



High Efficiency Synchronous Step-Up DC/DC Converter

FEATURES

- High Efficiency 93% (V_{IN}=2.4V, V_{OUT}=3.3V, I_{OUT}=200mA).
- Output Current up to 500mA. (AIC1620 at $V_{\text{IN}}\text{=}2.4\text{V}$ and $V_{\text{OUT}}\text{=}3.3\text{V})$
- 20µA Quiescent Supply Current.
- Power-Saving Shutdown Mode (0.1µA typical).
- Internal Synchronous Rectifier (no external diode)
- On-Chip Low Battery Detector.
- Low Battery Hysteresis.
- Space-Saving Package: MSOP-8

APPLICATIONS

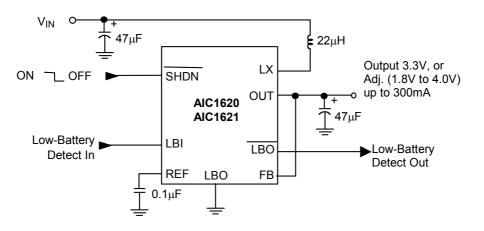
- Palmtop & Notebook Computers.
- PDAs
- Wireless Phones
- Pocket Organizers.
- Digital Cameras.
- Hand-Held Devices with 1 to 2-Cell of NiMH/NiCd Batteries.

DESCRIPTION

The AIC1620/AIC1621 are high efficiency step up DC-DC converter. The start-up voltage is as low as 0.8V and operate with an input voltage down to 0.7V. Consuming only 20μ A of quiescent current. These devices offer a built-in synchronous rectifier that reduces size and cost by eliminating the need for an external Schottky diode and improves overall efficiency by minimizing losses.

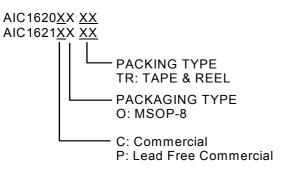
The switching frequency depends on the load and the input voltage can range up to 500KHz. The output voltage can be easily set by two external resistors from 1.8V to 4.0V, connecting FB to OUT to get 3.3V.The peak current of the internal switch is fixed at 0.8A (AIC1620), at 0.45A (AIC1621) for design flexibility.

TYPICAL APPLICATION CIRCUIT



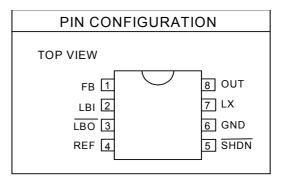


ORDERING INFORMATION



Example: AIC1620COTR

- → In MSOP-8 Package & Taping & Reel Packing Type
- AIC1620POTR
- → In MSOP-8 Lead Free Package & Taping & Reel Packing Type



ABSOLUTE MAXIMUM RATINGS

Supply Voltage (OUT to GND)	8.0V
Switch Voltage (LX to GND)	Vout+ 0.3V
SHDN, LBO to GND	6.0V
LBI, REF, FB, to GND	Vout+0.3V
Switch Current (LX)	-1.5A to +1.5A
Output Current (OUT)	-1.5A to +1.5A
Operating Temperature Range	-40°C ~ +85°C
Maximum Junction Temperature	125°C
Storage Temperature Range	-65°C ~150°C
Lead Temperature (Soldering 10 Sec.)	260°C
Absolute Maximum Ratings are those values beyond which the	life of a device may be impaired.

TEST CIRCUIT

Refer to Typical Application Circuit.



ELECTRICAL CHARACTERISTICS (VIN=2.0V, VOUT=3.3V (FB=OUT), RL=∞, TA=25°C, unless otherwise specified.) (Note1)

PARAMETER	TEST CONDITIONS		MIN.	TYP.	MAX.	UNIT
Minimum Input Voltage				0.7		V
Operating Voltage			1.1		4.0	V
Start-Up Voltage	R _L =3KΩ (N	lote2)		0.8	1.1	V
Start-Up Voltage Tempco				-2		mV/°C
Output Voltage Range			1.8		4.0	
Output Voltage	FB = V _{OUT}		3.17	3.3	3.43	V
Steady State Output Current	FB=OUT	AIC1620	300	400		
(Note 3)	(V _{OUT} =3.3V)	AIC1621	150	220		mA
Reference Voltage	I _{REF} = 0	I _{REF} = 0		1.23	1.261	V
Reference Voltage Tempco				0.024		mV/°C
Reference Load Regulation	I _{REF} = 0 to		10	30	mV	
Reference Line Regulation	V _{OUT} = 1.8V to 4V			5	10	mV/V
FB , LBI Input Threshold			1.199	1.23	1.261	V
Internal switch On-Resistance	I _{LX} = 100mA			0.3		Ω
LV Quitab Queent Limit	AIC1620	0.6 0.8		1.0		
LX Switch Current Limit	AIC1621	0.3	0.45	0.6	— A	
LX Leakage Current	V _{LX} =0V, 4V; V _{OUT} =4V			0.05	1	μA
Operating Current into OUT	V _{FB} = 1.4V , V _{OUT} = 3.3V			20	35	μA
(Note 4)						
Shutdown Current into OUT	SHDN = GND			0.1	1	μA
	V _{OUT} = 3.3V ,I _{LOAD} = 200mA			90		%
Efficiency	V _{OUT} = 2V ,I _{LOAD} = 1mA			85		
LX Switch On-Time	V _{FB} =1V , '	2	4	7	μS	
LX Switch Off-Time	V _{FB} =1V , '	0.6	0.9	1.3	μS	
FB Input Current	V _{FB} = 1.4V		0.03	50	nA	

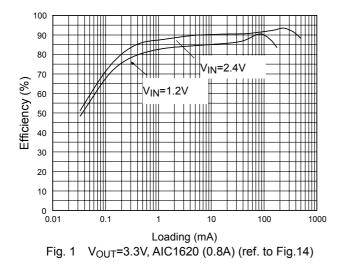


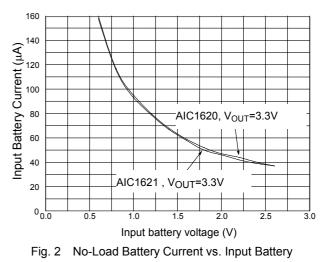
PARAMETER	TEST CONDITIONS MIN. TYP. MAX.		MAX.	UNIT	
LBI Input Current	V _{LBI} = 1.4V		1	50	nA
SHDN Input Current	$V_{\overline{SHDN}} = 0 \text{ or } V_{OUT}$		0.07	50	nA
LBO Low Output Voltage	V _{LBI} = 0, I _{SINK} = 1mA		0.2	0.4	μA
LBO Off Leakage Current	V_{LBO} = 5.5V, V_{LBI} = 5.5V		0.07	1	
LBI Hystereisis			50		mV
SHDN Input Voltage	V _{IL}			0.2Vout	V
Si Divinput voltage	V _{IH}	0.8V _{OUT}			v

ELECTRICAL CHARACTERISTICS (Continued)

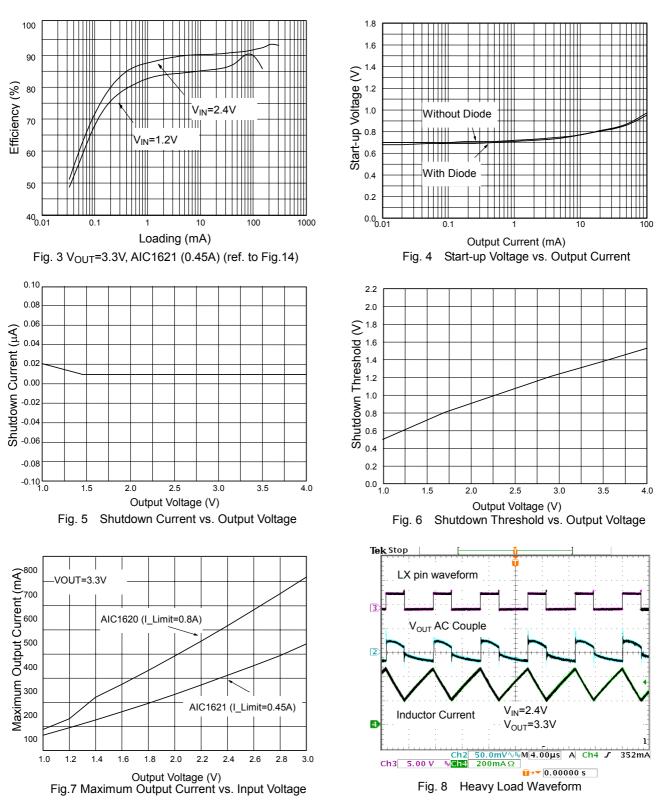
- **Note 1:** Specifications are production tested at T_A=25°C. Specifications over the -40°C to 85°C operating temperature range are assured by design, characterization and correlation with Statistical Quality Controls (SQC).
- **Note 2:** Start-up voltage operation is guaranteed without the addition of an external Schottky diode between the input and output.
- Note 3: Steady-state output current indicates that the device maintains output voltage regulation under load.
- Note 4: Device is bootstrapped (power to the IC comes from OUT). This correlates directly with the actual battery supply.

TYPICAL PERFORMANCE CHARACTERISTICS



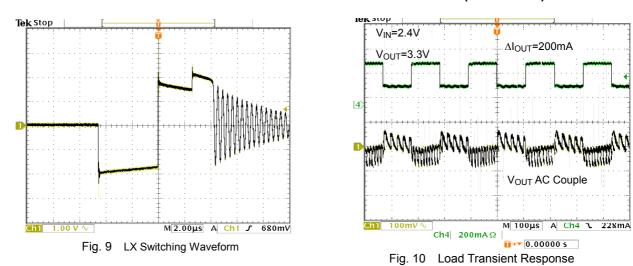






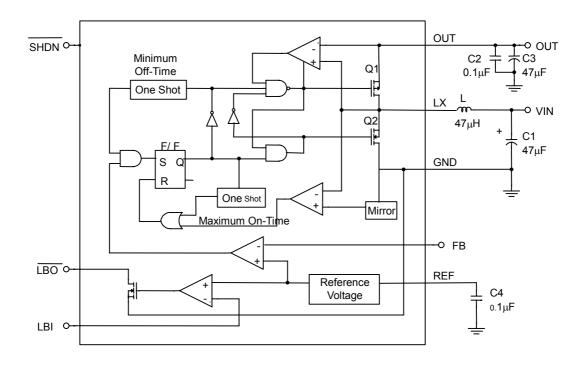
TYPICAL PERFORMANCE CHARACTERISTICS (Continued)





TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

BLOCK DIAGRAM







PIN DESCRIPTIONS

AIC1620/ AIC1621

- PIN 1: FB- Connect to OUT for +3.3V output. Use a resistor network to set the output voltage from +1.8V to +4.0V.
- PIN 2: LBI- Low-battery comparator input. Internally set to trip at +1.23V.
- PIN 3: \overline{LBO} Open-drain low battery comparator output. Output is low when V_{LBI} is <1.23V. \overline{LBO} is high impedance during shutdown.

APPLICATION INFORMATION

Overview

AIC1620/AIC1621 series are high efficiency, stepup DC-DC converters, designed to feature a built-in synchronous rectifier, which reduces size and cost by eliminating the need for an external Schottky diode. The start-up voltage of AIC1620/AIC1621 is as low as 0.8V and it operates with an input voltage down to 0.7V. Quiescent supply current is only 20µA. The internal P-MOSFET on-resistance is typically 0.3Ω to improve overall efficiency by minimizing AC losses. The output voltage can be easily set by two external resistors from 1.8V to 4.0V, connecting FB to OUT to get 3.3V. The peak current of the internal switch is fixed at 0.8A (AIC1620) or 0.45A (AIC1621) for design flexibility. The current limit of AIC1620 and AIC1621 are 0.8A and 0.45A respectively. The lower current limit allows the use of a physically smaller inductor in space-sensitive applications.

PFM Control Scheme

The key feature of the AIC1620 series is a unique minimum-off-time, constant-on-time, current-limited, pulse-frequency-modulation (PFM) control scheme (see BLOCK DIAGRAM) with the ultra-low quies-cent current. The peak current of the internal N-MOSFET power switch can be fixed at 0.8A (AIC1620) or 0.45A (AIC1621). The switch frequency depends on either loading condition or input voltage, and can range up to 500KHz. It is

PIN 4: REF-	1.23V	reference	voltage.	Bypass
with a 0.1µF capacitor.				

- PIN 5: SHDN- Shutdown input. High=operating, low=shutdown.
- PIN 6: GND- Ground
- PIN 7: LX- N-channel and P-channel power MOSFET drain.
- PIN 8: OUT- Power output. OUT provides bootstrap power to the IC.

voltage, and can range up to 500KHz. It is governed by a pair of one-shots that set a minimum off-time $(1\mu S)$ and a maximum on-time $(4\mu S)$.

Synchronous Rectification

Using the internal synchronous rectifier eliminates the need for an external Schottky diode. Therefore, the cost and board space is reduced. During the cycle of off-time, the P-MOSFET turns on and shunts the N-MOSFET. Due to the low turn-on resistance of MOSFET, the synchronous rectifier significantly improves efficiency without the addition of an external component. Thus, the conversion efficiency can be as high as 93%.

Reference Voltage

The reference voltage (REF) is nominally 1.23V for excellent T.C. performance. In addition, REF pin can source up to 100 μ A to external circuit with good load regulation (<10mV). A bypass capacitor of 0.1 μ F is required for proper operation and good performance

Shutdown

The whole circuit is shutdown when V_{SHDN} is low. At shutdown mode, the current can flow from battery to output due to body diode of the P-MOSFET. V_{OUT} falls to approximately Vin-0.6V and LX remains high



impedance. The capacitance and load at OUT determine the rate at which V_{OUT} decays. Shutdown can be pulled as high as 6V. Regardless of the voltage at OUT.

Selecting the Output Voltage

 V_{OUT} can be simply set to 3.3V by connecting FB pin to OUT due to the use of internal resistor divider in the IC (Fig. 14). In order to adjust output voltage, a resistor divider is connected to V_{OUT} , FB, GND (Fig. 15). Vout can be calculated by the following equation:

R5=R6 [(V_{OUT} / V_{REF})-1]

Where V_{REF} =1.23V and V_{OUT} may range from 1.8V to 4V. The recommended R6 is 240K Ω .

Low-Battery Detection

AIC1620 series contains an on-chip comparator with 50mV internal hysteresis (REF, REF+50mV) for low battery detection. If the voltage at LBI falls below the internal reference voltage. LBO (an open-drain output) sinks current to GND.

Component Selection

1. Inductor Selection

An inductor value of 22μ H performs well in most applications. The AIC1620 series also work with inductors in the 10μ H to 47μ H range. An inductor with higher peak inductor current tends a higher output voltage ripple (I_{PEAK}×output filter capacitor ESR). The inductor's DC resistance significantly affects efficiency. We can calculate the maximum output current as follows:

$$I_{OUT(MAX)} = \frac{V_{IN}}{V_{OUT}} \Bigg[I_{LIM} - t_{OFF} \Bigg(\frac{V_{OUT} - V_{IN}}{2 \times L} \Bigg) \Bigg] \eta$$

where $I_{OUT(MAX)}$ =maximum output current in amps

VIN=input voltage

L=inductor value in μ H η =efficiency (typically 0.9) t_{OFF}=LX switch' off-time in μ S I_{LIM}=0.45A or 0.8A

2. Capacitor Selection

The output voltage ripple relates with the peak inductor current and the output capacitor ESR. Besides output ripple voltage, the output ripple current also needs to be concerned. The smaller the capacitor ESR is, the higher the ripple current will be. A filter capacitor with low ESR is helpful to the efficiency and steady state output current of AIC1620 series. Therefore NIPPON tantalum capacitor MCM series with 100μ F/6V is recommended. A smaller capacitor (down to 47 μ F with higher ESR) is acceptable for light loads or in applications that can tolerate higher output ripple.

3. PCB Layout and Grounding

Since AIC1620's switching frequency can range up to 500kHz, it makes AIC1620 become very sensitive. So careful printed circuit layout is important for minimizing ground bounce and noise. IC's OUT pin should be as clear as possible. And the GND pin should be placed close to the ground plane. Keep the IC's GND pin and the ground leads of the input and output filter capacitors less than 0.2in (5mm) apart. In addition, keep all connection to the FB and LX pins as short as possible. In particular, when using external feedback resistors, locate them as close to the FB as possible. To maximize output power and efficiency and minimize output ripple voltage, use a ground plane and solder the IC's GND directly to the ground plane. Following are the recommended layout diagrams.





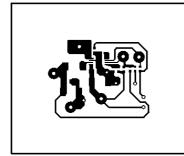


Fig.11 Top layer

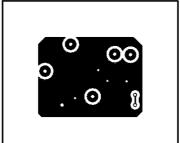


Fig.12 Bottom layer

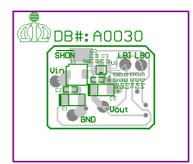
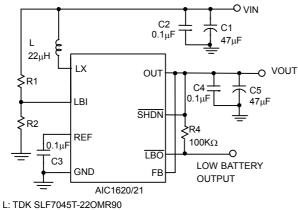


Fig.13 Placement

APPLICATION EXAMPLES



C1, C5: NIPPON Tantalum Capacitor 6MCM476MB2TER

Fig.14 V_{OUT} = 3.3V Application Circuit.

-о_{VIN} C2 C1 47μF 0.1µF T L 22μΗ Þ LX VOUT OUT -0 ŚR1 C4 C5 0.1µF LBI 47μF SHDN 100KΩ ₹R5 REF ₹R4 0.1μF LBO 0 С3 LOW BATTERY GND FΒ OUTPUT AIC1620/21 ₹R6 L: TDK SLF7045T-22OMR90 C1, C5: NIPPON Tantalum Capacitor 6MCM476MB2TER $V_{OUT}=V_{REF}*(1+R5/R6)$

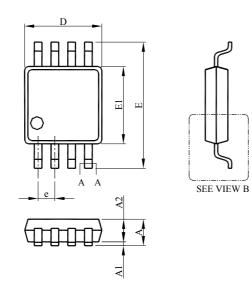
Fig.15 An Adjustable Output Application Circuit



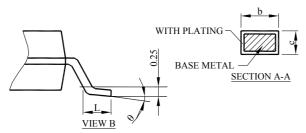
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PHYSICAL DIMENSION (unit: mm)

MSOP-8



s	MSOP-8		
S Y B O L	MILLIMETERS		
O L	MIN.	MAX.	
А		1.10	
A1	0.05	0.15	
A2	0.75	0.95	
b	0.25	0.40	
с	0.13	0.23	
D	2.90	3.10	
E	4.90 BSC		
E1	2.90	3.10	
е	0.65 BSC		
L	0.40	0.70	
θ	0°	6°	



Note:

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