

# ICs for Communications

Analog Front End for Telephone Systems  
SAM-AFE

PSB 4851 Version 2.1

Data Sheet 09.99

DS 1

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## **1 Overview**

The PSB 4851 integrates all amplifiers to directly connect the transducers to the chip. It features two completely independent channels. An integrated analog multiplexer allows the connection of three signal sources (handset microphone, speakerphone microphone, analog line) to the two channels.

Furthermore the PSB 4851 supports a sophisticated power management and a loop mode in the analog domain. These features can be used to implement a line powered mode for emergency operation of the phone.

The chip is programmed by a simple four wire serial control interface.

# Analog Front End for Telephone Systems SAM-AFE

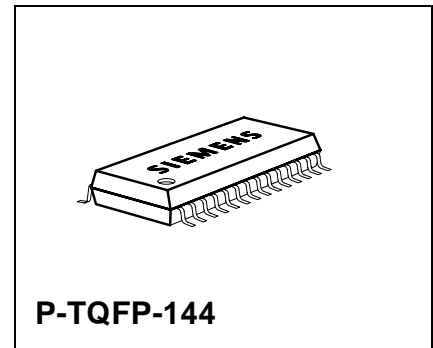
PSB 4851

Version 2.1

CMOS

## 1.1 Features

- Direct connection to handset
- Direct connection to microphone
- Direct connection to loudspeaker (50  $\Omega$ )
- Low power emergency operation
- Serial control interface for programming
- 3.3V or 5V operation (full operating mode)
- 3V-5V voltage range (emergency mode)
- 2.4 V reference voltage
- Two differential inputs
- Support for controlled loudhearing
- Compliant to G.712



Type	Package
PSB 4851	P-TQFP-144

## 1.2 Pin Configuration

(top view)

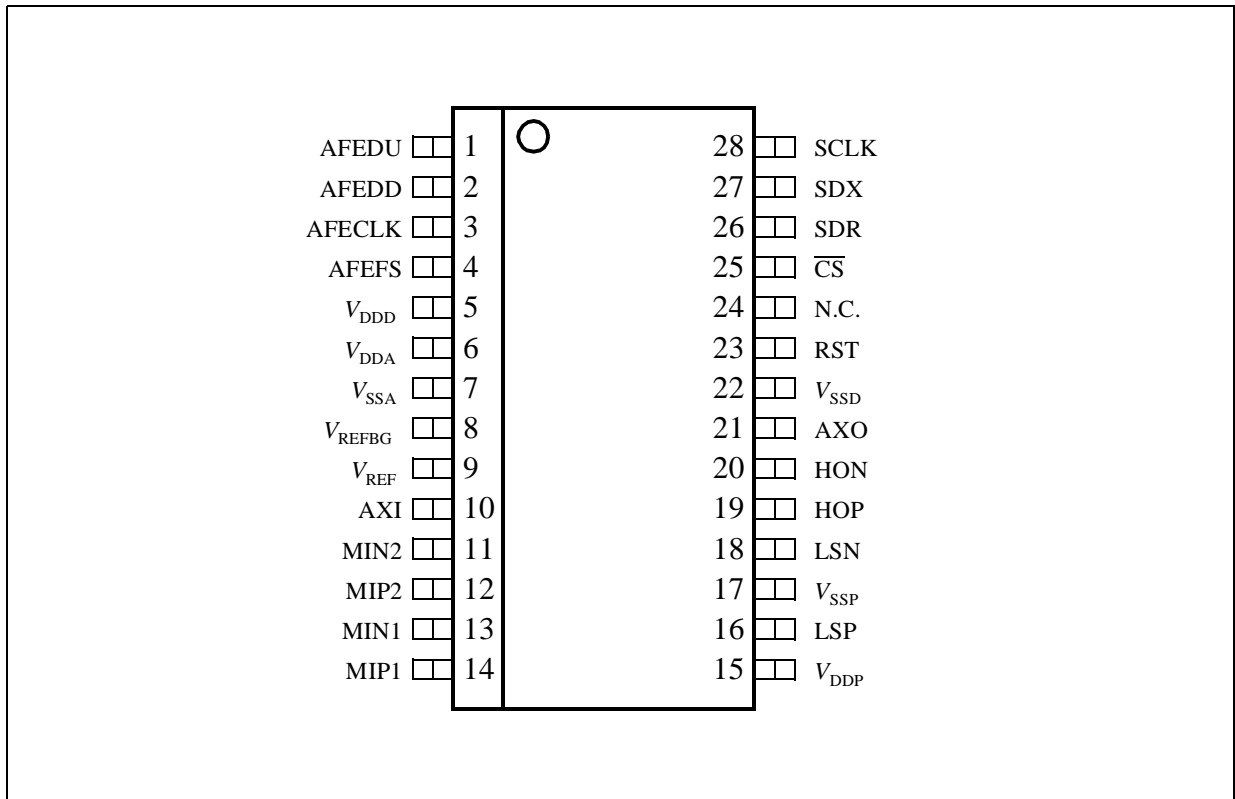


Figure 1-1 Pin Configuration

### 1.3 Pin Definitions and Functions

Table 1-1 Pin Definitions and Functions

Pin No. P-DSO-28-1	Symbol	Dir.	Reset	Function
5	$V_{DDD}$	-	-	<b>Power supply (3.0V - 3.6V or 4.75V - 5.25V)</b> Power supply for digital parts. Must be at the same level as $V_{DDA}$ and $V_{DDP}$ .
6	$V_{DDA}$	-	-	<b>Power supply (3.0V - 3.6V or 4.75V - 5.25V)</b> Power supply for analog parts. Must be at the same level as $V_{DDD}$ and $V_{DDP}$ .
15	$V_{DDP}$	-	-	<b>Power supply (3.0V - 3.6V or 4.75V - 5.25V)</b> Power supply for amplifiers. Must be at the same level as $V_{DDA}$ and $V_{DDD}$ .
22	$V_{SSD}$	-	-	<b>Power supply (0 V)</b> Ground for digital parts.
7	$V_{SSA}$	-	-	<b>Power supply (0 V)</b> Ground for analog parts.
17	$V_{SSP}$	-	-	<b>Power supply (0 V)</b> Ground for amplifiers.
1	AFEDU	O	L	<b>Data Upstream:</b> Data output to PSB 4860.
2	AFEDD	I	-	<b>Data Downstream:</b> Data input from PSB 4860.
3	AFECLK	I	-	<b>Data Clock:</b> 6.912 MHz clock.
4	AFEFS	I	-	<b>Frame Synchronization:</b> 8kHz frame synchronization from PSB 4860.
28	SCLK	I	-	<b>Serial Clock:</b> Clock for serial control interface (SCI).
27	SDX	OD	H	<b>Serial Data Transmit:</b> Data output for serial control interface (SCI).
26	SDR	I	-	<b>Serial Data Receive:</b> Data input for serial control interface (SCI).
25	$\overline{CS}$	I	-	<b>Chip Select:</b> Select signal for serial control interface (SCI).

Table 1-1 Pin Definitions and Functions

Pin No. P-DSO-28-1	Symbol	Dir.	Reset	Function
23	RST	I	-	<b>Reset:</b> Active high reset signal.
10	AXI	I	-	<b>Auxiliary Input:</b> Single ended analog input (e.g. line in)
13 14	MIN1 MIP1	I I	-	<b>Microphone Input 1:</b> This input provides a highly symmetrical differential input for commonly used telephone microphones.
11 12	MIN2 MIP2	I I	-	<b>Microphone Input 2:</b> This input provides a highly symmetrical differential input for commonly used telephone microphones.
21	AXO	O	0 V	<b>Auxiliary Output:</b> Single ended analog output (e.g line out).
19 20	HOP HON	O O	0 V 0 V	<b>Handset Earpiece Output:</b> Differential outputs which can drive common handset earpiece transducers (200Ω) directly.
16 18	LSP LSN	O	0 V 0 V	<b>Loudspeaker Output:</b> Differential outputs which can drive a 50Ω loudspeaker at 5V or a 25Ω loudspeaker at 3.3V directly. A piezo transducer can be used for ringing instead of the loudspeaker.
8	V <sub>REFBG</sub>	O	0 V	<b>Reference Bandgap Voltage</b> Connection to external 22 nF capacitor for low pass filtering.
9	V <sub>REF</sub>	O	0 V	<b>Reference Voltage (2.4 V):</b> Output for biasing external circuitry, e.g. electret microphone. Connection to external 100 nF capacitor.

## 1.4 Logic Symbol

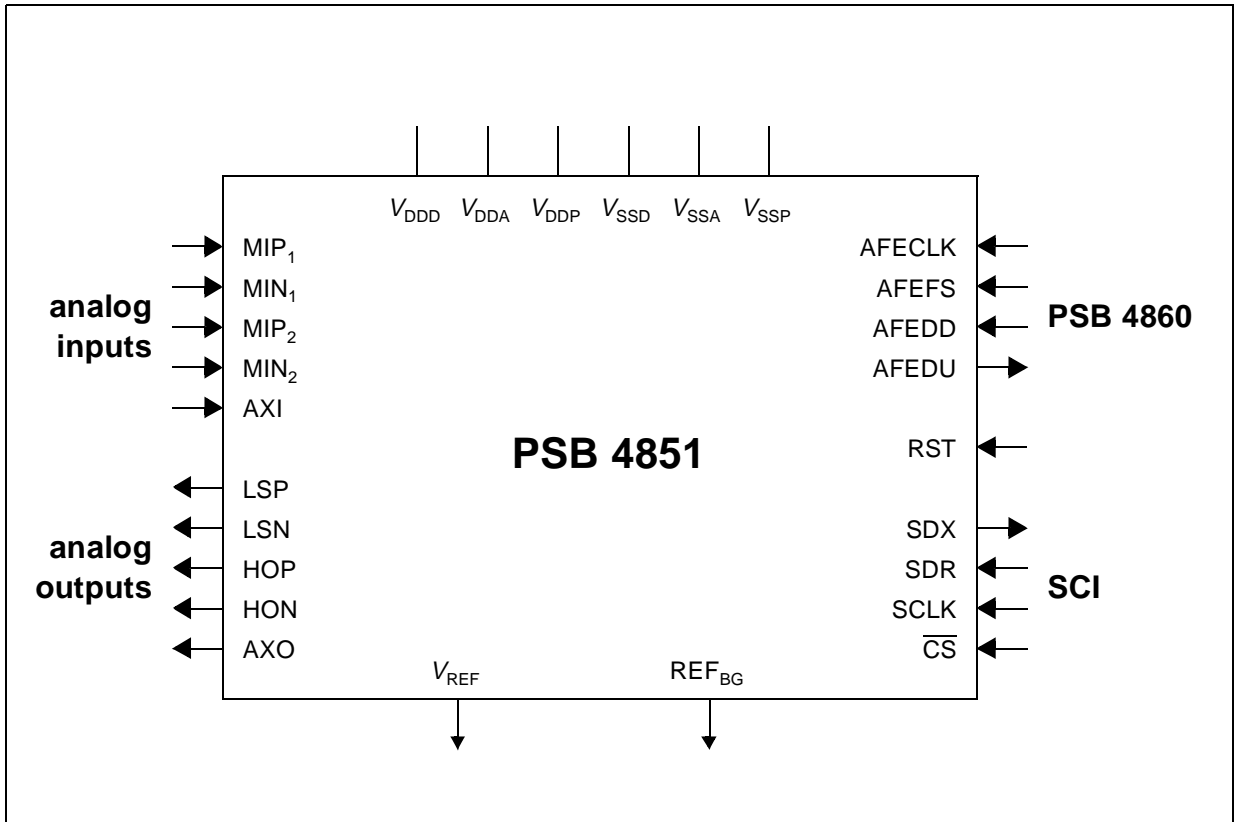
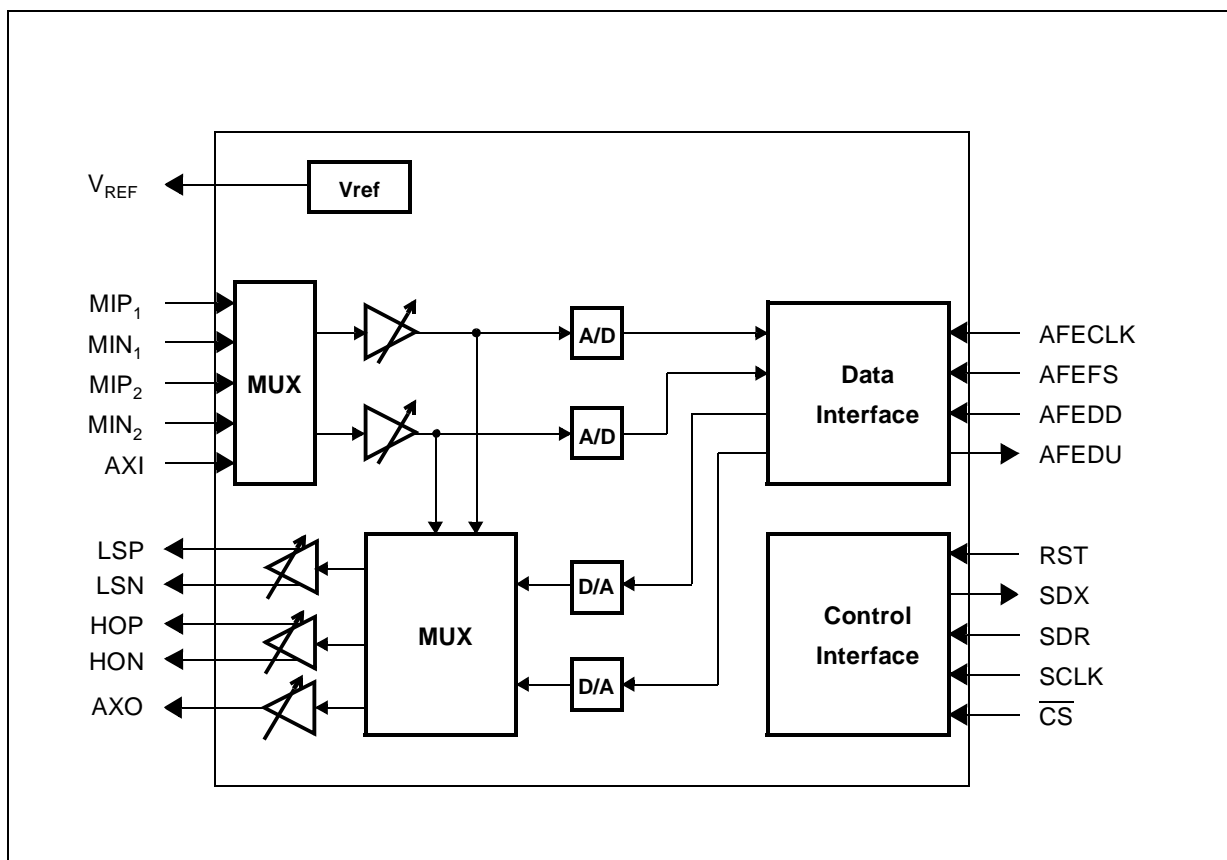


Figure 1-2 Logic Symbol of PSB 4851

## 1.5 Functional Block Diagram



**Figure 1-3 Block Diagram of PSB 4851**

## 1.6 System Integration

The PSB 4851 is the standard analog interface for several digital telecommunication ICs such as:

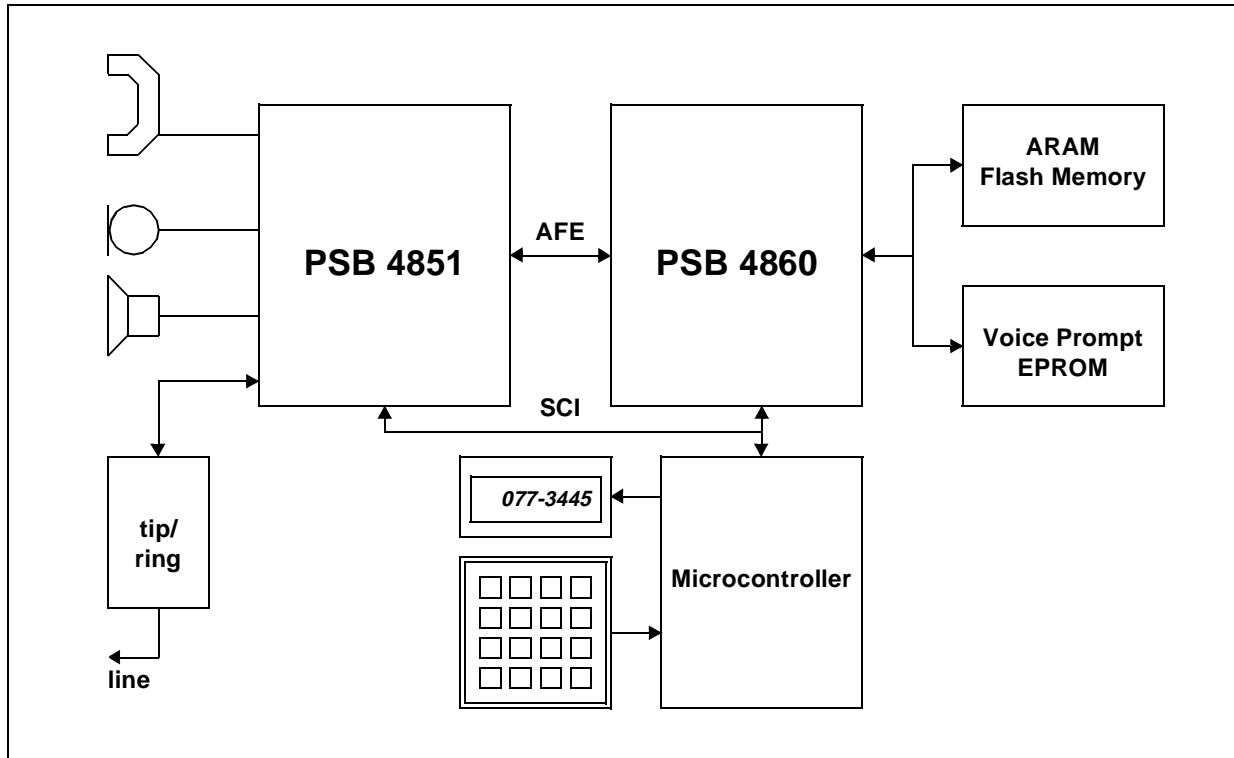
- PSB 4860 (digital answering machine)
- PSB 2170 (acoustic echo canceller)

The PSB 4851 is especially suited for applications that need two independent analog channels where one codec interfaces to a loudspeaker/microphone combination while the other codec serves the line.

### 1.6.1 Analog Featurephone with Digital Answering Machine

Figure 1-4 shows an example of an analog telephone system. The telephone can operate during power failure by line powering. In this case only the handset is active. All other parts of the chipset are shut down leaving enough power for the external microcontroller to perform basic tasks like keyboard monitoring.

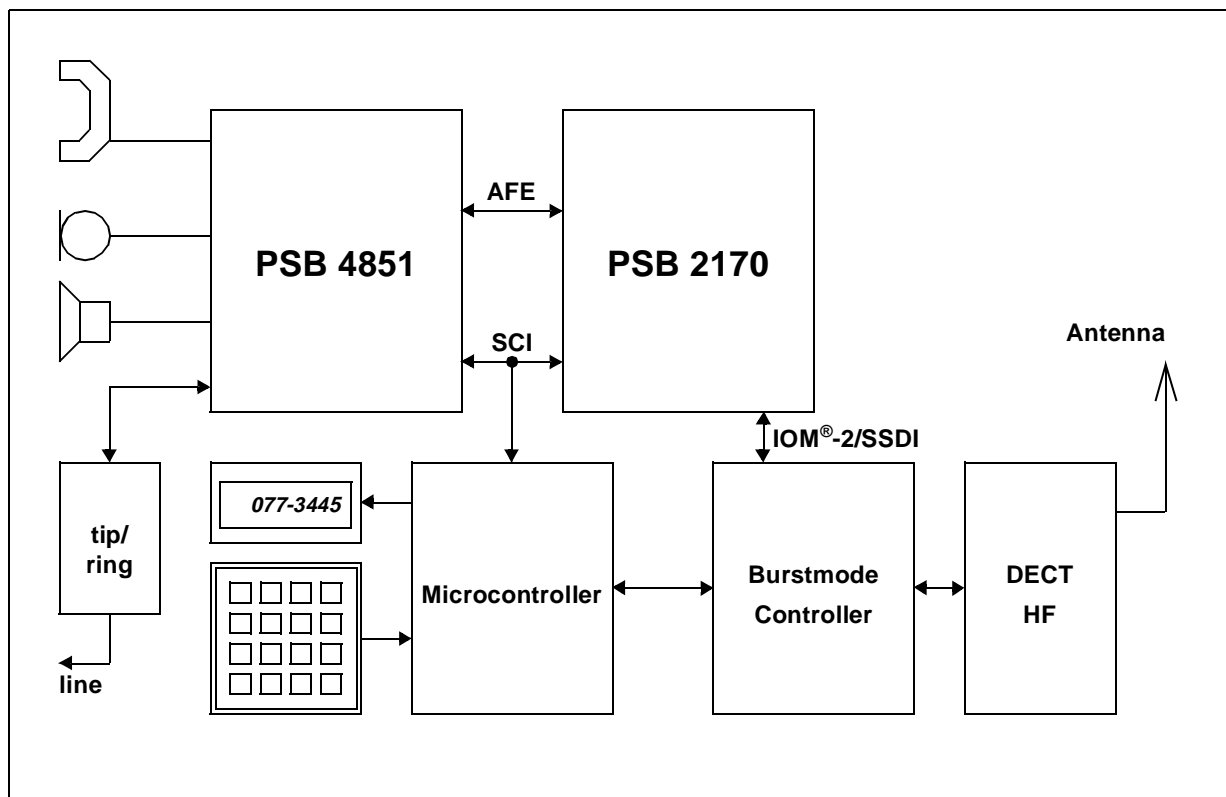
For answering machine operation the voice data is stored in ARAM or Flash Memory devices and voice prompts can be played back from an optional voice prompt EPROM. If Flash Memory is used the functionality of the voice EPROM can be realized by the Flash Memory devices.



**Figure 1-4 Analog Full Duplex Speakerphone with Digital Answering Machine**

### 1.6.2 DECT Basestation with Full Duplex Featurephone

Figure 1-5 shows a DECT basestation with acoustic echo cancellation based on the PSB 2170. The full duplex featurephone can be switched to the basestation or a mobile handset dynamically. For programming the serial control interface (SCI) is used while voice data is transferred via the strobed serial data interface (SSDI).



**Figure 1-5 DECT Basestation with Full Duplex Speakerphone**

## 2 Functional Description

The PSB 4851 provides two bidirectional channels from the analog domain to the digital domain, an internal loopback and a sophisticated power management for line-powered operation. The first section describes the signal paths and functional units of the PSB 4851 while the second section discusses the support of the line powered operation.

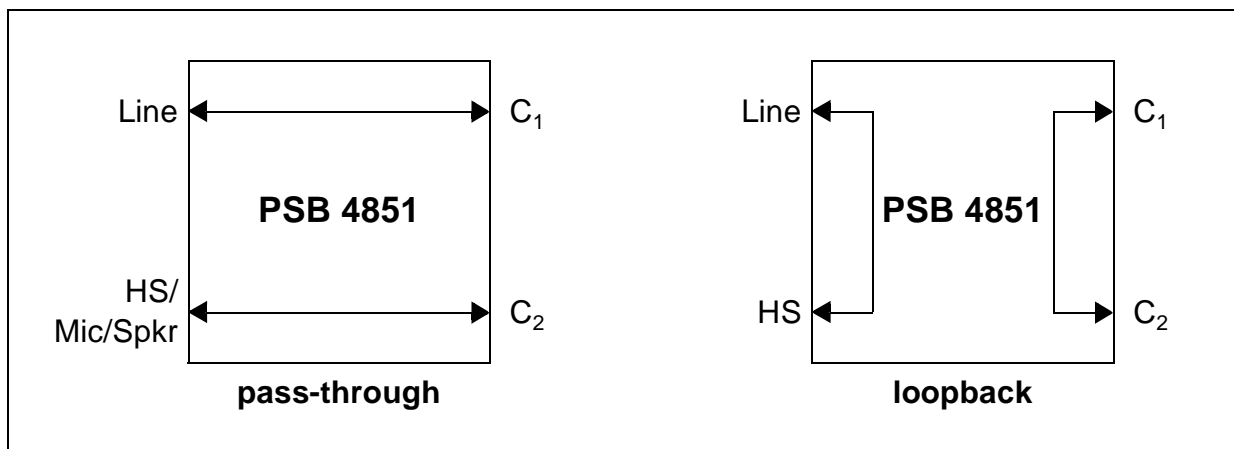
### 2.1 Signal Paths and Functional Units

The PSB 4851 supports three analog inputs, three analog outputs and two digital channels (Table 2-1).

**Table 2-1**

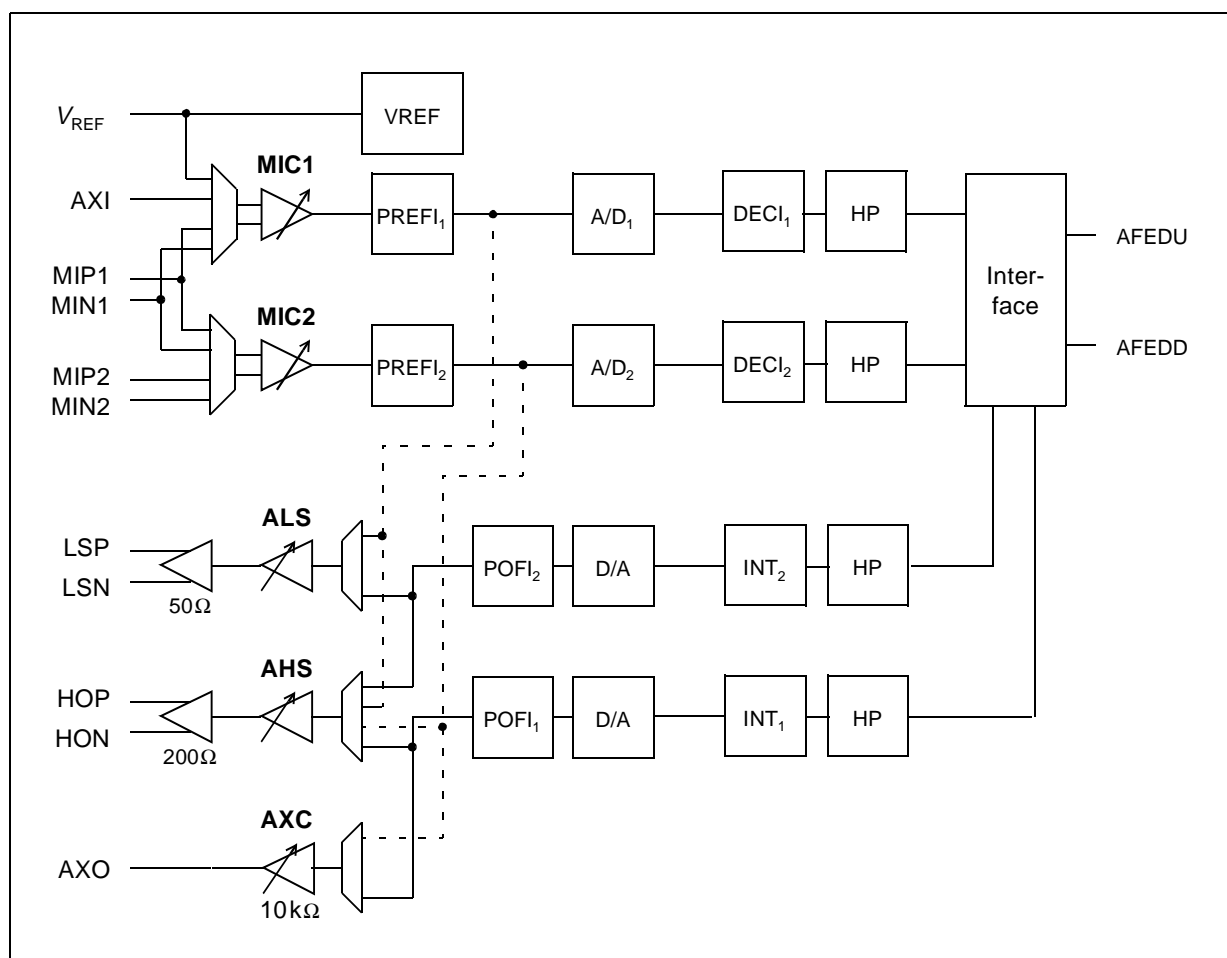
Analog Inputs	Pins	Comment
	AXI	line input from tip/ring interface
	MIP1, MIN1	mic. 1, e.g. speakerphone microphone
	MIP2, MIN2	mic. 2, e.g. handset microphone
Analog Outputs	AXO	line output to tip/ring interface
	HOP, HON	handset earpiece
	LSP, LSN	speakerphone loudspeaker
Digital Channels	AFEDD, AFEDU, AFEFS, AFECLK	Channel 1 and 2 of AFE interface (to/from PSB 2170, PSB 4860)

These signals can be routed in either **pass-through** or **loopback** mode (Fig. 2-1). In loopback mode different loops are available for test purposes and line powered operation. In loopback mode the digital part of the PSB 4851 can be completely shut down if it is not needed. The loop on the analog side remains fully functional.



**Figure 2-1 Basic Configurations of PSB 4851**

A detailed functional diagram of the PSB 4851 is shown in figure 2-2.



**Figure 2-2 Detailed Block Diagram of PSB 4851**

Two differential inputs for microphones and one single ended input for the tip/ring interface are fed to two analog input amplifiers (MIC2, MIC1). These amplifiers can be programmed for a gain of up to 42 dB in steps of 6 dB.

For both the loudspeaker and the handset transducer differential amplifiers (ALS, AHS) are provided. These amplifiers can be programmed within a range of 33 (24) dB in steps of 3 dB or muted.

A third programmable amplifier (AXC) is provided for the tip/ring interface.

The high passes (HP) have a cut-off frequency of 150 Hz. These filters can be disabled individually for the receive and transmit direction.

## 2.2 Line Powered Operation

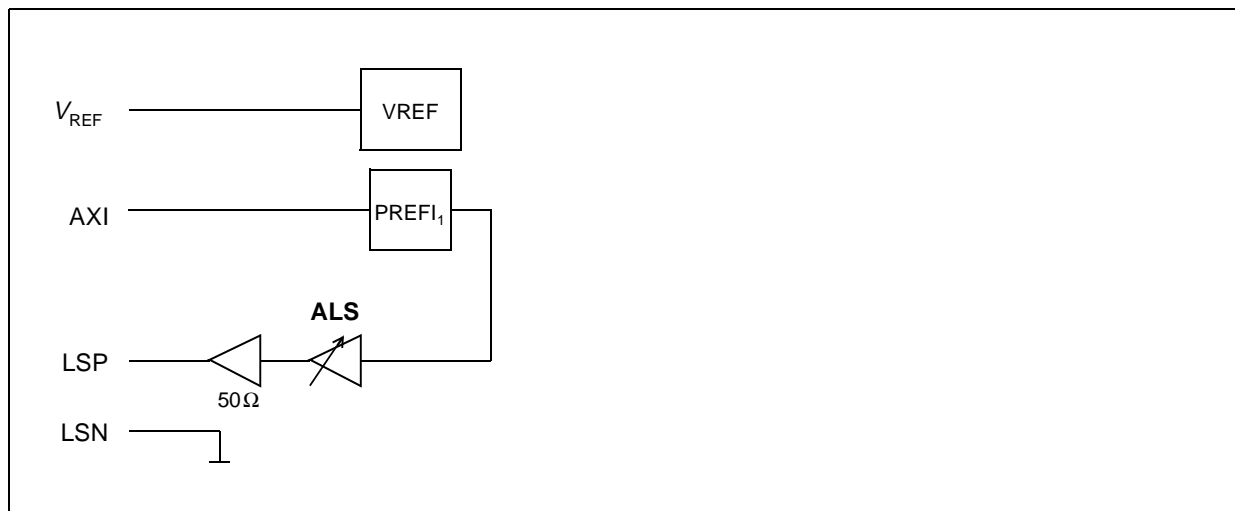
The PSB 4851 supports line powered operation by a flexible power management. The controller can power down all elements that are not needed for the current task. In particular, the following three states are useful in line powered operation:

### 1. Idle

All elements are powered down. The power consumption is minimal. This state is automatically entered by reset.

### 2. Ringing

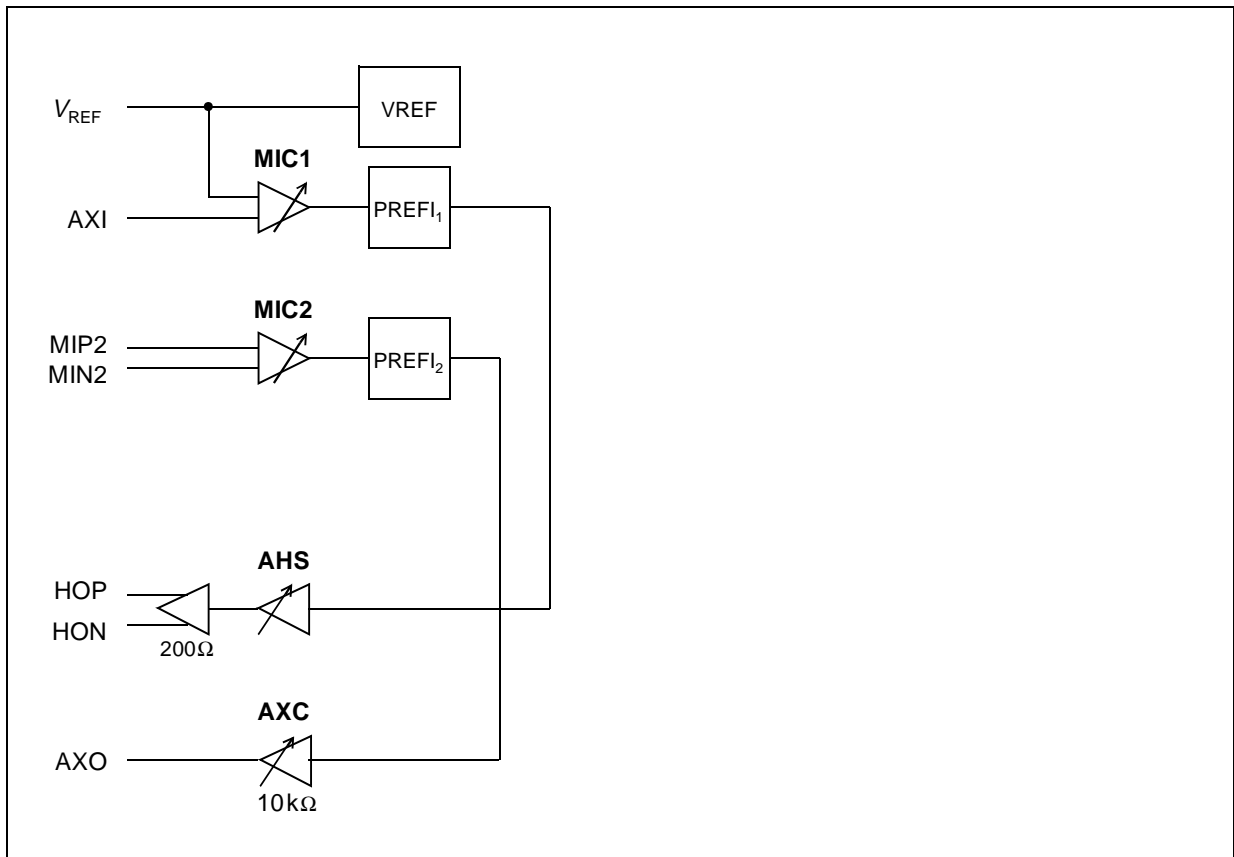
MIC1 is in bypass mode, PREFI1 is powered up and ALS is connected to PREFI1. Therefore a signal fed into AXI is amplified by ALS and output at LSP/LSN. In order to maximize the loudness of the ringing signal one of the output drivers of ALS (either LSP or LSN) can be forced to GND thus providing a single ended output. Figure 2-3 shows the signal routing and the remaining active elements in this mode (single ended mode).



**Figure 2-3 Emergency Ringing Mode**

### 3. Speech

MIC1, PREFI1, MIC2, PREFI2, AHS and AXC are powered up. AHS is connected to PREFI1 and AXC is connected to PREFI2. Therefore the signal fed into MIP2/MIN2 is amplified by MIC2 and AXC and output at AXO. The signal fed into AXI is amplified by MIC1 and AHS and output at HOP/HON. Figure 2-4 shows the signal routing and the active elements.



**Figure 2-4 Emergency Speech Mode**

*Note: An external circuitry should be provided to detect power failure and inform the controller. The controller in turn should reduce the gain of the ALS amplifier if necessary to avoid excessive power consumption.*

*Note: The serial control interface must remain operational even when some of the connected devices are without power supply. Some devices have clamping diodes at their inputs and might block the bus.*

## 2.3 Attenuation Plan

Figure 2-5 shows the attenuation plan for 3.3V and 5V operation. At the digital side the reference signal level is 3.14dBmO (maximum digital signal value). The stated gain settings at the amplifiers are the maximum gains for the guaranteed transmission characteristics. Values above the signal lines refer to 3.3V operation, values below the signal lines refer to 5V operation.

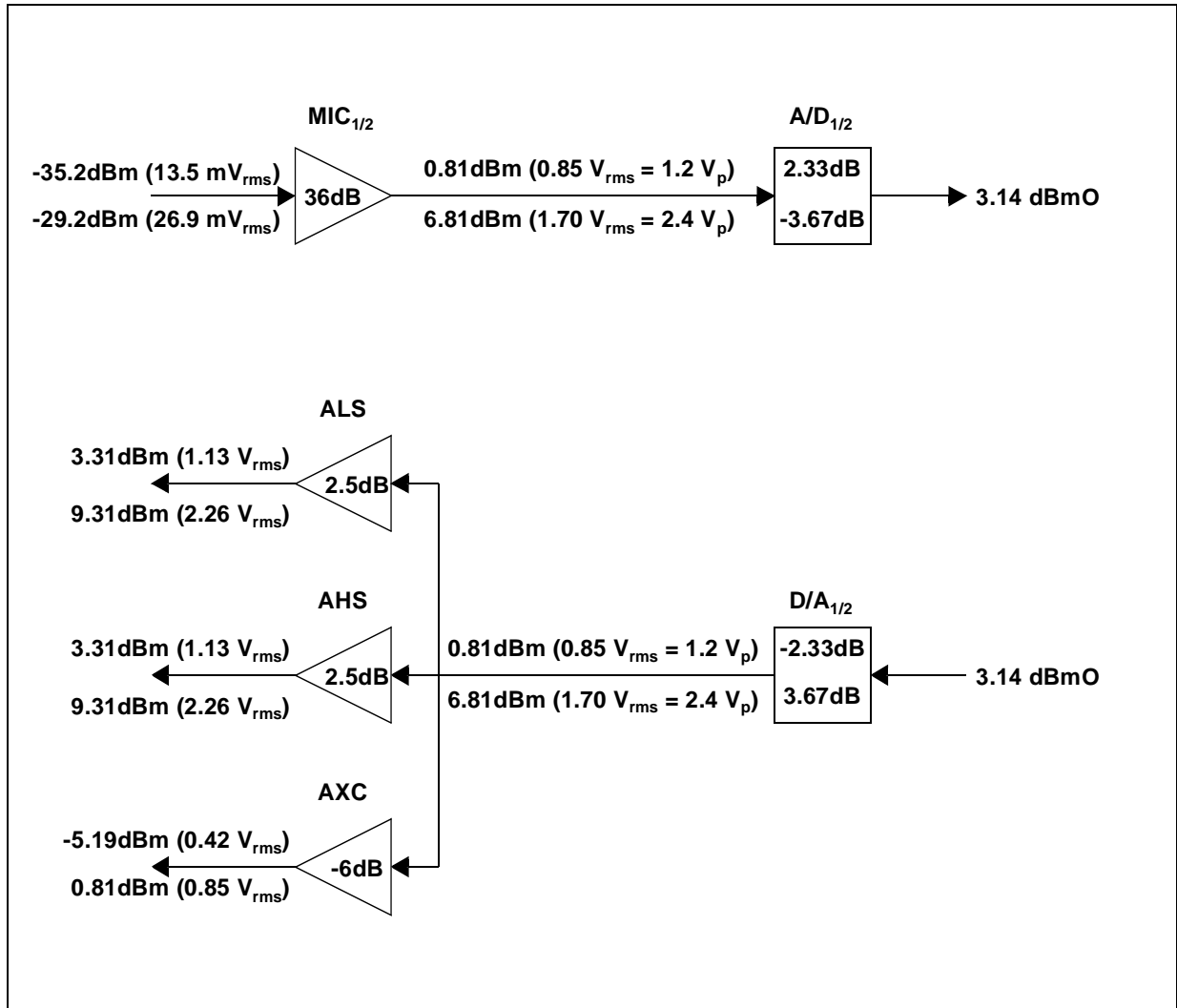


Figure 2-5 Attenuation Plan

## 2.4 Analog Front End Interface

The PSB 4851 uses a four wire interface similar to the IOM<sup>®</sup>-2 interface to exchange information in the digital domain. The main difference is that all timeslots and the channel assignments are fixed as shown in figure 2-6.

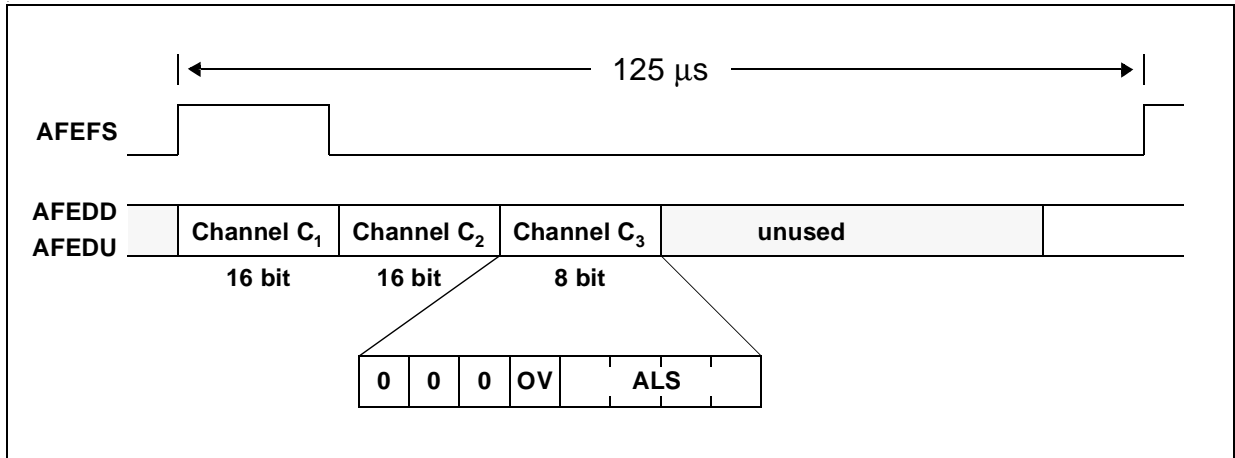


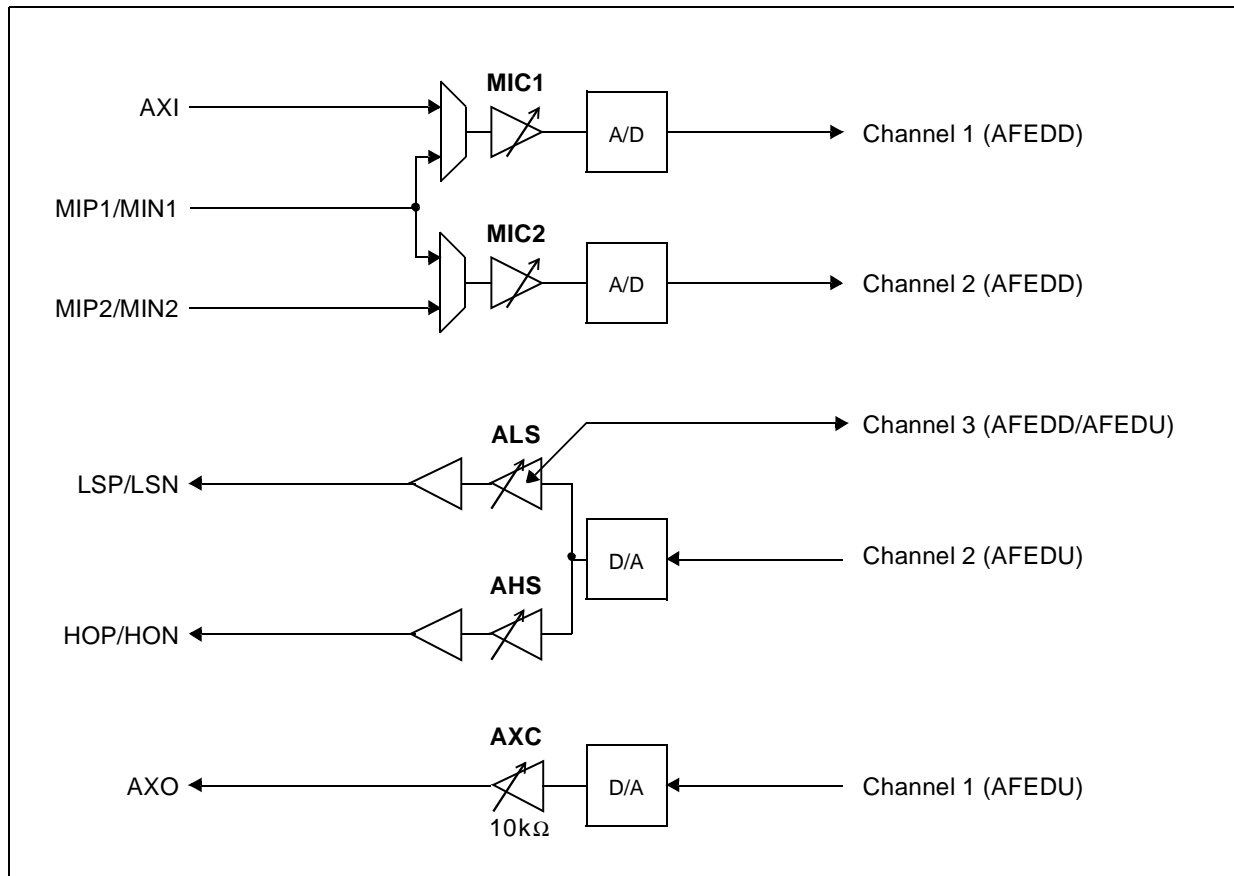
Figure 2-6 AFE Interface - Frame Structure

Voice data is transferred in 16 bit linear coding in two bidirectional channels  $C_1$  and  $C_2$ . For controlled loudhearing an auxiliary channel  $C_3$  is used to transfer the current setting of the loudspeaker amplifier ALS to the PSB 4860. The remaining bits are fixed to zero. In the other direction  $C_3$  transfers an override value for ALS from the PSB 4860 to the PSB 4851. An additional override bit OV determines if the currently transmitted value should override the AOAR:LSC setting. The AOAR:LSC setting is not affected by  $C_3$ :ALS override. Table 2-2 shows the source control of the gain for the ALS amplifier.

**Table 2-2** Source control of the gain for the ALS amplifier

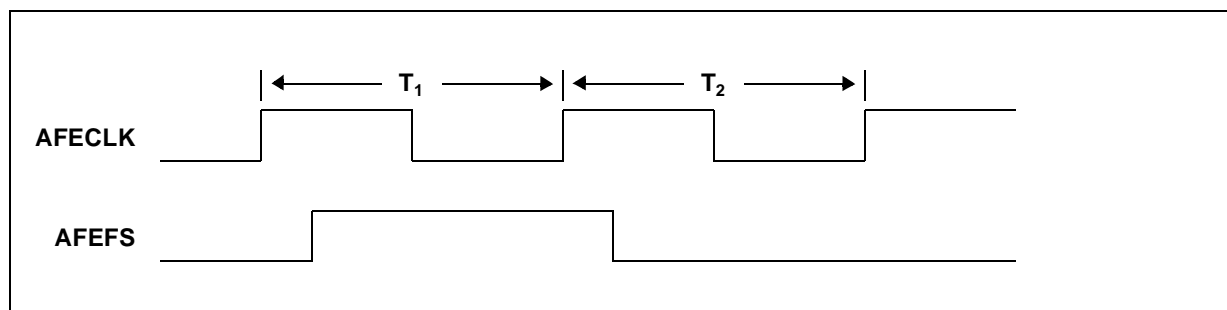
AOPR:OVRE	$C_3$ :OV	Gain of ALS amplifier
0	-	AOAR:LSC
1	0	AOAR:LSC
1	1	$C_3$ :ALS

Therefore the PSB 4860 can control the gain of the loudspeaker amplifier (ALS) independently from the gain of the handset amplifier (AHS) as shown in figure 2-7.



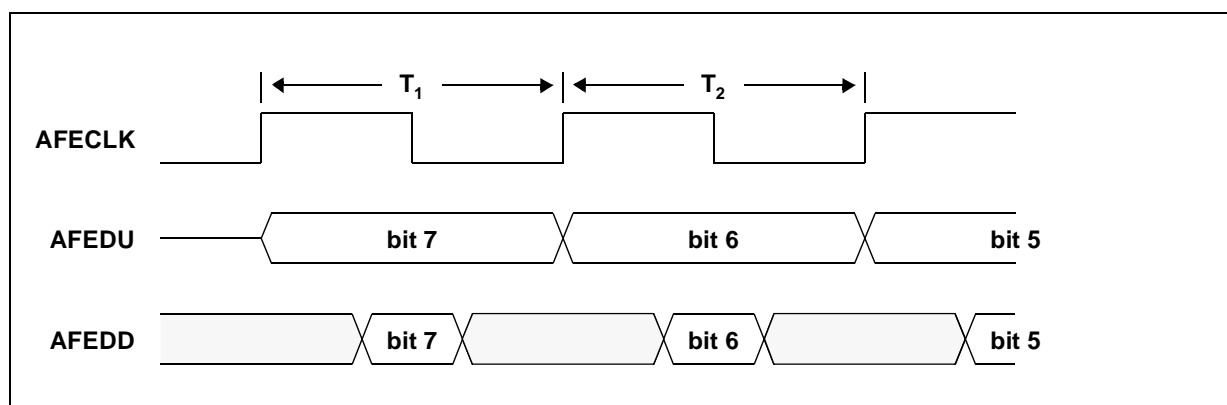
**Figure 2-7** AFE Interface - Signal Routing

Figure 2-8 shows the synchronization of a frame by AFEFS. The first clock of a new frame ( $T_1$ ) is indicated by AFEFS switching from low to high before the falling edge of  $T_1$ . AFEFS may remain high during subsequent cycles up to  $T_{32}$ .



**Figure 2-8 AFE Interface - Frame Start**

The data is shifted out with the rising edge of AFECLK and sampled at the falling edge of AFECLK (figure 2-9).



**Figure 2-9 AFE Interface - Data Transfer**

If AOPR:OVRE is not set, the channel  $C_3$  is not used by the PSB 4851. All values ( $C_1$ ,  $C_2$ ,  $C_3$ :ALS) are transferred MSB first. The data clock (AFECLK) rate is fixed at 6.912 MHz. Table 2-3 shows the clock cycles used for the three channels.

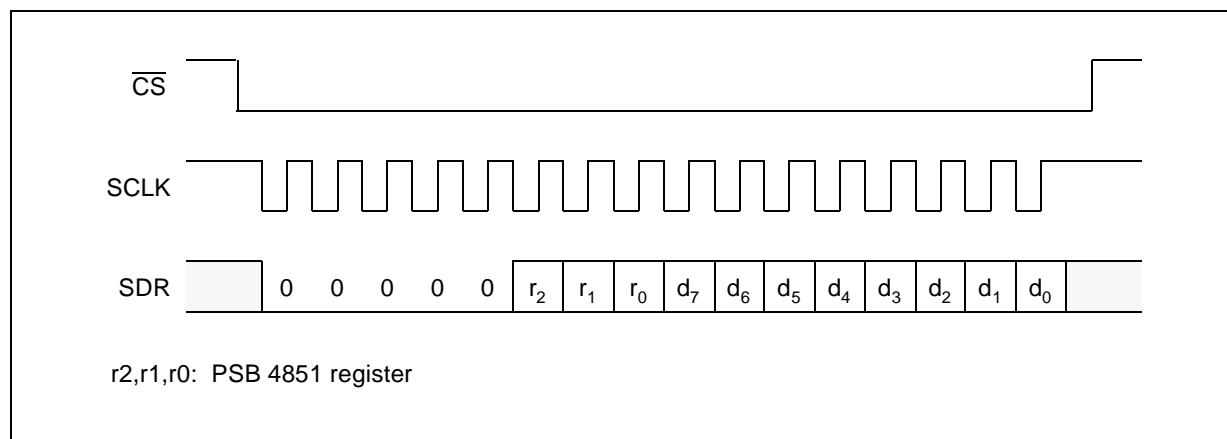
**Table 2-3 Clock cycles**

Clock Cycles	AFEDD (driven by PSB 4860)	AFEDU (driven by PSB 4851)
$T_1$ - $T_{16}$	$C_1$ data	$C_1$ data
$T_{17}$ - $T_{32}$	$C_2$ data	$C_2$ data
$T_{33}$ - $T_{40}$	$C_3$ data	$C_3$ data
$T_{41}$ - $T_{864}$	0	tristate

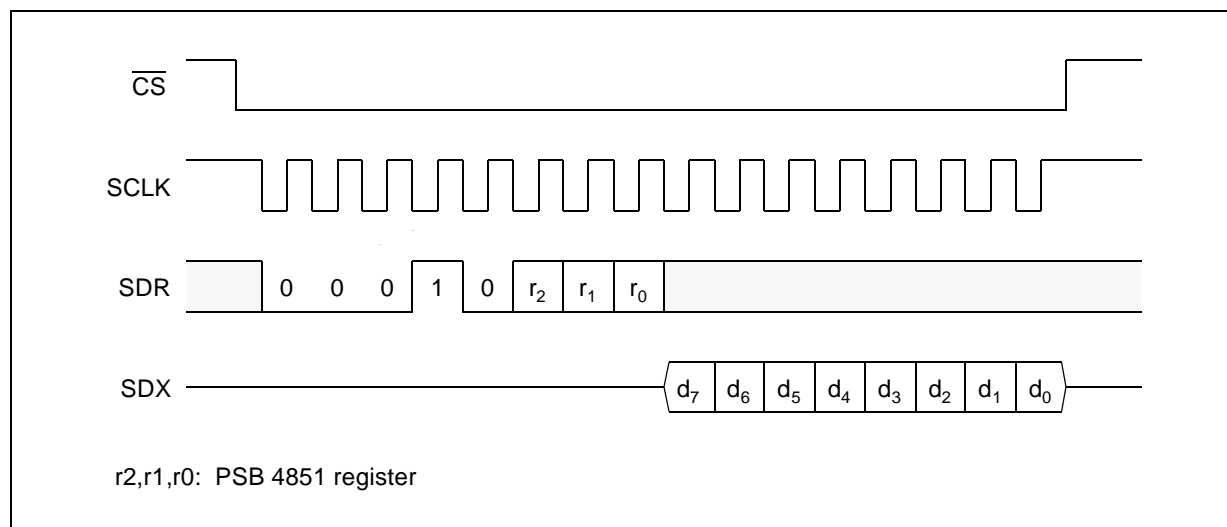
## 2.5 Serial Control Interface

The serial control interface (SCI) uses four lines. Data is transferred by the lines SDR and SDX at the rate given by SCLK. The falling edge of  $\overline{CS}$  indicates the beginning of an access. Data is sampled by the PSB 4851 at the rising edge of SCLK and shifted out at the falling edge of SCLK. Each access must be terminated by a rising edge of  $\overline{CS}$ .

Data is transferred in bytes (8 bits). Data from the controller is latched into a register at the rising edge of  $\overline{CS}$ . Figure 2-10 shows a write access to the PSB 4851 and figure 2-11 shows a read access to the PSB 4851.



**Figure 2-10 SCI Interface - Write Access**



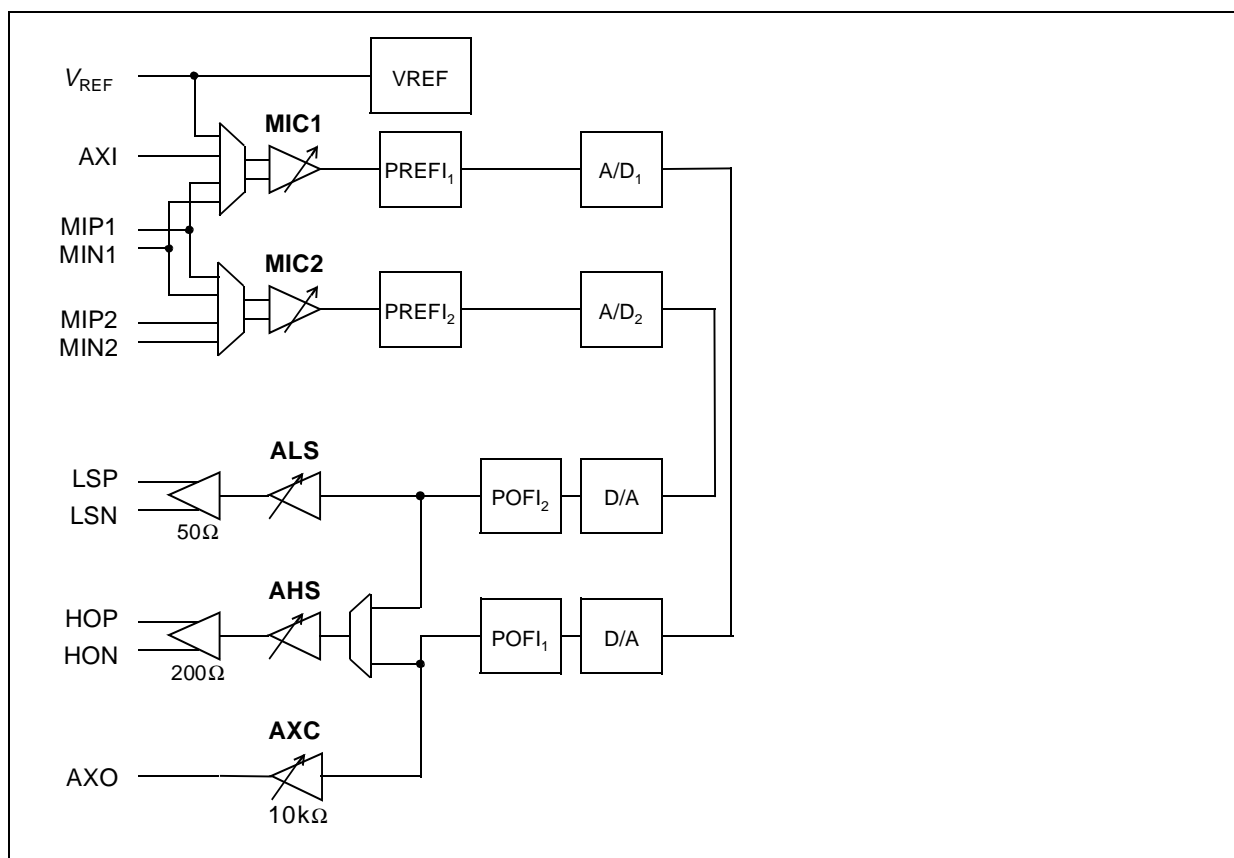
**Figure 2-11 SCI Interface - Read Access**

## 2.6 Test Loops

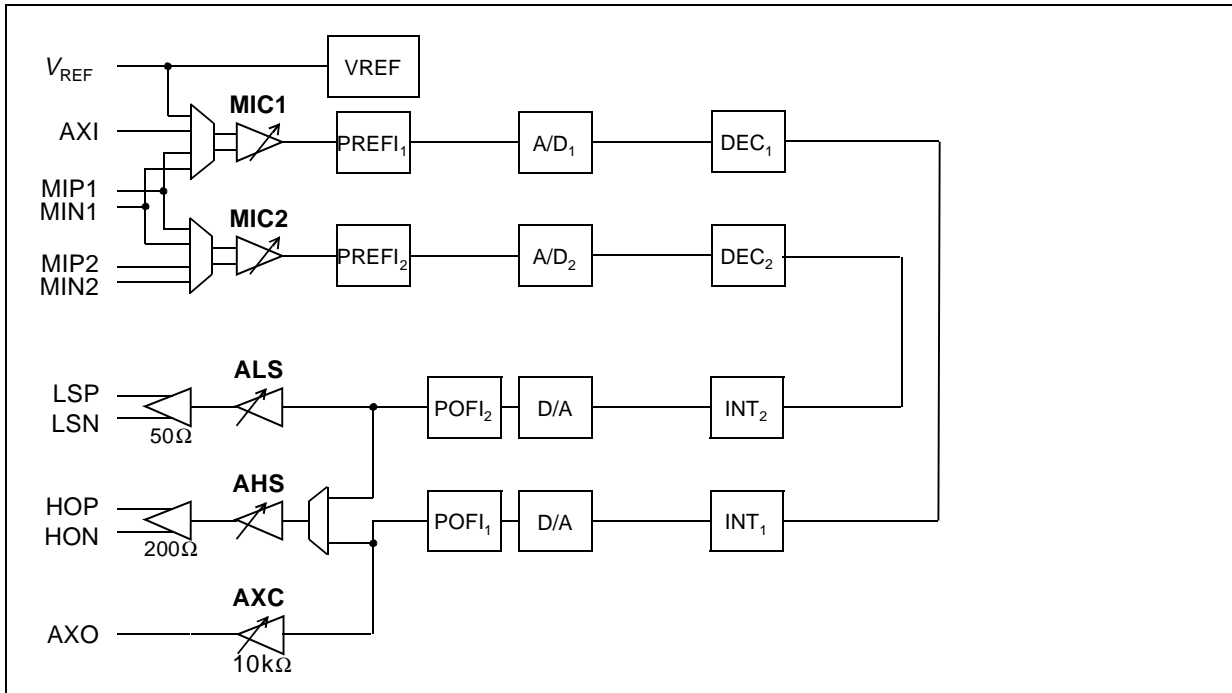
The PSB 4851 supports several internal loops for performance testing. There are two kinds of loops: analog loops and digital loops.

### 2.6.1 Analog Loops

Analog loops connect the analog inputs to the analog outputs within the PSB 4851. Emergency mode is the most basic loop involving a minimal amount of components within the chip. These loops can be programmed by the bit field ALTF of register TFCR. The next figures show the programmable loops besides emergency mode.

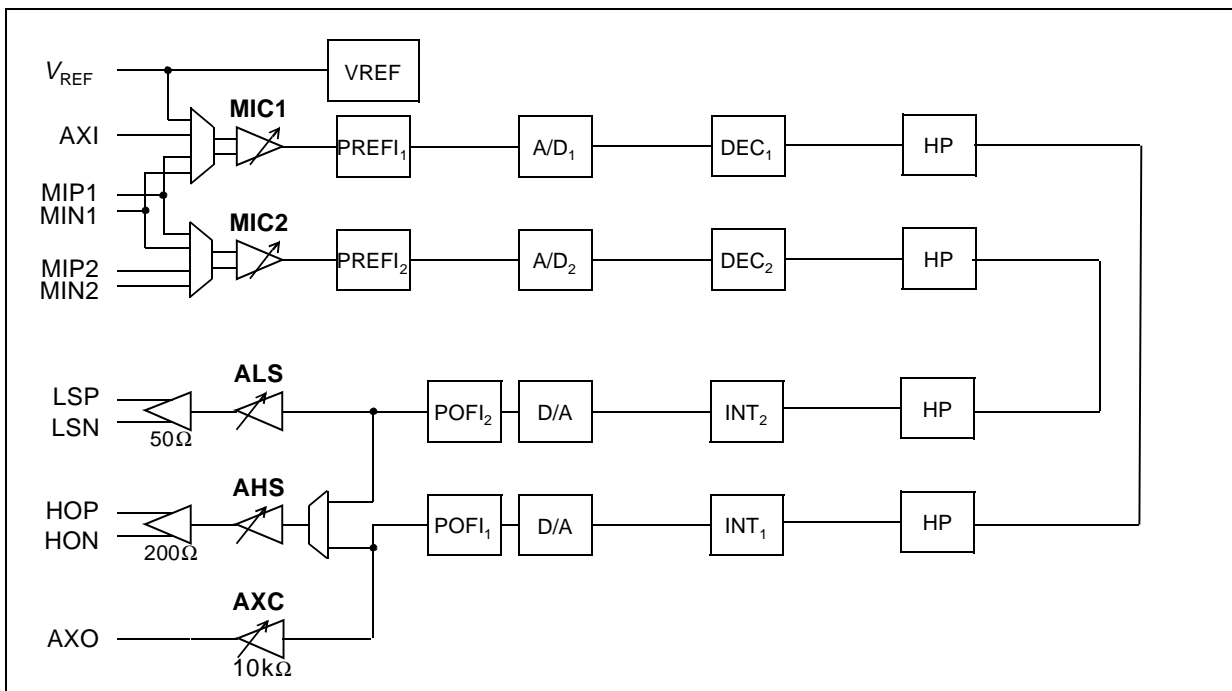


**Figure 2-12 Analog Loop via Converter**



**Figure 2-13 Analog Loop via 64 kHz**

The loop via 64 kHz incorporates only a part of the interpolation and decimation filters.



**Figure 2-14 Analog Loop via Interface**

### 2.6.2 Digital Loops

The digital loops can be programmed by bit field DLTF of register TFCR. These loops feed the input from the analog front end interface directly back to the analog front end interface. As no analog parts of the PSB 4851 are involved these loops are only useful for functional tests. The loops are as follows:

- Digital Loop via PCM register  
Received data is fed back without modification.
- Digital Loop via 64 kHz  
Received data is passed through the interpolation filters ( $INT_1$ ,  $INT_2$ ), fed back into the decimation filters ( $DEC_1$ ,  $DEC_2$ ) and then sent back.

### 3 Register Description

A summary of the registers of the PSB 4851 is presented in table 3-1.

**Table 3-1** Summary of the registers

Name	Reg	7							0
AIAR	1	MIC2				MIC1			
AIPR	2	0	ADC2	ADC1	EVREF	0	0	ISS	
AOAR	3	HOC				LSC			
AOCR	4	SEM	AXC			DHOP	DHON	DLSP	DLSN
AOPR	5	OSS			DAC2	DAC1	PSS	0	OVRE
TFCR	6	DHPR	DHPX	ALTF			DLTF		0
TMR	7	-	- <sup>1)</sup>	0	0	0	0	0	0

<sup>1)</sup> undefined

All registers except TMR are set to 0 after reset.

## REG 1: AIAR - AFE Input Amplification Register

7

0

MIC2	MIC1
------	------

### MIC1 MIC1 amplifier control

3	2	1	0	Description
0	0	0	0	MIC1 and PREFI1 are in power down mode
0	0	0	1	0 dB amplification
0	0	1	0	6 dB amplification
0	0	1	1	12 dB amplification
0	1	0	0	18 dB amplification
0	1	0	1	24 dB amplification
0	1	1	0	30 dB amplification
0	1	1	1	36 dB amplification
1	0	0	0	42 dB amplification
1	1	1	1	MIC1 is in bypass mode, PREFI1 is powered up

### MIC2 MIC2 amplifier control

7	6	5	4	Description
0	0	0	0	MIC2 and PREFI2 are in power-down mode
0	0	0	1	0 dB amplification
0	0	1	0	6 dB amplification
0	0	1	1	12 dB amplification
0	1	0	0	18 dB amplification
0	1	0	1	24 dB amplification
0	1	1	0	30 dB amplification
0	1	1	1	36 dB amplification
1	0	0	0	42 dB amplification
1	1	1	1	MIC2 is in bypass mode, PREFI2 is powered up

## REG 2: AIPR - AFE Input Path Register

7						0
0	ADC2	ADC1	EVREF	0	0	ISS

### ISS Input Source Selection

1	0	Description
0	0	reserved
0	1	AXI connected to A/D <sub>1</sub> , MIP1/MIN1 connected to A/D <sub>2</sub>
1	0	MIP1/MIN1 connected to A/D <sub>1</sub> , MIP2/MIN2 connected to A/D <sub>2</sub>
1	1	AXI connected to A/D <sub>1</sub> , MIP2/MIN2 connected to A/D <sub>2</sub>

### EVREF Enable VREF

- 0: VREF module is enabled when any other module needs the reference voltage
- 1: VREF module always enabled

### ADC1 A/D Control 1

- 0: A/D<sub>1</sub> is in power down mode
- 1: A/D<sub>1</sub> active

### ADC2 A/D Control 2

- 0: A/D<sub>2</sub> is in power down mode
- 1: A/D<sub>2</sub> active

*Note: If ADC1 and ADC2 are set to 0 then DEC<sub>1</sub>, DEC<sub>2</sub>, INT<sub>1</sub>, INT<sub>2</sub> and the timing generation are also forced into power down mode.*

### REG 3: AOAR - AFE Output Amplification Register

7

0

HOC	LSC
-----	-----

#### LSC Loudspeaker Amplifier Control

3	2	1	0	Description
0	0	0	0	ALS is in power-down mode
0	0	0	1	11.5 dB amplification
0	0	1	0	8.5 dB amplification
0	0	1	1	5.5 dB amplification
0	1	0	0	2.5 dB amplification
0	1	0	1	-0.5 dB amplification
0	1	1	0	-3.5 dB amplification
0	1	1	1	-6.5 dB amplification
1	0	0	0	-9.5 dB amplification
1	0	0	1	-12.5 dB amplification
1	0	1	0	-15.5 dB amplification
1	0	1	1	-18.5 dB amplification
1	1	0	0	-21.5 dB amplification
1	1	1	1	reserved

#### HOC Handset Amplifier Control

7	6	5	4	Description
0	0	0	0	AHS is in power-down mode
0	0	0	1	2.5 dB amplification
0	0	1	0	-0.5 dB amplification
0	0	1	1	-3.5 dB amplification
0	1	0	0	-6.5 dB amplification
0	1	0	1	-9.5 dB amplification
0	1	1	0	-12.5 dB amplification
0	1	1	1	-15.5 dB amplification

7	6	5	4	Description
1	0	0	0	-18.5 dB amplification
1	0	0	1	-21.5 dB amplification
1	1	1	1	reserved

## REG 4: AOCR - AFE Output Configuration Register

7					0
SEM	AXC	DHOP	DHON	DLSP	DLSN

### DLSN Disable Loudspeaker Amplifier Output N

- 0: LSN output of ALS amplifier controlled by LSC setting
- 1: LSN controlled by SEM setting

### DLSP Disable Loudspeaker Amplifier Output P

- 0: LSP output of ALS amplifier controlled by LSC setting
- 1: LSP controlled by SEM setting

### DHON Disable Handset Amplifier Output N

- 0: HON output of AHS amplifier controlled by HOC setting
- 1: HON output of AHS amplifier disabled (power down)

### DHOP Disable Handset Amplifier Output P

- 0: HOP output of AHS amplifier controlled by HOC setting
- 1: HOP output of AHS amplifier disabled (power down)

### AXC Auxiliary Output Control

6	5	4	Description
0	0	0	AXO is in power-down mode
0	0	1	-6 dB amplification
0	1	0	-9 dB amplification
0	1	1	-12 dB amplification
1	0	0	-15 dB amplification
1	0	1	-18 dB amplification
1	1	0	-21 dB amplification
1	1	1	-24 dB amplification

### SEM Single Ended Mode

- 0: LSN (LSP) fixed to GND
- 1: LSN (LSP) tristated

## REG 5: AOPR - AFE Output Path Register

7						0
OSS	DAC2	DAC1	PSS	0	OVRE	

### OVRE Override Enable

- 0: Gain for ALS is always defined by LSC
- 1: Gain for ALS can be overridden by interchip communication

### DAC1 D/A Control 1

- 0: POFl<sub>1</sub> and D/A<sub>1</sub> are in power down mode
- 1: POFl<sub>1</sub> and D/A<sub>1</sub> are active

### PSS Power Supply Selection

- 0: 3.3V power supply
- 1: 5V power supply

### DAC2 D/A Control 2

- 0: POFl<sub>2</sub> and D/A<sub>2</sub> are in power down mode
- 1: POFl<sub>2</sub> and D/A<sub>2</sub> are active

### OSS Output Source Selection

7	6	5	Description
0	0	0	ALS and AHS are connected to PREFI <sub>1</sub> , AXC is connected to PREFI <sub>2</sub> , POFl <sub>1</sub> and POFl <sub>2</sub> must be set to power down
0	0	1	ALS is connected to PREFI <sub>1</sub> , AHS and AXC are connected to PREFI <sub>2</sub> , POFl <sub>1</sub> and POFl <sub>2</sub> must be set to power down
0	1	-	reserved
1	0	-	reserved
1	1	0	ALS and AHS are connected to POFl <sub>2</sub> , AXC is connected to POFl <sub>1</sub>
1	1	1	ALS is connected to POFl <sub>2</sub> , AHS and AXC are connected to POFl <sub>1</sub>

## REG 6: TFCR - Test Function Configuration Register

7				0
DHPR	DHPX	ALTF	DLTF	0

### DHPR Disable High-Pass (Receive Direction)

0: High Pass activated (Receive)

1: High Pass disabled (Receive)

### DHPX Disable High-Pass (Transmit Direction)

0: High Pass activated (Transmit)

1: High Pass disabled (Transmit)

### ALTF Analog Loop Test Function

5	4	3	Description
0	0	0	Normal Mode
0	0	1	Analog Loop via Front End
0	1	0	Analog Loop via Converter
0	1	1	Analog Loop via 64kHz
1	0	0	Analog Loop via Interface

### DLTF Digital Loop Test Function

2	1	Description
0	0	Normal Mode
0	1	Digital Loop via PCM register
1	0	Digital Loop via 64kHz
1	1	reserved

## REG 7: TMR - Test Mode Register

7							0
- <sup>1)</sup>	-	0	0	0	0	0	0

<sup>1)</sup> undefined

This register is reserved for factory tests. Do not write this register.

## 4 Electrical Characteristics

### 4.1 Absolute Maximum Ratings

Parameter	Symbol	Limit Values	Unit
Ambient temperature under bias	$T_A^{1)}$	– 40 to 85	°C
Storage temperature	$T_{STG}$	– 65 to 125	°C
Voltage on any pin with respect to ground	$V_S$	– 0.3 to $V_{DD} + 0.3$	V
Maximum voltage on any pin	$V_{max}$	7	V

<sup>1)</sup> Reduced performance e.g. noise and gain tracking

ESD-integrity (according MIL-Std 883D, method 3015.7): 1000 V

exception: The pins #16, #18, #19 and #20 are not protected against voltage stress > 630 V

*Note: Stresses above those listed here may cause permanent damage to the device. Exposure to absolute maximum ratings conditions for extended periods may affect device reliability.*

## 4.2 DC Characteristics

The performance is guaranteed for 3.0V-3.6V or 4.75V-5.25V only. In emergency mode the supply voltage range is 3.0V-5.25V for  $V_{DD}/V_{DDA}/V_{DDP}$ .

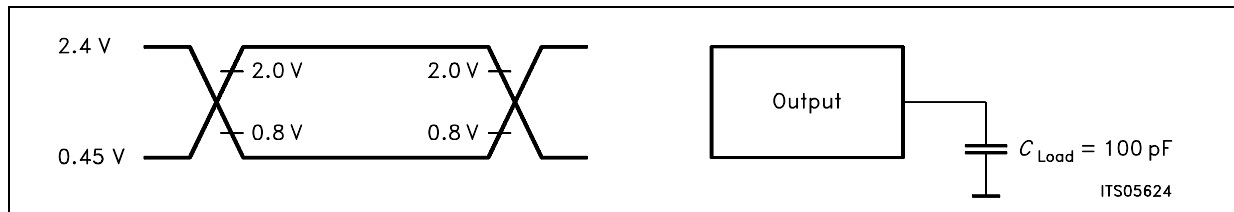
$V_{DD}/V_{DDA}/V_{DDP} = 3.0\text{V}-3.6\text{V}$  or  $4.75\text{V}-5.25\text{V}$ ;  $V_{SSD}/V_{SSA}/V_{SSP} = 0\text{ V}$ ;  $T_A = 0$  to  $70\text{ }^{\circ}\text{C}$

Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.		
Input leakage current	$I_{IL}$	- 1.0		1.0	$\mu\text{A}$	$0\text{ V} \leq V_{IN} \leq V_{DD}$
H-input level	$V_{IH}$	2.0		$V_{DD} + 0.3$	V	
L-input level	$V_{IL}$	- 0.3		0.8	V	
H-output level	$V_{OH}$	$V_{DD} - 0.45$			V	$I_O = 2\text{ mA}$
L-output level	$V_{OL}$			0.45	V	$I_O = - 2\text{ mA}$
Input capacitance	$C_I$			10	pF	
Output capacitance	$C_O$			15	pF	
$V_{DD}$ standby supply current	$I_{DDS1}$			200	$\mu\text{A}$	$V_{dd}=5\text{ V}$ , power down (after reset), no clock on AFECLK,
				50	$\mu\text{A}$	$V_{dd}=3.3\text{ V}$ , power down (after reset), no clock on AFECLK,
	$I_{DDS2}$			1.8	mA	$V_{VREF} = \text{ON}$
$V_{DD}$ supply current operating <sup>1)</sup>	$I_{DDO1}$			3.8	mA	emergency ringing via ALS (single ended mode)
	$I_{DDO2}$			5.1	mA	emergency speech mode (AHS in differential mode)
	$I_{DDO3}$			20.0	mA	full operation (loudhearing)

<sup>1)</sup> Operating power dissipation is measured with all analog outputs open. All analog inputs are set to  $V_{REF}$ .

### 4.3 AC Characteristics

Digital inputs are driven to 2.4 V for a logical “1” and to 0.45 V for a logical “0”. Timing reference points are 2V and 0.8 V. The AC-testing waveforms are shown below.



**Figure 38**  
**Input/Output Waveforms for AC-Tests**

The performance is guaranteed for 3.0V-3.6V aor 4.75V-5.25V only. In emergency mode the supply voltage range is 3.0V-5.25V for  $V_{DD0}/V_{DDA}/V_{DDP}$ .

#### Analog Front End Input Characteristics <sup>1)</sup>

Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.		
AXI-input impedance	$Z_{AXI}$	12.5	20.5		k $\Omega$	300 – 3400 Hz
AXI-input voltage swing	$V_{AXI}$			19	mVpk	42 dB, $V_{dd}=5$ V
AXI-input voltage swing	$V_{AXI}$			1.67	Vpk	0 dB, $V_{dd}=5$ V
AXI-input voltage swing	$V_{AXI}$			9.53	mVpk	42 dB, $V_{dd}=3.3$ V
AXI-input voltage swing	$V_{AXI}$			0.75	Vpk	0 dB, $V_{dd}=3.3$ V
AXI-gain	$G_{AXI}$			42	dB	9.55 mV @ 1 kHz
AXI-input impedance in bypass-mode	$Z_{AXI}$	160	270		k $\Omega$	300 – 3400 Hz
AXI-gain in bypass-mode	$G_{AXI}$			0	dB	1V @ 1 kHz
MIP/MIN1,2-input voltage swing	$V_{MIP/MIN}$			19	mVpk	42 dB, $V_{dd}=5$ V
MIP/MIN1,2-input voltage swing	$V_{MIP/MIN}$			9.53	mVpk	42 dB, $V_{dd}=3.3$ V
MIP/MIN1,2-gain	$G_{MIP/MIN}$			42	dB	9.55 mV @ 1 kHz

<sup>1)</sup> The maximum voltage swing at the internal paths corresponds to the maximum PCM-code ( $\pm 127$ ).

Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.		
RST input	$t_{RSTI}$	1			$\mu s$	

### Analog Front End Output Characteristics

AHO-output impedance	$Z_{AHO}$			2	$\Omega$	300 – 3400 Hz
AHO-output voltage swing <sup>1)</sup>	$V_{AHO}$			3.2	Vpk	$V_{dd} = 5V$ , load measured from HOP to HON
AHO-output voltage swing <sup>1)</sup>	$V_{AHO}$			1.6	Vpk	$V_{dd} = 3.3V$ , load measured from HOP to HON

ALS-output impedance	$Z_{ALS}$			2	$\Omega$	300 – 3400 Hz
ALS-output voltage swing <sup>1)</sup>	$V_{ALS}$			3.2	Vpk	$V_{dd} = 5V$ , load measured from LSP to LSN
ALS-output voltage swing <sup>1)</sup>	$V_{ALS}$			1.6	Vpk	$V_{dd} = 3.3V$ , load measured from LSP to LSN

AXO-output impedance	$Z_{AXO}$		15	21	$\Omega$	300 – 3400 Hz
AXO-output voltage swing <sup>1)</sup>	$V_{AXO}$			1.2	Vpk	$V_{dd} = 5V$ , load measured from AXO to GND
AXO-output high voltage <sup>1)</sup>	$V_{AXOH}$			3.6	Vpk	$V_{dd} = 5V$ , input load – 0.12 mA reference: GND
AXO-output low voltage <sup>1)</sup>	$V_{AXOL}$			1.2	Vpk	$V_{dd} = 5V$ , input load + 0.12 mA reference: GND
AXO-output voltage swing <sup>1)</sup>	$V_{AXO}$			0.7	Vpk	$V_{dd} = 3.3V$ , load measured from AXO to GND

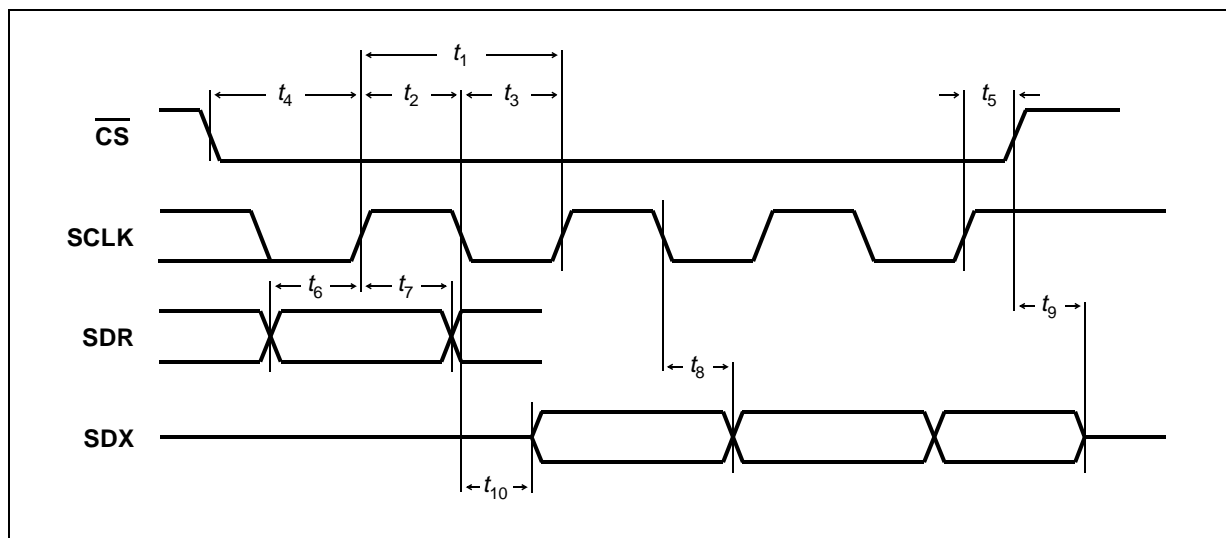
Parameter	Symbol	Limit Values			Unit	Test Condition
		min.	typ.	max.		
$V_{REF}$ output impedance	$Z_{VREF}$		3	5	$\Omega$	Load measured from $V_{REF}$ to $V_{SSA}$
$V_{REF}$ output voltage	$V_{VREF}$	2.3	2.4	2.5	V	input load – 2 mA
$V_{REFBG}$ output voltage	$V_{VREFBG}$		1.2		V	with ext. capacitor

## Transmission Characteristics

$V_{DD}/V_{DDA}/V_{DDP} = 3.0V-3.6V$  or  $4.75V-5.25V$ ;  $V_{SSD}/V_{SSA}/V_{SSP} = 0 V$ ;  $T_A = 0$  to  $70\text{ }^{\circ}\text{C}$

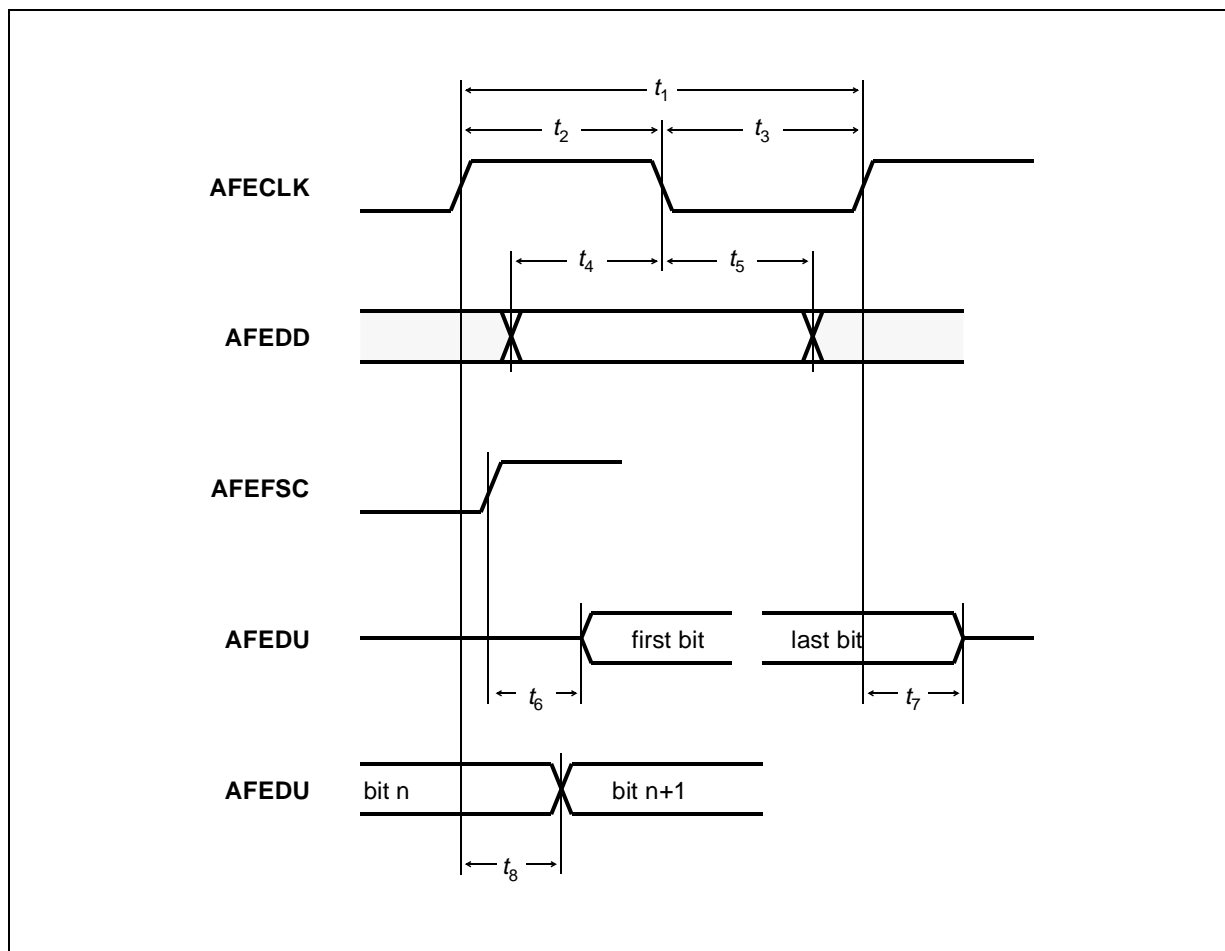
Parameter	Limit Values		Unit	Test Condition
	min.	max.		
Attenuation Distortion @ 0 dBmO	0		dB	< 200 Hz
	– 0.25		dB	200 – 300 Hz
	– 0.25	0.25	dB	300 – 2400 Hz
	– 0.25	0.45	dB	2400 – 3000 Hz
	– 0.25	0.9	dB	3000 – 3400 Hz
	0		dB	> 3400 Hz
Out-of-band signals		– 35	dB	receive signal filtering: 4.6 kHz
		– 45	dB	>8.0 kHz
		– 35	dB	transmit: 4.6 kHz
		– 40	dB	>8.0 kHz
Group delay distortion @ 0 dBmO		750	μs	500 – 600 Hz
		380	μs	600 – 1000 Hz
		130	μs	1000 – 2600 Hz
		750	μs	2600 – 2800 Hz
Signal-to-total distortion (sine signal)	50		dB	0 to – 20dBm0 <sup>1)</sup>
	39		dB	– 30 dBm0
	29		dB	– 40 dBm0
	24			– 45 dBm0
Gain tracking (sine signal) @ – 10 dBmO	– 0.3	0.3	dB	3 to – 40 dB
	– 0.6	0.6	dB	– 40 to – 50 dB
	– 1.6	1.6	dB	– 50 to – 55 dB
Idle-channel noise (Psophometric)		– 75	dBmO	receive
		– 66	dBmO	transmit
Channel crosstalk		– 75	dB	Reference: 0 dBmO
Programmable gain	– 0.5	0.5	dB	step accuracy
	– 1.0	1.0	dB	overall accuracy

<sup>1)</sup> For single ended inputs only within gain settings 0 dB to 24 dB at  $V_{dd} = 5V$  and within gain settings 0 dB to 18 dB at  $V_{dd} = 3.3V$ .  
For differential inputs 0 dB to 36 dB.

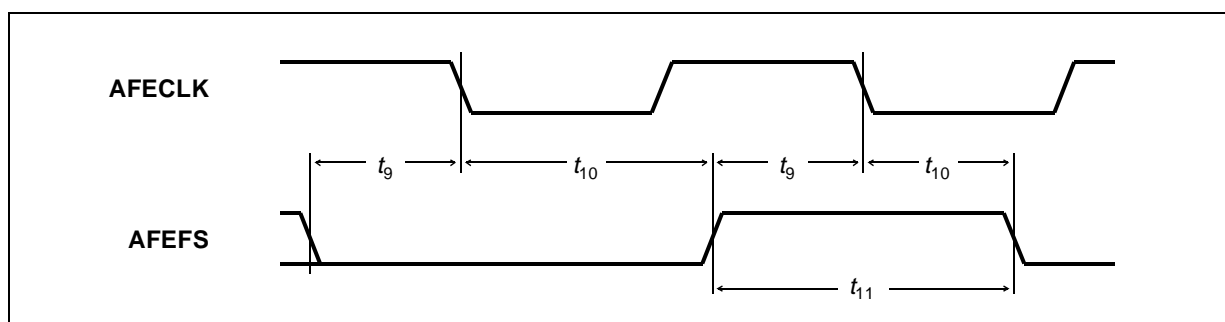


**Figure 4-1 SCI Interface**

Parameter SCI Interface	Symbol	Limit values		Unit
		Min	Max	
SCLK cycle time	$t_1$	500		ns
SCLK high time	$t_2$	100		ns
SCLK low time	$t_3$	100		ns
$\overline{\text{CS}}$ setup time	$t_4$	0		ns
$\overline{\text{CS}}$ hold time	$t_5$	10		ns
SDR setup time	$t_6$	40		ns
SDR hold time	$t_7$	40		ns
SDX data out delay	$t_8$		80	ns
$\overline{\text{CS}}$ high to SDX tristate	$t_9$		40	ns
SCLK to SDX active	$t_{10}$		80	ns

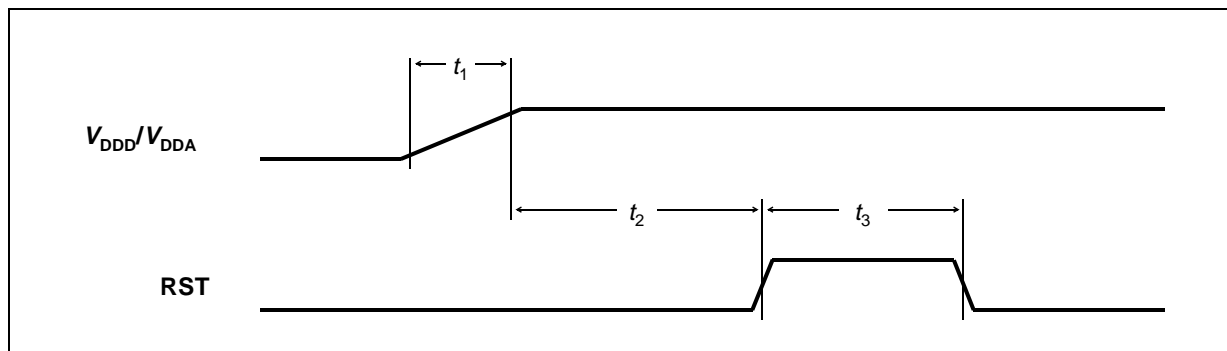


**Figure 4-2 AFE Interface - Bit Synchronization Timing**



**Figure 4-3 AFE Interface - Frame Synchronization Timing**

Parameter AFE Interface	Symbol	Limit values		Unit
		Min	Max	
AFECLK period	$t_1$	125	165	ns
AFECLK high	$t_2$	50		ns
AFECLK low	$t_3$	50		ns
AFEDD setup	$t_4$	20		ns
AFEDD hold	$t_5$	20		ns
AFEDU high impedance to active	$t_6$		20	ns
AFEDU from active to high impedance	$t_7$		20	ns
AFEDU output delay	$t_8$		20	ns
AFEFS setup	$t_9$	20		ns
AFEFS hold	$t_{10}$	20		ns
AFEFS high	$t_{11}$	1		$t_1$

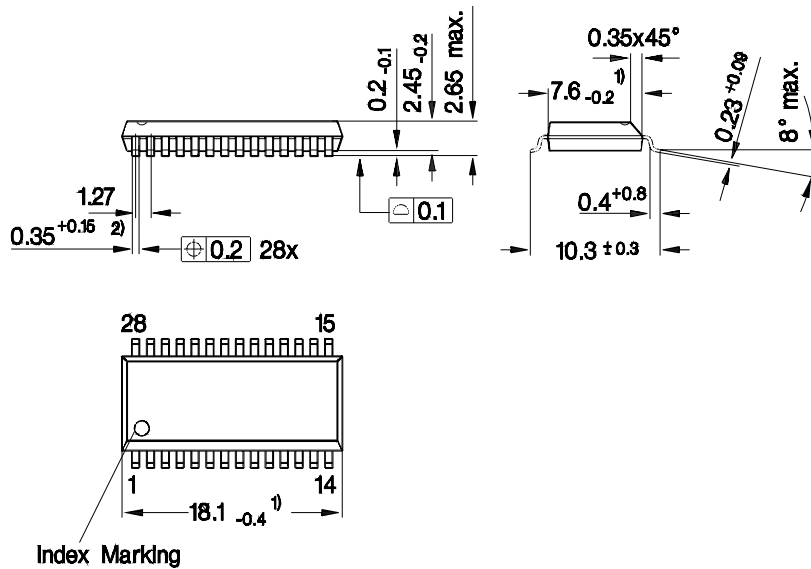


**Figure 4-4 Power and Reset Timing**

Parameter Power and Reset Timing	Symbol	Limit values		Unit
		Min	Max	
$V_{DDDD}/V_{DDDA}$ rise time 5%-95%	$t_1$		20	ms
Supply voltages stable to RST high	$t_2$		100	ns
RST high	$t_3$	100		ns

## 5 Package Outlines

### Plastic Package, P-DSO-28-1 (SMD) (Dual Small Outline)



- 1) Does not include plastic or metal protrusion of 0.15 max. per side
- 2) Does not include dambar protrusion

Sorts of Packing

Package outlines for tubes, trays etc. are contained in our

SMD = Surface Mounted Device

Dimensions in mm