November 2012



# FAN48630 — 2.5 MHz, 1500 mA, Synchronous TinyBoost<sup>™</sup> Regulator with Bypass Mode

#### **Features**

- Few External Components: 0.47 µH Inductor and 0603 Case Size Input and Output Capacitors
- Input Voltage Range: 2.35 V to 5.5 V
- Fixed Output Voltage Options: 3.0 V to 5.0 V
- Maximum Continuous Load Current: 1500 mA at V<sub>IN</sub> of 2.6 V Boosting V<sub>OUT</sub> to 3.5 V
- Up to 96% Efficient
- True Bypass Operation when V<sub>IN</sub> > Target V<sub>OUT</sub>
- Internal Synchronous Rectifier
- Soft-Start with True Load Disconnect
- Forced Bypass Mode
- V<sub>SEL</sub> Control to Optimize Target V<sub>OUT</sub>
- Short-Circuit Protection
- Low Operating Quiescent Current
- 16-Bump, 0.4 mm Pitch WLCSP

## **Applications**

- Boost for Low-Voltage Li-ion Batteries, Brownout Prevention, Boosted Audio, USB OTG, and LTE / 3G RF Power
- Cell Phones, Smart Phones, Portable Instruments

## Description

The FAN48630 allows systems to take advantage of new battery chemistries that can supply significant energy when the battery voltage is lower than the required voltage for system power ICs. By combining built-in power transistors, synchronous rectification, and low supply current; this IC provides a compact solution for systems using advanced Li-lon battery chemistries.

The FAN48630 is a boost regulator designed to provide a minimum output voltage ( $V_{\text{OUT(MIN)}}$ ) from a single-cell Li-lon battery, even when the battery voltage is below system minimum. Output voltage regulation is guaranteed to a maximum load current of 1500 mA. Quiescent current in Shutdown Mode is less than 3  $\mu$ A, which maximizes battery life. The regulator transitions smoothly between Bypass and normal Boost Mode. The device can be forced into Bypass Mode to reduce quiescent current.

The FAN48630 is available in a 16-bump, 0.4 mm pitch, Wafer-Level Chip-Scale Package (WLCSP).

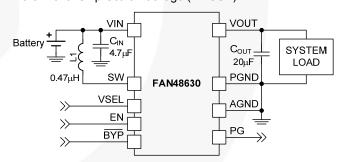


Figure 1. Typical Application

## **Ordering Information**

Part Number	Output Voltage <sup>(1)</sup> V <sub>SEL0</sub> / V <sub>SEL1</sub>	Soft- Start	Forced Bypass	Operating Temperature	Package <sup>(2)</sup>	Packing
FAN48630UC315X	3.15 / 3.33	FAST	Low I <sub>Q</sub>			
FAN48630UC33X	3.30 / 3.49	FAST	Low I <sub>Q</sub>		16-Ball, 4x4 Array, 0.4 mm Pitch,	Tana and Daal
FAN48630UC35X	3.50 / 3.70	FAST	Low I <sub>Q</sub>	-40°C to 85°C	250 µm Ball,	
FAN48630UC37AX	3.70 / 3.77	FAST	Low I <sub>Q</sub>	-40 C to 65 C	Wafer-Level Chip-	Tape and Reel
FAN48630UC45X	4.50 / 4.76	SLOW	OCP On		Scale Package (WLCSP)	
FAN48630UC50X	5.00 / 5.29	SLOW	OCP On		, ,	

#### Notes:

- 1. Other output voltages are available on request. Please contact a Fairchild Semiconductor representative.
- Includes backside laminate.

## **Block Diagram**

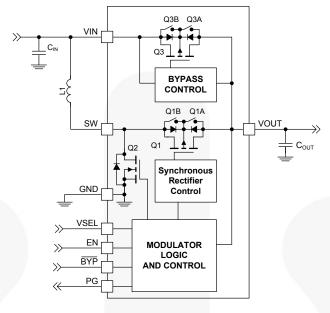


Figure 2. Block Diagram

**Table 1. Recommended Components** 

Component	Description	Vendor	Parameter	Тур.	Unit
		Toko: DFE201612C	L	0.47	μH
L1	0.47 μH, 30%	μH, 30% DFR201612C Cyntec: PIFE20161B		40	mΩ
C <sub>IN</sub>	4.7 μF, 10%, 6.3 V, X5R, 0603	Murata: GRM188R60J475K TDK: C1608X5R0J475K	С	4.7	μF
C <sub>OUT</sub>	2 x 10 μF, 20%, 10 V, X5R, 0603	TDK: C1608X5R1A106M	С	20	μF

## **Pin Configuration**

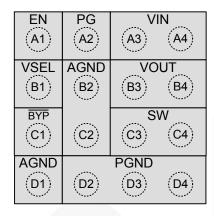


Figure 3. Top Through View (Bumps Down)

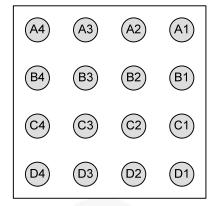


Figure 4. Bottom View (Bumps Up)

## **Pin Definitions**

Pin#	Name	Description			
A1	EN	Enable. When this pin is HIGH, the circuit is enabled.			
A2	PG	ver Good. This is an open-drain output. PG is actively pulled LOW if output falls out of ulation due to overload or if thermal protection threshold is exceeded.			
A3-A4	VIN	out Voltage. Connect to Li-Ion battery input power source.			
B1	VSEL	Output Voltage Select. When boost is running, this pin can be used to select output voltage.			
B2, C2 D1	AGND	Analog Ground. This is the signal ground reference for the IC. All voltage levels are measured with respect to this pin.			
B3-B4	VOUT	Output Voltage. Place C <sub>OUT</sub> as close as possible to the device.			
C1	BYP	<b>Bypass</b> . This pin can be used to activate Forced Bypass Mode. When this pin is LOW, the bypass switches (Q3 and Q1) are turned on and the IC is otherwise inactive.			
C3-C4	SW	Switching Node. Connect to inductor.			
D2-D4	PGND	<b>ower Ground</b> . This is the power return for the IC. The $C_{OUT}$ bypass capacitor should be sturned with the shortest path possible to these pins.			

## **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol		Parameter	Min.	Max.	Unit
V <sub>IN</sub>	VIN Input Voltage		-0.3	6.5	V
V <sub>OUT</sub>	V <sub>OUT</sub> Output Voltage	V <sub>оит</sub> Output Voltage		6.0	V
	SW Node	DC	-0.3	8.0	V
	SW Node	Transient: 10 ns, 3 MHz	-1.0	8.0	V
	Other Pins		-0.3	6.5 <sup>(3)</sup>	V
ESD	Electrostatic Discharge	Human Body Model per JESD22-A114	3.0		kV
ESD	Protection Level	Charged Device Model per JESD22-C101	1.5		kV
$T_J$	Junction Temperature		-40	+150	°C
T <sub>STG</sub>	Storage Temperature		-65	+150	°C
T <sub>L</sub>	Lead Soldering Temperature,	10 Seconds		+260	°C

#### Note:

3. Lesser of 6.5 V or  $V_{IN}$  + 0.3 V.

## **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to absolute maximum ratings.

Symbol	Parameter	Min.	Max.	Unit
V <sub>IN</sub>	Supply Voltage	2.35	5.50	V
l <sub>out</sub>	Output Current	0	1500	mA
T <sub>A</sub>	Ambient Temperature	-40	+85	°C
TJ	Junction Temperature	-40	+125	°C

## **Thermal Properties**

Junction-to-ambient thermal resistance is a function of application and board layout. This data is measured with four-layer Fairchild evaluation boards (1 oz copper on all layers). Special attention must be paid not to exceed junction temperature  $T_{J(max)}$  at a given ambient temperate  $T_A$ .

Symbol	Parameter		Unit
$\theta_{\sf JA}$	Junction-to-Ambient Thermal Resistance	80	°C/W
$\theta_{\sf JB}$	Junction-to-Board Thermal Resistance	42	C/VV

## **Electrical Specifications**

Recommended operating conditions, unless otherwise noted, circuit per Figure 1,  $V_{IN}$  = 2.35 V to  $V_{OUT}$ ,  $T_A$  = -40°C to 85°C. Typical values are given  $V_{IN}$  = 3.0 V and  $T_A$  = 25°C.

Symbol	Parameter	Conditions		Min.	Тур.	Max.	Unit
		Bypass Mode V <sub>OUT</sub> =3.5 V, V <sub>IN</sub> =4.2 V			140	190	μΑ
		Boost Mode V <sub>OUT</sub> =3.5 V,	V <sub>IN</sub> =2.5 V		150	250	μΑ
$I_Q$	V <sub>IN</sub> Quiescent Current	Shutdown: EN=0, V <sub>IN</sub> =3.	0 V		1.5	5.0	μA
		Forced Bypass Mode	Low I <sub>Q</sub>		4	10	μA
		V <sub>OUT</sub> =3.5 V, V <sub>IN</sub> =3.5 V	OCP On		45	90	μA
I <sub>LK</sub>	VOUT to VIN Reverse Leakage	V <sub>OUT</sub> =5 V, EN=0	l .		0.2	1.0	μ <b>A</b>
I <sub>LK OUT</sub>	V <sub>OUT</sub> Leakage Current	V <sub>OUT</sub> =0, EN=0, V <sub>IN</sub> =4.2 \	/		0.1	1.0	μA
V <sub>UVLO</sub>	Under-Voltage Lockout	V <sub>IN</sub> Rising			2.20	2.35	V
V <sub>UVLO_HYS</sub>	Under-Voltage Lockout Hysteresis				200		mV
$V_{PG(OL)}$	PG Low	I <sub>PG</sub> =5 mA				0.4	V
I <sub>PG_LK</sub>	PG Leakage Current	V <sub>PG</sub> =5 V				1	μΑ
V <sub>IH</sub>	Logic Level HIGH EN, VSEL, BYP			1.2			V
$V_{IL}$	Logic Level LOW EN, VSEL, BYP					0.4	V
$R_{LOW}$	Logic Control Pin Pull Downs (LOW Active)	BYP, VSEL, EN			300		kΩ
$I_{PD}$	Weak Current Source Pull-Down	BYP, VSEL, EN			100		nA
$V_{REG}$	Output Voltage Accuracy	Referred to GND, DC, V <sub>OUT</sub> -V <sub>IN</sub> > 100 mV		-2		4	%
$V_{TRSP}$	Load Transient Response	$500 - 1250 \text{ mA}, V_{IN} = 3.6 \text{ V}, V_{OUT} = 5.0 \text{ V}$			±4		%
t <sub>ON</sub>	On-Time	V <sub>IN</sub> =3.0 V, V <sub>OUT</sub> =3.5 V, Load >1000 mA			80		ns
f <sub>SW</sub>	Switching Frequency	V <sub>IN</sub> =3.6 V, V <sub>OUT</sub> =5.0 V, Load=1000 mA		2.0	2.5	3.0	MH
$I_{V\_LIM}$	Boost Valley Current Limit	V <sub>IN</sub> =2.6 V		2.6	2.9	3.1	Α
I <sub>V_LIM_SS</sub>	Boost Valley Current Limit During SS	V <sub>IN</sub> =2.6 V			1.6		Α
		V <sub>OUT</sub> =5.0 V, T <sub>J</sub> < 120°C			3.0		V
V <sub>MIN_1.5A</sub>	Minimum V <sub>IN</sub> for 1500 mA Load	V <sub>OUT</sub> =4.5 V, T <sub>J</sub> < 120°C			2.8		V
• IVIII_1.5A	(Short Term)	V <sub>OUT</sub> =3.5 V, T <sub>J</sub> < 120°C			2.35		V
		$V_{OUT}$ =3.15 V, $T_{J}$ < 120°C			2.35		V
		LIN1	Slow		350		mA
I <sub>SS_PK</sub>	Soft-Start Input Peak Current Limit		Fast		800		mA
35 <u>-</u> 1 IX		LIN2	Slow		700		mA
			Fast		1600		mA
t <sub>SS</sub>	Soft-Start EN HIGH to Regulation	Slow, 50 Ω Load			1300		μS
		Fast, 50 Ω Load			600		μS
$V_{OCP}$	OCP Comparator Threshold	V <sub>IN</sub> =5.0 V, V <sub>IN</sub> -V <sub>OUT</sub>			200		m∨
V <sub>OVP</sub>	Output Over-Voltage Protection Threshold				6.0	6.3	V
$V_{\text{OVP\_HYS}}$	Output Over-Voltage Protection Hysteresis				300		mV

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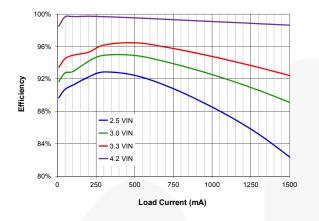
## **Electrical Specifications** (Continued)

Recommended operating conditions, unless otherwise noted, circuit per Figure 1,  $V_{IN}$  = 2.35 V to  $V_{OUT}$ ,  $T_A$  = -40°C to 85°C. Typical values are given  $V_{IN}$  = 3.0 V and  $T_A$  = 25°C.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
R <sub>DS(ON)N</sub>	N-Channel Boost Switch R <sub>DS(ON)</sub>	V <sub>IN</sub> =3.5 V, V <sub>OUT</sub> =3.5 V		85	120	mΩ
R <sub>DS(ON)P</sub>	P-Channel Sync Rectifier R <sub>DS(ON)</sub>	V <sub>IN</sub> =3.5 V, V <sub>OUT</sub> =3.5 V		65	85	mΩ
R <sub>DS(ON)P_BYP</sub>	P-Channel Bypass Switch R <sub>DS(ON)</sub>	V <sub>IN</sub> =3.5 V, V <sub>OUT</sub> =3.5 V		65	85	mΩ
T <sub>120A</sub>	T120 Activation Threshold			120		°C
T <sub>120R</sub>	T120 Release Threshold			100		°C
T <sub>150T</sub>	T150 Threshold			150		°C
T <sub>150H</sub>	T150 Hysteresis			20		°C
t <sub>RST</sub>	FAULT Restart Timer			20		ms

## **Typical Characteristics**

Unless otherwise specified;  $V_{IN}$  = 3.6 V, and  $V_{OUT}$  = 5 V, and  $T_A$  = 25°C; circuit and components according to Figure 1.



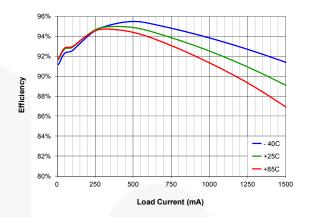
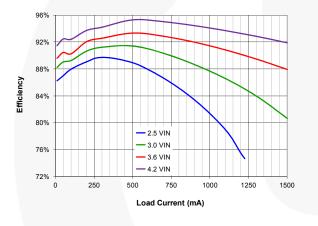


Figure 5. Efficiency vs. Load Current and Input Voltage,  $V_{\text{OUT}}$ =3.5 V

Figure 6. Efficiency vs. Load Current and Temperature,  $V_{\text{IN}}$ =3.0V,  $V_{\text{OUT}}$ =3.5 V



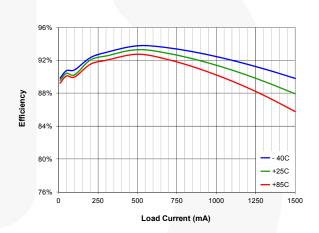
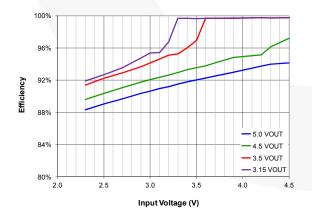


Figure 7. Efficiency vs. Load Current and Input Voltage

Figure 8. Efficiency vs. Load Current and Temperature



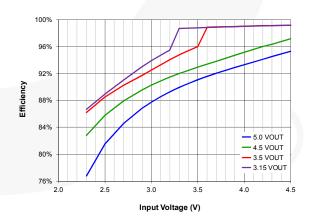
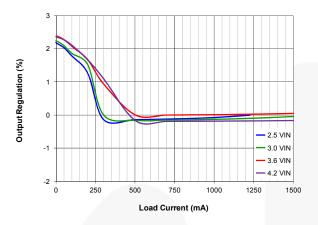


Figure 9. Efficiency vs. Input Voltage and Output Voltage, Figure 10. Efficiency vs. Input Voltage and Output Voltage, 200 mA Load

1000 mA Load

## Typical Characteristics (Continued)

Unless otherwise specified;  $V_{IN} = 3.6 \text{ V}$ ,  $V_{OUT} = 5 \text{ V}$ , and  $T_A = 25^{\circ}\text{C}$ ; circuit and components according to Figure 1.

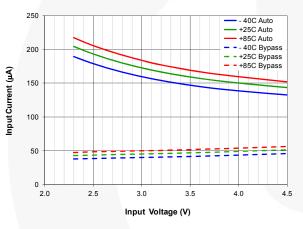


3
2
1
0
-1-40C
-+25C
-+85C
-2
0
250
500
750
1000
1250
1500

Load Current (mA)

Figure 11. Output Regulation vs. Load Current and Input Voltage (Normalized to 3.6 V<sub>IN</sub>, 500 mA Load)

Figure 12. Output Regulation vs. Load Current and Temperature (Normalized to 3.6  $V_{IN}$ , 500 mA Load,  $T_A=25^{\circ}C$ )



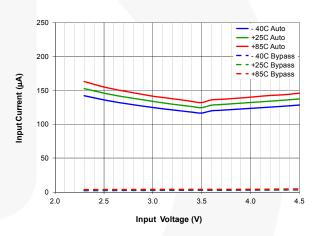
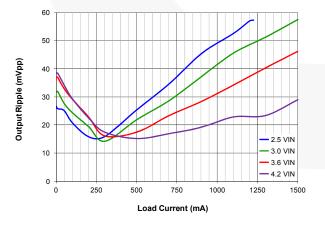


Figure 13. Quiescent Current vs. Input Voltage, Temperature and Mode, V<sub>OUT</sub>=5.0 V, Forced Bypass, OCP Active

Figure 14. Quiescent Current vs. Input Voltage, Temperature and Mode,  $V_{\text{OUT}}$ =3.5 V, Forced Bypass, Low  $I_{\text{Q}}$ 



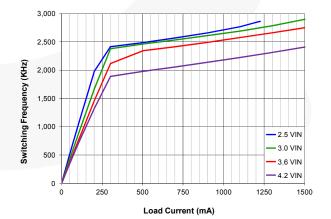
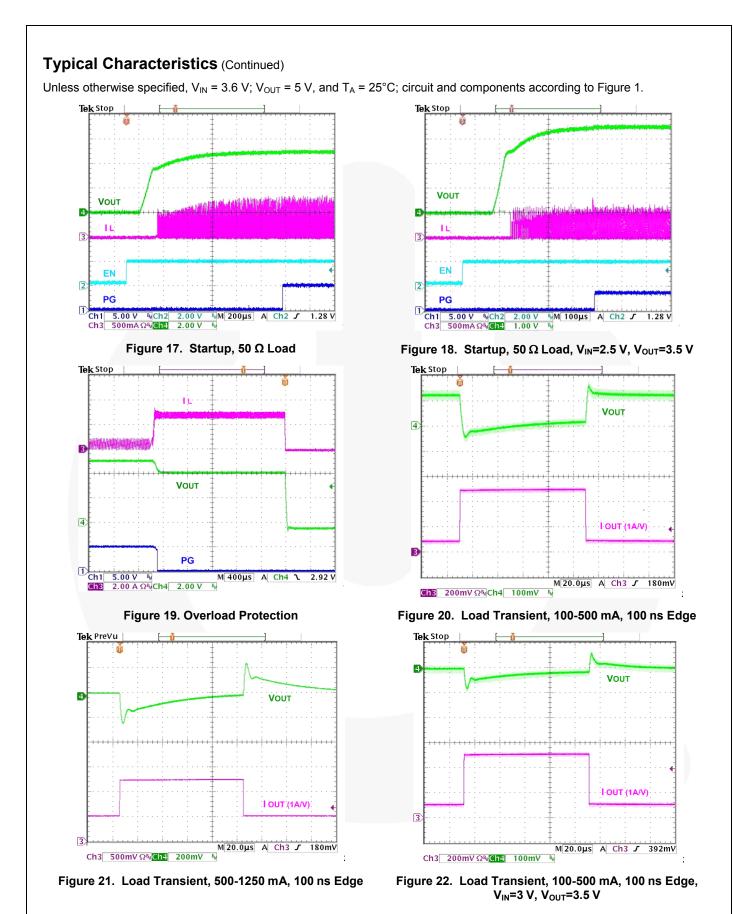


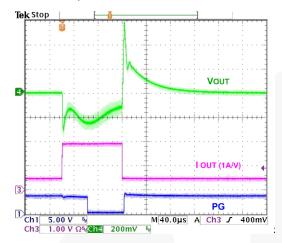
Figure 15. Output Ripple vs. Load Current and Input Voltage

Figure 16. Frequency vs. Load Current and Input Voltage



## **Typical Characteristics** (Continued)

Unless otherwise specified,  $V_{IN}$  = 3.6 V,  $V_{OUT}$  = 5 V,  $T_A$  = 25°C; circuit and components according to Figure 1.



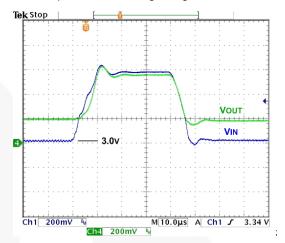
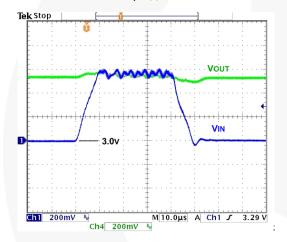


Figure 23. Load Transient, 500-1950 mA, 100 ns Edge,  $V_{IN}$ =3 V,  $V_{OUT}$ =3.5 V

Figure. 24 Line Transient, 3.0-3.6  $V_{IN}$ , 10  $\mu s$  Edge, 500 mA Load,  $V_{OUT}$ =3.15 V



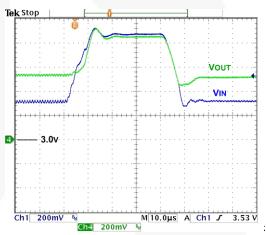
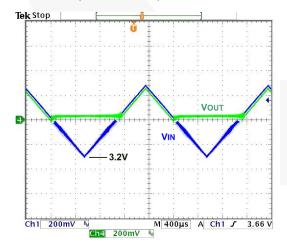


Figure 25. Line Transient, 3.0-3.6 V<sub>IN</sub>, 10 µs Edge, 1.000 mA Load, V<sub>OUT</sub>=3.5 V

Figure 26. Line Transient, 3.3-3.9  $V_{IN}$ , 10  $\mu s$  Edge, 500 mA load,  $V_{OUT}$ =3.5 V



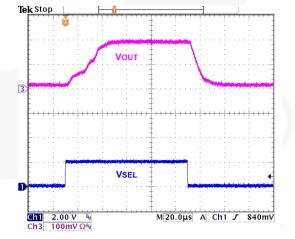


Figure. 27 Bypass Entry / Exit, Slow  $V_{\text{IN}}$  Ramp 1 ms Edge, 500 mA Load,  $V_{\text{OUT}}\text{=}3.5$  V, 3.2 - 3.8  $V_{\text{IN}}$ 

Figure 28.  $V_{SEL}$  Step,  $V_{IN}$ =3 V,  $V_{OUT}$ =3.5 V, 500 mA Load

## **Circuit Description**

FAN48630 is a synchronous boost regulator, typically operating at 2.5 MHz in Continuous Conduction Mode (CCM), which occurs at moderate to heavy load current and low  $V_{\text{IN}}$  voltages. The regulator includes a Bypass Mode that activates when  $V_{\text{IN}}$  is above the boost regulator's setpoint.

In anticipation of a heavy load transition, the setpoint can be adjusted upward by fixed amounts with the VSEL pin to reduce the required system headroom during lighter-load operation to save power.

**Table 2. Operating States** 

Mode	Description	Invoked When
LIN	Linear Startup	$V_{IN} > V_{OUT}$
SS	Boost Soft-Start	$V_{OUT} < V_{OUT(MIN)}$
BST	Boost Operating Mode	$V_{OUT} = V_{OUT(MIN)}$
BPS	True Bypass Mode	$V_{IN} > V_{OUT(MIN)}$

#### **Boost Mode**

The FAN48630 uses a current-mode modulator to achieve excellent transient response and smooth transitions between CCM and Discontinuous Conduction Mode (DCM) operation. During CCM operation, the device maintains a switching frequency of about 2.5 Mhz. In light-load operation (DCM), frequency is reduced to maintain high efficiency.

Table 3. Boost Startup Sequence

Start State	Entry	Exit	End State	Timeout (µs)
LIN1	V <sub>IN</sub> > UVLO, EN=1	V <sub>OUT</sub> > V <sub>IN</sub> - 300 mV	SS	
			LIN2	512
LIN2	LIN1 Exit	V <sub>OUT</sub> > V <sub>IN</sub> - 300 mV	SS	
		TIMEOUT	FAULT	1024
SS	LIN1 or LIN2 Exit	V <sub>OUT</sub> =V <sub>OUT(MIN)</sub>	BST	
		OVERLOAD TIMEOUT	FAULT	64

## **Shutdown and Startup**

If EN is LOW, all bias circuits are off and the regulator is in Shutdown Mode. During shutdown, current flow is prevented from  $V_{\text{IN}}$  to  $V_{\text{OUT}}$ , as well as reverse flow from  $V_{\text{OUT}}$  to  $V_{\text{IN}}$ . During startup, it is recommended to keep DC current draw below 500 mA.

#### **LIN State**

When EN is HIGH and  $V_{\text{IN}} > \text{UVLO}$ , the regulator attempts to bring  $V_{\text{OUT}}$  within 300 mV of  $V_{\text{IN}}$  using the internal fixed current source from  $V_{\text{IN}}$  (Q3). The current is limited to LIN1 set point.

If  $V_{OUT}$  reaches  $V_{IN}$ -300 mV during LIN1 Mode, the SS state is initiated. Otherwise, LIN1 times out after 512  $\mu s$  and LIN2 Mode is entered.

In LIN2 Mode, the current source is incremented to 2A. If  $V_{OUT}$  fails to reach  $V_{IN}$ -300 mV after 1024  $\mu s$ , a fault condition is declared.

#### **SS State**

Upon the successful completion of the LIN state (V<sub>OUT</sub>≥V<sub>IN</sub>-300 mV), the regulator begins switching with boost pulses current limited to 50% of nominal level.

During SS state,  $V_{\text{OUT}}$  is ramped up by stepping the internal reference. If  $V_{\text{OUT}}$  fails to reach regulation during the SS ramp sequence for more than 64  $\mu$ s, a fault condition is declared. If large  $C_{\text{OUT}}$  is used, the reference is automatically stepped slower to avoid excessive input current draw.

#### **BST State**

This is a normal operating state of the regulator.

#### BPS State

If  $V_{\text{IN}}$  is above  $V_{\text{REG}}$  when the SS Mode successfully completes, the device transitions directly to BPS Mode.

#### **FAST and SLOW Soft-Start Options**

FAN48630UC315X, FAN48630UC33X, FAN48630UC35X, and FAN48630UC37AX feature fast startup with EN to regulation time of 500 µs. LIN1 and LIN2 phase currents are doubled compared to SLOW options, SS phase is also faster.

FAN48630UC45X and FAN48630UC50XS feature low startup with EN to regulation time of 1300  $\mu$ s to reduce inrush current.

#### **FAULT State**

The regulator enters the FAULT state under any of the following conditions:

- V<sub>OUT</sub> fails to achieve the voltage required to advance from LIN state to SS state.
- V<sub>OUT</sub> fails to achieve the voltage required to advance from SS state to BST state.
- Boost current limit triggers for 2 ms during the BST state.
- V<sub>DS</sub> protection threshold is exceeded during BPS state.

Once a fault is triggered, the regulator stops switching and presents a high-impedance path between  $V_{\text{IN}}$  and  $V_{\text{OUT}}$ . After waiting 20 ms, a restart is attempted.

#### **Power Good**

Power good is 0 FAULT, 1 POWER GOOD, open-drain input.

The Power good pin is provided for signaling the system when the regulator has successfully completed soft-start and no faults have occurred. Power good also functions as an early warning flag for high die temperature and overload conditions.

- PG is released HIGH when the soft-start sequence is successfully completed.
- PG is pulled LOW when PMOS current limit has triggered for 64 µs OR the die the temperature exceeds 120°C. PG is re-asserted when the device cools below to 100°C.
- Any FAULT condition causes PG to be de-asserted.

### **Over-Temperature**

The regulator shuts down when the die temperature exceeds 150°C. Restart occurs when the IC has cooled by approximately 20°C.

#### **Bypass Operation**

In normal operation, the device automatically transitions from Boost Mode to Bypass Mode, if V<sub>IN</sub> goes above target V<sub>OUT</sub>. In Bypass Mode, the device fully enhances both Q1 and Q3 to provide a very low impedance path from VIN to VOUT. Entry to the Bypass Mode is triggered by condition where V<sub>IN</sub> > V<sub>OUT</sub> and no switching has occurred during past 5  $\mu$ s. To soften the entry to Bypass Mode, Q3 is driven as a linear current source for the first 5  $\mu$ s. Bypass Mode exit is triggered when V<sub>OUT</sub> reaches the target V<sub>OUT</sub> voltage. During Automatic Bypass Mode, the device is short-circuit protected by voltage comparator tracking the voltage drop from V<sub>IN</sub> to V<sub>OUT</sub>; if the drop exceeds 200 mV, FAULT is declared.

## **Forced Bypass**

Entry to Forced Bypass Mode initiates with a current limit on Q3 and then proceeds to a true bypass state. To prevent reverse current to the battery, the device waits until output discharges below  $V_{\text{IN}}$  before entering Forced Bypass Mode.

Low-IQ Forced Bypass Mode is available for FAN48630UC315X, FAN48630UC33X, FAN48630UC35X, and FAN48630UC37AX. After the transition is complete, most of the internal circuitry is disabled to minimize quiescent current draw. Short-circuit, UVLO, output OVP and overtemperature protections are inactive in Forced Bypass Mode.

OCP in Forced Bypass Mode is available for FAN48630UC45X and FAN48630UC50X. During Forced Bypass Mode, the device is short-circuit protected by a voltage comparator tracking the voltage drop from VIN to VOUT. If the drop exceeds 200 mV, a FAULT is declared. The over-temperature protection is also active.

In Forced Bypass Mode, V<sub>OUT</sub> can follow V<sub>IN</sub> below V<sub>OUT(MIN)</sub>.

#### **VSEL**

 $V_{SEL}$  can be asserted in anticipation of a positive load transient. Raising  $V_{SEL}$  increases  $V_{OUT(MIN)}$  by a fixed amount and  $V_{OUT}$  is stepped to the corresponding target output voltage in 20  $\mu$ s. The functionality can also be utilized to mitigate undershoot during severe line transients, while minimizing  $V_{OUT}$  during more benign operating conditions to save power.

## **Application Information**

## Output Capacitance (Cout)

#### **Stability**

The effective capacitance (C<sub>EFF</sub>) of small, high-value, ceramic capacitors decreases as bias voltage increases.

FAN48630 is guaranteed for stable operation with the minimum value of  $C_{\text{EFF}}\left(C_{\text{EFF}(MIN)}\right)$  outlined in Table 4 below.

Table 4. Minimum C<sub>EFF</sub> Required for Stability

Operating	C <sub>EFF(MIN)</sub>	
V <sub>OUT</sub> (V)	V <sub>OUT</sub> (V) I <sub>LOAD</sub> (mA)	
3.15	0 to 1500	12
3.5	0 to 1500	9
4.5 and 5	0 to 1500	6

C<sub>EFF</sub> varies with manufacturer, material, and case size.

#### Inductor selection

Recommended nominal inductance value is 0.47  $\mu H$ .

FAN48630 employs valley-current limiting; peak inductor current can reach 3.8 A for a short duration during overload conditions. Saturation effects cause the inductor current ripple to become higher under high loading as only valley of the inductor current ripple is controlled.

For FAN48630UC315X and FAN48630UC33X, a  $0.33~\mu H$  inductor can be used for improved transient performance.

#### Startup

Input current limiting is in effect during soft-start, which limits the current available to charge  $C_{\text{OUT}}$  and any additional capacitance on the  $V_{\text{OUT}}$  line. If the output fails to achieve regulation within the limits described in the Startup section, a FAULT occurs, causing the circuit to shut down then restart after a significant time period. If the total combined output capacitance is very high, the circuit may not start on the first attempt, but eventually achieves regulation if no load is present. If a high-current load and high capacitance are both present during soft-start, the circuit may fail to achieve regulation and continually attempts soft-start, only to have the output capacitance discharged by the load when in a FAULT state.

#### **Output Voltage Ripple**

Output voltage ripple is inversely proportional to  $C_{\text{OUT}}$ . During  $t_{\text{ON}}$ , when the boost switch is on, all load current is supplied by  $C_{\text{OUT}}$ . Output ripple is calculated as:

$$V_{RIPPLE(P-P)} = t_{ON} \bullet \frac{I_{LOAD}}{C_{
m OUT}}$$
 EQ. 1

and

$$t_{ON} = t_{SW} \bullet D = t_{SW} \bullet \left(1 - \frac{V_{IN}}{V_{OUT}}\right)$$
 EQ. 2

therefore:

$$V_{RIPPLE(P-P)} = t_{SW} \bullet \left(1 - \frac{V_{IN}}{V_{OUT}}\right) \bullet \frac{I_{LOAD}}{C_{OUT}}$$
 EQ. 3

and

$$t_{SW} = \frac{1}{f_{SW}}$$
 EQ. 4

As can be seen from EQ. 3, the maximum  $V_{\text{RIPPLE}}$  occurs when  $V_{\text{IN}}$  is minimum and  $I_{\text{LOAD}}$  is maximum.

## **Layout Recommendations**

The layout recommendations below highlight various top-copper pours using different colors.

To minimize spikes at  $V_{\text{OUT}}$ ,  $C_{\text{OUT}}$  must be placed as close as possible to PGND and VOUT, as shown in Figure 29.

For thermal reasons, it is suggested to maximize the pour area for all planes other than SW. Especially the ground pour should be set to fill all available PCB surface area and tied to internal layers with a cluster of thermal vias.

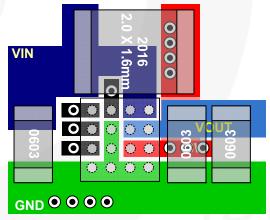
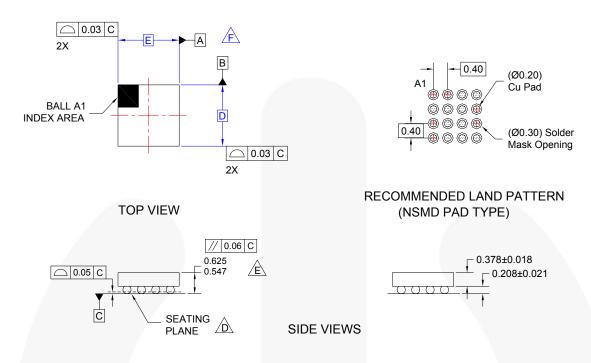


Figure 29. Layout Recommendation

## **Physical Dimensions**



⊕ 0.005(M) C A B

Ø0.260±0.02

(Y) ±0.018

 $(X) \pm 0.018$ 

/F\

#### NOTES:

- A. NO JEDEC REGISTRATION APPLIES.
- B. DIMENSIONS ARE IN MILLIMETERS.
- C. DIMENSIONS AND TOLERANCE PER ASME Y14.5M, 1994.
- DATUM C IS DEFINED BY THE SPHERICAL CROWNS OF THE BALLS.
- E PACKAGE NOMINAL HEIGHT IS 586 MICRONS ±39 MICRONS (547-625 MICRONS).
- F. FOR DIMENSIONS D, E, X, AND Y SEE PRODUCT DATASHEET.
- G. DRAWING FILNAME: MKT-UC016AArev2.

Figure 30. 16-Ball, 4x4 Array, 0.4 mm Pitch, 250 µm Ball, Wafer-Level Chip-Scale Package (WLCSP)

### **Product-Specific Dimensions**

0.40

0.40

 $\oplus$   $\oplus$   $\bigcirc$   $\oplus$ 

0000

3

**BOTTOM VIEW** 

Product	D	E	X	Y
FAN48630UCX	1.780 ±0.030	1.780 ±0.030	0.290	0.290

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wer

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Rev. 162