



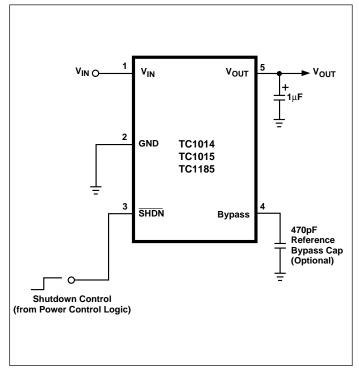
## **FEATURES**

- Extremely Low Supply Current (50µA, Typ.)
- Very Low Dropout Voltage
- 50mA, 100mA, and 150mA Output (TC1014, TC1015, and TC1185, Respectively)
- High Output Voltage Accuracy
- Standard or Custom Output Voltages
- Power-Saving Shutdown Mode
- Reference Bypass Input for Ultra Low-Noise Operation
- Over-Current and Over-Temperature Protection
- Space-Saving 5-Pin SOT-23A Package
- Pin Compatible Upgrades for Bipolar Regulators

## **APPLICATIONS**

- Battery Operated Systems
- Portable Computers
- Medical Instruments
- Instrumentation
- Cellular / GSM / PHS Phones
- Linear Post-Regulator for SMPS
- Pagers

### **TYPICAL APPLICATION**



## **GENERAL DESCRIPTION**

The TC1014, TC1015, and TC1185 are high accuracy (typically  $\pm 0.5\%$ ) CMOS upgrades for older (bipolar) low dropout regulators such as the LP2980. Designed specifically for battery-operated systems, the devices' CMOS construction eliminates wasted ground current, significantly extending battery life. Total supply current is typically 50µA at full load (20 to 60 times lower than in bipolar regulators!).

Key features for the devices include ultra low-noise operation (plus optional Bypass input), fast response to step changes in load, and very low dropout voltage, typically 85mV (TC1014), 180mV (TC1015), and 270mV (TC1185) at full load. Supply current is reduced to  $0.5\muA$  (max) and  $V_{OUT}$  falls to zero when the shutdown input is low. The devices also incorporate both over-temperature and over-current protection.

The TC1014, TC1015, and TC1185 are stable with an output capacitor of only 1 $\mu$ F and have a maximum output current of 50mA, 100mA, and 150mA, respectively. For higher output versions, see the TC1107, TC1108, and TC1173 (I<sub>OUT</sub> = 300 mA) data sheets.

## **ORDERING INFORMATION**

Part Number	Package	Junction Temp. Range
TC1014-xxVCT	5-Pin SOT-23A*	– 40°C to +125°C
TC1015-xxVCT	5-Pin SOT-23A*	- 40°C to +125°C
TC1185-xxVCT	5-Pin SOT-23A*	- 40°C to +125°C
TC1015EV Eva	luation Kit for CMO	S LDO Family
NOTE *5-Pin SOT-2	3A is equivalent to FIALS	C-74A

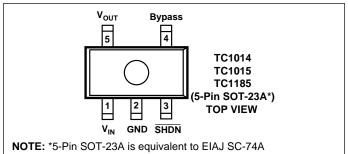
NOTE: \*5-Pin SOT-23A is equivalent to EIAJ SC-74A.

#### **Available Output Voltages:**

1.8, 2.5, 2.7, 2.8, 2.85, 3.0, 3.3, 3.6, 4.0, 5.0 xx indicates ouput voltages

Other output voltages are available. Please contact Microchip Technologies for details.

### **PIN CONFIGURATION**



## TC1014 TC1015 TC1185

## **ABSOLUTE MAXIMUM RATINGS\***

Maximum Voltage On Any Pin ......  $V_{IN}$  + 0.3V to - 0.3V Lead Temperature (Soldering, 10 Sec.) ..... +260°C

Input Voltage	6.5V
Output Voltage	(- 0.3) to (V <sub>IN</sub> + 0.3)
Power Dissipation	Internally Limited
Operating Temperature	$-40^{\circ}C < T_{J} < 125^{\circ}C$
Storage Temperature	– 65°C to +150°C

\*Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

<b>ELECTRICAL CHARACTERISTICS:</b> $V_{IN} = V_R + 1V$ , $IL = 100\mu$ A, $CL = 3.3\mu$ F, SHDN > $V_{IH}$ , $T_A = 25^{\circ}$ C, unless otherwise noted.
<b>Boldface</b> type specifications apply for junction temperatures of – 40°C to +125°C.

Symbol	Parameter	Test Conditions	Min	Тур	Max	Units
V <sub>IN</sub>	Input Operating Voltage	Note 1	2.7	_	6.0	V
I <sub>OUTMAX</sub>	Maximum Output Current	TC1014 TC1015	50 100	_	_	mA
		TC1185	150	_	_	
Vout	Output Voltage	Note 2	V <sub>R</sub> – 2.5%	V <sub>R</sub> ±0.5%	V <sub>R</sub> + 2.5%	V
TCV <sub>OUT</sub>	V <sub>OUT</sub> Temperature Coefficient	Note 3	_	20 <b>40</b>	_	ppm/°C
$\Delta V_{OUT} / \Delta V_{IN}$	Line Regulation	$(V_R + 1V) \le V_{IN} \le 6V$	_	0.05	0.35	%
$\Delta V_{OUT}/V_{OUT}$	Load Regulation TC1014;TC1015 TC1185	$I_L = 0.1$ mA to $I_{OUTMAX}$ $I_L = 0.1$ mA to $I_{OUTMAX}$ Note 4	_	0.5 0.5	2 3	%
V <sub>IN</sub> – V <sub>OUT</sub>	Dropout Voltage TC1015; TC1185 TC1185	$I_{L} = 100\mu A$ $I_{L} = 20m A$ $I_{L} = 50m A$ $I_{L} = 100m A$ $I_{L} = 150m A$	 	2 65 85 180 270	 120 250 400	mV
I <sub>IN</sub>	Supply Current (Note 8)	Note 5 $\overline{\text{SHDN}} = V_{\text{IH}}, I_{\text{L}} = 0$		50	80	μA
	Shutdown Supply Current	$\overline{SHDN} = 0V$		0.05	0.5	μΑ
PSRR	Power Supply Rejection Ratio	F <sub>RE</sub> ≤ 1KHz		64		dB
I <sub>OUTSC</sub>	Output Short Circuit Current	V <sub>OUT</sub> = 0V	—	300	450	mA
ΔV <sub>OUT</sub> /ΔP <sub>D</sub>	Thermal Regulation	Notes 6, 7	_	0.04	_	V/W
T <sub>SD</sub>	Thermal Shutdown Die Temperature		—	160	—	°C
$\Delta T_{SD}$	Thermal Shutdown Hysteresis		—	10	_	°C
eN	Output Noise	$I_L = I_{OUTMAX}$ , F = 10kHz 470pF from Bypass to GND	-	600	—	nV/√ <del>Hz</del>

### **SHDN** Input

VIH	SHDN Input High Threshold	V <sub>IN</sub> = 2.5V to 6.5V	45	_		%V <sub>IN</sub>
V <sub>IL</sub>	SHDN Input Low Threshold	$V_{IN} = 2.5V$ to 6.5V	—	—	15	%V <sub>IN</sub>

**NOTES:** 1. The minimum V<sub>IN</sub> has to meet two conditions: V<sub>IN</sub>  $\ge$  2.7V and V<sub>IN</sub>  $\ge$  V<sub>R</sub> + V<sub>DROPOUT</sub>.

2.  $V_R$  is the regulator output voltage setting. For example:  $V_R = 1.8V$ , 2.5V, 2.7V, 2.8V, 2.85V, 3.0V, 3.3V, 3.6V, 4.0V, 5.0V.

3. TCV<sub>OUT</sub> =  $(V_{OUT_{MAX}} - V_{OUT_{MIN}}) \times 10^6$ 

V<sub>OUT</sub> χ ΔΤ

4. Regulation is measured at a constant junction temperature using low duty cycle pulse testing. Load regulation is tested over a load range from 1.0mA to the maximum specified output current. Changes in output voltage due to heating effects are covered by the thermal regulation specification.

5. Dropout voltage is defined as the input to output differential at which the output voltage drops 2% below its nominal value at a 1V differential.

 Thermal Regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a current pulse equal to I<sub>LMAX</sub> at V<sub>IN</sub> = 6V for T = 10msec.

7. The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction-to-air (i.e. T<sub>A</sub>, T<sub>J</sub>, θ<sub>JA</sub>). Exceeding the maximum allowable power dissipation causes the device to initiate thermal shutdown. Please see **Thermal Considerations** section of this data sheet for more details.

8. Apply for Junction Temperatures of  $-40^{\circ}$ C to  $+85^{\circ}$ C.

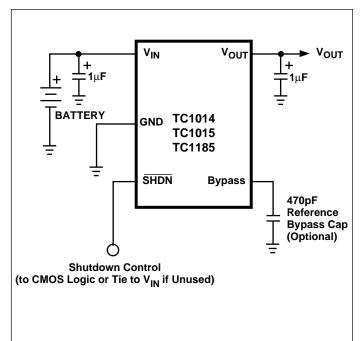
## **PIN DESCRIPTION**

Pin No. (5-Pin SOT-23A)	Symbol	Description
1	V <sub>IN</sub>	Unregulated supply input.
2	GND	Ground terminal.
3	SHDN	Shutdown control input. The regulator is fully enabled when a logic high is applied to this input. The regulator enters shutdown when a logic low is applied to this input. During shutdown, output voltage falls to zero, and supply current is reduced to 0.5µA (max).
4	Bypass	Reference bypass input. Connecting a 470pF to this input further reduces output noise.
5	V <sub>OUT</sub>	Regulated voltage output.

## **DETAILED DESCRIPTION**

The TC1014, TC1015, and TC1185 are precision fixed output voltage regulators. (If an adjustable version is desired, please see the TC1070, TC1071, or TC1187 data sheets.) Unlike bipolar regulators, the TC1014, TC1015, and TC1185 supply current does not increase with load current. In addition,  $V_{OUT}$  remains stable and within regulation at very low load currents (an important consideration in RTC and CMOS RAM battery back-up applications).

Figure 1 shows a typical application circuit. The regulator is enabled any time the shutdown input (SHDN) is at or above V<sub>IH</sub>, and shutdown (disabled) when SHDN is at or below V<sub>IL</sub>. SHDN may be controlled by a CMOS logic gate, or I/O port of a microcontroller. If the SHDN input is not required, it should be connected directly to the input supply. While in shutdown, supply current decreases to  $0.05\mu A$  (typical) and V<sub>OUT</sub> falls to zero volts.



## **Bypass Input**

A 470pF capacitor connected from the Bypass input to ground reduces noise present on the internal reference, which in turn significantly reduces output noise. If output noise is not a concern, this input may be left unconnected. Larger capacitor values may be used, but results in a longer time period to rated output voltage when power is initially applied.

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## **Output Capacitor**

A 1 $\mu$ F (min) capacitor from V<sub>OUT</sub> to ground is required. The output capacitor should have an effective series resistance of 5 $\Omega$  or less. A 1 $\mu$ F capacitor should be connected from V<sub>IN</sub> to GND if there is more than 10 inches of wire between the regulator and the AC filter capacitor, or if a battery is used as the power source. Aluminum electrolytic or tantalum capacitor types can be used. (Since many aluminum electrolytic capacitors freeze at approximately – 30°C, solid tantalums are recommended for applications operating below – 25°C.) When operating from sources other than batteries, supply-noise rejection and transient response can be improved by increasing the value of the input and output capacitors and employing passive filtering techniques.

## **Thermal Considerations**

### Thermal Shutdown

Integrated thermal protection circuitry shuts the regulator off when die temperature exceeds 160°C. The regulator remains off until the die temperature drops to approximately 150°C.

### **Power Dissipation**

The amount of power the regulator dissipates is primarily a function of input and output voltage, and output current. The following equation is used to calculate worst case power dissipation:

Figure 1. Typical Application Circuit

## TC1014 TC1015 TC1185

 $P_D \approx (V_{IN_{MAX}} - V_{OUT_{MIN}})I_{LOAD_{MAX}}$ 

Where:

$$\begin{split} P_D &= \text{Worst case actual power dissipation} \\ V_{\text{IN}_{MAX}} &= \text{Maximum voltage on } V_{\text{IN}} \\ V_{\text{OUT}_{MIN}} &= \text{Minimum regulator output voltage} \\ I_{\text{LOAD}_{MAX}} &= \text{Maximum output (load) current} \end{split}$$

#### Equation 1.

The maximum *allowable* power dissipation (Equation 2) is a function of the maximum ambient temperature  $(T_{AMAX})$ , the maximum allowable die temperature (125°C) and the thermal resistance from junction-to-air ( $\theta_{JA}$ ). The 5-Pin SOT-23A package has a  $\theta_{JA}$  of approximately *220°C/Watt* when mounted on a single layer FR4 dielectric copper clad PC board.

$$P_{D_{MAX}} = \frac{(T_{J_{MAX}} - T_{A_{MAX}})}{\theta_{JA}}$$

Where all terms are previously defined.

#### Equation 2.

Equation 1 can be used in conjunction with Equation 2 to ensure regulator thermal operation is within limits. For example:

Given:

$$\begin{array}{l} V_{\text{IN}_{\text{MAX}}} = 3.0V + 10\% \\ V_{\text{OUT}_{\text{MIN}}} = 2.7V - 2.5\% \\ I_{\text{LOAD}_{\text{MAX}}} = 40\text{mA} \\ T_{\text{JMAX}} = 125^{\circ}\text{C} \\ T_{\text{AMAX}} = 55^{\circ}\text{C} \end{array}$$

Find: 1. Actual power dissipation 2. Maximum allowable dissipation

Actual power dissipation:

$$\begin{split} \mathsf{P}_{\mathsf{D}} &\approx (\mathsf{V}_{\mathsf{IN}_{\mathsf{MAX}}} - \mathsf{V}_{\mathsf{OUT}_{\mathsf{MIN}}}) \mathsf{I}_{\mathsf{LOAD}_{\mathsf{MAX}}} \\ &= [(3.0 \text{ x } 1.1) - (2.7 \text{ x } .975)] 40 \text{ x } 10^{-3} \\ &= 26.7 \text{mW} \end{split}$$

Maximum allowable power dissipation:

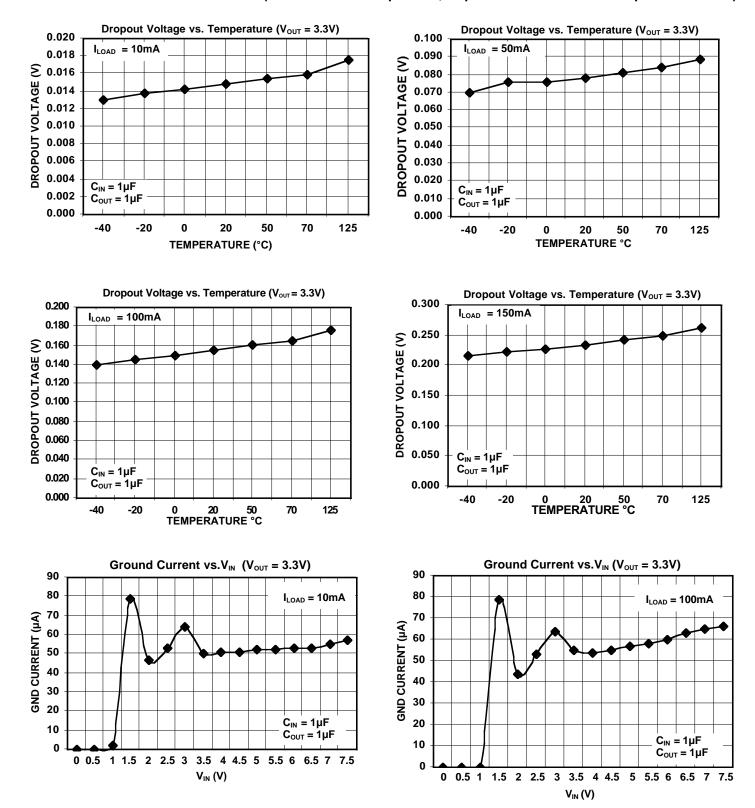
$$P_{D_{MAX}} = (\underbrace{T_{J_{MAX}} - T_{A_{MAX}}}_{\theta_{JA}})$$
$$= \underbrace{(125 - 55)}_{220}$$
$$= 318 \text{mW}$$

In this example, the TC1014 dissipates a maximum of only 26.7 mW; far below the allowable limit of 318 mW. In a similar manner, Equation 1 and Equation 2 can be used to calculate maximum current and/or input voltage limits.

### Layout Considerations

The primary path of heat conduction out of the package is via the package leads. Therefore, layouts having a ground plane, wide traces at the pads, and wide power supply bus lines combine to lower  $\theta_{JA}$  and, therefore, increase the maximum allowable power dissipation limit.

## TC1014 TC1015 TC1185



### **TYPICAL CHARACTERISTICS:** (Unless otherwise specified, all parts are measured at Temperature = 25°C)

 $V_{OUT}$  vs. $V_{IN}$  ( $V_{OUT}$  = 3.3V)

#### 80 3.5 $I_{LOAD} = 0$ $I_{LOAD} = 150 mA$ 70 3 GND CURRENT (µ A) 5 0 0 0 0 5 0 0 0 2.5 2 2 م 1 C<sub>IN</sub> = 1µF 10 0.5 $C_{IN} = 1 \mu F$ $C_{OUT} = 1 \mu F$ 0 С<sub>оит</sub> = 1µF 0 0.5 1 1.5 2 2.5 3 3.5 4 4.5 5 5.5 6 6.5 7 7.5 0 VIN (V) 0.5 1 1.5 2 2.5 3 3.5 4 4.5 5 5.5 6 6.5 7 0 V<sub>IN</sub>(V) Output Voltage vs. Temperature (Vout = 3.3V) V<sub>OUT</sub> vs. V<sub>IN</sub> (V<sub>OUT</sub> = 3.3V) 3.320 3.5 $I_{LOAD} = 100 \text{mA}$ I<sub>LOAD</sub> = 10mA 3.315 3.0 3.310 2.5 3.305 € <sup>2.0</sup> ^ 1.5 € 3.300 <sup>™</sup>> 3.295 3.290 1.0 $C_{IN} = 1 \mu F$ 3.285 $C_{OUT} = 1 \mu F$ 0.5 3.280 $C_{IN} = 1 \mu F$ $V_{IN} = 4.3V$ $C_{OUT} = \dot{1}\mu F$ 3.275 0.0 3 3.5 4 4.5 5 5.5 6 6.5 7 0 0.5 1 1.5 2 2.5 -40 -20 0 20 85 125 -10 40 TEMPERATURE °C V<sub>IN</sub>(V) Output Voltage vs. Temperature (Vout = 3.3V) 3.290 I<sub>LOAD</sub> = 150 mA 3.288 3.286 € <sup>3.284</sup> > <sup>5</sup> 3.282 3.280 3.278 $C_{IN} = 1 \mu F$ $C_{OUT} = 1 \mu F$ 3.276 $V_{IN} = 4.3V$ 3.274 -40 -10 0 20 40 -20 85 125

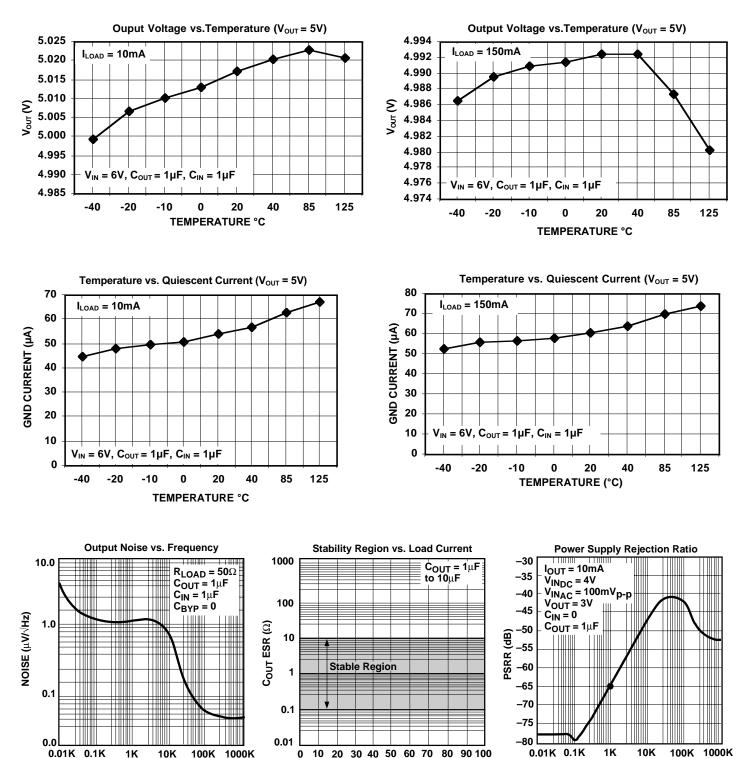
## TYPICAL CHARACTERISTICS: (Unless otherwise specified, all parts are measured at Temperature = 25°C)

Ground Current vs.V<sub>IN</sub> (V<sub>OUT</sub> = 3.3V)

**TEMPERATURE °C** 

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## **TYPICAL CHARACTERISTICS**



FREQUENCY (Hz)

FREQUENCY (Hz)

TC1014 TC1015 TC1185

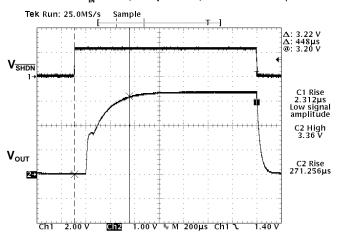
LOAD CURRENT (mA)

## TC1014 TC1015 TC1185

## **TYPICAL CHARACTERISTICS**

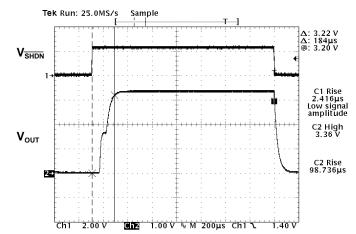
Measure Rise Time of 3.3V LDO with Bypass Capacitor

Conditions:  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $C_{BYP} = 470 pF$ ,  $I_{LOAD} = 100 mA$  $V_{IN} = 4.3V$ , Temp = 25°C, Rise Time = 448 $\mu$ S



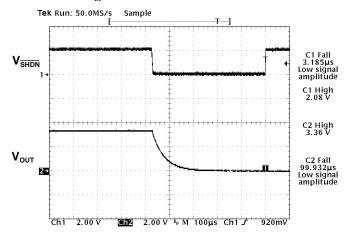
Measure Rise Time of 3.3V LDO without Bypass Capacitor

Conditions:  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $C_{BYP} = 0pF$ ,  $I_{LOAD} = 100mA$  $V_{IN} = 4.3V$ , Temp = 25°C, Rise Time = 184 $\mu$ S



Measure Fall Time of 3.3V LDO with Bypass Capacitor

Conditions:  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $C_{BYP} = 470pF$ ,  $I_{LOAD} = 50mA$  $V_{IN} = 4.3V$ , Temp = 25°C, Fall Time = 100 $\mu$ S



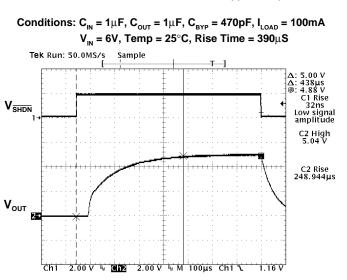
Measure Fall Time of 3.3V LDO without Bypass Capacitor

Conditions: C<sub>IN</sub> = 1µF, C<sub>OUT</sub> = 1µF, C<sub>BYP</sub> = 0pF, I<sub>LOAD</sub> = 100mA

 $V_{IN} = 4.3V, Temp = 25^{\circ}C, Fall Time = 52\mu S$ Tek Run: 50.0MS/s Sample  $V_{SHDN}$   $I_{I}$   $V_{OUT}$   $V_{OUT}$  C1 Fall 2.08V C2 High 3.36V C2 High 3.36V  $C2 Fall 51.392\mu S Low signal amplitude$  C1 Fall 2.08V C2 High 3.36V  $C2 Fall 51.392\mu S Low signal amplitude$ 

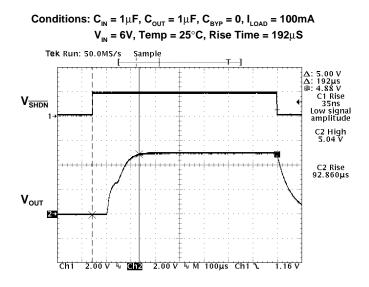
## **TYPICAL CHARACTERISTICS**

Measure Rise Time of 5.0V LDO with Bypass Capacitor



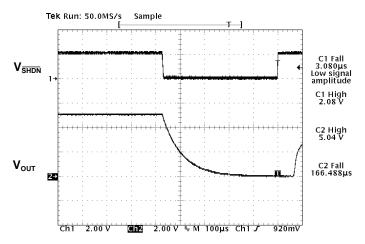
Measure Rise Time of 5.0V LDO without Bypass Capacitor

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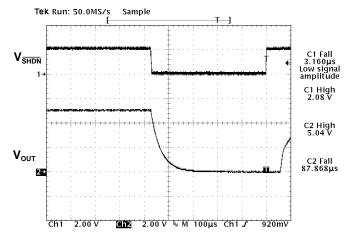
Measure Fall Time of 5.0V LDO with Bypass Capacitor

Conditions:  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $C_{BYP} = 470pF$ ,  $I_{LOAD} = 50mA$  $V_{IN} = 6V$ , Temp = 25°C, Fall Time = 167 $\mu$ S



Measure Fall Time of 5.0V LDO without Bypass Capacitor

Conditions:  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $C_{BYP} = 0pF$ ,  $I_{LOAD} = 100mA$  $V_{IN} = 6V$ , Temp = 25°C, Fall Time = 88 $\mu$ S



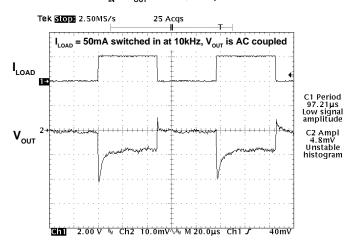
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## TC1014 TC1015 TC1185

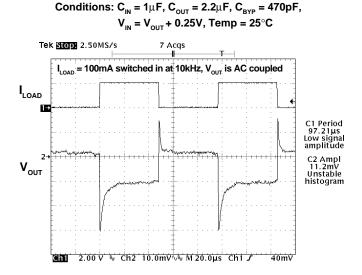
## **TYPICAL CHARACTERISTICS**

#### Load Regulation of 3.3V LDO

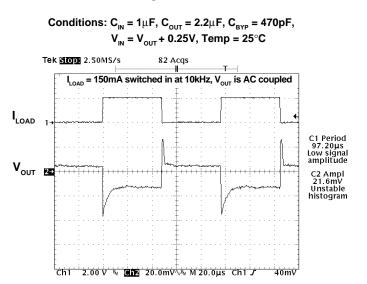
Conditions:  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 2.2\mu F$ ,  $C_{BYP} = 470pF$ ,  $V_{IN} = V_{OUT} + 0.25V$ , Temp = 25°C



## Load Regulation of 3.3V LDO

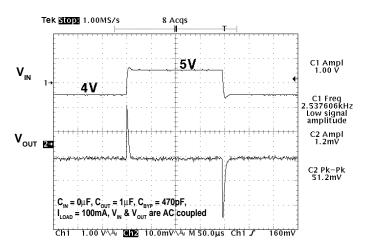


#### Load Regulation of 3.3V LDO



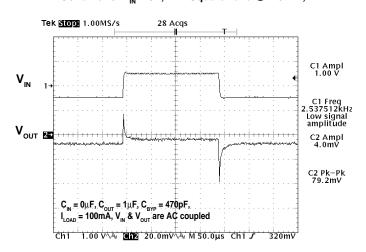
Line Regulation of 3.3V LDO

#### Conditions: $V_{IN} = 4V,+ 1V$ Squarewave @ 2.5kHz,



## **TYPICAL CHARACTERISTICS**

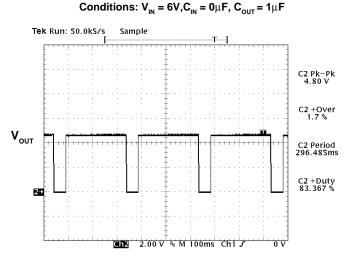




### Conditions: $V_{IN} = 6V, + 1V$ Squarewave @ 2.5kHz,

#### Thermal Shutdown Response of 5.0V LDO

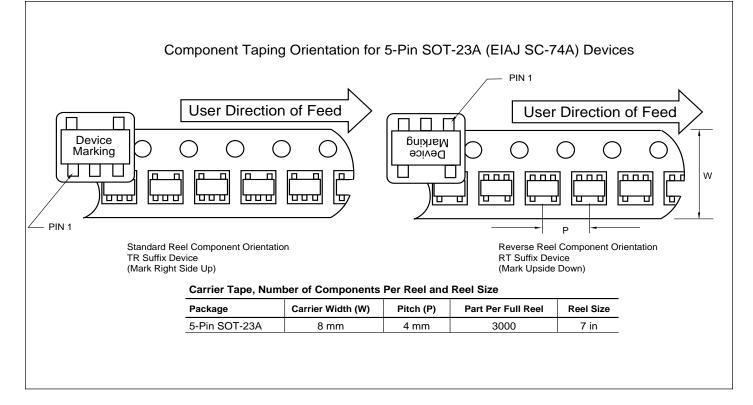
TC1014 TC1015 TC1185



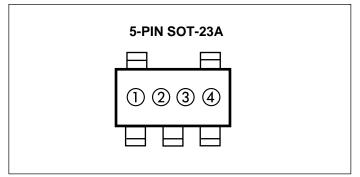
 ${\rm I}_{_{LOAD}}$  was increased until temperature of die reached about 160°C, at which time integrated thermal protection circuitry shuts the regulator off when die temperature exceeds approximately 160°C. The regulator remains off until die temperature drops to approximately 150°C.

## TC1014 TC1015 TC1185

### TAPING FORM



#### MARKING



(1) & (2) b = part number code + temperature range and voltage

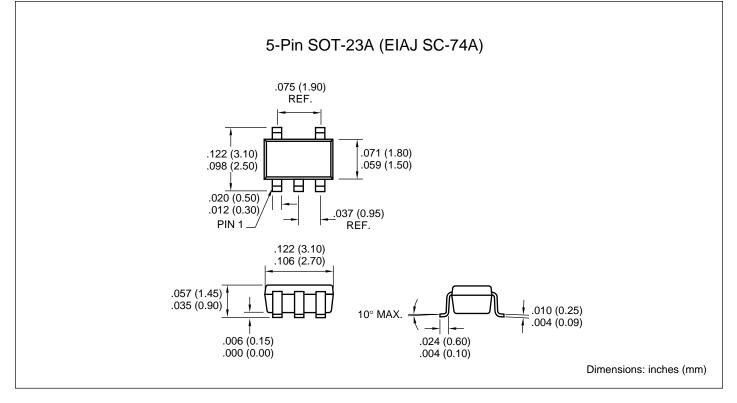
<u>(V)</u>	TC1014 Code	TC1015 Code	TC1185 Code
1.8	AY	BY	NY
2.5	A1	B1	N1
2.7	A2	B2	N2
2.8	AZ	BZ	NZ
2.85	A8	B8	N8
3.0	A3	B3	N3
3.3	A5	B5	N5
3.6	A9	B9	N9
4.0	A0	B0	N0
5.0	A7	B7	N7

③ represents date code

④ represents lot ID number

## TC1014 TC1015 TC1185

#### PACKAGE DIMENSIONS





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Microchip Technology Inc. 2107 North First Street, Suite 590 San Jose, CA 95131 Tel: 408-436-7950 Fax: 408-436-7955 Toronto 6285 Northam Drive, Suite 108 Mississauga, Ontario L4V 1X5, Canada Tel: 905-673-0699 Fax: 905-673-6509

#### ASIA/PACIFIC

#### China - Beijing Microchip Technology Beijing Office Unit 915 New China Hong Kong Manhattan Bldg. No. 6 Chaoyangmen Beidajie Beijing, 100027, No. China Tel: 86-10-85282100 Fax: 86-10-85282104 China - Shanghai Microchip Technology Shanghai Office Room 701, Bldg. B Far East International Plaza No. 317 Xian Xia Road Shanghai, 200051 Tel: 86-21-6275-5700 Fax: 86-21-6275-5060 Hong Kong Microchip Asia Pacific RM 2101, Tower 2, Metroplaza 223 Hing Fong Road Kwai Fong, N.T., Hong Kong Tel: 852-2401-1200 Fax: 852-2401-3431 India Microchip Technology Inc. India Liaison Office **Divyasree Chambers** 1 Floor, Wing A (A3/A4) No. 11, OiShaugnessey Road Bangalore, 560 025, India Tel: 91-80-2290061 Fax: 91-80-2290062 Japan Microchip Technology Intl. Inc. Benex S-1 6F 3-18-20, Shinyokohama

Kohoku-Ku, Yokohama-shi Kanagawa, 222-0033, Japan Tel: 81-45-471- 6166 Fax: 81-45-471-6122 Korea Microchip Technology Korea

168-1, Youngbo Bldg. 3 Floor Samsung-Dong, Kangnam-Ku Seoul, Korea Tel: 82-2-554-7200 Fax: 82-2-558-5934

#### **ASIA/PACIFIC** (continued)

Singapore Microchip Technology Singapore Pte Ltd. 200 Middle Road #07-02 Prime Centre Singapore, 188980 Tel: 65-334-8870 Fax: 65-334-8850 Taiwan Microchip Technology Taiwan 11F-3, No. 207 Tung Hua North Road Taipei, 105, Taiwan Tel: 886-2-2717-7175 Fax: 886-2-2545-0139

#### EUROPE

Australia Microchip Technology Australia Pty Ltd Suite 22, 41 Rawson Street Epping 2121, NSW Australia Tel: 61-2-9868-6733 Fax: 61-2-9868-6755 Denmark Microchip Technology Denmark ApS **Regus Business Centre** Lautrup hoj 1-3 Ballerup DK-2750 Denmark Tel: 45 4420 9895 Fax: 45 4420 9910 France Arizona Microchip Technology SARL Parc díActivite du Moulin de Massy 43 Rue du Saule Trapu Batiment A - ler Etage 91300 Massy, France Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79 Germany Arizona Microchip Technology GmbH Gustav-Heinemann Ring 125 D-81739 Munich, Germany Tel: 49-89-627-144 0 Fax: 49-89-627-144-44 Germany Analog Product Sales Lochhamer Strasse 13 D-82152 Martinsried, Germany Tel: 49-89-895650-0 Fax: 49-89-895650-22 Italy Arizona Microchip Technology SRL Centro Direzionale Colleoni Palazzo Taurus 1 V. Le Colleoni 1 20041 Agrate Brianza Milan, Italy Tel: 39-039-65791-1 Fax: 39-039-6899883 United Kingdom Arizona Microchip Technology Ltd. 505 Eskdale Road Winnersh Triangle Wokingham Berkshire, England RG41 5TU Tel: 44 118 921 5869 Fax: 44-118 921-5820

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