
HA12181FP

AM Radio Noise Reduction System

REJ03F0130-0200
(Previous: ADE-207-171A)
Rev.2.00
Jun 15, 2005

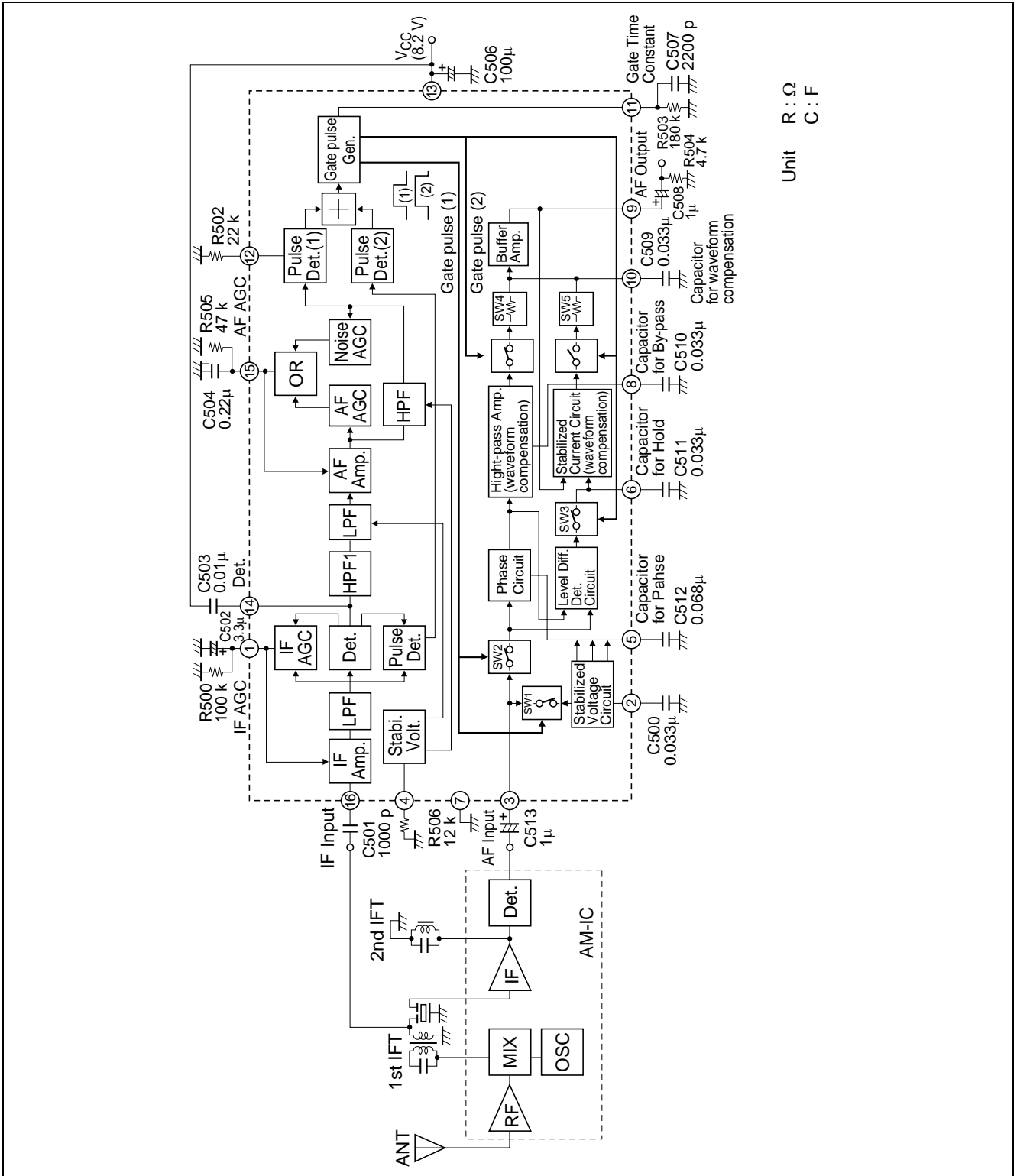
Functions

- Buffer amp. for audio
- Linear approximate circuit for noise reduction
- IF Amp., detector, audio amp. and AGC circuit for noise detection
- Gate pulse generator

Features

- High noise cancelling capacity: 46 dB typ.
- Less gain loss: $G_V = -0.5$ dB typ.
- Low total harmonic distortion and high signal-to noise ratio: THD = 0.06% typ., S/N = 75 dB typ.
- Operation supply voltage range: 7.0 V to 10 V (8.2 V typ.)
- Less external parts count

Block Diagram



Unit R : Ω
C : F


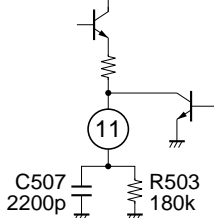
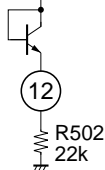
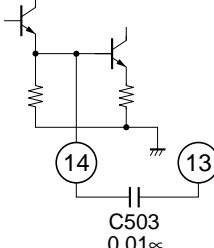
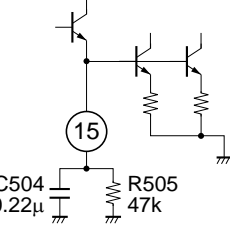
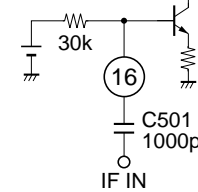
Table of Pin Description and External Parts

No. of pin	Name	Function	DC voltage (V) (No input)	Equivalent circuit	External parts		Influence of External parts	
					No.	recommended value	Larger than recommended value	Smaller than recommended value
1	IF AGC	Time constant for IF AGC.	2.7		R500 C502	100 K 3.3 μ	Longer time to stabilize AGC.	Longer distortion of recover.
2	Bias1	Bypass for voltage Stabi.	3.2		C500	0.033 μ	—	Increased noise.
3	AF input	Input of AF.	3.3		C513	1 μ	—	—
4	Bias2	Decide the current of filter network.	1.3		R506	12 K	Cut off frequency of L·P·F and H·P·F shifted lower.	Cut off frequency of L·P·F and H·P·F shifted higher.
5	Phase	Phase circuit	3.3		C512	0.068 μ	Must be used on the recommended value.	

Table of Pin Description and External Parts (cont.)

No. of pin	Name	Function	DC voltage (V) (No input)	Equivalent circuit	External parts		Influence of External parts	
					No.	recommended value	Larger than recommended value	Smaller than recommended value
6	Hold	Hold of level difference.	3.3		C511	0.033 μ	Must be used on the recommended value.	
7	GND	GND	—	—	—	—	—	—
8	High-Pass.	High-Pass AMP. (Waveform Compensation)	3.3		C510	0.033 μ	Must be used on the recommended value.	
9	AF out	Output of AF	3.3		C508	1 μ	Output DC cut	
					R504	4.7 K	Output load	
10	Wave form	Waveform Compensation	3.3		C509	0.033 μ	Must be used on the recommended value.	

Table of Pin Description and External Parts (cont.)

No. of pin	Name	Function	DC voltage (V) (No input)	Equivalent circuit	External parts		Influence of External parts	
					No.	recommended value	Larger than recommended value	Smaller than recommended value
11	Gate	Gate pulse generation			R503	180 K	Gate pulse width become wider.	Gate pulse width become narrow.
					C507	2200 P		
12	Vth	Determination of noise detection sensitivity	1.1		R502	22 K	Higher noise detection sensitivity.	Lower noise detection sensitivity.
13	V _{CC}	V _{CC}	8.2	—	—	—	—	—
14	IF Det.	IF AGC detector	3.3		C503	0.01 μ	—	—
15	AF AGC	Time constant for AF AGC	0		R505	47 K	Longer time to stabilize AGC.	Miss-operation in noise detector.
					C504	0.22 μ		
16	IF in	IF input	1.3				IF Input	Coupling Instability
							—	

Absolute Maximum Ratings

(Ta = 25°C)

Item	Symbol	Ratings	Unit
Supply voltage	V _{CC}	16	V
Power dissipation	P _d	400*1	mW
Operating temperature	T _{opr}	-40 to +85	°C
Storage temperature	T _{stg}	-55 to +125	°C

Note: 1. Value at Ta = 85°C

Electrical Characteristics (Tentative)

(V_{CC} = 8.2 V, Ta = 25°C, Pin 3 input: Vin = 100 mVrms, f = 1 KHz, Pin 16 input: Vin = 74 dBμ, fc = 450 KHz, fm = 1 KHz, m = 30%)

Item	Symbol	Min	Typ	Max	Unit	Test Conditions
Supply current	I _{CC}	—	11.0	—	mA	No input signal, IC only
Output voltage	V _{out}	70	95	120	mVrms	Pin 3 input only
Total harmonic distortion	THD1	—	0.06	0.3	%	
Signal-to-noise ratio	S/N (1)	60	75	—	dB	Pin 3 input Vin = 100 mVrms (Reference), Rg = 10 KΩ
Strong input total harmonic distortion	THD2	—	1.0	2.5	%	Pin 3 input Vin = 500 mVrms
Recovered output voltage	V _O (AF)	50	78	120	mVrms	Pin 16 input only
Recovered output signal-to-noise-ratio	S/N (2)	35	45	—	dB	
Noise suppression ratio	NSR	35	46	—	dB	Input the waveform below. Pin 3 input Vin = 100 mVrms (Reference) no input sine wave

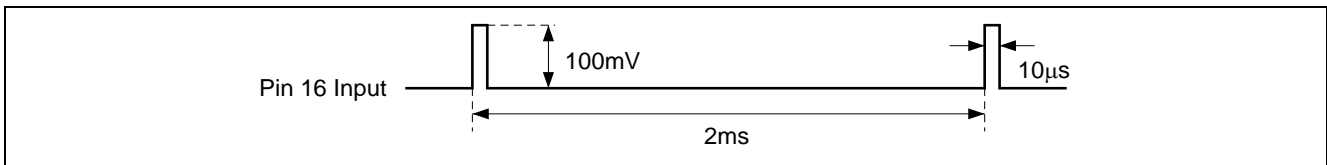
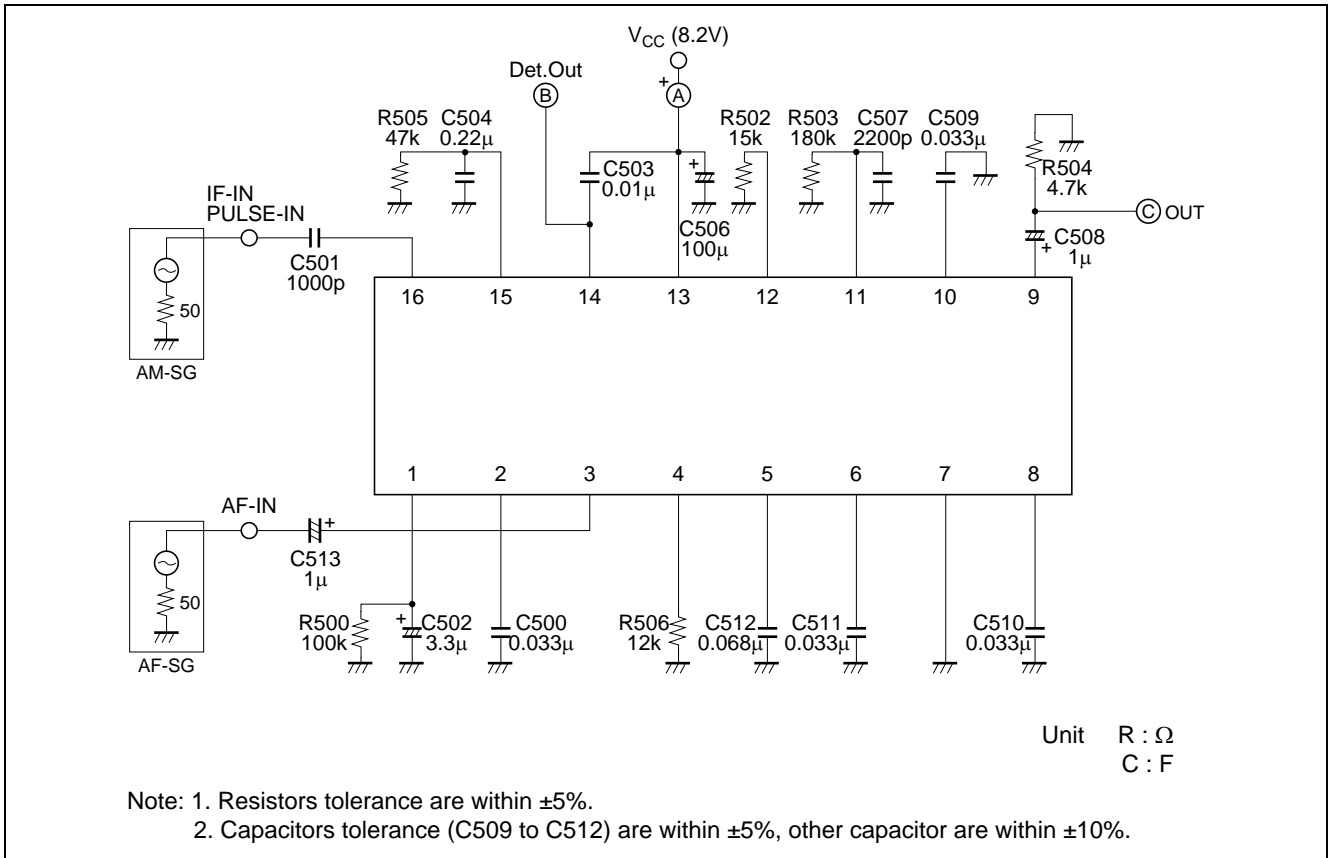


Figure 1 Input Waveform at Measurement of Noise Suppression Ratio

Test Circuit



Operation Principle

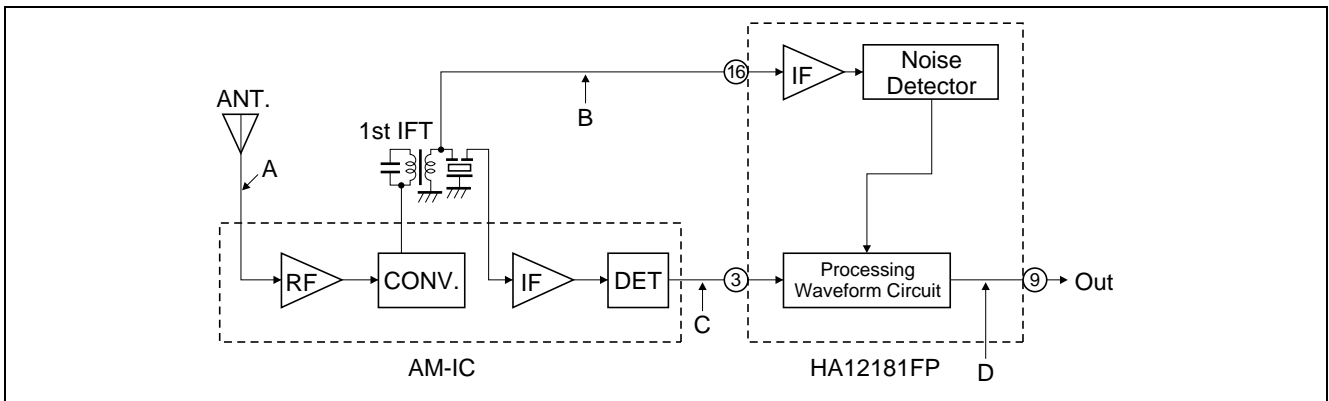


Figure 2 System Block Diagram of AM Radio

A system block diagram of AM Radio using the HA12181FP is shown in Figure 2 and waveforms at each point in the system are illustrated in Figure 3. For AM wave with impulse noise from ANT, the pulse spreads its width each time when the AM wave passes through a selection filter.

The pulse width becomes the order of several hundred microseconds at detector output (Point C).

A radio without a noise canceller produces large noise to the audience. This IC perfectly detects every noise by using the signals from 1st IFT (Point B) in front of the narrow band filter.

The wave process circuit approximates the voltage linearly at the pulse to reduce the noise in the output.

The principle for wave processing follows. Further investigation make it clear that the pulse width of impulse noise is constant (several hundred microseconds) and independent of the waveform or waveheight.

Therefore the former and later voltage (V_A , V_B) of the pulse can be found at the same time (T_1) by means of the wave and the delayed one for this time, as shown in the right figure.

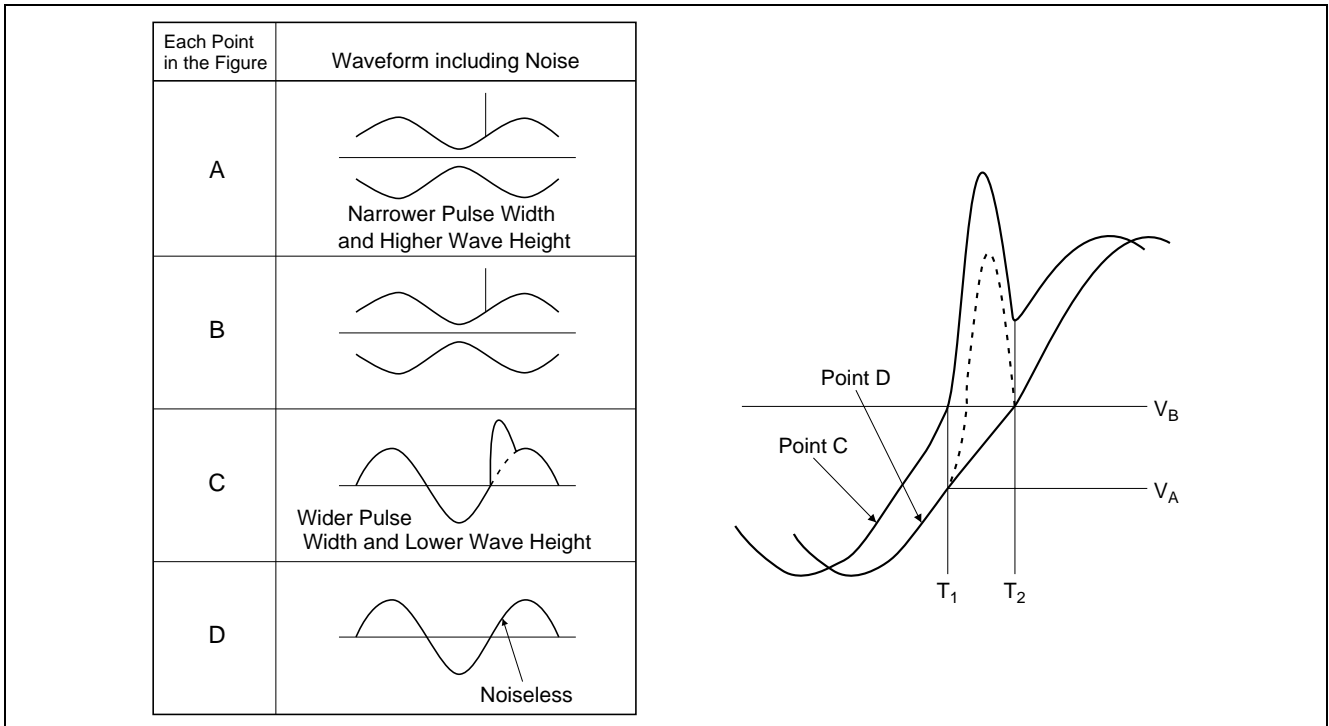


Figure 3 Waveforms at Each Point in the System

In an actual circuit, the differential voltage between input and output of phase shift circuit is changed to the capacitor C511 at pin 6.

At the time of T_1 , when the switch turns to the noise processing mode (the switch positions in Figure 4 are inverted), the voltage difference ($V_A - V_B$) is held in C511.

C509 at pin 10 is changed by the differential voltage between the held voltage and the output voltage at pin 9 (V_A):
 $V_A - (V_A - V_B) = V_B$.

As the initial voltage of C509 is equal to the output voltage (VA) before the switch change, the voltage between terminals of C509 is changed from VA to VB.

The waveform which change up to C509 becomes the output, because the voltage of C509 appears at pin 9 through the buffer.

The changed up waveform of C509 is almost linearly approximated because of the constant current change by the feedback from the output at pin 9.

At the time of T2 when the awitches change to the normal mode (the switch position in Figure 4), the output recovers smoothly as the voltage of C509 is VB.

However the unmatched wave delay time due to the pulse width or the phase circuit and the offset of circuit make a slight step difference on the waveform at the moment of switch change.

LPF, consisting of R1 and C509 make it smooth.

The frequency characteristics, which is deteriorated by LPF in the normal mode, is compensated so that it might become flat. C509 and C510 should have the same capacity, and the tolerance must be within $\pm 5\%$.

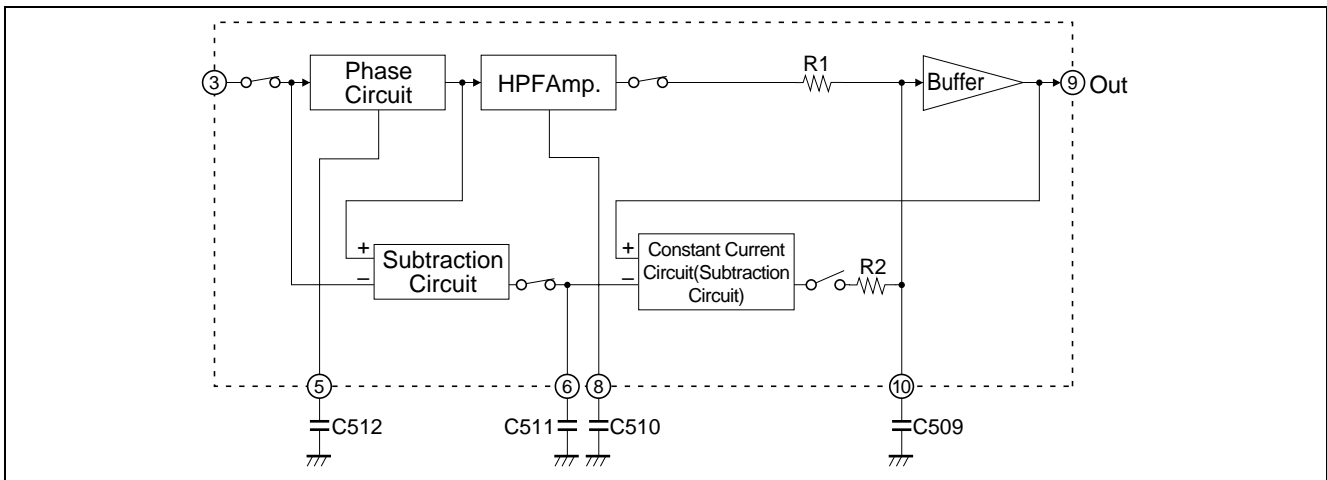
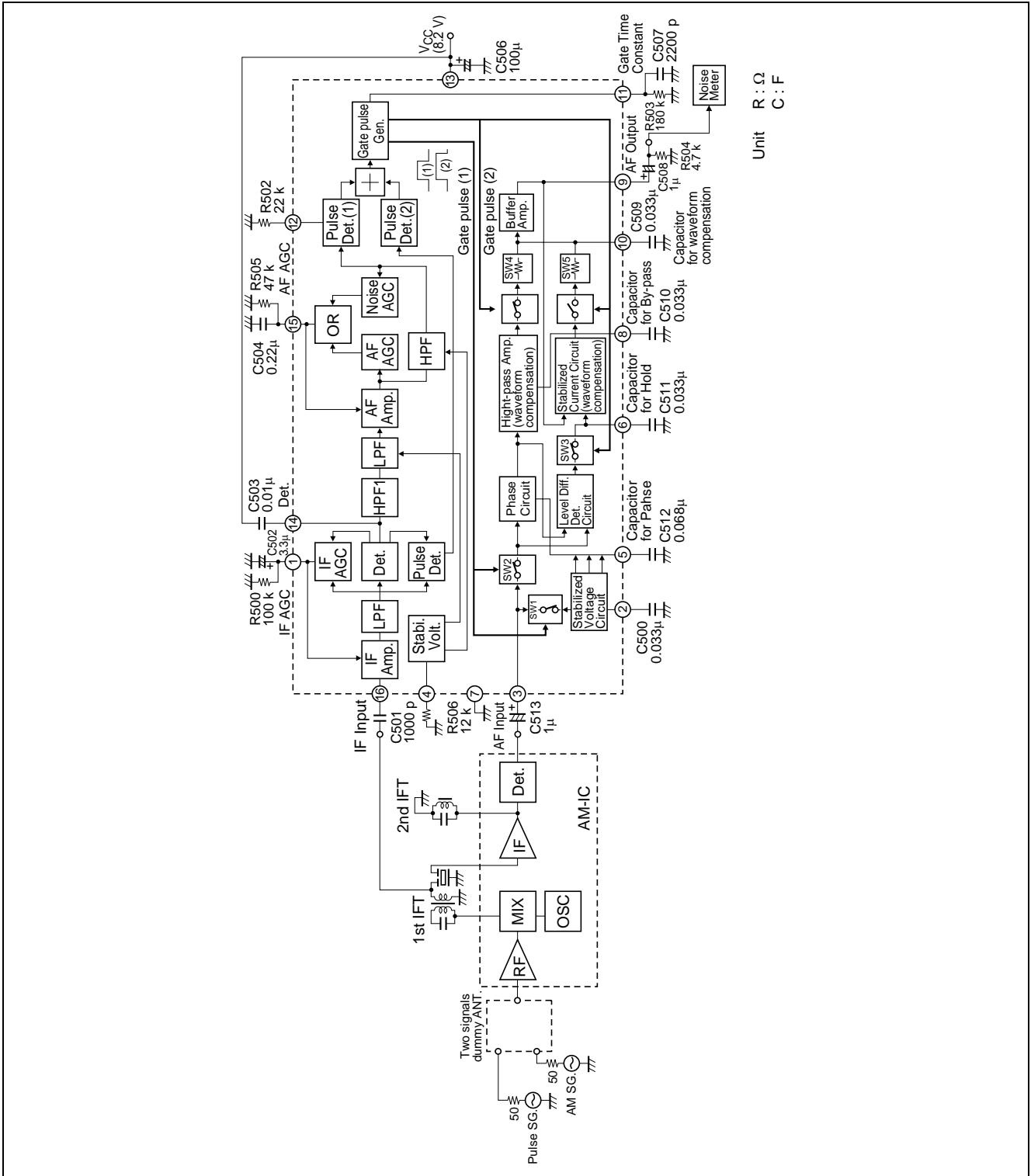
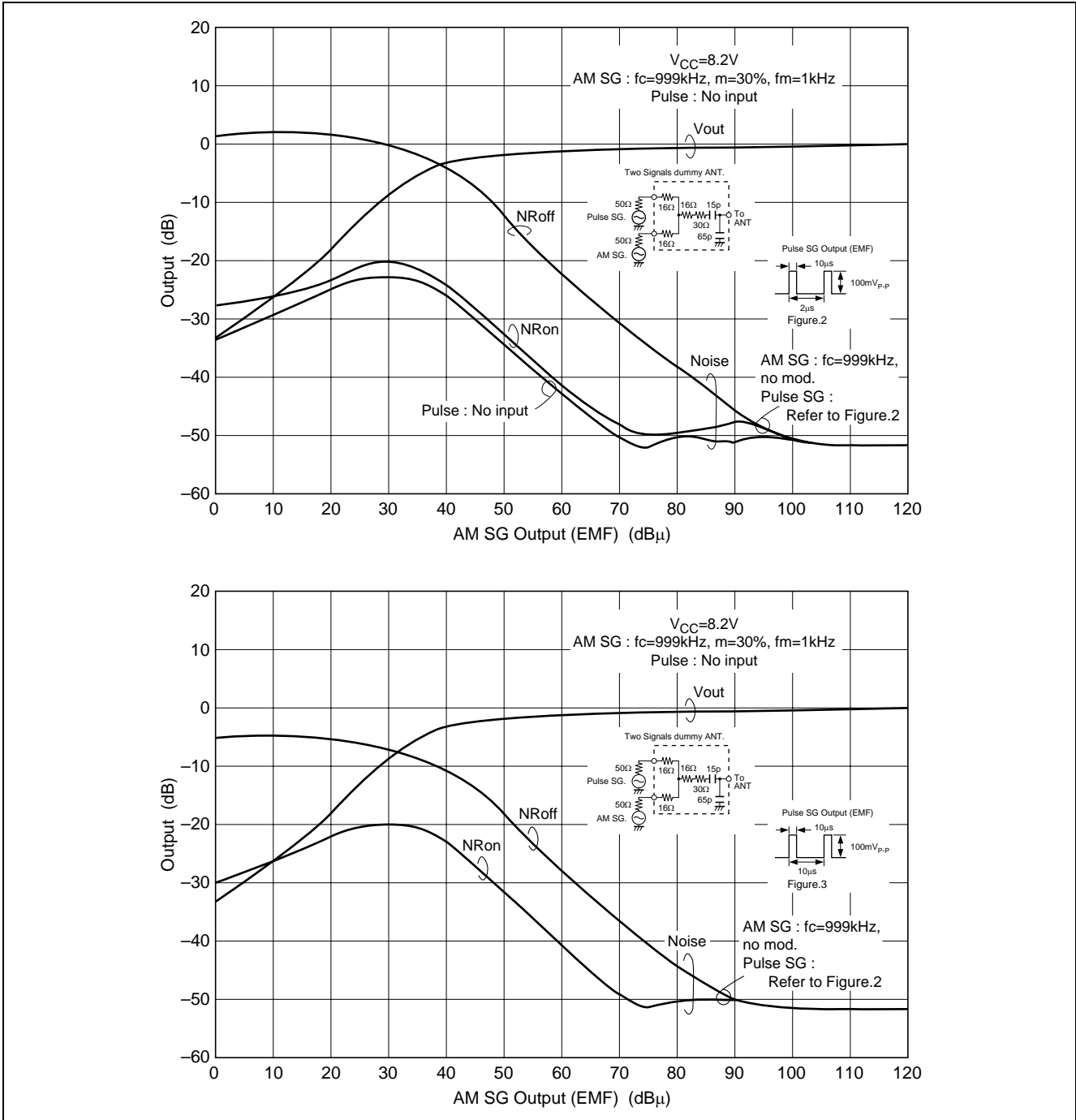


Figure 4 Waveform Processing Circuit

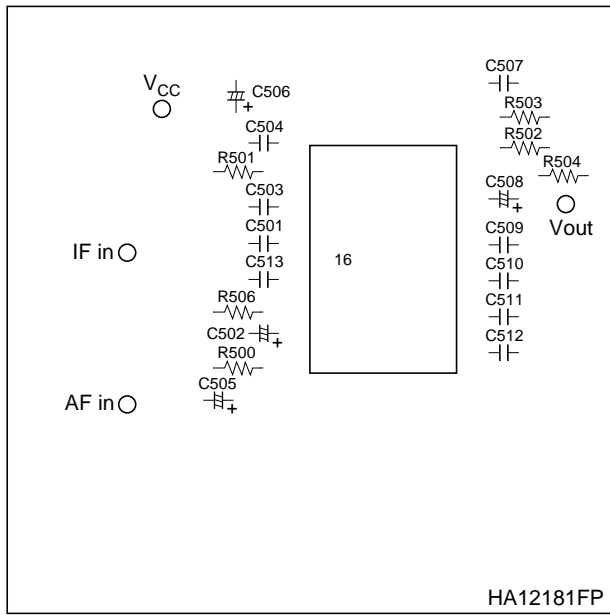
Evaluation Circuit for Noise Reduction Effect



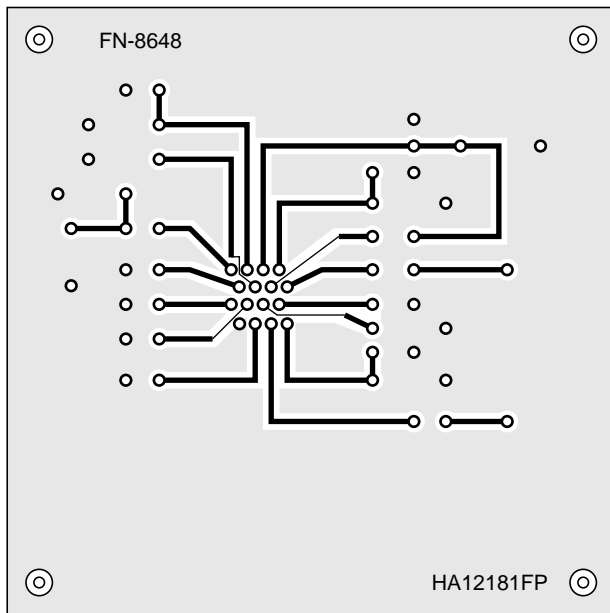
Example of Noise Reduction Effect



PC Board Layout Pattern

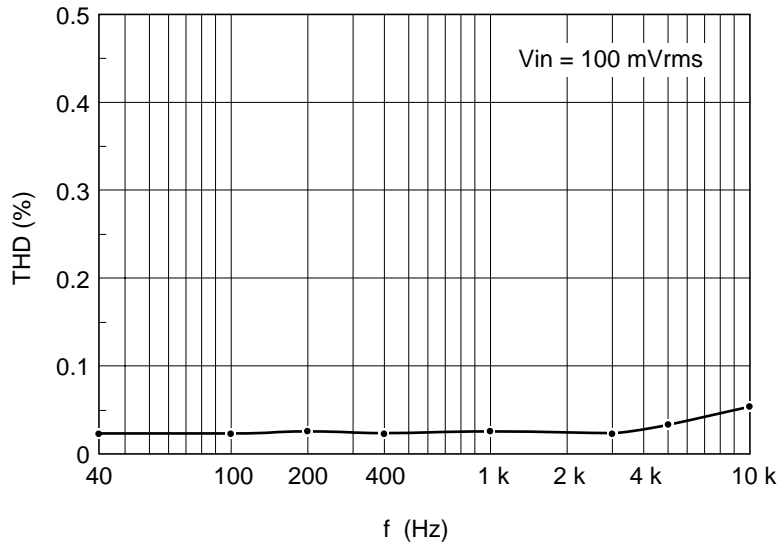
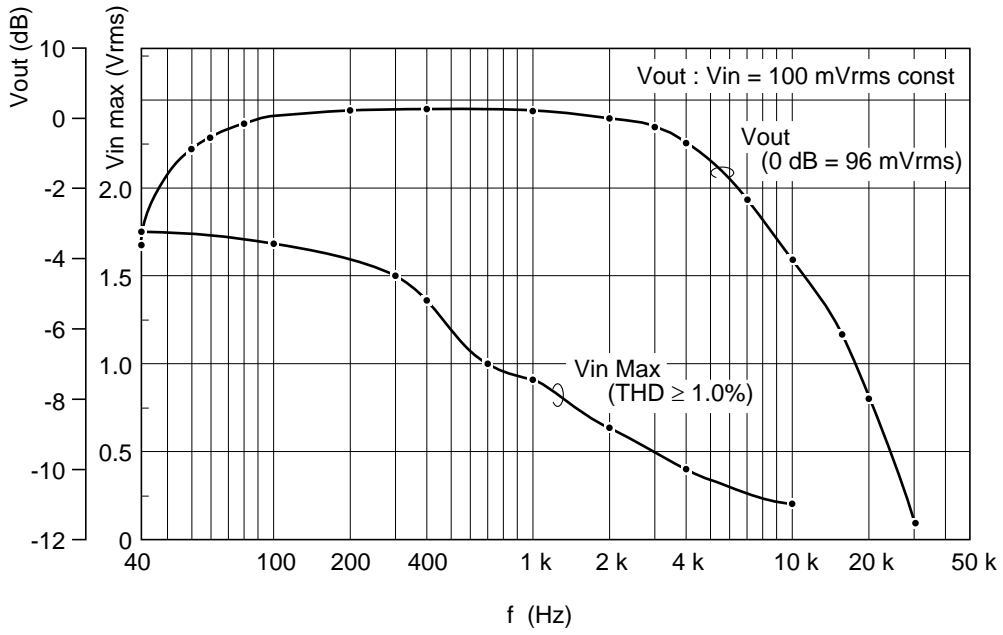


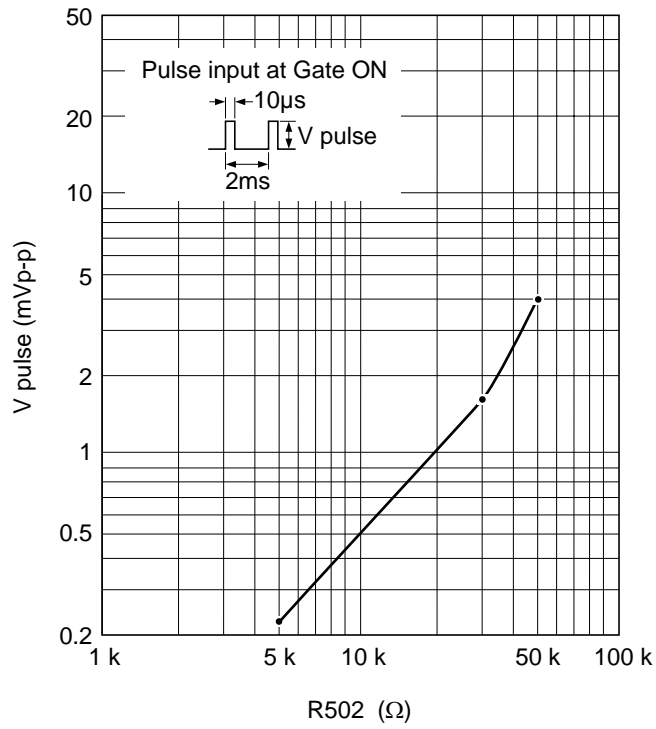
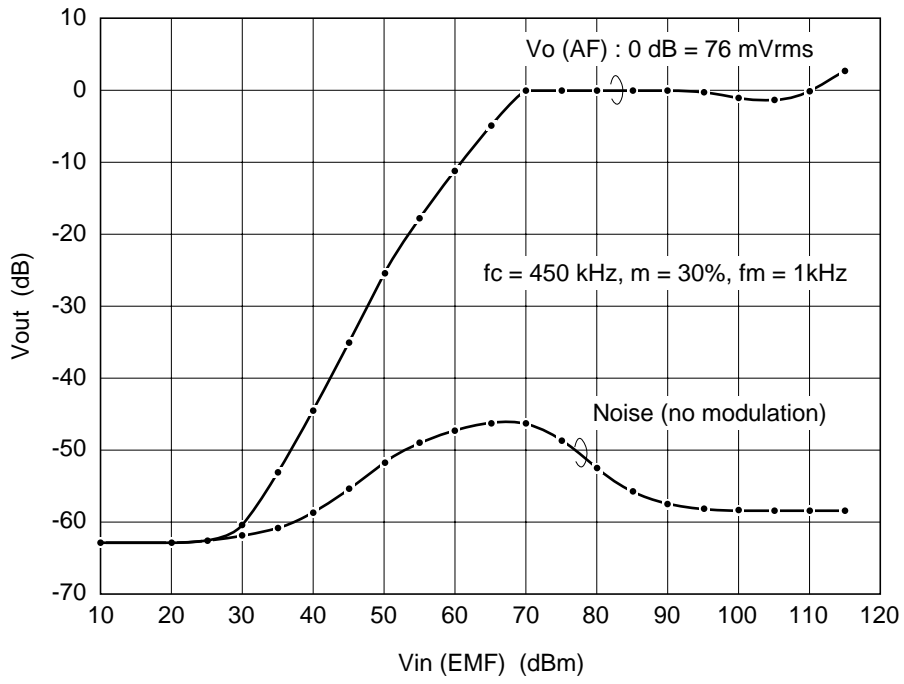
(Top view)

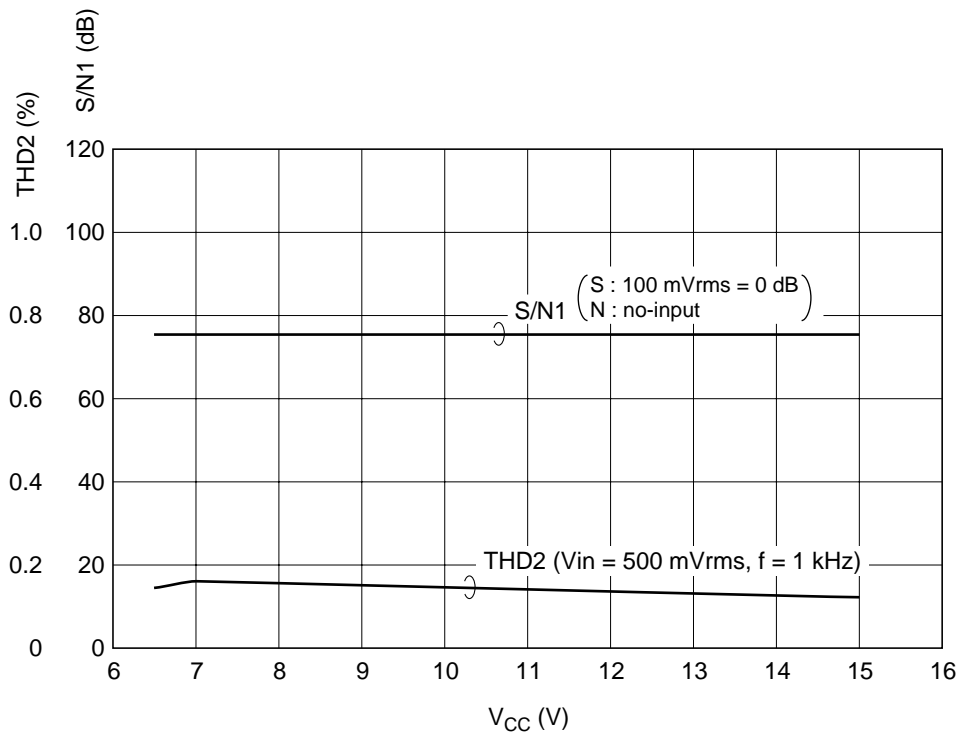
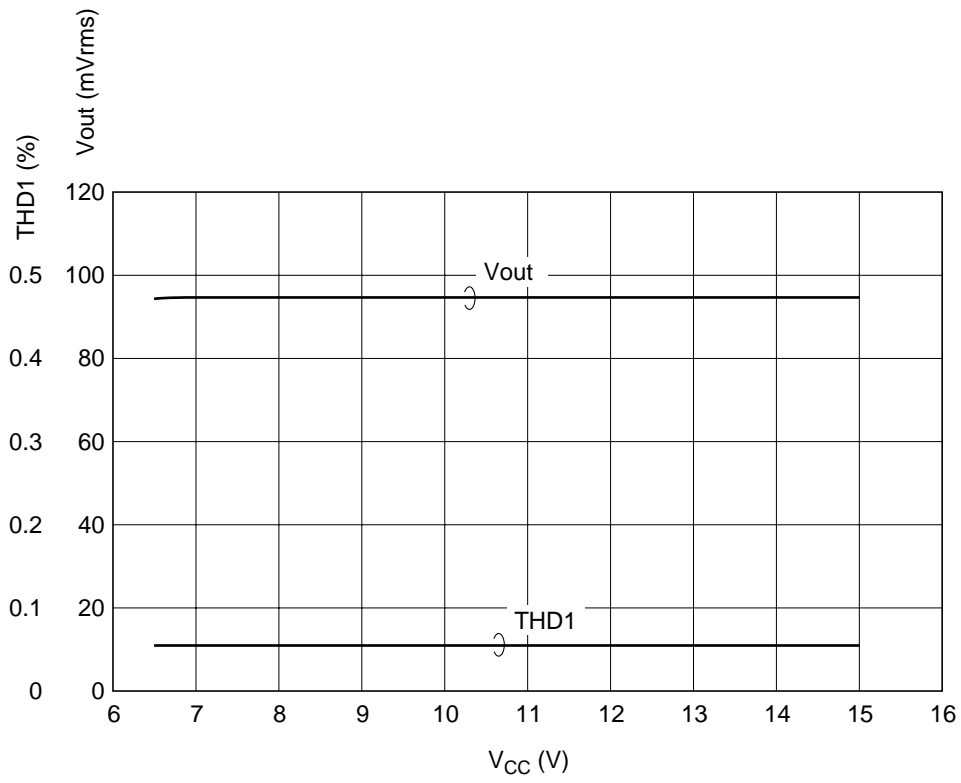


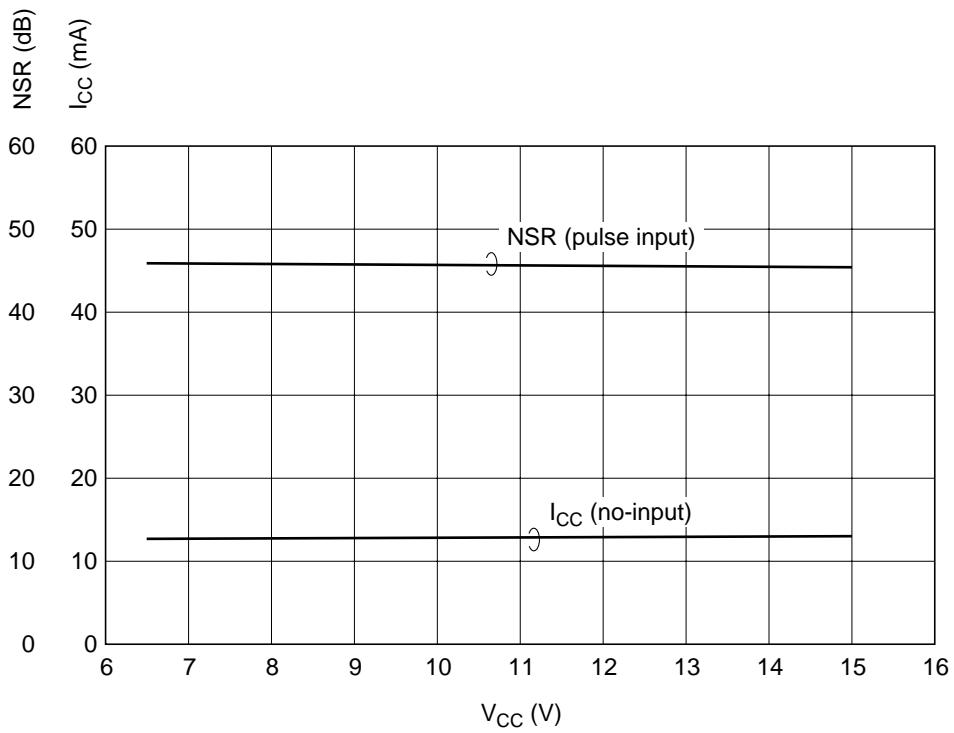
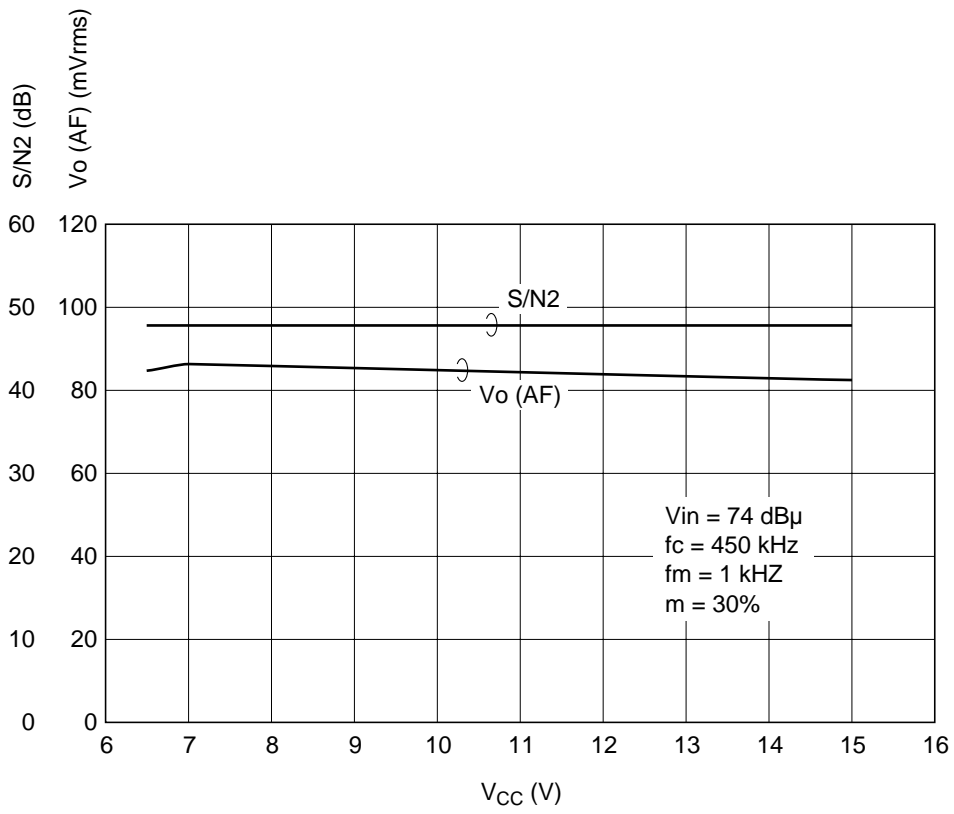
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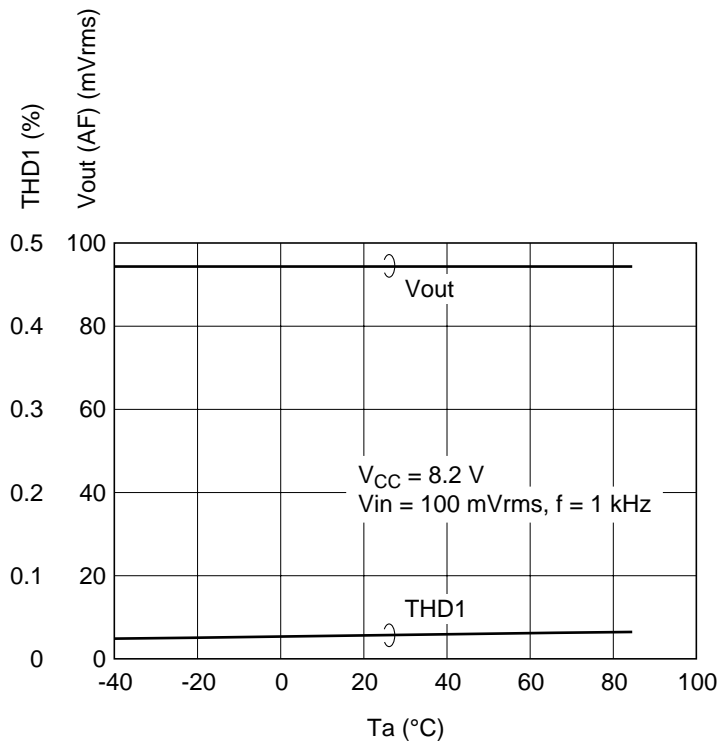
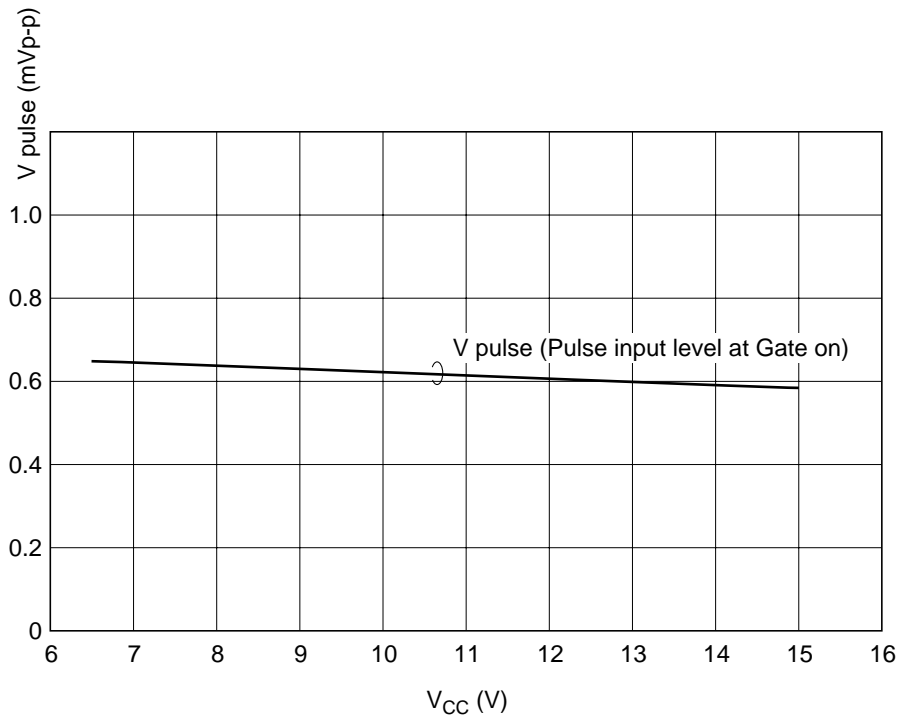
Main Characteristics

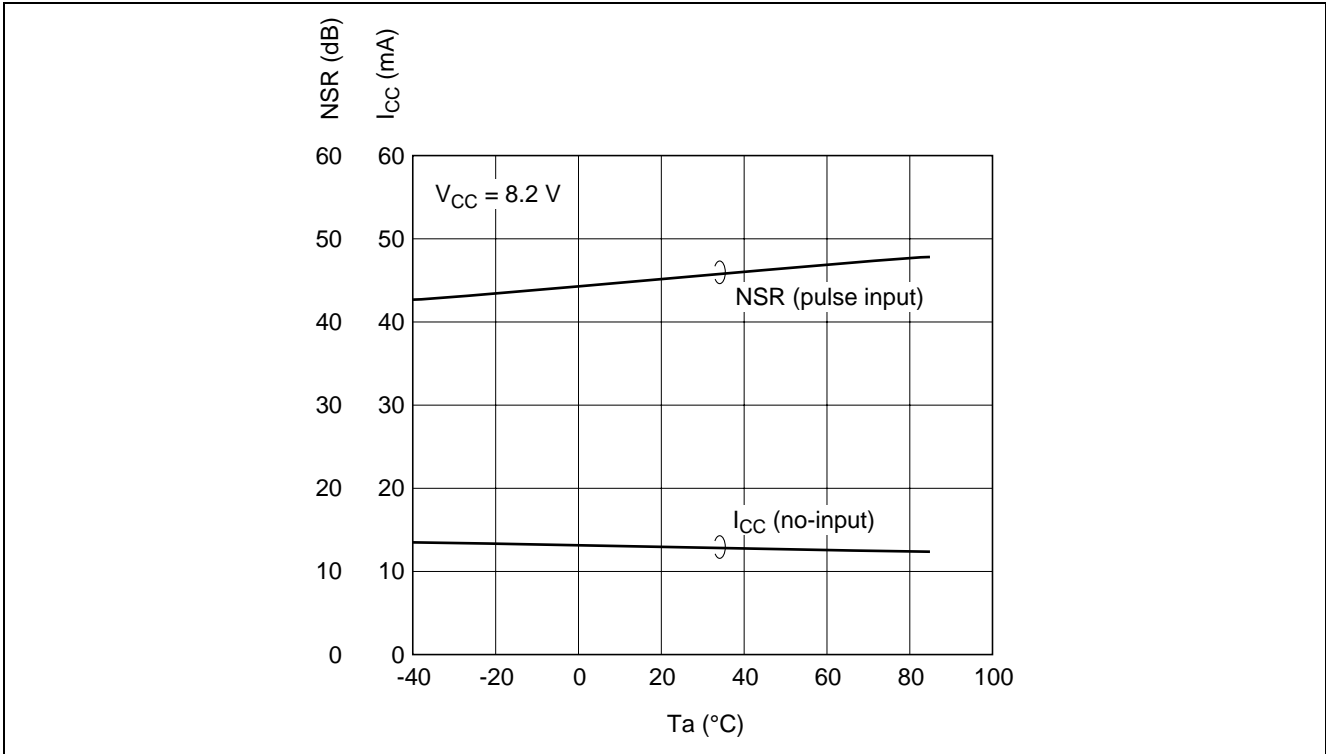






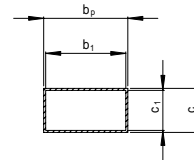
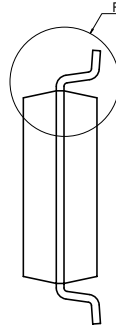
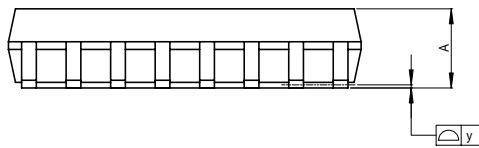
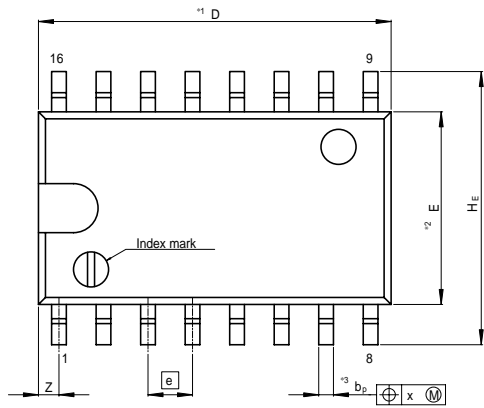




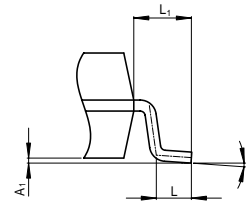


Package Dimensions

JEITA Package Code P-SOP16-5.5x10.06-1.27	RENESAS Code PRSP0016DH-A	Previous Code FP-16DA	MASS[Typ.] 0.24g
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Terminal cross section



Detail F

NOTE)
1. DIMENSIONS*1 (Nom)*AND*2*
DO NOT INCLUDE MOLD FLASH.
2. DIMENSION*3*DOES NOT
INCLUDE TRIM OFFSET.

Reference Symbol	Dimension in Millimeters		
	Min	Nom	Max
D	—	10.06	10.5
E	—	5.5	—
A ₂	—	—	—
A ₁	0.00	0.10	0.20
A	—	—	2.20
b _p	0.34	0.42	0.50
b ₁	—	0.40	—
c	0.17	0.22	0.27
c ₁	—	0.20	—
θ	0°	—	8°
H _E	7.50	7.80	8.00
e	—	1.27	—
x	—	—	0.12
y	—	—	0.15
Z	—	—	0.80
L	0.50	0.70	0.90
L ₁	—	1.15	—

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