are written into the THDLC Transmit Data register and the EOM is set to logic 1.

# T1/J1 THDLC #1, #2 Upper Transmit Threshold (035H, 0B5H, 135H, 1B5H, 235H, 2B5H, 335H, 3B5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	Reserved	UTHR[6]	UTHR[5]	UTHR[4]	UTHR[3]	UTHR[2]	UTHR[1]	UTHR[0]
Туре		R/W						
Default		1	0	0	0	0	0	0

Selection of the THDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, T1/J1-00DH).

# UTHR[6:0]:

These bits define the upper fill level of the FIFO. Once the fill level exceeds the UTHR[6:0] value, the data stored in the FIFO will start to transmit. The transmission will not stop until the last complete packet is transmitted and the THDLC FIFO fill level is below UTHR[6:0] + 1.

It should be greater than the value of the LINT[6:0] unless both are equal to 00H.

# **T1 / J1 THDLC #1, #2 Lower Interrupt Threshold** (036H, 0B6H, 136H, 1B6H, 236H, 2B6H, 336H, 3B6H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		LINT[6]	LINT[5]	LINT[4]	LINT[3]	LINT[2]	LINT[1]	LINT[0]
Туре	Reserved	R/W						
Default		0	0	0	0	1	1	1

Selection of the THDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, T1/J1-00DH).

### LINT[6:0]:

These bits define the fill level of the FIFO that can cause an interrupt. That is, when the fill level of the FIFO is below the LINT[6:0], an interrupt will be generated. It should be less than the value of the UTHR[6:0] unless both are equal to 00H.

# T1 / J1 THDLC #1, #2 Interrupt Enable (037H, 0B7H, 137H, 1B7H, 237H, 2B7H, 337H, 3B7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				FULLE	OVRE	UDRE	LFILLE	
Туре		Rese	rved		R/W	R/W	R/W	R/W
Default					0	0	0	0

Selection of the THDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, T1/J1-00DH).

#### FULLE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the FULLI (b3, T1/J1-038H) is logic 1.
- = 1: enable the interrupt on the INT pin when the FULLI (b3, T1/J1-038H) is logic 1.

### OVRE:

- = 0: disable the interrupt on the INT pin when the OVRI (b2, T1/J1-038H) is logic 1.
- = 1: enable the interrupt on the INT pin when the OVRI (b2, T1/J1-038H) is logic 1.

# UDRE:

- = 0: disable the interrupt on the INT pin when the UDRI (b1, T1/J1-038H) is logic 1.
- = 1: enable the interrupt on the INT pin when the UDRI (b1, T1/J1-038H) is logic 1.

# LFILLE:

- = 0: disable the interrupt on the INT pin when the LFILLI (b0, T1/J1-038H) is logic 1.
- = 1: enable the interrupt on the INT pin when the LFILLI (b0, T1/J1-038H) is logic 1.

# T1 / J1 THDLC #1, #2 Interrupt Status / UDR Clear (038H, 0B8H, 138H, 1B8H, 238H, 2B8H, 338H, 3B8H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		FULL	BLFILL		FULLI	OVRI	UDRI	LFILLI
Туре	Reserved	R	R	Reserved	R	R	R	R
Default		Χ	Χ		Х	Х	Х	Х

Selection of the THDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, T1/J1-00DH).

#### FULL:

- = 0: the THDLC FIFO is not full.
- = 1: the THDLC FIFO is full (128 bits).

### BLFILL:

- = 0: the data in the THDLC FIFO are not below the value of the LINT[6:0] (b6~0, T1/J1-036H).
- = 1: the data in the THDLC FIFO are empty or below the value of the LINT[6:0] (b6~0, T1/J1-036H).

# FULLI:

- = 0: there is no transition (from 0 to 1) on the FULL.
- = 1: there is a transition (from 0 to 1) on the FULL.

This bit is clear to 0 after the bit is read.

### OVRI:

The Over-Written is that the THDLC FIFO was already full when another data byte was written to the THDLC Transmit Data register.

- = 0: the THDLC FIFO is not overwritten.
- = 1: the THDLC FIFO is overwritten (more than 128 bits).

This bit is clear to 0 after the bit is read.

# UDRI:

The Under-Run is that the THDLC was in the process of transmitting a packet when it ran out of data to be transmitted.

- = 0: the THDLC FIFO is not under-run.
- = 1: the THDLC FIFO is under-run.

This bit is clear to 0 after the bit is read.

### LFILLI:

- = 0: there is no transition (from 0 to 1) on the BLFILL.
- = 1: there is a transition (from 0 to 1) on the BLFILL.

This bit is clear to 0 after the bit is read.

# T1/J1 THDLC #1, #2 Transmit Data (039H, 0B9H, 139H, 1B9H, 239H, 2B9H, 339H, 3B9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	TD[7]	TD[6]	TD[5]	TD[4]	TD[3]	TD[2]	TD[1]	TD[0]
Туре	R/W							
Default	Х	X	Χ	X	X	X	X	Χ

Selection of the THDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the THDLCSEL[1:0] (b5~4, T1/J1-00DH).

The content is the data to be transmitted. It is serially transmitted (TD[0] is the first).

# T1 / J1 IBCD Configuration (03CH, 0BCH, 13CH, 1BCH, 23CH, 2BCH, 33CH, 3BCH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			IBCD_ERR[1]	IBCD_ERR[0]	DSEL1	DSEL0	ASEL1	ASEL0
Туре	Reserved		R/W	R/W	R/W	R/W	R/W	R/W
Default			0	0	0	0	0	0

# IBCD\_ERR[1:0]:

The IBCD\_ERR[1:0] sets the error tolerance in the received activate/deactivate code within 39.8ms:

IBCD_ERR[1:0]	Error Tolerance			
0 0	0 bit			
0 1	200 bits			
10	20 bits			
11	2 bits			

# DSEL[1:0]:

The DSEL[1:0] define the length of the received loopback deactivate code, meanwhile, it define the valid code in the DACT[7:0] (b7~0, T1/J1-03FH):

DSEL[1:0]	Deactivate Code Length & Valid Code In the DACT[7:0]
0 0	5-bit length & the code in the DACT[7:3] is valid
0 1	6-bit or 3-bit length & the code in the DACT[7:2] is valid
1 0	7-bit length & the code in the DACT[7:1] is valid
11	8-bit or 4-bit length & the code in the DACT[7:0] is valid

# ASEL[1:0]:

The ASEL[1:0] define the length of the received loopback activate code, meanwhile, it define the valid code in the ACT[7:0] (b7~0, T1/J1-03EH):

ASEL[1:0]	Activate Code Length & Valid Code In the ACT[7:0]
0 0	5-bit length & the code in the ACT[7:3] is valid
0 1	6-bit or 3-bit length & the code in the ACT[7:2] is valid
1 0	7-bit length & the code in the ACT[7:1] is valid
11	8-bit or 4-bit length & the code in the ACT[7:0] is valid

## T1 / J1 IBCD Interrupt Enable / Status (03DH, 0BDH, 13DH, 1BDH, 23DH, 2BDH, 33DH, 3BDH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	LBACP	LBDCP	LBAE	LBDE	LBAI	LBDI	LBA	LBD
Туре	R	R	R/W	R/W	R	R	R	R
Default	0	0	0	0	0	0	0	0

# LBACP:

- = 0: no loopback activate code is present for 39.8ms.
- = 1: the loopback activate code is present for 39.8ms.

### LBDCP:

- = 0: no loopback deactivate code is present for 39.8ms.
- = 1: the loopback deactivate code is present for 39.8ms.

### LBAE:

- = 0: disable the interrupt on the INT pin when the loopback activate code status changes (i.e., the LBAI is logic one).
- = 1: enable the interrupt on the INT pin when the loopback activate code status changes (i.e., the LBAI is logic one).

#### LBDE:

- = 0: disable the interrupt on the INT pin when the loopback deactivate code status changes (i.e., the LBDI is logic one).
- = 1: enable the interrupt on the INT pin when the loopback deactivate code status changes (i.e., the LBDI is logic one).

### LBAI:

- = 0: the loopback activate code status does not change.
- = 1: the loopback activate code status changes (i.e., there is a transition from 0 to 1 or from 1 to 0 on the LBA). This bit is cleared to 0 after the register is read.

## LBDI:

- = 0: the loopback deactivate code status does not change.
- = 1: the loopback deactivate code status changes (i.e., there is a transition from 0 to 1 or from 1 to 0 on the LBD). This bit is cleared to 0 after the register is read.

### LBA:

- = 0: no loopback activate code is present for 5.1s.
- = 1: the loopback activate code is present for 5.1s.

#### LBD:

- = 0: no loopback deactivate code is present for 5.1s.
- = 1: the loopback deactivate code is present for 5.1s.

## T1/J1 IBCD Activate Code (03EH, 0BEH, 13EH, 1BEH, 23EH, 2BEH, 33EH, 3BEH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	ACT7	ACT6	ACT5	ACT4	ACT3	ACT2	ACT1	ACT0
Туре	R/W							
Default	0	0	0	0	0	0	0	0

The ACT[7:X] defines the content of the activate code. 'X' is 3, 2, 1 or 0 and depends on the length defined by the ASEL[1:0] (b1~0, T1/J1-03CH). The unused bits should be ignored. The ACT[7] is the MSB and compares with the first received code bit.

# T1/J1 IBCD Deactivate Code (03FH, 0BFH, 13FH, 1BFH, 23FH, 2BFH, 33FH, 3BFH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DACT7	DACT6	ADCT5	DACT4	DACT3	DACT2	DACT1	DACT0
Туре	R/W							
Default	0	0	0	0	0	0	0	0

The DACT[7:X] defines the content of the deactivate code. 'X' is 3, 2, 1 or 0 and depends on the length defined by the DSEL[1:0] (b3~2, T1/J1-03CH). The unused bits should be ignored. The DACT[7] is the MSB and compares with the first received code bit.

# T1/J1 RCRB Configuration (COSS = 0) (040H, 0C0H, 140H, 1C0H, 240H, 2C0H, 340H, 3C0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		COSS	SIGE			ESF		PCCE
Туре	Reserved	R/W	R/W	Reserved		R/W	Reserved	R/W
Default		0	0			0		0

### COSS:

- = 0: allow the RCRB registers to access the indirect registers.
- = 1: allow the RCRB registers to reflect the change of the signaling of its corresponding channel.

## SIGE:

- = 0: disable generation of an interrupt on the INT pin when there is signaling change in any one of the 24 channels.
- = 1: enable generation of an interrupt on the INT pin when there is signaling change in any one of the 24 channels.

# ESF:

This bit selects the SF or ESF format in the Receive CAS/RBS Buffer block.

- = 0: SF format is selected.
- = 1: ESF format is selected.

# PCCE:

- = 0: the per-channel functions in RCRB are disabled.
- = 1: the per-channel functions in RCRB are enabled.

## T1/J1 RCRB Channel Indirect Status (COSS = 0) (041H, 0C1H, 141H, 1C1H, 241H, 2C1H, 341H, 3C1H)

Bit No.	7	6	5	4	3	2	1	0			
Bit Name	BUSY										
Туре	R		Reserved								
Default	0										

#### BUSY:

- = 0: no reading or writing operation on the indirect registers is occurring.
- = 1: an internal indirect register is being accessed, any new operation on the internal indirect register is not allowed.

This bit goes low timed to an internal high-speed clock rising edge after the operation has been completed. The operation cycle is 650ns. No more operations to the indirect registers could be done until this bit is cleared.

### T1/J1 RCRB Channel Indirect Address / Control (COSS = 0) (042H, 0C2H, 142H, 1C2H, 242H, 2C2H, 342H, 3C2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	R/WB	A6	A5	A4	A3	A2	A1	A0
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

#### R/WB:

- = 0: write the data to the specified indirect register.
- = 1: read the data from the specified indirect register.

# A[6:0]:

Specifies the address of the indirect registers (from 20H to 57H) for the microprocessor access.

# T1/J1 RCRB Channel Indirect Data Buffer (COSS = 0) (043H, 0C3H, 143H, 1C3H, 243H, 2C3H, 343H, 3C3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Туре	R/W							
Default	Χ	Χ	Χ	Х	X	X	Χ	X

This register holds the value which will be read from or written to the indirect registers (from 20H to 57H). If data are to be written to the indirect registers, the byte to be written must be written into this register before the target indirect register's address and R/WB=0 is written into the Address/Control register, initiating the access. If data are to be read from the indirect registers, only the target indirect register's address and R/WB=1 is written into the Address/Control register, initiating the request. After 640 ns, this register will contain the requested data byte.

## T1/J1 RCRB Configuration (COSS = 1) (040H, 0C0H, 140H, 1C0H, 240H, 2C0H, 340H, 3C0H)

Bit No.	7	6	5	4	3	2	1	0			
Bit Name		COSS									
Туре	Reserved	R/W		Reserved							
Default		0									

### COSS:

- = 0: allow the RCRB registers to access the indirect registers.
- = 1: allow the RCRB registers to reflect the change of the signaling of its corresponding channel.

# T1 / J1 RCRB Signaling State Change Channels 17-24 (COSS = 1) (041H, 0C1H, 141H, 1C1H, 241H, 2C1H, 341H, 3C1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	COSS[24]	COSS[23]	COSS[22]	COSS[21]	COSS[20]	COSS[19]	COSS[18]	COSS[17]
Type	R	R	R	R	R	R	R	R
Default	X	X	Χ	Χ	Χ	Х	Х	Χ

### COSSn:

- = 0: the signaling in its corresponding channel is not changed.
- = 1: the signaling in its corresponding channel is changed.

These bits are cleared to 0 after the register is read. COSS[24:17] correspond to channels 24 to 17.

# T1/J1 RCRB Signaling State Change Channels 9-16 (COSS = 1) (042H, 0C2H, 142H, 1C2H, 242H, 2C2H, 342H, 3C2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	COSS[16]	COSS[15]	COSS[14]	COSS[13]	COSS[12]	COSS[11]	COSS[10]	COSS[9]
Туре	R	R	R	R	R	R	R	R
Default	X	Χ	Χ	Χ	X	X	X	Х

# COSSn:

- = 0: the signaling in its corresponding channel is not changed.
- = 1: the signaling in its corresponding channel is changed.

These bits are cleared to 0 after the register is read. COSS[16:9] correspond to channels 16 to 9.

# T1 / J1 RCRB Signaling State Change Channels 1-8 (COSS = 1) (043H, 0C3H, 143H, 1C3H, 243H, 2C3H, 343H, 3C3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	COSS[8]	COSS[7]	COSS[6]	COSS[5]	COSS[4]	COSS[3]	COSS[2]	COSS[1]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Х	Х	Х	Х	Х	Х	Х

### COSSn:

- = 0: the signaling in its corresponding channel is not changed.
- = 1: the signaling in its corresponding channel is changed.

These bits are cleared to 0 after the register is read. COSS[8:1] correspond to channels 8 to 1.

# **RCRB Indirect Registers Map**

01H ~ 18H / 21H ~ 38H	Channel Signaling Data Register for Channel 1 ~ 24
19H ~ 20H, 39H ~ 40H	-
41H ~ 58H	Per-Channel Configuration Register for Channel 1 ~ 24

# T1/J1 RCRB Channel Signaling Data Registers (COSS = 0) (RCRB Indirect Registers 01H ~ 18H/21H – 38H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					Α	В	С	D
Туре	Reserved				R	R	R	R
Default					Χ	Χ	Χ	X

# A, B, C, D:

They contain the signaling of the corresponding channel.

There is a maximum 2 ms delay between the transition of the COSS[n] bit (T1/J1-041H & T1/J1-042H & T1/J1-043H) and the updating of the A, B, C, D code in the corresponding indirect registers 21H  $\sim$  38H. To avoid this 2ms delay, users can read the corresponding b3 $\sim$ 0 in the indirect registers 01H  $\sim$  18H first. If the value of these four bits are different from the previous A, B, C, D code, then the content of b3 $\sim$ 0 in the 01H  $\sim$  18H is the updated A, B, C, D code. If the conternt of the four bits is the same as the previous A, B, C, D code, then users should read the b3 $\sim$ 0 in the 21H  $\sim$  38H to get the updated A, B, C, D code.

In SF format, the C and D are the repetition of the A and B respectively.

# T1/J1 RCRB Per-Channel Configuration Registers (COSS = 0) (RCRB Indirect Registers 41H – 58H)

Bit No.	7	6	5	4	3	2	1	0	
Bit Name								DEB	
Туре	Reserved								
Default								Х	

#### DEB:

- = 0: disable signaling debounce.
- = 1: enable signaling debounce (valid only if the PCCE is logic 1). That is, the signaling is acknowledged only when 2 consecutive signaling bits of a channel are the same.

## T1/J1 FRMG Configuration (044H, 0C4H, 144H, 1C4H, 244H, 2C4H, 344H, 3C4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	MTRK	J1_CRC	J1_YEL	ESF	Reserved		GZCS[1]	GZCS[0]
Туре	R/W	R/W	R/W	R/W			R/W	R/W
Default	0	0	0	0			0	0

# MTRK:

Valid when the PCCE (b0, T1/J1-030H) is logic 1.

- = 0: normal operation.
- = 1: replace the data on all channels with the data set in the IDLE[7:0] (b7~0, T1/J1-TPLC-indirect registers-19~30H); replace the signaling on all channels with the data on the TSSIGn/MTSSIG pin or the data in the A, B, C, D (b3~0, T1/J1-TPLC-indirect registers-31~48H) according to the setting in the SIGC[1:0] (b7~6, T1/J1-TPLC-indirect registers-31~48H).

### J1 CRC:

This bit selects the T1 or J1 CRC-6 algorithm when the ESF (b4, T1/J1-044H) is 1.

- = 0: the CRC-6 algorithm meets T1 standard.
- = 1: the CRC-6 algorithm meets J1 standard.

### J1 YEL:

This bit selects the T1 or J1 Yellow alarm pattern to be transmitted.

- = 0: the Yellow alarm transition meets T1 standard.
- = 1: the Yellow alarm transition meets J1 standard.

The Yellow alarm pattern is:

- In T1 SF format: Transmit the logic 0 on the 2nd bit of each channel.
- In T1 ESF format: Transmit the 'FF00' on each FDL link.
- In J1 SF format: Transmit the logic 1 on the 12th F-bit.
- In J1 ESF format: Transmit the 'FFFF' on each FDL link.

The SF or ESF format is selected by the ESF (b4, T1/J1-044H).

# ESF:

This bit selects the SF or ESF format in the Frame Generator block.

- = 0: the SF format is selected.
- = 1: the ESF format is selected.

#### GZCS[1:0]:

These bits select the Zero Code Suppression format to be used. They are logically ORed with the ZCS[1:0] (b1~0, T1/J1-TPLC-indirect registers-01~18H).

GZCS[1:0]	Zero Code Suppression
0 0	No zero code suppression.
0 1	GTE Zero Code Suppression – Every bit 8 (or bit 7 in signaling frames) is forced to be logic one when the bits in a channel are all zeros.
10	Reserved.
11	Bell Zero Code Suppression – Every bit 7 is forced to be logic one when the bits in a channel are all zeros.

## T1/J1 FRMG Alarm Transmit (045H, 0C5H, 145H, 1C5H, 245H, 2C5H, 345H, 3C5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name							XYEL	
Туре			R/W	Reserved				
Default							0	

# XYEL:

- = 0: disable generating Yellow alarm manually.
- = 1: enable generating Yellow alarm manually.

# **T1/J1 IBCG Control** (046H, 0C6H, 146H, 1C6H, 246H, 2C6H, 346H, 3C6H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	EN	UF			CL1	CL0		
Туре	R/W	R/W		Rese	R/W	R/W		
Default	0	0					0	0

### EN:

- = 0: disable transmiting the inband loopback code.
- = 1: enable transmiting the inband loopback code.

### UF:

- = 0: the Frame Generator block operates normally. It transmits the inband loopback code in framed mode, that is, only the 192 bits are replaced with the inband loopback while the F-bit is occupied by Frame Alignment Pattern, DL or CRC-6.
- = 1: disable the Frame Generator block, that is, disable to form the SF/ESF frame. It transmits the inband loopback code in un-framed mode, that is, all 193 bits are replaced with the inband loopback code.

### CL[1:0]:

The CL[1:0] define the length of the loopback code to be transmitted, meanwhile, they define the valid code in the IBC[7:0] (b7~0, T1/J1-047H):

CL[1:0]	Loopback Code Length & valid code in the IBC[7:0]
0 0	5-bit length & the code in the IBC[7:3] is valid
0 1	6-bit or 3-bit length & the code in the IBC[7:2] is valid
10	7-bit length & the code in the IBC[7:1] is valid
1 1	8-bit or 4-bit length & the code in the IBC[7:0] is valid

### T1/J1 IBCG Loopback Code (047H, 0C7H, 147H, 1C7H, 247H, 2C7H, 347H, 3C7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	IBC7	IBC6	IBC5	IBC4	IBC3	IBC2	IBC1	IBC0
Туре	R/W							
Default	X	X	X	X	X	X	X	Х

The IBC[7:X] defines the content of the inband loopback code. 'X' is one of 3 to 0 which is depending on the length defined by the CL[1:0] ( $b1\sim0$ , T1/J1-046H). The IBC[7] is the MSB.

# T1 / J1 PMON Interrupt Enable / Status (049H, 0C9H, 149H, 1C9H, 249H, 2C9H, 349H, 3C9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name						INTE	XFER	OVR
Туре			Reserved			R/W	R	R
Default						0	0	0

#### INTE:

- = 0: disabled the interrupt on the INT pin when the counter data has been transferred into the Error Count registers.
- = 1: enabled the interrupt on the INT pin when the counter data has been transferred into the Error Count registers.

### XFER:

- = 0: indicate that the counter data has not been transferred to the Error Count registers.
- = 1: indicate that the counter data has been transferred to the Error Count registers.

This bit is clear to 0 after the bit is read.

# OVR:

- = 0: indicate that no overwritten on the Error Count registers has occurred.
- = 1: indicate that one of the Error Count registers is overwritten.

This bit is clear to 0 after the bit is read.

# Registers 04A-04FH, 0CA-0CFH, 14A-14FH, 1CA-1CFH, 24A-24FH, 2CA-2CFH, 34A-34FH, 3CA-3CFH:

The PMON Error Count registers for a single framer are updated as a group by writing to any of the PMON count registers or updated every 1 second when the AUTOUPDATE (b0, T1/J1-000H) is set. The PMON Error Count registers for eight framers are updated by writing to the Chip ID/ Global PMON Update register (T1/J1-00CH).

When the chip is reset, the contents of the PMON Error Count registers are unknown until the first latching of performance data is performed.

# T1 / J1 PMON BEE Count (LSB) (04AH, 0CAH, 14AH, 1CAH, 24AH, 2CAH, 34AH, 3CAH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	BEE7	BEE6	BEE5	BEE4	BEE3	BEE2	BEE1	BEE0
Туре	R	R	R	R	R	R	R	R
Default	Χ	Χ	Χ	Х	Χ	Χ	Χ	Χ

# T1 / J1 PMON BEE Count (MSB) (04BH, 0CBH, 14BH, 1CBH, 24BH, 2CBH, 34BH, 3CBH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name						BEE10	BEE9	BEE8
Туре		Rese	rved		R	R	R	R
Default						Х	X	X

In the ESF format, the BEE[11:0] represent the number of the CRC-6 errors, that is, the differences between the received CRC-6 and the local calculated CRC-6

In the SF format, the BEE[11:0] represent the number of the bit errors in the Frame Alignment Pattern.

This register is updated on the defined intervals.

# T1 / J1 PMON FER Count (LSB) (04CH, 0CCH, 14CH, 1CCH, 24CH, 2CCH, 34CH, 3CCH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FER7	FER6	FER5	FER4	FER3	FER2	FER1	FER0
Type	R	R	R	R	R	R	R	R
Default	Χ	Χ	Χ	X	Χ	Χ	Х	X

# T1/J1 PMON FER Count (MSB) (04DH, 0CDH, 14DH, 1CDH, 24DH, 2CDH, 34DH, 3CDH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name								FER8
Туре				Reserved				R
Default								Х

The FER[8:0] represent the number of the bit errors in the Frame Alignment Pattern.

This register is updated on the defined intervals.

# T1/J1 PMON OOF Count (04EH, 0CEH, 14EH, 1CEH, 24EH, 2CEH, 34EH, 3CEH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				OOF4	OOF3	OOF2	OOF1	OOF0
Туре		Reserved	'	R	R	R	R	R
Default				Χ	Χ	Χ	Х	Х

The OOF[4:0] represent the number of the out of SF/ESF sync events and update on the defined intervals

# T1/J1 PMON COFA Count (04FH, 0CFH, 14FH, 1CFH, 24FH, 2CFH, 34FH, 3CFH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name						COFA2	COFA1	COFA0
Туре			Reserved			R	R	R
Default							Χ	Χ

The COFA[2:0] represent the number of the changes of the Frame Alignment Pattern position and update on the defined intervals

# **T1/J1 RPLC Configuration** (050H, 0D0H, 150H, 1D0H, 250H, 2D0H, 350H, 3D0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name								PCCE
Туре				Reserved				R/W
Default								0

# PCCE:

- = 0: the per-channel functions in RPLC are disabled.
- = 1: the per-channel functions in RPLC are enabled.

## T1 / J1 RPLC mP Access Status (051H, 0D1H, 151H, 1D1H, 251H, 2D1H, 351H, 3D1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	BUSY							
Туре	R				Reserved			
Default	0							

# BUSY:

- = 0: no reading or writing operation on the indirect registers.
- = 1: an internal indirect register is being accessed, any new operation on the internal indirect register is not allowed.

This bit goes low timed to an internal high-speed clock rising edge after the operation has been completed. The operation cycle is 640ns. No more operations to the indirect registers could be done until this bit is cleared.

# T1/J1 RPLC Channel Indirect Address / Control (052H, 0D2H, 152H, 1D2H, 252H, 2D2H, 352H, 3D2H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	R/WB	A6	A5	A4	A3	A2	A1	A0
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

Writing to this register with a valid address and R/WB bit initiates an internal operation cycle to the indirect registers.

#### R/WB:

- = 0: write the data to the specified indirect register.
- = 1: read the data from the specified indirect register.

### A[6:0]:

Specify the address of the indirect registers (from 01H to 48H) for the microprocessor access.

# T1 / J1 RPLC Channel Indirect Data Buffer (053H, 0D3H, 153H, 1D3H, 253H, 2D3H, 353H, 3D3H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	D7	D6	D5	D4	D3	D2	D1	D0
Туре	R/W							
Default	0	0	0	0	0	0	0	0

This register holds the value which will be read from or written to the indirect registers (from 01H to 48H). If data are to be written to the indirect registers, the byte to be written must be written into this register before the target indirect register's address and R/WB=0 is written into the Address/Control register, initiating the access. If data are to be read from the indirect registers, only the target indirect register's address and R/WB=1 is written into the Address/Control register, initiating the request. After 640 ns, this register will contain the requested data byte.

	RPLC Indirect Registers Map					
01H ~ 18H	Per-Channel Configuration for Channel 1 ~ 24					
19H ~ 30H	Data Trunk Conditioning Code for Channel 1 ~ 24					
31H ~ 48H	Signaling Trunk Conditioning for Channel 1 ~ 24					

## T1/J1 RPLC Per-Channel Configuration Registers (RPLC Indirect Registers 01H – 18H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	INVERT	DTRKC	DMW	SIGNINV	TEST	EXTRACT	FIX	POL
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	Χ	X	Χ	X	Χ	X	X	X

#### INVERT:

This bit, together with the SIGNINV (b4, T1/J1-RPLC-indirect register - 01~18H), determines the bit inversion of the corresponding channel when output from the RSDn/MRSD pin.

INVERT	SIGNINV	Bit Inversion
0	0	No bit inversion
0	1	Invert the MSB of the corresponding channel
1	0	Invert all the bits of the corresponding channel
1	1	Invert all the bits except the MSB of the corresponding channel

### DTRKC:

- = 0: disable the data in the corresponding channel to be replaced by the data set in the DTRK[7:0] (b7~0, T1/J1-19~30H) when output on the RSDn/MRSD pin.
- = 1: enable the data in the corresponding channel to be replaced by the data set in the DTRK[7:0] (b7~0, T1/J1-19~30H) when output on the RSDn/MRSD pin.

#### DMW:

- = 0: disable the data in the corresponding channel to be replaced with a digital milliwatt pattern when output on the RSDn/MRSD pin.
- = 1: enable the data in the corresponding channel to be replaced with a digital milliwatt pattern when output on the RSDn/MRSD pin.

### SIGNINV:

Refer to the INVERT (b7, T1/J1-RPLC-indirect register - 01~18H)

#### TEST:

- = 0: disable the data in the corresponding channel to be tested by PRGD.
- = 1: enable the data in the corresponding channel to be extracted to PRGD for test (when the RXPATGEN [b2, T1/J1-00FH] is logic 0), or enable the test pattern from PRGD to replace the data in the corresponding channel for test (when the RXPATGEN [b2, T1/J1-00FH] is logic 1).

All the channels that are extracted to the PRGD are concatenated and treated as a continuous stream in which pseudo random are searched for. Similarly, all channels set to be replaced with PRGD test pattern data are concatenated replaced by the PRBS.

#### EXTRACT:

This bit is valid in Receive Clock Slave Fractional T1/J1 mode:

- = 0: RSCKn is held in its inactivated state.
- = 1: RSCKn is clocked for the corresponding channel.

# FIX:

- = 0: disable the signaling bit of the corresponding channel to be fixed with the value set by the POL when output on the RSDn/MRSD pin.
- = 1: enable the signaling bit of the corresponding channel to be fixed with the value set by the POL when output on the RSDn/MRSD pin.

#### POL:

Valid when the FIX is logic 1:

- = 0: fix the signaling bit of the corresponding channel to be logic 0.
- = 1: fix the signaling bit of the corresponding channel to be logic 1.

The priority of the RPLC operation of the corresponding channel on the RSDn/MRSD pin from high to low is:

Extract data to PRGD for test; Replace the data with the value in the DTRK[7:0]; Replace the data with the milliwatt pattern; Replace the data with the pattern generated in the PRGD; Invert the bits in the channel; Fix the signaling bit.

# T1/J1 RPLC Data Trunk Conditioning Code Byte Registers (RPLC Indirect Registers 19H – 30H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DTRK7	DTRK6	DTRK5	DTRK4	DTRK3	DTRK2	DTRK1	DTRK0
Туре	R/W							
Default	X	X	X	X	X	X	X	X

These indirect registers contain the data that will replace the data output on the RSDn/MRSD pin when the corresponding DTRKC (b6, T1/J1-RPLC-indirect registers-01~18H) is logic 1. DTRK7 is the MSB.

# T1/J1 RPLC Signaling Trunk Conditioning Byte Registers (RPLC Indirect Registers 31H – 48H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	STRKC				Α	В	С	D
Туре	R/W		Reserved			R/W	R/W	R/W
Default	X				Χ	Χ	Χ	Χ

### STRKC:

- = 0: disable the signaling of the corresponding channel to be replaced by the data set in the A, B, C, D (b3~0, T1/J1-RPLC-indirect registers-31~48H) when output on the RSSIGn/MRSSIG pin.
- = 1: enable the signaling of the corresponding channel to be replaced by the data set in the A, B, C, D (b3~0, T1/J1-RPLC-indirect registers-31~48H) when output on the RSSIGn/MRSSIG pin.

### A, B, C, D:

These bits contain the data that will replace the data output on the RSSIGn/MRSSIG pin when the corresponding STRKC (b7, T1/J1-RPLC-indirect registers-31~48H) is logic 1. They are in the least significant nibble.

# T1/J1 RHDLC #1, #2 Configuration (054H, 0D4H, 154H, 1D4H, 254H, 2D4H, 354H, 3D4H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					MEN	MM	TR	EN
Туре		Reserved				R/W	R/W	R/W
Default					0	0	0	0

Selection of the RHDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, T1/J1-00DH).

### MEN, MM:

The MEN & MM define the address matching mode:

MEN	MM	Address Matching Mode
0	X	No address matching is needed. All the HDLC data are stored in the FIFO.
1	0	The HDLC data are stored in the FIFO when the first byte is all ones or the same as the setting in the PA[7:0] (b7~0, T1/J1-
		058H) or the SA[7:0] (b7~0, T1/J1-059H).
1	1	The HDLC data are stored in the FIFO when the most significant 6 bits in the first byte are all ones or the same as the
		setting in the PA[7:2] (b7~2, T1/J1-058H) or the SA[7:2] (b7~2, T1/J1-059H).

#### TR:

- = 0: Normal operation.
- = 1: forces the RHDLC to immediately terminate the reception of the current data frame, empty the FIFO buffer, clear the interrupts and initiate a new HDLC searching.

This bit is clear to 0 after a rising and falling edge occur on the internal clock or after the register is read.

# EN:

- = 0: disabled the operation of the RHDLC block and all the FIFO buffer and interrupts are cleared.
- = 1: enabled the operation of the RHDLC block and the HDLC opening flag will be searched immediately.

If the EN is set from logic 1 to logic 0 and back to logic 1, the RHDLC will immediately terminate the reception of the current data frame, empty the FIFO buffer, clear the interrupts and initiate a new HDLC searching.

## T1/J1 RHDLC #1, #2 Interrupt Control (055H, 0D5H, 155H, 1D5H, 255H, 2D5H, 355H, 3D5H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	INTE	INTC[6]	INTC[5]	INTC[4]	INTC[3]	INTC[2]	INTC[1]	INTC[0]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

Selection of the RHDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, T1/J1-00DH).

#### INTE:

- = 0: disable the interrupt on the INT pin when there is a transition from 0 to 1 on the INTR (b0, T1/J1-056H).
- = 1: enable the interrupt on the INT pin when there is a transition from 0 to 1 on the INTR (b0, T1/J1-056H).

# INTC[6:0]:

These bits set the interrupt set point of the FIFO buffer. Exceeding the set point will cause an interrupt, and the interrupt will persist until the FIFO is empty. The set point is decimal 128 when the INTC[6:0] is all zeros.

The contents of this register should only be changed when the EN (b0, T1/J1-054H) is logic 0. This prevents any erroneous interrupt generation.

T1/J1 RHDLC #1, #2 Status (056H, 0D6H, 156H, 1D6H, 256H, 2D6H, 356H, 3D6H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	FE	OVR	COLS	PKIN	PBS[2]	PBS[1]	PBS[0]	INTR
Type	R	R	R	R	R	R	R	R
Default	Χ	Χ	X	Χ	Χ	X	X	X

Selection of the RHDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, T1/J1-00DH).

#### FE:

- = 0: the FIFO is loaded with data.
- = 1: the FIFO is empty.

#### OVR:

The overwritten condition occurs when data are written over unread data in the FIFO buffer. This bit is cleared to 0 after the register is read.

- = 0: no overwritten occurs.
- = 1: the FIFO is overwritten, and then the FIFO is reset, which cause the COLS and PKIN to be reset to logic 0.

#### COLS:

This bit reflects the HDLC link status change.

- = 0: normal operation.
- = 1: the first HDLC opening flag sequence (7E) activated the HDLC or the HDLC abort sequence (7F) deactivated the HDLC is detected.

This bit is cleared to 0 after the bit is read, or after the OVR transitions to logic 1, or after the EN is cleared.

#### PKIN:

- = 0: the last byte of a non-aborted packet is not written into the FIFO.
- = 1: the last byte of a non-aborted packet is written into the FIFO.

This bit is cleared to 0 after the bit is read, or after the OVR transitions to logic 1.

### PBS[2:01:

The PBS[2:0] indicate the status of the last byte read from the FIFO.

PBS[2:0]	Status of the Data
000	Normal data
001	A dummy byte to indicate the first HDLC opening flag sequence (7E) was detected, which means the HDLC link became active.
010	A dummy byte to indicate the HDLC abort sequence (7F) was detected, which means the HDLC link became inactive.
011	Reserved
100	The last byte of a non-aborted HDLC packet was received. The HDLC packet is in an integer number of bytes and has no FCS
	error.
101	The last byte of a non-aborted HDLC packet was received and a non-integer number of bytes is in the packet.
110	The last byte of a non-aborted HDLC packet was received. The HDLC packet is in an integer number of bytes and has FCS error.
111	The last byte of a non-aborted HDLC packet was received. The HDLC packet is in a non-integer number of bytes and has FCS
	error.

### INTR:

- = 0: no interrupt sources in the HDLC Receiver block occurred.
- = 1: any one of the interrupt sources in the HDLC Receiver block occurred. The interrupt sources in the HDLC Receiver are: 1. Receiving the first 7E opening flag sequence which activates the HDLC link; 2. A packet was received; 3. Change of link status; 4. Exceeding the set point of the FIFO which is defined in the INTC[6:0] (b6~0, T1/J1-055H); 5. Over-writting the FIFO.

This bit is cleared to 0 after the bit is read.

# T1/J1 RHDLC #1, #2 Data (057H, 0D7H, 157H, 1D7H, 257H, 2D7H, 357H, 3D7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	RD[7]	RD[6]	RD[5]	RD[4]	RD[3]	RD[2]	RD[1]	RD[0]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Χ	Χ	Χ	Χ	Х	Χ	Χ

Selection of the RHDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, T1/J1-00DH).

# RD[7:0]:

These bits represent the bytes read from the FIFO. These bits should not be accessed at a rate greater than 1/15 of the XCK rate. The RD[0] corresponds to the first bit of the serial received data from the FIFO.

# T1/J1 RHDLC #1, #2 Primary Address Match (058H, 0D8H, 158H, 1D8H, 258H, 2D8H, 358H, 3D8H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PA[7]	PA[6]	PA[5]	PA[4]	PA[3]	PA[2]	PA[1]	PA[0]
Туре	R/W							
Default	1	1	1	1	1	1	1	1

Selection of the RHDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, T1/J1-00DH).

# PA[7:0]:

These bits stipulate the primary address pattern.

PA[0] compares to the first bit of the serial data.

# T1/J1 RHDLC #1, #2 Second Address Match (059H, 0D9H, 159H, 1D9H, 259H, 2D9H, 359H, 3D9H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	SA[7]	SA[6]	SA[5]	SA[4]	SA[3]	SA[2]	SA[1]	SA[0]
Туре	R/W							
Default	1	1	1	1	1	1	1	1

Selection of the RHDLC block (#1 or #2) whose registers are visible on the microprocessor interface is done via the RHDLCSEL[1:0] (b7~6, T1/J1-00DH).

### SA[7:0]:

These bits stipulate the secondary address pattern.

SA[0] compares to the first bit of the serial data.

T1 / J1 TBOM Code (05DH, 0DDH, 15DH, 1DDH, 25DH, 2DDH, 35DH, 3DDH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name			BOC[5]	BOC[4]	BOC[3]	BOC[2]	BOC[1]	BOC[0]
Туре	Reserved		R/W	R/W	R/W	R/W	R/W	R/W
Default			1	1	1	1	1	1

When the BOC[5:0] are written with any 6-bit code other than the '111111', the code will be transmitted as the Bit Oriented Message (BOM), overwriting any HDLC packets currently being transmitted. The BOM pattern is '111111110BOC[0]BOC[1]BOC[2]BOC[3]BOC[4]BOC[5]0', that is, the BOC[0] is transmitted first.

### T1/J1 PRGD Control (060H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PDR[1]	PDR[0]		PS	TINV	RINV	AUTOSYNC	MANSYNC
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	1	0

# PDR[1:0]:

The PDR[1:0] define the function of the four PRGD Pattern Detector registers:

PDR[1:0]	PRGD Pattern Detector Registers (#1 ~ #4)			
00, 01	Pattern Receive			
10	Error Count			
11	Bit Count			
(The #1 is the LSB, while the #4 is the MSB.)				

# PS:

- = 0: a pseudo-random pattern is generated/detected by the PRGD.
- = 1: a repetitive pattern is generated/detected by the PRGD.

This bit should be set first of all the PRGD registers.

# TINV:

- = 0: disable inverting the generated pattern before being transmitted.
- = 1: enable inverting the generated pattern before being transmitted.

### RINV:

- = 0: disable inverting the received pattern before being processed.
- = 1: enable inverting the received pattern before being processed.

### AUTOSYNC:

- = 0: disable automatic re-search for the sync of the pattern after the pattern is out of synchronization.
- = 1: enable automatic re-search for the sync of the pattern after the pattern is out of synchronization.

# MANSYNC:

Trigger on the rising edge. A transition from logic 0 to logic 1 on this bit manually initiates a re-search for the sync of a pattern.

Each time the value of the PRGD registers is changed or the detector data source changes, a manual sync operation is recommended to ensure that the detector works correctly.

### T1 / J1 PRGD Interrupt Enable / Status (061H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	SYNCE	BEE	XFERE	SYNCV	SYNCI	BEI	XFERI	OVR
Туре	R/W	R/W	R/W	R	R	R	R	R
Default	0	0	0	Χ	Χ	Χ	X	Χ

### SYNCE:

- = 0: disable the interrupt on the  $\overline{\text{INT}}$  pin when the SYNCI is logic one.
- = 1: enable the interrupt on the INT pin when the SYNCI is logic one.

### BEE:

- = 0: disable the interrupt on the INT pin when at least one bit error has been detected in the received pattern.
- = 1: enable the interrupt on the INT pin when at least one bit error has been detected in the received pattern.

### XFERE:

- = 0: disable the interrupt on the INT pin when the the data in the PRGD pattern detector register is updated.
- = 1: enable the interrupt on the INT pin when the the data in the PRGD pattern detector register is updated.

#### SYNCV:

- = 0: the pattern is out of sync (the pattern detector has detected 10 or more bit errors in a fixed 48-bit window).
- = 1: the pattern is in sync (the pattern detector has observed at least 48 consecutive error-free bit periods).

### SYNCI:

- = 0: there is no transition on the SYNCV.
- = 1: there is a transition (from 0 to 1 or from 1 to 0) on the SYNCV.

This bit is cleared to 0 after the bit is read.

### BEI:

- = 0: no bit error is detected in the received pattern.
- = 1: at least one bit error has been detected in the received pattern.

This bit is cleared to 0 after the bit is read.

### XFERI:

- = 0: the data in the PRGD pattern detector register is not updated.
- = 1: the data in the PRGD pattern detector register is updated.

This bit is cleared to 0 after the bit is read.

# OVR:

- = 0: the PRGD pattern detector register is not overwritten.
- = 1: the PRGD pattern detector register is overwritten.

This bit is cleared to 0 after the bit is read.

## T1/J1 PRGD Shift Register Length (062H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				PL[4]	PL[3]	PL[2]	PL[1]	PL[0]
Type	Reserved			R/W	R <i>₩</i>	R <i>₩</i>	R <i>W</i>	R <i>W</i>
Default				0	0	0	0	0

These bits determine the length of the valid data in the PRGD pattern insertion register. The length is equal to the value of PL[4:0] + 1.

# **T1 / J1 PRGD Tap** (063H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name				PT[4]	PT[3]	PT[2]	PT[1]	PT[0]
Туре	Reserved			R/W	R/W	R/W	R/W	R/W
Default				0	0	0	0	0

These bits determine the feedback tap position of the generated pseudo random pattern before it is transmitted. The feedback tap position is equal to the value of PT[4:0] + 1. In application, the PT is always less than the PL.

# T1 / J1 PRGD Error Insertion (064H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name					EVENT	EIR[2]	EIR[1]	EIR[0]
Туре	Reserved			R/W	R/W	R/W	R/W	
Default					0	0	0	0

### **EVENT:**

A single bit error is generated when the state of this bit is changed from 0 to 1. To insert another bit error, this bit must be cleared to 0, and then set from 0 to 1 again.

# EIR[2:0]:

The EIR[2:0] bits determine the bit error rate that will be inserted in the PRGD test pattern. If the bit error rate is changed from one non-zero value to another non-zero value, it is recommended to set the EIR[2:0] to '000' first, then set the EIR[2:0] to the desired value.

EIR[2:0]	Bit error rate
0 0 0	No error inserted
0 0 1	No error inserted
010	10 <sup>-2</sup>
0 1 1	10 <sup>-3</sup>
100	10-4
101	10⁻⁵
110	10-6
111	10 <sup>-7</sup>

### T1 / J1 PRGD Pattern Insertion #1 (068H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[7]	PI[6]	PI[5]	PI[4]	PI[3]	PI[2]	PI[1]	PI[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

### T1 / J1 PRGD Pattern Insertion #2 (069H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[15]	PI[14]	PI[13]	PI[12]	PI[11]	PI[10]	PI[9]	PI[8]
Туре	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default	0	0	0	0	0	0	0	0

#### T1 / J1 PRGD Pattern Insertion #3 (06AH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[23]	PI[22]	PI[21]	PI[20]	PI[19]	PI[18]	PI[17]	PI[16]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

### T1 / J1 PRGD Pattern Insertion #4 (06BH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PI[31]	PI[30]	PI[29]	PI[28]	PI[27]	PI[26]	PI[25]	PI[24]
Type	R/W							
Default	0	0	0	0	0	0	0	0

When a repetitive pattern is selected to transmit, the data in these registers are the repetitive pattern.

When a pseudo random pattern is selected to transmit, the data in these registers should be set to FFFFFFFH. They are the initial value for the pseudo random pattern.

Writing to PI[31:24] updates the PRGD configuration.

When a repetitive pattern is transmitted, the PI[31] is transmitted first, followed by the remaining bits in sequence down to PI[0]. The length of the valid data in these four registers is determined by the PL[4:0]. When the length is less than 31, the bits in higher PI are not used.

#### T1 / J1 PRGD Pattern Detector #1 (06CH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[7]	PD[6]	PD[5]	PD[4]	PD[3]	PD[2]	PD[1]	PD[0]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Χ	Χ	Х	Χ	Х	Χ	Х

#### T1 / J1 PRGD Pattern Detector #2 (06DH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[15]	PD[14]	PD[13]	PD[12]	PD[11]	PD[10]	PD[9]	PD[8]
Туре	R	R	R	R	R	R	R	R
Default	Χ	Χ	Χ	X	Х	Х	X	Х

#### T1 / J1 PRGD Pattern Detector #3 (06EH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[23]	PD[22]	PD[21]	PD[20]	PD[19]	PD[18]	PD[17]	PD[16]
Type	R	R	R	R	R	R	R	R
Default	X	X	X	X	X	X	X	X

#### T1 / J1 PRGD Pattern Detector #4 (06FH)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	PD[31]	PD[30]	PD[29]	PD[28]	PD[27]	PD[26]	PD[25]	PD[24]
Туре	R	R	R	R	R	R	R	R
Default	Х	X	X	X	X	X	X	Х

When the PDR[1:0] (b7~6, T1/J1-060H) are set to 00 or 01, the four PRGD pattern detector registers are configured as Pattern Receive registers. They reflect the content of the received pattern.

When the PDR[1:0] (b7~6, T1/J1-060H) are set to 10, the four PRGD pattern detector registers are configured as Error Counter registers. The value in these registers represent the number of bit errors. The bit errors are not accumulated when the pattern is out of sync.

When the PDR[1:0] (b7~6, T1/J1-060H) are set to 11, the four PRGD pattern detector registers are configured as Bit Counter registers. The value in these registers represent the total received bit number.

These registers are updated each second automatically, or by writing to any of these four registers, or to the Revision / Chip ID / Global PMON register.

#### T1/J1 RHDLC Receive Data Link 2 Control (TXCISEL = 0) (070H, 0F0H, 170H, 1F0H, 270H, 2F0H, 370H, 3F0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_EVEN	DL2_ODD		DL2_TS[4]	DL2_TS[3]	DL2_TS[2]	DL2_TS[1]	DL2_TS[0]
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	0	0

When the TXCISEL (b3, T1/J1-00DH) is 0, this register is used for the Receive HDLC #2.

#### DL2 EVEN:

- = 0: the data is not extracted from the even frames.
- = 1: the data is extracted from the even frames.

#### DL2 ODD:

- = 0: the data is not extracted from the odd frames.
- = 1: the data is extracted from the odd frames.

### DL2\_TS[4:0]:

These bits binary define one channel of even and/or odd frames to extract the data from. They are invalid when the DL2\_EVEN and the DL2\_ODD are both logic 0.

#### T1/J1 RHDLC Data Link 2 Bit Select (TXCISEL = 0) (071H, 0F1H, 171H, 1F1H, 271H, 2F1H, 371H, 3F1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_BIT[7]	DL2_BIT[6]	DL2_BIT[5]	DL2_BIT[4]	DL2_BIT[3]	DL2_BIT[2]	DL2_BIT[1]	DL2_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, T1/J1-00DH) is 0, this register is used for the Receive HDLC #2.

#### DL2\_BITn:

- = 0: the data is not extracted from the corresponding bit.
- = 1: the data is extracted from the corresponding bit of the assigned channel.

These bits are invalid when the DL2\_EVEN and the DL2\_ODD are both logic 0.

The DL1\_BIT[7] corresponds to the first bit (MSB) of the selected channel.

## T1 / J1 THDLC Transmit Data Link 2 Control (TXCISEL = 1) (070H, 0F0H, 170H, 1F0H, 270H, 2F0H, 370H, 3F0H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_EVEN	DL2_ODD		DL2_TS[4]	DL2_TS[3]	DL2_TS[2]	DL2_TS[1]	DL2_TS[0]
Туре	R/W	R/W	Reserved	R/W	R/W	R/W	R/W	R/W
Default	0	0		0	0	0	0	0

When the TXCISEL (b3, T1/J1-00DH) is 1, this register is used for the Transmit HDLC #2.

### DL2 EVEN:

- = 0: the data is not inserted to the even frames.
- = 1: the data is inserted to the even frames.

#### DL2 ODD:

- = 0: the data is not inserted to the odd frames.
- = 1: the data is inserted to the odd frames.

### DL2\_TS[4:0]:

These bits binary define one channel of even and/or odd frames to insert the data to. They are invalid when the DL2\_EVEN and the DL2\_ODD are both logic 0.

## T1 / J1 THDLC Data Link 2 Bit Select (TXCISEL = 1) (071H, 0F1H, 171H, 1F1H, 271H, 2F1H, 371H, 3F1H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name	DL2_BIT[7]	DL2_BIT[6]	DL2_BIT[5]	DL2_BIT[4]	DL2_BIT[3]	DL2_BIT[2]	DL2_BIT[1]	DL2_BIT[0]
Туре	R/W							
Default	0	0	0	0	0	0	0	0

When the TXCISEL (b3, T1/J1-00DH) is 1, this register is used for the Transmit HDLC #2.

#### DL2 BITn:

- = 0: the data is not inserted to the corresponding bit.
- = 1: the data is inserted to the corresponding bit of the assigned channel.

These bits are invalid when the DL2 EVEN and the DL2 ODD are both logic 0.

The DL1\_BIT[0] corresponds to the first bit (MSB) of the selected channel.

T1 / J1 RESI Timeslot Offset (077H, 0F7H, 177H, 1F7H, 277H, 2F7H, 377H, 3F7H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		TSOFF[6]	TSOFF[5]	TSOFF[4]	TSOFF[3]	TSOFF[2]	TSOFF[1]	TSOFF[0]
Туре	Reserved	R/W						
Default		0	0	0	0	0	0	0

In the Receive Clock Slave mode, when the data rate on the system side is 2.048M bit/s (the RSCCK2M [b4, T1/J1-001H] and RSCCK8M [b3, T1/J1-001H] are set to '10'), these bits determine the channel offset between the RSCFS and the start of the corresponding frame on the RSDn (and RSSIGn).

In the Receive Multiplexed mode, these bits determine the channel offset between the MRSCFS and the start of the corresponding frame on the MRSD and MRSSIG.

In the Receive Clock Slave mode, when the data rate on the system side is 1.544M bit/s, and in Receive Clock Master mode, the channel offset is disabled. Thus, the TSOFF must be set to 0.

They define a binary number. The offset can be set from 0 to 127 channels.

#### T1/J1 RESI Timeslot Offset (078H, 0F8H, 178H, 1F8H, 278H, 2F8H, 378H, 3F8H)

Bit No.	7	6	5	4	3	2	1	0
Bit Name		FPINV	RSD_RSCFS_EDGE	CMS	BOFF_EN	BOFF[2]	BOFF[1]	BOFF[0]
Туре	Reserved	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Default		0	0	0	0	0	0	0

#### FPINV:

- = 0: The receive framing pulse RSCFS and RSFSn/MRSFS are active high.
- = 1: The receive framing pulse RSCFS and RSFSn/MRSFS are active low.

When the bit indicates the RSCFS and MRSFS polarity, the bits of all eight framers must have the same value.

#### RSD RSCFS EDGE:

Valid when the CMS (b4, T1/J1-078H) is logic 1 and the setting in the RSCFSFALL (b1, T1/J1-003H) and that in the RSCCKRISE (b0, T1/J1-003H) are equal.

- = 0: select the second active edge of the RSCCK to update the signal on the RSDn, RSSIGn and RSFSn pins, or select the first active edge of the MRSCCK to update the signal on the MRSD, MRSSIG and MRSFS pins.
- = 1: select the first active edge of the RSCCK to update the signal on the RSDn, RSSIGn and RSFSn pins, or select the second active edge of the MRSCCK to update the signal on the MRSD, MRSSIG and MRSFS pins.

(The signal on the RSCFS/MRSCFS pin is always sampled on the first active edge.)

In Receive Multiplexed mode, the RSD\_RSCFS\_EDGE in all eight framers should be set to the same value.

#### CMS:

- = 0: the bit rate of RSCCK/MRSCCK is the same as the bit rate of the backplane.
- = 1: the bit rate of RSCCK/MRSCCK is twice the bit rate of the backplane.

The CMS in all eight framers should be set to the same value.

#### BOFF\_EN:

Valid when the CMS (b4, T1/J1-078H) is 0.

- = 0: disable the bit offset.
- = 1: enable the bit offset.

#### BOFF[2:0]:

Valid when the CMS (b4, T1/J1-078H) is 0 and the BOFF\_EN is logic 1.

In Receive Clock Slave mode, when the data rate in the system side is 2.048M bit/s (the RSCCK2M [b4, T1/J1-001H] and RSCCK8M [b3, T1/J1-001H] are set to '10'), these bits determine the bit offset between the RSCFS and the start of the corresponding frame on the RSDn (and RSSIGn).

In Receive Multiplexed mode, these bits determine the bit offset between the MRSCFS and the start of the corresponding frame on the MRSD and MRSSIG.

In Receive Clock Slave mode, when the data rate in the system side is 1.544M bit/s, and in Receive Clock Master mode, the bit offset is disabled. These bits define a binary number. Programming of the Bit Offsets is consistent with the convention established by the Concentration Highway Interface (CHI) specification. Refer to the Functional Description for details.

# 6 IEEE STD 1149.1 JTAG TEST ACCESS PORT

The IDT82V2108 supports the digital Boundary Scan Specification as described in the IEEE 1149.1 standards.

The boundary scan architecture consists of data and instruction registers plus a Test Access Port (TAP) controller. Control of the TAP is achieved through signals applied to the Test Mode Select (TMS) and

Test Clock (TCK) input pins. Data are shifted into the registers via the Test Data Input (TDI) pin, and shifted out of the registers via the Test Data Output (TDO) pin. Both TDI and TDO are clocked at a rate determined by TCK.

The JTAG boundary scan registers includes BSR (Boundary Scan Register), IDR (Device Identification Register), BR (Bypass Register) and IR (Instruction Register). These will be described in the following pages. *Refer to Figure -* 83 for architecture.

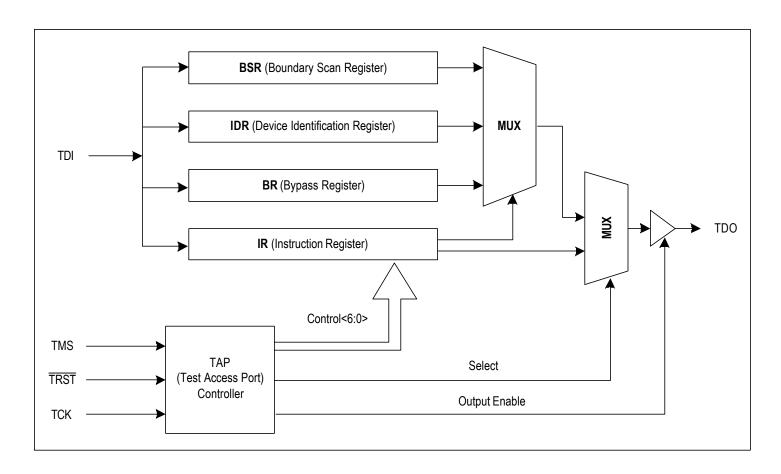


Figure - 83. JTAG Architecture

Table - 64. IR Code

IR CODE	INSTRUCTION	COMMENTS
000	EXTEST	The external test instruction allows testing of the interconnection to other devices. When the current instruction is the EXTEST instruction, the boundary scan register is placed between TDI and TDO. The signal on the input pins can be sampled by loading the boundary scan register using the Capture-DR state. The sampled values can then be viewed by shifting the boundary scan register using the Shift-DR state. The signal on the output pins can be controlled by loading patterns shifted in through input TDI into the boundary scan register using
		the Update-DR state.
010	SAMPLE / PRELOAD	The SAMPLE/PRELOAD instruction is used to allow scanning of the boundary-scan register without causing interference to the normal operation of the on-chip system logic. Data received at system input pins is supplied without modification to the on-chip system logic; data from the on-chip system logic is driven without modification through the system output pins. SAMPLE allows a snapshot to be taken of the data flowing from the system pins to the on-chip system logic or vice versa, without interfering with the normal operation of the assembled board. PRELOAD allows an initial data pattern to be placed at the latched parallel outputs of boundary-scan register cells prior to selection of another boundary-scan test operation.
001	IDCODE	The identification instruction is used to connect the identification register between TDI and TDO. The device's identification code can then be shifted out using the Shift-DR state.
111	BYPASS	The BYPASS instruction shifts data from input TDI to output TDO with one TCK clock period delay. The instruction is used to bypass the device.
100	CLAMP	This instruction allows the state of the signals driven from device pins to be determined from the boundary-scan register while the bypass register is selected as the serial path between TDI and TDO. The signals driven from the device pins will not change while the CLAMP instruction is selected.
101	HIGHZ	Use of the HIGHZ instruction places the device in a state in which all of its system logic outputs are placed in an inactive drive state (e.g., high impedance). In this state, and in-circuit test system may drive signals onto the connections normally driven by a device output without incurring the risk of damage to the device.
011		(for manufactory test)

# 6.1 JTAG INSTRUCTIONS AND INSTRUCTION REGISTER (IR)

The IR (Instruction Register) with instruction decode block is used to select the test to be executed or the data register to be accessed or both.

The instructions are shifted in LSB first to this 3-bit register. See Table - 64 for details of the codes and the instructions related.

#### 6.2 JTAG DATA REGISTER

#### 6.2.1 DEVICE IDENTIFICATION REGISTER (IDR)

The IDR can be set to define the Vision, the Part Number, the Manufacturer Identity and a fixed bit. The IDR is 32 bits long and is partitioned as in *Table - 65*. Data from the IDR is shifted out to the TDO LSB first.

Table - 65. IDR

BIT No.	COMMENTS			
0	Set to "1"			
1~11	Manufacturer Identity (033H)			
12~27	Part Number (04D0H)			
28~31	Version (2H)			

#### 6.2.2 BYPASS REGISTER (BYR)

The BYR consists of a single bit. It can provide a serial path between the TDI input and TDO output, bypassing the BYR to reduce test access times.

#### 6.2.3 BOUNDARY SCAN REGISTER (BSR)

The scan chain uses 3 types of cells:

- Input / Output cells: When used as input, the cells are able to sample and control the state of an external signal during BS tests. When used as output, the cells are able to control the state of an external signal during BS tests.
- In/Out or Tri-state output cells: When configured as input, the cells are able to sample and control the state of an external signal. When configured as output, the cells are able to control the state of an external signal.
- Control cell: This cell provides a signal for direction control of bidirectional or tri-state output pins during BS tests.

The Boundary Scan sequence and the I/O Pad Cell type are illustrated in Table - 66:

#### 6.3 TEST ACCESS PORT CONTROLLER

The TAP controller is a 16-state synchronous state machine. *Figure - 84* shows its state diagram. A description of each state follows. Note that the figure contains two main branches to access either the data or instruction registers. The value shown next to each state transition in this figure states the value present at TMS at each rising edge of TCK. *Please refer to Table - 67 for details of the state description.* 

Table - 66. Boundary Scan Sequence and the I/O Pad Cell Type

Pin_name	Cell Type	BS*
LRD[1]	Input	128
LRCK[1]	Input	127
LRD[2]	Input	126
LRCK[2]	Input	125
LRD[3]	Input	124
LRCK[3]	Input	123
LRD[4]	Input	122
LRCK[4]	Input	121
LTD[1]	Output	120
LTCK[1]	Output	119
LTD[2]	Output	118
LTCK[2]	Output	117
LTD[3]	Output	116
LTCK[3]	Output	115
LTD[4]	Output	114
LTCK[4]	Output	113
LTD[5]	Output	112
LTCK[5]	Output	111
LTD[6]	Output	110
LTCK[6]	Output	109
LTD[7]	Output	108

Pin_name	Cell Type	BS *
LTCK[7]	Output	107
LTD[8]	Output	106
LTCK[8]	Output	105
LRD[5]	Input	104
LRCK[5]	Input	103
LRD[6]	Input	102
LRCK[6]	Input	101
LRD[7]	Input	100
LRCK[7]	Input	99
LRD[8]	Input	98
LRCK[8]	Input	97
RST	Input	96
ĪNT	Output	95
D[7:0]_EN	Control	94
D[0]	In/Out	93
D[1]	In/Out	92
D[2]	In/Out	91
D[3]	In/Out	90
D[4]	In/Out	89
D[5]	In/Out	88
D[6]	In/Out	87

Pin_name	Cell Type	BS *
D[7]	In/Out	86
ALE	Input	85
A[0]	Input	84
A[1]	Input	83
A[2]	Input	82
A[3]	Input	81
A[4]	Input	80
A[5]	Input	79
A[6]	Input	78
A[7]	Input	77
A[8]	Input	76
A[9]	Input	75
A[10]	Input	74
CS	Input	73
WR	Input	72
RD	Input	71
RSFS[8]	Output	70
RSCK[8]/RSSIG[8]	Tri-state Output	69
RSCK[8]/RSSIG[8]_EN	Control	68
RSD[8]	Tri-state Output	67
RSD[8]_EN	Control	66
RSFS[7]	Output	65
RSCK[7]/RSSIG[7]	Tri-state Output	64
RSCK[7]/RSSIG[7]_EN	Control	63
RSD[7]	Tri-state Output	62
RSD[7]_EN	Control	61
RSFS[6]	Output	60
RSCK[6]/RSSIG[6]	Tri-state Output	59
RSCK[6]/RSSIG[6]_EN	Control	58
RSD[6]	Tri-state Output	57
RSD[6]_EN	Control	56
RSFS[5]	Output	55
RSCK[5]/RSSIG[5]	Tri-state Output	54
RSCK[5]/RSSIG[5]_EN	Control	53
RSD[5]	Tri-state Output	52
RSD[5]_EN	Control	51
RSFS[4]	Output	50
RSCK[4]/RSSIG[4]	Tri-state Output	49
RSCK[4]/RSSIG[4]_EN	Control	48
RSD[4]	Tri-state Output	47
RSD[4]_EN	Control	46
RSFS[3]	Output	45
RSCK[3]/RSSIG[3]	Tri-state Output	44
RSCK[3]/RSSIG[3]_EN	Control	43
RSD[3]	Tri-state Output	42
RSD[3]_EN	Control	41
RSFS[2]/MRSFS[2]	Output	40
RSCK[2]/RSSIG[2]/MRSSIG[2]	Tri-state Output	39
RSCK[2]/RSSIG[2]/MRSSIG[2]_EN	Control	38

Pin_name	Cell Type	BS *
RSD[2]/MRSD[2]	Tri-state Output	37
RSD[2]/MRSD[2]_EN	Control	36
RSFS[1]/MRSFS[1]	Output	35
RSCK[1]/RSSIG[1]/MRSSIG[1]	Tri-state Output	34
RSCK[1]/RSSIG[1]/MRSSIG[1]_EN	Control	33
RSD[1]/MRSD[1]	Tri-state Output	32
RSD[1]/MRSD[1]_EN	Control	31
TSSIG[8]/TSFS[8]	In/Out	30
TSSIG[8]/TSFS[8]_EN	Control	29
TSD[8]	Input	28
TSSIG[7]/TSFS[7]	In/Out	27
TSSIG[7]/TSFS[7]_EN	Control	26
TSD[7]	Input	25
TSSIG[6]/TSFS[6]	In/Out	24
TSSIG[6]/TSFS[6]_EN	Control	23
TSD[6]	Input	22
TSSIG[5]/TSFS[5]	In/Out	21
TSSIG[5]/TSFS[5]_EN	Control	20
TSD[5]	Input	19
TSSIG[4]/TSFS[4]	In/Out	18
TSSIG[4]/TSFS[4]_EN	Control	17
TSD[4]	Input	16
TSSIG[3]/TSFS[3]	In/Out	15
TSSIG[3]/TSFS[3]_EN	Control	14
TSD[3]	Input	13
TSFS[2]/TSSIG[2]/MTSSIG[2]	In/Out	12
TSFS[2]/TSSIG[2]/MTSSIG[2]_EN	Control	11
TSD[2]/MTSD[2]	Input	10
TSFS[1]/TSSIG[1]/MTSSIG[1]	In/Out	9
TSFS[1]/TSSIG[1]/MTSSIG[1]_EN	Control	8
TSD[1]/MTSD[1]	Input	7
XCK	Input	6
RSCFS/MRSCFS	Input	5
RSCCK/MRSCCK	Input	4
TSCFS/MTSCFS	Input	3
TSCCKB/MTSCCKB	Input	2
TSCCKA	Input	1

Note: \* BS means Boundary Scan Sequence

Table - 67. TAP Controller State Description

STATE	DESCRIPTION
Test Logic Reset	In this state, the test logic is disabled to continue normal operation of the device. During initialization, the device initializes the instruction register with the IDCODE instruction.  Regardless of the original state of the controller, the controller enters the Test-Logic-Reset state when the TMS input is held high for at least 5 rising edges of TCK. The controller remains in this state while TMS is high.
Run-Test/Idle	This is a controller state between scan operations. Once in this state, the controller remains in the state as long as TMS is held low. The instruction register and all test data registers retain their previous state. When TMS is high and a rising edge is applied to TCK, the controller moves to the Select-DR state.
Select-DR-Scan	This is a temporary controller state and the instruction does not change in this state. The test data register selected by the current instruction retains its previous state. If TMS is held low and a rising edge is applied to TCK when in this state, the controller moves into the Capture-DR state and a scan sequence for the selected test data register is initiated. If TMS is held high and a rising edge applied to TCK, the controller moves to the Select-IR-Scan state.
Capture-DR	In this state, the Boundary Scan Register captures input pin data if the current instruction is EXTEST or SAMPLE/PRELOAD. The instruction does not change in this state. The other test data registers, which do not have parallel input, are not changed. When the TAP controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-DR state if TMS is high or the Shift-DR state if TMS is low.
Shift-DR	In this controller state, the test data register connected between TDI and TDO as a result of the current instruction shifts data on stage toward its serial output on each rising edge of TCK. The instruction does not change in this state. When the TAP controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-DR state if TMS is high or remains in the Shift-DR state if TMS is low.
Exit1-DR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-DR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Pause-DR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Pause-DR	The pause state allows the test controller to temporarily halt the shifting of data through the test data register in the serial path between TDI and TDO. For example, this state could be used to allow the tester to reload its pin memory from disk during application of a long test sequence. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state. The controller remains in this state as long as TMS is low. When TMS goes high and a rising edge is applied to TCK, the controller moves to the Exit2-DR state.
Exit2-DR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-DR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Shift-DR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Update-DR	The Boundary Scan Register is provided with a latched parallel output to prevent changes while data is shifted in response to the EXTEST and SAMPLE/PRELOAD instructions. When the TAP controller is in this state and the Boundary Scan Register is selected, data is latched into the parallel output of this register from the shift-register path on the falling edge of TCK. The data held at the latched parallel output changes only in this state. All shift-register stages in the test data register selected by the current instruction retain their previous value and the instruction does not change during this state.
Select-IR-Scan	This is a temporary controller state. The test data register selected by the current instruction retains its previous state. If TMS is held low and a rising edge is applied to TCK when in this state, the controller moves into the Capture-IR state, and a scan sequence for the instruction register is initiated. If TMS is held high and a rising edge is applied to TCK, the controller moves to the Test-Logic-Reset state. The instruction does not change during this state.
Capture-IR	In this controller state, the shift register contained in the instruction register loads a fixed value of '100' on the rising edge of TCK. This supports fault-isolation of the board-level serial test data path. Data registers selected by the current instruction retain their value and the instruction does not change during this state. When the controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-IR state if TMS is held high, or the Shift-IR state if TMS is held low.

STATE	DESCRIPTION
Shift-IR	In this state, the shift register contained in the instruction register is connected between TDI and TDO and shifts data one stage towards its serial output on each rising edge of TCK. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state. When the controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-IR state if TMS is held high, or remains in the Shift-IR state if TMS is held low.
Exit1-IR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-IR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Pause-IR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Pause-IR	The pause state allows the test controller to temporarily halt the shifting of data through the instruction register. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state. The controller remains in this state as long as TMS is low. When TMS goes high and a rising edge is applied to TCK, the controller moves to the Exit2-IR state.
Exit2-IR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-IR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Shift-IR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Update-IR	The instruction shifted into the instruction register is latched into the parallel output from the shift-register path on the falling edge of TCK. When the new instruction has been latched, it becomes the current instruction. The test data registers selected by the current instruction retain their previous value.

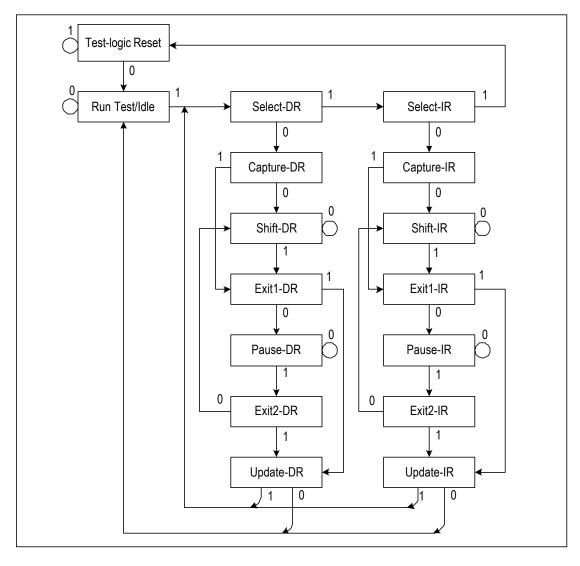


Figure - 84. JTAG State Diagram

# 7 PHYSICAL AND ELECTRICAL SPECIFICATIONS

# 7.1 ABSOLUTE MAXIMUM RATINGS

	Min	Max	
Storage temperature	-65°C	+150°C	
Voltage on VDD w.r.t. GND	-0.3V	4.6V	
Voltage on BIAS w.r.t. GND	VDD-0.3V	5.5V	
Voltage on any pin	-0.3V	BIAS+0.3V	
Maximum lead temperature		230°C	during TBC seconds
ESD Performance (HBM)	2000V		
ESD Performance (CDM)	1000V		
Latch-up current on any pin	100ma		
Maximum DC current on any pin			
Maximum lead temperature			
Maximum junction temperature			

# 7.2 OPERATING CONDITIONS

@ TA = -40 to +85  $^{\circ}$ C, VDD = 3.3V  $\pm$  10%, VDD $\leq$ BIAS $\leq$ 5.5V

# 7.3 D.C. CHARACTERISTICS

Parameter	Description	Min	Тур	Max	Unit	Test Conditions
VDDC, VDDIO	Core Power Supply	2.97	3.3	3.63	V	
BIAS	5V Tolerant Bias	VDD	5.0	5.5	V	
IBIAS	Current into 5V Bias	0	1	3	mΑ	VBIAS=5.5V
VIL	Input Low Voltage			0.8	V	
VIH	Input High Voltage	2.0		BIAS	V	
VOL	Output Low Voltage			0.4	V	VDD=min, IOL= 2mA, 3mA
VOH	Output High Voltage	2.4			V	VDD=min, IOL= 3mA, 3mA
VT+	Reset Input High Voltage	1.50	1.75	2.0	V	
VT-	Reset Input Low Voltage	0.83	1.10	1.33	V	
VTH	Reset Input Hysteresis Voltage	0.17	0.65	1.17	V	
IILPU	Input Low Current	-70	-330	-450	uA	VIL=GND
IIL	Input Low Current	-1	0	+1	uA	VIL=GND
IIH	Input High Current	-10	0	+10	uA	VIH=VBIAS
I <sub>DDOP1</sub>	Operating current		160		mΑ	E1 mode, XCK=49.152MHz,
DBOI 1						TSCCKB=2.048MHz, output unloaded, Vdd=3.63V.
I <sub>DDOP2</sub>	Operating current		170		mΑ	E1 mode, XCK=49.152MHz,
55012						TSCCKB=8.192MHz, output unloaded. Vdd=3.63V.
DDOP3	Operating current		120		mΑ	T1 mode, XCK=37.056MHz,
						TSCCKB=1.544MHz, output unloaded. Vdd=3.63V.

### 7.4 CLOCK AND RESET TIMING

The RST must be asserted for a minimum of 100 ns after XCLK is stable to ensure that the chip is completely reset.

#### 7.4.1 CLOCK PARAMETERS E1 CONFIGURATION

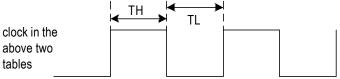
	Min. Frequency (MHz)	Max. Frequency (MHz)	ppm	T <sub>L min (ns)</sub> *	T *
XCK	49.	152	±50		
LRCK	2.0	2.1		100	100
MRSCCK	8.0	8.4		40	40
TSCCKA	2.0	2.1		100	100
RSCCK	2.0	2.1		140	140
TSCCKB	2.0	2.1		140	140

#### 7.4.2 CLOCK PARAMETERS T1/J1 CONFIGURATION

	Min. Frequency (MHz)	Max. Frequency (MHz)	ppm	T <sub>L min</sub> 1	T <sub>H min</sub> 1
XCK <sup>3</sup>	37.05	6	±32 <sup>2</sup>		
LRCK	1.534	1.545		100	100
MRSCCK	8.00	8.40		40	40
TSCCKA	1.534	1.545		100	100
RSCCK	1.50	2.0584		140	140
TSCCKB	1.50	2.10 <sup>4</sup>		140	140

#### NOTE:

1. The T  $_{\rm L}$  and T  $_{\rm H}$  are defined in the figure.



- 2. An XCK input accuracy of ±100 ppm is only acceptable if an accurate line rate reference is provided. If TJAT is left to free-run without a reference, or referenced to a derivative of XCK, then XCK accuracy must be ±32 ppm. The accuracy of XCK affect the performance of TJAT/RJAT.
- 3. An XCK input accuracy of  $\pm 100$  ppm is only acceptable if an accurate line rate reference is provided. If TJAT is left to free-run without a reference, or referenced to a derivative of XCLK, then XCLK accuracy must be  $\pm 32$  ppm. The accuracy of XCLK affect the performance of TJAT/DJAT.
- 4. For T1 mode with 2.048Mb/s back-plane data rate only.

#### 7.5 MICROPROCESSOR READ ACCESS TIMING

Symbol	Parameter		Min	Max	Units
tS <sub>AR</sub>	Address to Valid Read Set-up Time		0		ns
tH <sub>AR</sub>	Address to Valid Read Hold Time		0		ns
tS <sub>ALR</sub>	Address to Latch Set-up Time		5		ns
tH <sub>ALR</sub>	Address to Latch Hold Time		5		ns
tV <sub>L</sub>	Valid Latch Pulse Width		10		ns
tS <sub>LR</sub>	Latch to Read Set-up Time		0		ns
tH <sub>LR</sub>	Latch to Read Hold		0		ns
tZ <sub>RD</sub>	Valid Read Negated to Output Tri-state	Valid Read Negated to Output Tri-state		20	ns
tP <sub>RD</sub>	Valid Read to Valid Data Propagation Delay	E1		130	ns
		T1		160	ns
tZ <sub>INTH</sub>	Valid Read Negated to INT Inactive	E1		145	ns
		T1		185	ns
tV <sub>RD</sub>	Valid Read Width	E1	120		ns
		T1	150		ns
t <sub>w2R</sub>	Valid interval from last write to next read	E1	120		ns
		T1	150		ns

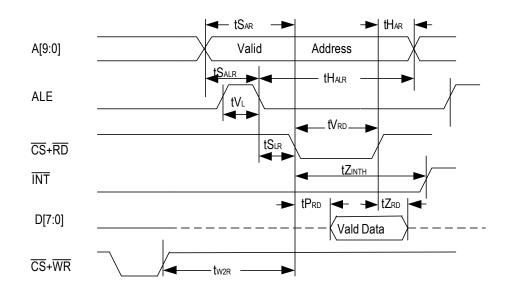


Figure - 85. Read Access Timing

#### Notes:

- Output propagation delay time is the time from the  $V_{DD}/2$  point of the reference signal to the 1.4V point of the output. Maximum output propagation delays are measured with a 100pF load on the MPIF data bus D[7:0]. 1.
- 2.
- 3.
- 4.
- All the set-up time or hold time are defined as the time between the  $V_{\rm D}/2$  point of the reference signal. In non-multiplexed mode, ALE can be held high, tSA<sub>LR</sub>, tH<sub>ALR</sub>, tV<sub>L</sub>, tS<sub>LR</sub> and tH<sub>LR</sub> are not applicable. Parameter tH<sub>AR</sub> is not applicable when address latching is used. The interval of read accesses should > = 180 ns.

#### 7.6 MICROPROCESSOR WRITE ACCESS TIMING

Symbol	Parameter		Min	MAX	Units
tS <sub>AW</sub>	Address to Valid Write Set-up Time		5		ns
tS <sub>DW</sub>	Data to Valid Write Set-up Time		0		ns
tS <sub>ALW</sub>	Address to Latch Set-up Time		5		ns
tH <sub>ALW</sub>	Address to Latch Hold Time		5		ns
tV <sub>L</sub>	Valid Latch Pulse Width		5		ns
tH <sub>LW</sub>	Latch to Write Hold		5		ns
tH <sub>DW</sub>	Data to Valid Write Hold Time		5		ns
tH <sub>AW</sub>	Address to Valid Write Hold Time		5		ns
t <sub>w2W</sub>	Write to write interval	E1	80		ns
		T1	100		ns

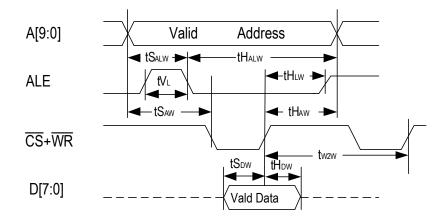


Figure - 86. Write Access Timing

#### Notes:

- Output propagation delay time is the time from the  $V_{DD}/2$  point of the reference signal to the  $V_{DD}/2$  point of the output.
- All the set-up time or hold time are defined as the time between the  $V_{nn}/2$  point of the reference signal.
- In non-multiplexed mode, ALE can be held high,  $tSA_{LW}$ ,  $tH_{ALW}$ ,  $tV_L$ ,  $tS_{LW}$  and  $tH_{LW}$  are not applicable. Parameter  $tH_{AW}$  and  $tS_{AW}$  are not applicable when address latching is used. 3.

# 7.7 I/O TIMING CHARACTERISTICS

### 7.7.1 TRANSMIT SYSTEM INTERFACE TIMING

Note that timing information can refer to the positive or negative edge of the reference clock. The active clock edge is selected by configuration flags.

Symbol	Parameter	Min	Тур	Max	Unit
Tprop	Propagation delay	0		20	ns
Ts	Set up time	15			ns
Thold	Hold time	10			ns

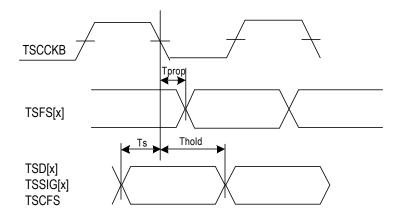


Figure - 87. Transmit Interface Timing (Transmit System Common Clock #B)

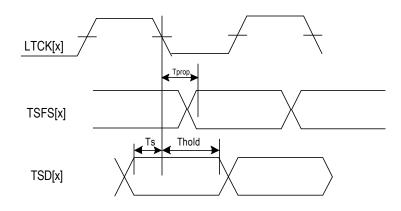


Figure - 88. Transmit Interface Timing (Line Transmit Clock)

### 7.7.2 RECEIVE SYSTEM INTERFACE TIMING

Note that timing information can refer to the positive or negative edge of the reference clock. The active clock edge is selected by configuration flags.

Symbol	Parameter	Min	Тур	Max	Unit
Tprop	Propagation delay	0		20	ns
Ts	Set up time	10			ns
Thold	Hold time	10			ns

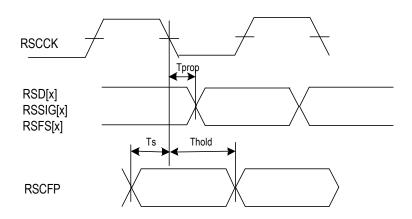


Figure - 89. Receive Interface Timing (Receive System Common Clock)

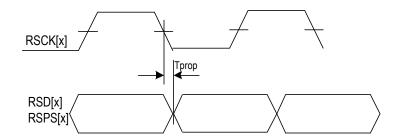


Figure - 90. Receive Interface Timing (Receive System Clock)

### 7.7.3 RECEIVE & TRANSMIT LINE TIMING

Note that timing information can refer to the positive or negative edge of the reference clock. The active clock edge is selected by configuration flags.

# 7.7.3.1 Receive Line Interface Timing

Symbol	Parameter	Min	Тур	Max	Unit
Ts	Setup Time	10			ns
Th	Hold Time	10			ns

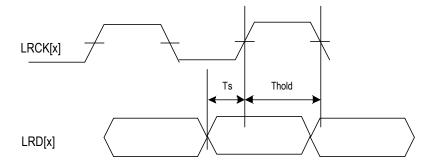


Figure - 91. Receive Line Interface Timing

# 7.7.3.2 Transmit Line Interface Timing

Symbol	Parameter	Min	Тур	Max	Unit
Tprop	Propagation delay	-10		10	ns

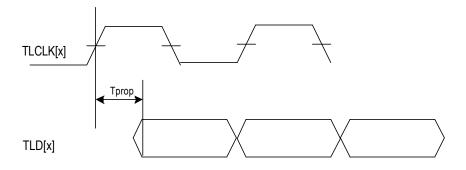
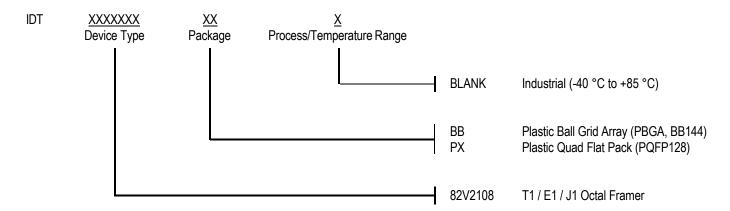


Figure - 92. Transmit Line Interface Timing

# **ORDERING INFORMATION**



# **Data Sheet Document History**

07/30/2002 pgs 48, 50, 199 09/09/2002 pgs 32, 33, 123, 125, 190, 236 01/15/2003 pgs 1, 272



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