## ADC1443D series

Dual 14-bit ADC; 125, 160 or 185 Msps; JESD204A/B serial outputs
Rev. 4.4 - 6 June 2014
Product data sheet

## 1. General description

The ADC1443D is a dual channel 14-bit Analog-to-Digital Converter (ADC) with JESD204B interface (which is backward compatible with the JESD204A interface) optimized for high dynamic performance and low power consumption at sample rates up to 185 Msps . Pipelined architecture and output error correction guarantee zero missing codes over the entire operating range.

Supplied from a single 1.8 V source, the ADC1443D has JESD204B serial outputs over a configurable number of lanes (1 or 2). Multiple Device Synchronization (MDS) allows sample-accurate synchronization of the data outputs of multiple ADC devices. It guarantees a maximum skew of one clock period between as many as 16 output lanes from up to eight ADC1443D devices.

An integrated Serial Peripheral Interface (SPI) allows easy configuration of the ADC. The device also includes a programmable full-scale to allow a flexible input voltage range of $1 \mathrm{~V}(\mathrm{p}-\mathrm{p})$ to $2 \mathrm{~V}(\mathrm{p}-\mathrm{p})$.

With an analog banwidth from the baseband to input frequencies of up to 1 GHz (typical), the ADC1443D is ideal for use in undersampled multi-carrier, multistandard communication system applications.

## 2. Features and benefits

|  | Dual channel 14-bit resolution ADC |  | SNR = 70.6 dBFS (typical); $\mathrm{f}_{\mathrm{s}}=154 \mathrm{Msps} ; \mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ |
| :---: | :---: | :---: | :---: |
| $\square$ | Sampling rate up to 185 Msps | - | $\begin{aligned} & \text { SFDR }=86 \mathrm{dBc} \text { (typical); } \mathrm{f}_{\mathrm{s}}=154 \mathrm{Msps} ; \\ & \mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz} \end{aligned}$ |
| $\square$ | JESD204B Device Subclass 0, 1 and 2 with harmonic clocking and deterministic latency support | $\square$ | $\begin{aligned} & \text { IMD3 }=88 \mathrm{dBc} \text { (typical); } \mathrm{f}_{\mathrm{s}}=154 \mathrm{Msps} ; \\ & \mathrm{f}_{\mathrm{i} 1}=188.5 \mathrm{MHz} ; \mathrm{f}_{\mathrm{i} 2}=191.5 \mathrm{MHz} \end{aligned}$ |
| $\square$ | ADC Multiple Device Synchronization (MDS) | - | Analog input bandwidth of 1 GHz (typical) |
| $\square$ | Two JESD204B serial output lanes, up to 5 Gbps typical | $\square$ | Pin to pin compatible with ADC1413D series |
| $\square$ | Single 1.8 V supply | $\square$ | $\begin{aligned} & \text { Typical power dissipation }=0.8 \mathrm{~W} \text {; } \\ & \mathrm{f}_{\mathrm{s}}=154 \mathrm{Msps} \end{aligned}$ |
|  | Offset binary, two's complement and Gray output data | - | Power-down and sleep modes |

- Flexible input voltage range from $1 \mathrm{~V}(\mathrm{p}-\mathrm{p})$ to $2 \mathrm{~V}(\mathrm{p}-\mathrm{p})$ by 1 dB steps
- Clock input divider from 1 to 8 supports harmonic clocking
- Duty Cycle Stabilizer (DCS)

Industrial temperature range from $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$

- Serial Peripheral Interface (SPI) for configuration control and status monitoring
- VFQFPN56 package; $8 \times 8 \mathrm{~mm}$


## 3. Applications

- Wireless infrastructure: LTE, TD-LTE, WiMAX, MC-GSM, CDMA, WCDMA, TD-SCDMA
- Software defined radio
- Medical non-invasive scanners
- Scientific particle detectors
- Microwave backhaul transceivers
- Aerospace and defense communications and radar systems
- Industrial signal analysis instruments
- General-purpose high-speed applications


## 4. Ordering information

Table 1. Ordering information

| Type number | $\mathbf{f}_{\mathbf{s}}$ (Msps) | Package |  |  |
| :--- | :---: | :--- | :--- | :--- | :--- |
|  |  | Name | Description | Version |
| ADC1443D200HD | 185 | VFQFPN | plastic thermal enhanced low profile quad flat package; <br> no leads; 56 terminals; resin based; body $8 \times 8 \times 1.35 \mathrm{~mm}$ | PSC-4449 |
| ADC1443D160HD | 160 | VFQFPN | plastic thermal enhanced low profile quad flat package; <br> no leads; 56 terminals; resin based; body $8 \times 8 \times 1.35 \mathrm{~mm}$ | PSC-4449 |
| ADC1443D125HD | 125 | VFQFPN | plastic thermal enhanced low profile quad flat package; <br> no leads; 56 terminals; resin based; body $8 \times 8 \times 1.35 \mathrm{~mm}$ | PSC-4449 |

## 5. Block diagram



Fig 1. Block diagram

## 6. Pinning information

### 6.1 Pinning



Fig 2. Pin configuration (PSC-4449)

### 6.2 Pin description

Table 2. Pin description

| Symbol | Pin | Type ${ }^{[1]}$ | Description |
| :---: | :---: | :---: | :---: |
| INAM | 1 | 1 | channel A complementary analog input |
| INAP | 2 | 1 | channel A analog input |
| VCMA | 3 | O | channel A output common voltage |
| DNC | 4 | - | do not connect |
| DNC | 5 | - | do not connect |
| AGND | 6 | G | analog ground |
| CLKP | 7 | I | clock input |
| CLKN | 8 | I | complementary clock input |
| AGND | 9 | G | analog ground |
| DNC | 10 | - | do not connect |
| DNC | 11 | - | do not connect |
| VCMB | 12 | O | channel B output common voltage |
| INBP | 13 | I | channel B analog input |
| INBM | 14 | I | channel B complementary analog input |
| VDDA | 15 | P | analog power supply |
| VDDA | 16 | P | analog power supply |
| SCLK | 17 | I | SPI clock. Internally connected to $50 \mathrm{k} \Omega$ pull-down |
| SDIO | 18 | I/O | SPI data IO. Internally connected to $50 \mathrm{k} \Omega$ pull-down (when used as input) |
| SCS_N | 19 | 1 | SPI chip select. Internally connected to $50 \mathrm{k} \Omega$ pull-up |
| AGND | 20 | G | analog ground |
| DNC | 21 | - | do not connect |
| SCR_EN | 22 | I | scrambler enable. Internally connected to $50 \mathrm{k} \Omega$ pull-up |
| CFGO/OTRA | 23 | I/O | configuration pin 0/OuT of Range A (OTRA). Internally connected to $50 \mathrm{k} \Omega$ pull-down (when used as input) |
| CFG1/OTRB | 24 | I/O | configuration pin 1/OuT of Range B (OTRB). Internally connected to $50 \mathrm{k} \Omega$ pull-down (when used as input) |
| CFG2 | 25 | I/O | configuration pin 2 . Internally connected to $50 \mathrm{k} \Omega$ pull-down (when used as input) |
| CFG3 | 26 | I/O | configuration pin 3 . Internally connected to $50 \mathrm{k} \Omega$ pull-down (when used as input) |
| VDDO | 27 | P | digital output power supply |
| AGND | 28 | G | analog ground |
| OGND | 29 | G | digital output ground |
| OGND | 30 | G | digital output ground |
| VDDO | 31 | P | digital output power supply |
| CMLBP | 32 | O | channel B output |
| CMLBN | 33 | O | channel B complementary output |
| VDDO | 34 | P | digital output power supply |
| OGND | 35 | G | digital output ground |
| OGND | 36 | G | digital output ground |

Table 2. Pin description ...continued

| Symbol | Pin | Typel ${ }^{[1]}$ | Description |
| :--- | :--- | :--- | :--- |
| VDDO | 37 | P | digital output power supply |
| CMLAN | 38 | O | channel A complementary output |
| CMLAP | 39 | O | channel A output |
| VDDO | 40 | P | digital output power supply |
| OGND | 41 | G | digital output ground |
| OGND | 42 | G | digital output ground |
| SYNCBP | 43 | I | JESD204B SYNC synchronization signal from receiver |
| SYNCBN | 44 | I | complementary SYNC from receiver |
| AGND | 45 | G | analog ground |
| VDDO | 46 | P | digital output power supply |
| DNC | 47 | - | do not connect |
| SYSREFP | 48 | I | positive clock synchronization |
| SYSREFN | 49 | I | negative clock synchronization |
| VDDA | 50 | P | analog power supply |
| AGND | 51 | G | analog ground |
| AGND | 52 | G | analog ground |
| VDDA | 53 | P | analog power supply |
| DNC | 54 | - | do not connect |
| DNC | 55 | - | do not connect |
| VDDA | 56 | P | analog power supply |
| AGND | EXP | G | Expose PAD |

[1] P: power supply; G: ground; I: input; O: output; I/O: input/output.

### 6.2.1 Start-up Configuration

Because the maximum sampling clock of the ADC1443D is 200 Msps , care should be taken in case of harmonic clocking. If the input clock frequency is higher than 200 MHz , the clock divider must be set before providing the clock.

In order to avoid any issue, it is recommended to start the device in power-down mode by setting the configuration pins to logic level '1' (see Table 19). This can be done by adding for example a $1 \mathrm{k} \Omega$ pull-up resistor on CFG0, CFG1, CFG2 and CFG3.

When the power supplies are set, the divider can be programmed by the use of the SPI registers. Then the device is powered on and the JESD204B configuration is set by the use of the SPI registers (bits CFG_SETUP[3:0] in Table 42).

## 7. Limiting values

Table 3. Limiting values In accordance with the Absolute Maximum Rating System (IEC 60134).

| Symbol | Parameter | Conditions | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{\text {DDA }}$ | analog supply voltage |  | -0.3 | +2.1 | V |
| $\mathrm{V}_{\text {DDO }}$ | output supply voltage |  | -0.3 | +2.1 | V |
| $\Delta \mathrm{V}_{\mathrm{DD}}$ | supply voltage difference | $V_{\text {DDA }}-V_{\text {DDO }}$ | -0.8 | +0.8 | V |
| $V_{1}$ | input voltage | pins INP, INM, CLKP and CLKM; referenced to AGND | -0.3 | $V_{\text {DDA }}+0.3$ | V |
|  |  | pins OTR, SCS_N, SDIO, SCLK, CFG, SCR_EN, SYSREFP, SYSREFN, SYNCBP, and SYNCBN; referenced to AGND | -0.3 | $V_{\text {DDO }}+0.3$ | V |
| $\mathrm{V}_{\mathrm{O}}$ | output voltage | pin VCM; referenced to AGND | -0.3 | $V_{\text {DDA }}+0.3$ | V |
|  |  | pins CMLP, and CMLN; referenced to OGND | -0.3 | $V_{\text {DDO }}+0.3$ | V |
| $\mathrm{T}_{\text {stg }}$ | storage temperature |  | -55 | +125 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {amb }}$ | ambient temperature |  | -40 | +85 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{j}}$ | junction temperature |  | - | 125 | ${ }^{\circ} \mathrm{C}$ |

## 8. Thermal characteristics

Table 4. Thermal characteristics

| Symbol | Parameter | Conditions | Typ | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $R_{\text {th(j-a) }}$ | thermal resistance from junction to ambient | 66 vias | [1] 22.7 | K/W |
| $R_{\text {th(j-c) }}$ | thermal resistance from junction to case | 66 vias | [1] 9.3 | K/W |

[1] In compliance with JEDEC test board, in free air.

## 9. Static characteristics

Table 5. Static characteristics[1]

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Supplies |  |  |  |  |  |  |
| $V_{\text {DDA }}$ | analog supply voltage |  | 1.7 | 1.8 | 1.9 | V |
| V $_{\text {DDO }}$ | output supply voltage |  | 1.7 | 1.8 | 1.9 | V |
| IDDA | analog supply current | $\mathrm{f}_{\mathrm{s}}=185 \mathrm{Msps} ;$ |  |  |  |  |
| $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | 321 | 391 | mA |  |  |
| IDDO | output supply current | $\mathrm{f}_{\mathrm{s}}=185 \mathrm{Msps} ;$ <br> $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | 169 | 198 | mA |

Table 5. Static characteristics [1] ...continued

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | total power dissipation | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ |  |  |  |  |
|  |  | ADC1443D125; $\mathrm{f}_{\mathrm{s}}=125 \mathrm{Msps}$ | - | 0.71 | 0.9 | W |
|  |  | ADC1443D160; $\mathrm{f}_{\mathrm{s}}=154 \mathrm{Msps}$ | - | 0.8 | 1.0 | W |
|  |  | ADC1443D200; $\mathrm{f}_{\mathrm{s}}=185 \mathrm{Msps}$ | - | 0.9 | 1.1 | W |
|  |  | Power-down mode | - | 10 | - | mW |
|  |  | Sleep mode | - | 115 | - | mW |
| Clock inputs: pins CLKP and CLKM (AC-coupled; peak-to-peak) |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{i}(\mathrm{clk})}$ | clock input voltage | LVPECL | - | $\pm 0.8$ | - | V |
|  |  | LVDS | - | $\pm 0.35$ | - | V |
|  |  | SINE differential | $\pm 0.5$ | $\pm 1.25$ | - | V |
|  |  | LVCMOS single | - | $\pm 0.6$ | - | V |
| $\mathrm{Cl}_{1}$ | input capacitance |  | - | 1.2 | - | pF |
| Logic inputs |  |  |  |  |  |  |
| $\mathrm{I}_{\mathrm{LL}}$ | LOW-level input current | absolute value | - | 30 | - | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\mathrm{H}}$ | HIGH-level input current | absolute value | - | 70 | - | $\mu \mathrm{A}$ |
| $\mathrm{Cl}_{1}$ | input capacitance |  | - | 1.2 | - | pF |
| pins SYSREFP, SYSREFN, SYNCBP, and SYNCBN (Diffrential Pins) |  |  |  |  |  |  |
| $V_{i(c m)}$ | common-mode input voltage |  | 0.925 | 1.2 | 1.475 | V |
| $V_{i(\text { dif })}$ | differential input voltage |  | 0.2 | 0.7 | - | V |
| pins SCS_N, SDIO, SCLK, SCR_EN ,CFG, SYNCBPand SYSREFP (Single Ended) |  |  |  |  |  |  |
| $\mathrm{V}_{\text {IL }}$ | LOW-level input voltage |  | 0 | - | $0.3 \mathrm{~V}_{\text {DDO }}$ | V |
| $\mathrm{V}_{\text {IH }}$ | HIGH-level input voltage |  | $0.7 \mathrm{~V}_{\text {DDO }}$ | - | $\mathrm{V}_{\text {DDO }}$ | V |
| Logic output: pins OTRA, OTRB and SDIO |  |  |  |  |  |  |
| $\mathrm{V}_{\text {OL }}$ | LOW-level output voltage |  | 0 | - | 0.2 | V |
| $\mathrm{V}_{\mathrm{OH}}$ | HIGH-level output voltage |  | $\mathrm{V}_{\text {DDO-0. }} 0$ | - | $\mathrm{V}_{\text {DDO }}$ | V |
| Digital outputs: pins CMLAP, CMLAN, CMLBP, and CMLBN |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{O}(\mathrm{cm})}$ | common-mode output voltage | default current | - | 1.4 | - | V |
| $V^{\text {O(dif) }}$ | differential output voltage | default current; peak-to-peak | - | 800 | - | mV |
| Analog inputs: pins INP and INM |  |  |  |  |  |  |
| 1 | input current |  | - | $\pm 5$ | - | $\mu \mathrm{A}$ |
| $\mathrm{R}_{\mathrm{I}}$ | input resistance | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | 400 | - | $\Omega$ |
| $\mathrm{C}_{1}$ | input capacitance | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | 5 | - | pF |
| $\mathrm{V}_{1(\mathrm{~cm})}$ | common-mode input voltage | $\mathrm{V}_{\text {INP }}=\mathrm{V}_{\text {INM }} ; \mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$ | 0.8 | 0.9 | 1.0 | V |
| $\mathrm{B}_{\mathrm{i}}$ | input bandwidth |  | - | 1 | - | GHz |
| $V_{1(\text { dif })}$ | differential input voltage | peak-to-peak; full-scale | 1 | - | 2 | V |
| Common-mode output voltage: pins VCMA and VCMB |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{O}(\mathrm{cm})}$ | common-mode output voltage | $\mathrm{l}_{\mathrm{O}(\mathrm{cm})}=1 \mathrm{~mA}$ | - | 0.9 | - | V |
| $\mathrm{l}_{(1 \mathrm{~cm})}$ | common-mode output current | $\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$ | - | - | 1 | mA |

Table 5. Static characteristics[1] ...continued

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Accuracy |  |  |  |  |  |  |
| INL | integral non-linearity | $\begin{aligned} & \mathrm{f}_{\mathrm{s}}=153.6 \mathrm{Msps} ; \\ & \mathrm{f}_{\mathrm{i}}=4.43 \mathrm{MHz} \end{aligned}$ | - | $\pm 1.43$ | $\pm 5.2$ | LSB |
| DNL | differential non-linearity | $\begin{aligned} & \mathrm{f}_{\mathrm{s}}=185 \mathrm{Msps} ; \\ & \mathrm{f}_{\mathrm{i}}=4.43 \mathrm{MHz} ; \text { guaranteed } \end{aligned}$ no missing codes |  |  |  |  |
|  |  | Minimum typical value | -0.49 | -0.30 |  | LSB |
|  |  | Maximum typical value |  | +0.25 | +0.37 | LSB |
| $\mathrm{E}_{\text {offset }}$ | offset error |  | -20 | - | +20 | mV |
| $\mathrm{E}_{G}$ | gain error | full-scale | - | 4.1 | - | \% |
| $\mathrm{M}_{\mathrm{G} \text { (CTC) }}$ | channel-to-channel gain matching |  | - | 3.5 | - | \% |
| OS | Offset Spur | measured at $\mathrm{FS} / 2$ with $\mathrm{FS}=160 \mathrm{Msps}$ |  | -82 |  | dBc |
| Supply |  |  |  |  |  |  |
| PSRR | power supply rejection ratio | $\begin{aligned} & 100 \mathrm{mV}(\mathrm{p}-\mathrm{p}) \text { on } \mathrm{V}_{\mathrm{DDA}}, 0.5 \\ & \text { to } 2 \mathrm{MHz} \end{aligned}$ | - | -47 | - | dB |

[1] Typical values measured at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V} ; \mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}$. Minimum and maximum values are across the full temperature range $\mathrm{T}_{\text {amb }}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V} ; \mathrm{V}_{\text {I(dif) }}=2 \mathrm{~V} ; \mathrm{V}_{\text {INP }}-\mathrm{V}_{\text {INM }}=-1 \mathrm{dBFS}$; unless otherwise specified.
10.1 Dynamic characteristics

Table 6. Dynamic characteristics[1]

| Symbol | Parameter | Conditions | $\begin{aligned} & \text { ADC1443D125 } \\ & \left(\mathrm{f}_{\mathrm{s}}=125 \mathrm{Msps}\right) \end{aligned}$ |  |  | $\begin{aligned} & \text { ADC1443D160 } \\ & \left(\mathrm{f}_{\mathrm{s}}=154 \mathrm{Msps}\right) \end{aligned}$ |  |  | $\begin{aligned} & \text { ADC1443D200 } \\ & \left(\mathrm{f}_{\mathrm{s}}=185 \mathrm{Msps}\right) \end{aligned}$ |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max | Min | Typ | Max | Min | Typ | Max |  |
| $\alpha_{2 \mathrm{H}}$ | second harmonic level | $\mathrm{f}_{\mathrm{i}}=5 \mathrm{MHz}$ | - | -87 | - | - | -84 | - | - | -85 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=70 \mathrm{MHz}$ | - | -85 | - | - | -82 | - | - | -75 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=140 \mathrm{MHz}$ | - | -92 | - | - | -85 | - | - | -80 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$ | - | -83 | - | - | -83 | - | - | -87 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | -82 | - | - | -86 | - | - | -83 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=230 \mathrm{MHz}$ | - | -78 | - | - | -80 | - | - | -86 | - | dBc |
| $\alpha_{3 H}$ | third harmonic level | $\mathrm{f}_{\mathrm{i}}=5 \mathrm{MHz}$ | - | -100 | - | - | -88 | - | - | -85 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=70 \mathrm{MHz}$ | - | -97 | - | - | -90 | - | - | -81 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=140 \mathrm{MHz}$ | - | -88 | - | - | -89 | - | - | -85 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$ | - | -94 | - | - | -90 | - | - | -83 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | -96 | - | - | -87 | - | - | -80 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=230 \mathrm{MHz}$ | - | -95 | - | - | -85 | - | - | -80 | - | dBc |
| SFDR | spurious-free dynamic range | $\mathrm{f}_{\mathrm{i}}=5 \mathrm{MHz}$ | - | 87 | - | - | 84 | - | - | 79 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=70 \mathrm{MHz}$ | - | 85 | - | - | 82 | - | - | 75 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=140 \mathrm{MHz}$ | - | 92 | - | - | 85 | - | - | 80 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$ | - | 83 | - | - | 83 | - | - | 83 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | 82 | - | - | 86 | - | - | 80 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=230 \mathrm{MHz}$ | - | 78 | - | - | 80 | - | - | 80 | - | dBc |
| THD | total harmonic distortion | $\mathrm{f}_{\mathrm{i}}=5 \mathrm{MHz}$ | - | -86.5 | - | - | -82.3 | - | - | -80 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=70 \mathrm{MHz}$ | - | -84.2 | - | - | -80 | - | - | -72 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=140 \mathrm{MHz}$ | - | -85.3 | - | - | -82.8 | - | - | -77 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$ | - | -81.8 | - | - | -81.7 | - | - | -80 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | -81.4 | - | - | -81.9 | - | - | -78 | - | dBc |
|  |  | $\mathrm{f}_{\mathrm{i}}=230 \mathrm{MHz}$ | - | -77.5 | - | - | -78.5 | - | - | -78 | - | dBc |


|  | Symbol | Parameter | Conditions | $\begin{aligned} & \text { ADC1443D125 } \\ & \left(\mathrm{f}_{\mathrm{s}}=125 \mathrm{Msps}\right) \end{aligned}$ |  |  | ADC1443D160 <br> ( $\mathrm{f}_{\mathrm{s}}=154 \mathrm{Msps}$ ) |  |  | $\begin{aligned} & \text { ADC1443D200 } \\ & \left(\mathrm{f}_{\mathrm{s}}=185 \mathrm{Msps}\right) \end{aligned}$ |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Min | Typ | Max | Min | Typ | Max | Min | Typ | Max |  |
|  | IMD3 | third-order intermodulation distortion | $\begin{aligned} & \mathrm{f}_{\mathrm{i} 1}=3.5 \mathrm{MHz} ; \\ & \mathrm{f}_{\mathrm{i} 2}=6.5 \mathrm{MHz} \end{aligned}$ | - | 91 | - | - | 90 | - | - | 85 | - | dBc |
|  |  |  | $\begin{aligned} & \mathrm{f}_{\mathrm{i} 1}=68.5 \mathrm{MHz} ; \\ & \mathrm{f}_{\mathrm{i} 2}=71.5 \mathrm{MHz} \end{aligned}$ | - | 90 | - | - | 89 | - | - | 85 | - | dBc |
|  |  |  | $\begin{aligned} & \mathrm{f}_{\mathrm{i} 1}=138.5 \mathrm{MHz} ; \\ & \mathrm{f}_{\mathrm{i} 2}=141.5 \mathrm{MHz} \end{aligned}$ | - | 89 | - | - | 88 | - | - | 84 | - | dBc |
|  |  |  | $\begin{aligned} & \mathrm{f}_{\mathrm{i} 1}=168.5 \mathrm{MHz} ; \\ & \mathrm{f}_{\mathrm{i} 2}=171.5 \mathrm{MHz} \end{aligned}$ | - | 91 | - | - | 88 | - | - | 83 | - | dBc |
|  |  |  | $\begin{aligned} & \mathrm{f}_{\mathrm{i} 1}=188.5 \mathrm{MHz} ; \\ & \mathrm{f}_{\mathrm{i} 2}=191.5 \mathrm{MHz} \end{aligned}$ | - | 88 | - | - | 87 | - | - | 87 | - | dBc |
|  |  |  | $\begin{aligned} & \mathrm{f}_{\mathrm{i} 1}=228.5 \mathrm{MHz} ; \\ & \mathrm{f}_{\mathrm{i} 2}=231.5 \mathrm{MHz} \end{aligned}$ | - | 87 | - | - | 87 | - | - | 88 | - | dBc |
|  | SNR | signal-to-noise ratio | $\mathrm{f}_{\mathrm{i}}=5 \mathrm{MHz}$ | - | 72.6 | - | - | 71.9 | - | - | 68.8 | - | dBFS |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=70 \mathrm{MHz}$ | - | 72.4 | - | - | 71.7 | - | - | 68 | - | dBFS |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=140 \mathrm{MHz}$ | - | 72.1 | - | - | 71.3 | - | - | 68.3 | - | dBFS |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$ | - | 71.6 | - | - | 70.8 | - | - | 68.5 | - | dBFS |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | 71.2 | - | - | 70.6 | - | - | 68.5 | - | dBFS |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=230 \mathrm{MHz}$ | - | 70.6 | - | - | 70 | - | - | 67 | - | dBFS |
|  | ENOB | effective number of bits | $\mathrm{f}_{\mathrm{i}}=5 \mathrm{MHz}$ | - | 11.7 | - | - | 11.4 | - | - | 10.9 | - | bit |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=70 \mathrm{MHz}$ | - | 11.7 | - | - | 11.4 | - | - | 10.7 | - | bit |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=140 \mathrm{MHz}$ | - | 11.7 | - | - | 11.3 | - | - | 10.8 | - | bit |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$ | - | 11.5 | - | - | 11.3 | - | - | 10.9 | - | bit |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$ | - | 11.5 | - | - | 11.2 | - | - | 10.8 | - | bit |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=230 \mathrm{MHz}$ | - | 11.3 | - | - | 11.1 | - | - | 10.6 | - | bit |
|  | $\alpha_{\text {ct(ch }}$ | channel crosstalk | $\mathrm{f}_{\mathrm{i}}=140 \mathrm{MHz}$ | - | 95 | - | - | 95 | - | - | 92 | - | dBc |
|  |  |  | $\mathrm{f}_{\mathrm{i}}=230 \mathrm{MHz}$ | - | 90 | - | - | 90 | - | - | 92 | - | dBc |

[1] Typical values measured at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V} ; \mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$. Minimum and maximum values are across the full temperature range $\mathrm{T}_{\text {amb }}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ at $V_{D D A}=V_{D D O}=1.8 \mathrm{~V} ; \mathrm{V}_{\text {Idif) }}=2 \mathrm{~V} ; \mathrm{V}_{I N P}-\mathrm{V}_{\text {INM }}=-1 \mathrm{dBFS}$; unless otherwise specified.

Dual 14-bit ADC; 125, 160 or 185 Msps; JESD204A/B serial outputs

### 10.2 Timing

### 10.2.1 Clock timing

Table 7. Clock and digital output timing characteristics[1]

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{t}_{\text {lat(data) }}$ | data latency time | $\mathrm{F}=1$ | 54 | - | 55 | clock <br> cycles |
|  |  | $\mathrm{F}=2$ | 45.5 | - | 46 | clock <br> cycles |
|  |  | $\mathrm{F}=4$ | 41 | - | 41.25 | clock <br> cycles |
|  |  | from Power-down mode | - | 60 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {wake }}$ | wake-up time | from Sleep mode | - | 54 | - | $\mu \mathrm{s}$ |
| Clock timing |  |  |  |  |  |  |
| $\mathrm{f}_{\text {s }}$ | sampling rate | ADC1443D125 | 60 | - | 125 | MHz |
|  |  | ADC1443D160 | 125 | - | 160 | MHz |
|  |  | ADC1443D200 | 160 | - | 185 | MHz |
| $\mathrm{f}_{\text {clk }}$ | clock frequency |  | 60 | - | 800 | MHz |
| $\delta_{\text {Clk }}$ | clock duty cycle |  | 40 | - | 60 | $\%$ |

[1] Typical values measured at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V} ; \mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$. Minimum and maximum values are across the full temperature range $\mathrm{T}_{\mathrm{amb}}=-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V}$; $\mathrm{V}_{\text {I(dif) }}=2 \mathrm{~V}$; $\mathrm{V}_{\text {INP }}-\mathrm{V}_{\text {INM }}=-1 \mathrm{dBFS}$; unless otherwise specified.

### 10.2.2 SYSREFP/N and SYNCBP/N timings

Table 8. SYSREF timing

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{t}_{\mathrm{su}}$ | set-up time | 0.5 | - | - | ns |  |
| $\mathrm{t}_{\mathrm{h}}$ | hold time | $(\mathrm{tclk} / 2)$ | - | - | ns |  |
|  |  | -0.5 |  |  |  |  |

Table 9. SYNCB timing

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{t}_{\text {su }}$ | set-up time | 0.75 | - | - | ns |  |
| $\mathrm{t}_{\mathrm{h}}$ | hold time | $(\mathrm{tclk} / 2)$ | - | - | ns |  |
|  |  | -0.25 |  |  |  |  |



Fig 3. SYSREF timing

### 10.2.3 SPI timing

Table 10. SPI timing characteristics [1]

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{t}_{\mathrm{w}(\mathrm{SCLK})}$ | SCLK pulse width |  | 40 | - | - | ns |
| $\mathrm{t}_{\mathrm{w}(\mathrm{SCLKH})}$ | SCLK HIGH pulse width |  | 16 | - | - | ns |
| $\mathrm{t}_{\mathrm{w}(\mathrm{SCLKL})}$ | SCLK LOW pulse width |  | 16 | - | - | ns |
| $\mathrm{t}_{\text {su }}$ | set-up time | SDIO to SCLK HIGH | 5 | - | - | ns |
|  |  | SCS_N to SCLK HIGH | 5 | - | - | ns |
| $\mathrm{t}_{\mathrm{h}}$ | hold time | SDIO to SCLK HIGH | 2 | - | - | ns |
|  |  | SCS_N to SCLK HIGH | 2 | - | - | ns |
| $\mathrm{f}_{\text {clk }}$ | clock frequency |  | - | - | 25 | MHz |

[1] Typical values measured at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V} ; \mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}$. Minimum and maximum values are across the full temperature range $\mathrm{T}_{\mathrm{amb}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V}$


Fig 4. SPI timing

### 10.3 Typical dynamic performances ${ }^{1}$

### 10.3.1 Typical FFT at 122.88 Msps



Fig 5. 1-tone FFT: $-1 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=50 \mathrm{MHz}$; $\mathrm{f}_{\mathrm{s}}=122.88 \mathrm{Msps}$


Fig 7. 1-tone FFT: - $14 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=155 \mathrm{MHz} ;$ $\mathrm{f}_{\mathrm{s}}=122.88 \mathrm{Msps}$


Fig 6. 1-tone FFT: -1 dBFS; $\mathrm{f}_{\mathrm{i}}=155 \mathrm{MHz}$; $\mathrm{f}_{\mathrm{s}}=122.88 \mathrm{Msps}$


Fig 8. 2-tone FFT: $-7 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i} 1}=153.5 \mathrm{MHz}$; $\mathrm{f}_{\mathrm{i} 2}=156.5 \mathrm{MHz} ; \mathrm{f}_{\mathrm{s}}=122.88 \mathrm{Msps}$

[^0]
### 10.3.2 Typical FFT at 153.6 Msps



Fig 9. 1-tone FFT: $-1 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=50 \mathrm{MHz}$; $\mathrm{f}_{\mathrm{s}}=153.6 \mathrm{Msps}$


Fig 11. 1-tone FFT: - $14 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz} ;$ $\mathrm{f}_{\mathrm{S}}=153.6 \mathrm{Msps}$


Fig 10. 1-tone FFT: - $1 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$;

$$
\mathrm{f}_{\mathrm{s}}=153.6 \mathrm{Msps}
$$



Fig 12. 2-tone FFT: $-7 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i} 1}=188.5 \mathrm{MHz}$;
$\mathrm{f}_{\mathrm{i} 2}=191.5 \mathrm{MHz} ; \mathrm{f}_{\mathrm{s}}=153.6 \mathrm{Msps}$

### 10.3.3 Typical FFT at 184.32 Msps



Fig 13. 1-tone FFT: $-1 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=65 \mathrm{MHz}$; $\mathrm{f}_{\mathrm{s}}=184.32 \mathrm{Msps}$


Fig 15. 1-tone FFT: - $14 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$; $\mathrm{f}_{\mathrm{s}}=184.32 \mathrm{Msps}$


Fig 14. 1-tone FFT: $-1 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=190 \mathrm{MHz}$;

$$
\mathrm{f}_{\mathrm{s}}=184.32 \mathrm{Msps}
$$



Fig 16. 2-tone FFT: $-\mathbf{7} \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i} 1}=140 \mathrm{MHz}$;
$\mathrm{f}_{\mathrm{i} 2}=143 \mathrm{MHz} ; \mathrm{f}_{\mathrm{s}}=184.32 \mathrm{Msps}$

### 10.3.4 Typical SNR performances



Fig 17. SNR as a function of sampling frequency: $-1 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$


Fig 19. SNR as a function of input amplitude:
$\mathrm{V}_{\text {(dif) }}=2 \mathrm{~V}$


Fig 18. SNR as a function of input frequency: -1 dBFS


Fig 20. SNR as a function of full-scale amplitude: $-1 \mathrm{dBFS}$

### 10.3.5 Typical SFDR performances



Fig 21. SFDR as a function of sampling frequency: $-1 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i}}=170 \mathrm{MHz}$


Fig 23. SFDR as a function of input amplitude: $V_{\text {I(dif) }}=2 \mathrm{~V}$


Fig 22. SFDR as a function of input frequency: -1 dBFS


Fig 24. SFDR as a function of full-scale amplitude: $-1 \mathrm{dBFS}$

### 10.3.6 Typical IMD3 performances



Fig 25. IMD3 as a function of sampling frequency: $-7 \mathrm{dBFS} ; \mathrm{f}_{\mathrm{i} 1}=168.5 \mathrm{MHz} ; \mathrm{f}_{\mathrm{i} 2}=171.5 \mathrm{MHz}$


Fig 27. IMD3 as a function of input amplitude: 3 MHz spacing; $\mathrm{V}_{\mathrm{I}(\text { dif) }}=2 \mathrm{~V}$


Fig 26. IMD3 as a function of input frequency: -7 dBFS; 3 MHz spacing


Fig 28. IMD3 as a function of full-scale amplitude: $-7 \mathrm{dBFS} ; 3 \mathrm{MHz}$ spacing

## 11. Application information

### 11.1 Analog inputs

### 11.1.1 Input stage

The analog input of the ADC1443D supports a differential or a single-ended input drive. Optimal performance is achieved using differential inputs with respect to the common-mode input voltage $\left(\mathrm{V}_{\mathrm{I}(\mathrm{cm})}\right)$ on pins INP and INM.

The equivalent circuit of the sample and hold input stage, including ElectroStatic Discharge (ESD) protection circuit and package parasitics, is shown in Figure 29.


Fig 29. Input sampling circuit
The sample phase occurs when the internal sampling clock (derived from the clock signal on pin CLKP/CLKM) is HIGH. The voltage is then held on the sampling capacitors. When the sampling clock signal becomes LOW, the device enters the hold phase and the voltage information is transmitted to the ADC core.

### 11.1.2 Common-mode input voltage ( $\mathrm{V}_{\mathrm{I}(\mathrm{cm})}$ )

Set the common-mode input voltage ( $\mathrm{V}_{\mathrm{I}(\mathrm{cm})}$ ) on pins INP and INM externally to 0.9 V for optimal performance.

### 11.1.3 Pin VCM

When the input stage is AC-coupled, pin VCM can be used to set the common-mode reference for the analog inputs, for instance, via a transformer middle point. Connect a $0.1 \mu \mathrm{~F}$ filter capacitor between pin VCM and ground to ensure a low-noise common-mode output voltage.


Fig 30. Equivalent schematic of the common-mode reference circuit

### 11.1.4 Programmable full-scale

The full-scale analog input voltage range is configurable between $1 \mathrm{~V}(p-p)$ and $2 \mathrm{~V}(p-p)$ by programming internal reference gain between 0 dB and -6 dB in 1 dB steps. The full-scale range can be set independently via bits INTREF[2:0] of the SPI local registers (see Table 11 and Table 30).

Table 11. Reference gain control
Default values are shown highlighted.

| INTREF[2:0] | Level (dB) | Full-scale (V (p-p)) |
| :--- | :--- | :--- |
| $\mathbf{0 0 0}$ | $\mathbf{0}$ | $\mathbf{2}$ |
| 001 | -1 | 1.78 |
| 010 | -2 | 1.59 |
| 011 | -3 | 1.42 |
| 100 | -4 | 1.26 |
| 101 | -5 | 1.12 |
| 110 | -6 | 1 |
| 111 | reserved | $x$ |

### 11.1.5 Anti-kickback circuitry

An anti-kickback circuitry (RC-filter in Figure 31) is required to counteract the effects of the charge injection generated by the sampling capacitance.

The RC-filter is also used to filter noise from the signal before it reaches the sampling stage. It is recommended that the capacitor has a value that maximizes noise attenuation without degrading the settling time excessively.


Fig 31. Anti-kickback circuit
The input frequency determines the component values. Select values that do not affect the input bandwidth. The values given in the following table are advised for $50 \Omega$ impedance system.

Table 12. RC coupling versus input frequency; typical values

| Input frequency range $(\mathbf{M H z})$ | $\mathbf{R}(\Omega)$ | $\mathbf{C}(\mathbf{p F})$ |
| :---: | :---: | :---: |
| 0 to 50 | 25 | 12 |
| 50 to 200 | 10 | 3.9 |
| 200 to 300 | 6.8 | 3 |

### 11.1.6 Transformer

The input frequency determines the configuration of the transformer circuit. The configuration shown in Figure 32 is suitable for a baseband application.


Fig 32. Single transformer configuration (baseband)

The configuration shown in Figure 33 is recommended for high-frequency applications. In both cases, the choice of transformer is a compromise between cost and performance.


Fig 33. Dual transformer configuration (high IF)

### 11.2 Clock input

### 11.2.1 Drive modes

The ADC1443D series can be driven differentially (LVPECL, LVDS or SINE). A single-ended LVCMOS signal connected to either pin CLKP or pin CLKM can also drive the device (connect the complementary pin to ground using a capacitor). The LVPECL is recommended for an optimal performance.


Fig 34. LVPECL/LVDS differential clock input

a. Differential sine clock input

b. Single-ended sine clock input (with transformer)

Fig 35. Sine clock input

a. Rising edge LVCMOS
b. Falling edge LVCMOS

Fig 36. LVCMOS single-ended clock input
Single-ended or differential clock inputs can be selected via bit DIFF_SE of SPI. If single-ended is enabled, the input pin (pin CLKM or pin CLKP) is selected using control bit SE_SEL (see Table 29).

### 11.2.2 Equivalent input circuit

Figure 37 shows the equivalent circuit of the input clock buffer. The input signal must be AC-coupled and the common-mode voltage of the differential input stage is set via internal $5 \mathrm{k} \Omega$ resistors.


Fig 37. Equivalent input circuit

### 11.2.3 JESD204B harmonic clocking

The ADC1443D embeds an input clock divider that divides the incoming clock (clock frequency fclk) by a factor of 1 to 8 . The output of this divider is then used as sampling clock (sampling frequency fs) (see bits CLK_DIV[2:0] in Table 29). This feature allows a higher clock frequency to be delivered to the ADC1443D, which ultimately leads to better jitter performance and better SNR.
The ADC1443D should not be driven with a clock higher than 250Msps if the cloc is not enabled, otherwise there is a risk of metastability.

Caution must be taken to first, power the ADC1443 in « Power Down» mode by setting the CFG Pins to «1111» see Table 18, second, program the clock divider to the wanted value (see bits CLK_DIV[1:0] in Table 29) and finally, set the ADC using the SPI register IP_CFG_SETUP Table 42, to the wanted configuration.

### 11.2.4 JESD204B Deterministic Latency (pins SYSREFN and SYSREFP or SYNCBP and SYNCBN)

In the JESD204B standard 3 subclasses have been defined. Subclass 0: No deterministic latency is required (equivalent to the JESD204A)

Subclass 1: Deterministic latency is required and is realized through the dedicated SYSREFP/N pins.

The deterministic latency can be controlled with a single-ended or a differential SYSREF signal.

When SYSREF is active (High by default), it resets the clock divider phase registers. In a multi-device application and when the clock divider factor is higher than 1 , all sampling clock edges for multiple ADC1443D will be aligned (see Table 8 and Figure 3).

On top of this, the SYSREFP/N pins initiates an internal LMFC clock ( Local Multi-frame Clock), with a frequency of a multi-frame ( $\mathrm{Fs} / \mathrm{K}$ ) ( K : number of frames per multi-frame) see table Table 18 for examples.

At a SYNC request from the receiver (on pins SYNCBP/N), K28.5 comma characters are sent over the serial lanes. When the receiver releases the SYNC request, then the Initial Lane Alignement (ILA) will start at an edge of the LMFC
At the receiver side, the different lanes are aligned using the ILA start of frame ch and fetched at the next LMFC boundary.

This operation ensures a deterministic latency. see the JESD204B JEDEC standard for more information.

Subclass2 : Behaviour is similar to Subclass1, but, instead of using a dedicated SYSREF signal, the SYNCBP/N is used for both SYNC request and deterministic latency.

The rising edge of the SYNCBP/N start the LMFC, while the falling edge set the SYNC request and hence start the Initial Lane Alignement according to the JEDEC JESD204B standard.

Below is an example of a Subclass1 ADC1443D registers programming :
Table 13. Subclass1 path activation

| Register | value | Comment |
| :--- | :--- | :--- |
| DCS_CTRL (@0x043) | $0 \times C 7$ | Choose the SYSREFP/N on rising edge as DCS <br> Reset |
| JESD204B_CTRL1 (@810) | $0 \times C 0$ | Enable an LMFC periodic reset |
| JESD204B_CTRL2 (@811) | $0 \times 40$ | Enable a one shot DCS reset |
| JESD204B_CTRL3(@812) | $0 \times 0$ A | Activate a Sync fetch at LMFC boundary |
| SYSREF_CFG (@81E) | $0 \times 08$ | Enable SYSREFP/N on differential mode |

### 11.3 Digital outputs

### 11.3.1 Digital output buffers

The JESD204A/JESD204B standard specifies that both the receiver and the transmitter must share the same supply if they are connected in DC-coupling.


Fig 38. JESD204A/JESD204B serial output - DC-coupled


Fig 39. JESD204A/JESD204B serial output - AC-coupled

### 11.3.2 JESD204A/JESD204B serializer

### 11.3.2.1 Digital JESD204A/JESD204B formatter

The block placed after the ADC1443D cores implements all the JESD204A/JESD204B standard functionalities. This ensures signal integrity and guarantees the clock and the data recovery at the receiver side.

The block is highly configurable in various ways depending on the sampling frequency and the number of lanes used.All the processing and transmission are done with MSB first.



Fig 41. Detailed view of the JESD204A/JESD204B serializer with debug functionalities

### 11.3.2.2 Scrambler (SCR_EN)

The main purpose of scrambling is to avoid the spectral peaks that would be produced when the same data octet repeats from frame to frame. In general, scrambling makes the spectrum data-independent, so that possible frequency-selective effects on the electrical interface will not cause data-dependent errors. However, all digital operations in converters (including scrambling) cause some amount of switching noise, so there may be applications where it is of advantage to disable the scrambling.

The scrambler can be selected via the pin SCR_EN or the SPI registers (bit SCR_EN in Table 57).

Table 14. Scrambler configuration

| Pin SCR_EN | Scrambler |
| :--- | :--- |
| HIGH | enabled |
| LOW | disabled |

An internal pull-up resistor ( $50 \mathrm{k} \Omega$ ) sets pin SCR_EN to HIGH when no signal is connected to it. The pin SCR_EN is active only at start-up or after a JESD204B reset (bit SCR_EN in Table 41).

### 11.3.3 OuT-of-Range (OTR)

An out-of-range signal is provided on pins OTRA and OTRB. The OTR signal goes logic level HIGH when the input signal exceeds the maximum full scale range.

The latency of OTR is 31 clock cycles. The OTR response can be speeded up by enabling fast OTR using SPI local registers (bit FAST_OTR in Table 37). In this mode, the latency of OTR is reduced to only 11 clock cycles. The fast OTR detection threshold (below full-scale) can be programmed using the SPI local registers (bits FAST_OTR_DET[2:0] in Table 37).

Table 15. Fast OTR register threshold

| FAST_OTR_DET[2:0] | Detection level (dB) |
| :--- | :--- |
| 000 | -18.06 |
| 001 | -14.54 |
| 010 | -12.04 |
| 011 | -8.52 |
| 100 | -6.02 |
| 101 | -4.08 |
| 111 | -2.5 |

### 11.3.4 Digital offset

By default, the ADC1443D delivers an output code that corresponds to the analog input. However, it is possible to add a digital offset to the output code using the SPI local registers (bits DIG_OFFSET[5:0] in see Table 16 and Table 33). The digital offset adjustment is coded in two's complement.

Table 16. Digital offset adjustment
Default values are shown highlighted.

| DIG_OFFSET[5:0] | Digital offset adjustment (LSB) |
| :--- | :--- |
| 100000 | -32 |
| 100001 | -31 |
| $\ldots$ | $\ldots$ |
| 111111 | -1 |
| 000000 | $\mathbf{0}$ |
| 000001 | +1 |
| $\ldots$ | $\ldots$ |
| 011110 | +30 |
| 011111 | +31 |

### 11.3.5 Test patterns

The ADC1443D can be configured to transmit a number of predefined test patterns using the SPI local registers (bits TEST_PAT_SEL[2:0] in Table 17 and Table 34). The selected test pattern is transmitted regardless of the analog input.

Table 17. Digital test pattern selection
Default values are shown highlighted.

| TEST_PAT_SEL[2:0] | Digital test pattern |
| :--- | :--- |
| $\mathbf{0 0 0}$ | Off |
| 001 | Mid code |
| 010 | Min code |
| 011 | Max code |
| 100 | Toggle '1111..1111'/'0000..0000' |
| 101 | Custom test pattern |
| 110 | '0101..0101' |
| 111 | $' 1010 . .1010 '$ |

A custom test pattern can be defined using the SPI local registers (bits TEST_PAT_USER[13:6] in Table 35 and bits TEST_PAT_USER[5:0] in Table 36).

### 11.3.6 Output data format selection

The ADC1443D output data format can be selected (offset binary, two's complement or gray code) using the SPI local registers (bits DATA_FORMAT[1:0] in Table 32).

### 11.3.7 Output codes versus input voltage

Table 18. Output codes

| $\mathbf{V}_{\text {INP }}-\mathbf{V}_{\text {INM }}$ | Offset binary | Two's complement | Gray code | OTR |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $<-1$ | 00000000000000 | 10000000000000 | 00000000000000 | 1 |
| -1 | 00000000000000 | 10000000000000 | 00000000000000 | 0 |
| -0.99987793 | 00000000000001 | 10000000000001 | 00000000000001 | 0 |
| -0.99975586 | 00000000000010 | 00000000000010 | 00000000000011 | 0 |
| $\ldots$ | $\ldots$ | $\ldots$ | $\ldots$ | 0 |
| -0.00024414 | 01111111111110 | 11111111111110 | 01000000000001 | 0 |

Table 18. Output codes ...continued

| $\mathbf{V}_{\text {INP }}-\mathbf{V}_{\text {INM }}$ | Offset binary | Two's complement | Gray code | OTR |
| :--- | :--- | :--- | :--- | :--- |
| -0.00012207 | 01111111111111 | 11111111111111 | 01000000000000 | 0 |
| +0.00012207 | 10000000000000 | 00000000000000 | 11000000000000 | 0 |
| +0.0 .00024414 | 10000000000001 | 00000000000001 | 11000000000001 | 0 |
| $\ldots$ | $\ldots$ | $\ldots$ | $\ldots$ | 0 |
| +0.99975586 | 11111111111101 | 01111111111101 | 10000000000011 | 0 |
| +0.99987793 | 11111111111110 | 01111111111110 | 10000000000001 | 0 |
| +1 | 11111111111111 | 01111111111111 | 10000000000000 | 0 |
| $>+1$ | 11111111111111 | 01111111111111 | 10000000000000 | 1 |

### 11.4 Configuration pins (CFG0, CFG1, CFG2, CFG3)

The configuration pins are only active as inputs at start-up. The values on those pins are read once to set up the device. Then the pins become outputs (OTRA and OTRB). Any further modification must be applied via SPI registers.

Each of these pins is internally connected to a $50 \mathrm{k} \Omega$ pull-down resistor. In case of harmonic sampling, it is recommended to connect externally a $1 \mathrm{k} \Omega$ pull-up resistor in order to start in power-down mode.

Table 19. JESD204B configuration table

| CFG 3 | CFG 2 | CFG 1 | CFG 0 | ADC A | ADC B | Lane A | Lane B | F[1] | HD[1] | K[1] | M [1] | L[1] |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | ON | ON | ON | ON | 2 | 0 | 9 | 2 | 2 |
| 0 | 0 | 0 | 1 | ON | ON | ON | OFF | 4 | 0 | 5 | 2 | 1 |
| 0 | 0 | 1 | 0 | ON | ON | OFF | ON | 4 | 0 | 5 | 2 | 1 |
| 0 | 0 | 1 | 1 |  |  |  | reserved |  |  |  |  |  |
| 0 | 1 | 0 | 0 |  |  |  | reserved |  |  |  |  |  |
| 0 | 1 | 0 | 1 | ON | OFF | ON | OFF | 2 | 0 | 9 | 1 | 1 |
| 0 | 1 | 1 | 0 | ON | OFF | OFF | ON | 2 | 0 | 9 | 1 | 1 |
| 0 | 1 | 1 | 1 | OFF | ON | ON | OFF | 2 | 0 | 9 | 1 | 1 |
| 1 | 0 | 0 | 0 | OFF | ON | OFF | ON | 2 | 0 | 9 | 1 | 1 |
| 1 | 0 | 0 | 1 | ON | OFF | ON | ON | 1 | 1 | 17 | 1 | 2 |
| 1 | 0 | 1 | 0 | OFF | ON | ON | ON | 1 | 1 | 17 | 1 | 2 |
| 1 | 0 | 1 | 1 |  |  |  | reserved |  |  |  |  |  |
| 1 | 1 | 0 | 0 |  |  |  | reserved |  |  |  |  |  |
| 1 | 1 | 0 | 1 |  |  |  | reserved |  |  |  |  |  |
| 1 | 1 | 1 | 0 |  |  |  | reserved |  |  |  |  |  |
| 1 | 1 | 1 | 1 | OFF | OFF | OFF | OFF | 2 | 0 | 9 | 2 | 2 |

[1] F: Octets per frame clock cycle
HD: High-density mode
K: Frame per multi-frame
M: Converters per device
L: Lane per converter device

For all the configurations, the number of control bit per conversion sample (CS) is 1 , the number of control words per frame clock cycle and link (CF) is 0 , the number of samples transmitter per single converter per frame cycle $(S)$ is 1 and the formula $(F \times K) \geq 17$ is always verified.

### 11.5 Serial Peripheral Interface (SPI)

### 11.5.1 Register description

The ADC1443D serial interface is a synchronous serial communication port, which allows easy interfacing with many commonly used microprocessors. It provides access to the registers controlling the operation of the chip.

The register bits are either global or local functions:

- A global function operates over the full IC behavior. A local function operates on one or several previously selected channels only. If a channel is selected, the next WRITE command in the local registers applies to the selected channel. The WRITE command has no impact on channels that are not selected. This makes it possible to apply different configurations on each channel by first selecting a specific channel and then all the related settings.
- Select only one channel during a READ operation of the local registers. If several channels are selected, the READ operation occurs on the channel A.

Programming all registers at the same time is required:

- The IC allows the storage of a set of settings for the addresses 06h to 23 h , which enables the configuration of all registers simultaneously by setting bit TRANSFER to HIGH (see Table 39). This bit is autoclearing. This function can be disabled using SPI (bit TRANS_DIS in Table 39). The registers are then updated at each WRITE operation.
- The transfer function does not apply to a READ operation.

The SPI interface is configured as a 3-wire type: pin SDIO is the bidirectional pin, pin SCLK is the serial clock input and SCS_N is the chip select pin.

A LOW level on pin SCS_N initiates each READ/WRITE operation. A minimum of 3 bytes is transmitted (two instruction bytes and at least 1 DATA byte; see Table 21).

Table 20. Instruction bytes for the SPI

| Bit: | $\mathbf{7}(\mathbf{M S B})$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ (LSB) |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Description | $R / \bar{W}$ | W1 | W0 | A12 | A11 | A10 | A9 | A8 |
|  | A7 | A6 | A5 | A4 | A3 | A2 | A1 | A0 |

- Bit $R / \bar{W}$ indicates whether it is a READ (when HIGH) or a WRITE (when LOW) operation.
- Bits W1 and W0 indicate the number of bytes to be transferred after both instruction bytes (see Table 21).

Table 21. Number of data bytes transferred

| W1 | W0 | Number of bytes transferred |
| :--- | :--- | :--- |
| 0 | 0 | 1 byte |
| 0 | 1 | 2 bytes |
| 1 | 0 | 3 bytes |
| 1 | 1 | 4 or more bytes |

- Bits A12 to A0 indicate the address of the register being accessed. If it concerns a multiple byte transfer, this address is the first register accessed. An address counter is increased to access subsequent addresses.

The steps for a data transfer are:

1. Communication starts with the first rising edge on pin SCLK after a falling edge on pin SCS_N.
2. The first phase is the transfer of the 2-byte instruction.
3. The second phase is the transfer of the data. Its length varies, but it is always a multiple of 8 bits. The MSB is always sent first (for instruction and data bytes).
4. A rising edge on pin SCS_N indicates the end on data transmission.


Fig 42. SPI mode timing

### 11.5.2 Start-up programing

At power-up or after a reset by SPI, the device needs a start-up programming for optimum performances. This initialization is done in 3 steps:

Table 22. Step 1-Clock divider programming

| register address (hex) | value (hex) | comment |
| :--- | :--- | :--- |
| 0007 | CLK_DIV[2:0] in Table 29 | in case of harmonic clocking |

Table 23. Step 2 - JESD204B initialization

| register address (hex) | value (hex) | comment |
| :--- | :--- | :--- |
| 0803 | CFG_SETUP[3:0] in Table 42 | JESD204B configuration |
| 0802 | 08 | frame assembler subclock reset |

Table 24. Step 3 - ADC core initialization

| register address (hex) | ADC1443D125 value (hex) | $\begin{aligned} & \text { ADC1443D160 } \\ & \text { value (hex) } \end{aligned}$ | ADC1443D200 value (hex) | comment |
| :---: | :---: | :---: | :---: | :---: |
| 0100 | d1 | d1 | d1 |  |
| 0200 | 01 | 01 | 01 |  |
| 00ff | 80 | 80 | 80 | registers updated on each WRITE command |
| 0102 | 07 | 07 | 07 |  |
| 0103 | 63 | 65 | 66 |  |
| 0012 | 10 | 10 | 10 |  |
| 0108 | a7 | a3 | a3 |  |
| 010a | c0 | c0 | c0 |  |
| 0154 | 01 | 01 | 01 |  |
| 0155 | - | - | 03 |  |
| 0156 | 10 | - | d8 |  |
| 0160 | - | - | ff |  |
| 0161 | 17 | 07 | 17 |  |
| 0170 | - | - | 10 |  |
| 0171 | - | - | 10 |  |
| 0400 | b0 | b0 | 30 |  |
| - | - |  | - | wait for 400 ms |
| 0004 | 08 | 08 | 08 |  |
| - | - |  | - | wait for 400 ms |
| 0004 | 10 | 10 | 10 |  |
| - | - |  | - | wait for 400 ms |
| 0004 | 20 | 20 | 20 |  |

Those registers adjust some specific currents and timings. The programmed values should not be modified by the customer to ensure proper behavior over temperature and power supply variations.

### 11.5.3 Register allocation map

Table 25 shows an overview of all registers.
Table 25. Register allocation map


| ADC control registers |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0000h | CHIP_RST | RW | SW_RST[7:0] |  |  |  |  |  |  |  | 00000000 |
| 0001h | CHIP_ID | R | CHIP_ID[7:0][1] |  |  |  |  |  |  |  | 01000011 |
| 0005h | SW_RST | R/W | SW_RST | - | - | - | - | - | - | - | 00000000 |
| $\begin{aligned} & \text { 0006h } \\ & \text { [2] } \end{aligned}$ | OP_MODE | R/W | - | - | - | - | - | - | OP_MO | $E[1: 0]^{[3]}$ | 00000000 |
| 0007h | CLK_CFG | R/W | - | - | - | SE_SEL | DIFF_SE |  | K_DIV[2:0] |  | 00000000 |
| 0008h | INTERNAL_ REF | R/W | - | - | - | - | - |  | NTREF[2:0] |  | 00000000 |
| 0009h | CHANNEL_ SEL | R/W | - | - | - | - | - | - | ADC_B | ADC_A | 00001111 |
| 0011h | OUTPUT_ CFG | R/W | - | - | - | - | - | DATA SWAP | DATA_FO | RMAT[1:0] | 00000000 |
| 0013h | DIG_OFFSET | R/W | DIG_OFFSET[5:0] |  |  |  |  |  | - | - | 00000000 |
| 0014h | TEST_CFG_1 | R/W | - | - | - | - | - | TEST_PAT_SEL[2:0] |  |  | 00000000 |
| 0015h | TEST_CFG_2 | R/W | TEST_PAT_USER[13:6] |  |  |  |  |  |  |  | 00000000 |
| 0016h | TEST_CFG_3 | R/W | TEST_PAT_USER[5:0] |  |  |  |  |  | - | - | 00000000 |
| 0017h | OTR_CFG | R/W | - | - | - | RESERVED | $\begin{aligned} & \text { FAST_- } \end{aligned}$ | FAST_OTR_DET[2:0] |  |  | 00010100 |
| 0043h | DCS_CTRL | R/W | RESERVED[5:0] |  |  |  |  |  | $\begin{aligned} & \text { IV_RESET } \\ & \text { POL } \end{aligned}$ | DIV_RESE T_SEL | 11000100 |
| 00FFh | TRANS_CFG | R/W | $\begin{gathered} \text { TRANS_ } \\ \text { DIS } \end{gathered}$ | TRANSFER | - | - | - | - | - | - | 00000000 |

JESD204A/JESD204B control


Integrated Device Technology ADC1443D series

|  | Addr. <br> (hex) | Register name | R/W | Bit definition |  |  |  |  |  |  |  | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
|  | 0871h | $\begin{aligned} & \text { LANE_A_0_ } \\ & \text { CTRL } \end{aligned}$ | R/W | RESERVED[2:0] |  |  | LANE_MODE[1:0] |  | LANE_POL | RESERVED | LANE PD | 00000000 |
|  | 0872h | $\begin{aligned} & \text { LANE_B_0_ } \\ & \text { CTRL } \end{aligned}$ | R/W | RESERVED[2:0] |  |  | LANE_MODE[1:0] |  | LANE_POL | RESERVED | $\begin{gathered} \text { LANE_- }_{\text {PD }} \end{gathered}$ | 00000000 |
|  | 0890h | $\begin{aligned} & \text { ADC_A_0_ } \\ & \text { CTRL } \end{aligned}$ | R/W | - | - | ADC_MODE[1:0] |  | - | - | - | $\begin{gathered} \text { ADC- } \\ \text { PD } \end{gathered}$ | 00000000 |
|  | 0891h | $\begin{aligned} & \text { ADC_B_0_ } \\ & \text { CTRL } \end{aligned}$ | R/W | - | - | ADC | [1:0] | - | - | - | $\begin{gathered} \text { ADC } \\ \mathrm{PD}^{-} \end{gathered}$ | 00000000 |

[1] The READ-ONLY and RESERVED registers.
[2] The registers influenced by the TRANSFER function.
[3] The LOCAL registers.

### 11.5.4 Detailed register description

The tables in this section contain detailed descriptions of the registers.

### 11.5.4.1 ADC control registers

Table 26. CHIP_RESET register (address 0000h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 0 | SW_RST | R/W | - | resets global and local registers for any value "1" <br> written at any bit (autoclear). |

Table 27. SW_RESET register (address 0005h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | SW_RST | R/W |  | resets global and local registers |
|  |  | 0 | no reset |  |

Table 28. OP_MODE register (address 0006h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 2 | - | - | - | not used |
| 1 to 0 | OP_MODE[1:0] $\underline{[1]}$ | R/W |  | operating mode for the selected channel |
|  |  | $\mathbf{0 0}$ | normal (power-up) |  |
|  |  | 01 | power-down |  |
|  |  | 10 | sleep |  |
|  |  | 11 | not used |  |

[1] Local register.

Table 29. CLK_CFG register (address 0007h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 5 | - | - | - | not used |
| 4 | SE_SEL | R/W |  | single-ended clock input pin selection |
|  |  |  | 0 | CLKP |
|  |  | R/W |  | CLKM |
| 3 | DIFF_SE |  | 0 | differential/single-ended clock input selection |
|  |  | 1 | fully differential |  |
|  |  |  | single-ended |  |

Table 29. CLK_CFG register (address 0007h) bit description ...continued Default settings are shown highlighted.

| Bit | Symbol | Access | Value |
| :--- | :--- | :--- | :--- |
| 2 to 0 | CLK_DIV[2:0] | R/W |  |
|  |  | Description |  |
|  |  | 000 | clock divider selection |
|  |  | 001 | divide by 1 |
|  |  | 010 | divide by 2 |
|  |  | 011 | divide by 3 |
|  |  | 100 | divide by 4 |
|  |  | 101 | divide by 5 |
|  |  | 110 | divide by 6 |
|  |  | 111 | divide by 7 |
|  |  |  |  |
|  |  |  |  |
|  |  |  |  |

Table 30. INTERNAL_REF register (address 0008h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 3 | - | - | - | not used |
| 2 to 0 | INTREF[2:0] $\underline{[1]}$ | R/W | $\mathbf{0 0 0}$ | see Table 11 |

[1] Local register

Table 31. CHANNEL_SEL register (address 0009h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 2 | - | - | - | not used |
| 1 | ADC_B | R/W |  | channel B selection for next SPI operation in local <br> registers |
|  |  | 0 | not selected <br> selected |  |
|  |  | R/W | $\mathbf{1}$ | channel A selection for next SPI operation in local <br> registers |
| 0 | ADC_A |  | 0 | not selected <br>  |
|  |  | $\mathbf{1}$ | selected |  |

Table 32. OUTPUT_CFG register (address 0011h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 3 | - | - | - | not used |
| 2 | DATA_SWAP[1] | R/W |  | output data bits swapped |
|  |  |  | 0 | no swapping |
|  |  | 1 | MSBs swapped with LSBs |  |

Table 32. OUTPUT_CFG register (address 0011h) bit description ...continued Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 1 to 0 | DATA_FORMAT[1:0:][1] | R/W |  | output data format |
|  |  |  | $\mathbf{0 0}$ | offset binary |
|  |  | 01 | two's complement |  |
|  |  | 10 | gray code |  |
|  |  | 11 | offset binary |  |

[1] Local register

Table 33. DIG_OFFSET register (address 0013h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 2 | DIG_OFFSET[7:0] $\underline{[1]}$ | R/W | $\mathbf{0 0 0 0 0 0}$ | see Table 16 |
| 1 to 0 | - | - | - | not used |

[1] Local register

Table 34. TEST_CFG_1 register (address 0014h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 3 | - | - | - | not used |
| 2 to 0 | TEST_PAT_SEL[2:0] $\underline{[1]}$ | R/W | $\mathbf{0 0 0}$ | see Table 17 |

[1] Local register

Table 35. TEST_CFG_2 register (address 0015h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 0 | TEST_PAT_USER[13:6][] | R/W | $\mathbf{0 0 0 0 0 0 0 0}$ | custom digital test pattern (bits 13 to 6) |

[1] Local register

Table 36. TEST_CFG_3 register (address 0016h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 2 | TEST_PAT_USER[5:0] $\underline{[1]}$ | R/W | $\mathbf{0 0 0 0 0 0}$ | custom digital test pattern (bits 5 to 0 ) |
| 1 to 0 | - | - | - | not used |

[1] Local register

Table 37. OTR_CFG register (address 0017h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 5 | - | - | - | not used |
| 4 | RESERVED | R/W | 1 | reserved |

Table 37. OTR_CFG register (address 0017h) bit description ...continued Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 3 | FAST_OTR[1] | R/W |  | Selection OTR full-scale/ fast OTR |
|  |  |  | 0 | OTR full-scale |
|  |  |  | 1 | fast OTR |
| 2 to 0 | FAST_OTR_DET[2:0] $\underline{[1]}$ | R/W | $\mathbf{1 0 0}$ | see Table 15 |

[1] Local register

Table 38. DCS_CTRL register (address 0043h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 2 | RESERVED | R/W | 110001 | reserved |
| 1 | DIV_RESET_POL | R/W |  | Polarity of the DCS reset |
|  |  |  | $\mathbf{0}$ | falling edge (Subclass 2) |
|  |  | 1 | Rising edge (Subclass 1) |  |
| 0 | DIV_RESET_SEL | R/W |  | DCS reset selection |
|  |  |  | $\mathbf{0}$ | SYNCBP/N is used (Subclass 2) |
|  |  |  |  |  |
|  |  |  |  | SYSREFP/N is used (Subclass 1) |

Table 39. TRANS_CFG register (address 00FFh) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | TRANS_DIS | R/W |  | disable transfer function <br> transfer function active |
|  |  | 0 | 1 | registers updated on a WRITE command |

### 11.5.4.2 JESD204A/JESD204B control registers

Table 40. IP_STATUS register (address 0801 h ) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | RXSYNC_ERR_FLG | R |  | RX synchronization error |
|  |  |  | $\mathbf{0}$ | no error |
|  |  | R/W | 100001 | reserved |
| 6 to 1 | RESERVED | R |  | JEDEC PLL lock |
| 0 | PLL_LOCK |  | $\mathbf{0}$ | unlocked |
|  |  | 1 | locked |  |

Table 41. IP_RESET register (address 0802h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | SW_RST | R/W |  | resets All JESD204B sub-blocks and registers <br> no reset |
|  |  |  | 0 | 1 |

Table 42. IP_CFG_SETUP register (address 0803h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 4 | - | - | - | not used |
| 3 to 0 | CFG_SETUP[3:0] | R/W | $\mathbf{0 0 0 0}$ | see Table 43 |

Table 43. JESD204B configuration table

| CFG_SETUP [3:0] | ADC A | ADC B | Lane A | Lane B | F[1] | HD[1] | K[1] | M [1] | L[1] | Lane A serial frequency | Lane B serial frequency |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0000 | ON | ON | ON | ON | 2 | 0 | 9 | 2 | 2 | $20 \times \mathrm{f}_{\text {s }}$ | $20 \times \mathrm{f}_{\text {s }}$ |
| 0001 | ON | ON | ON | OFF | 4 | 0 | 5 | 2 | 1 | $40 \times \mathrm{f}_{\text {s }}$ | 0 |
| 0010 | ON | ON | OFF | ON | 4 | 0 | 5 | 2 | 1 | 0 | $40 \times \mathrm{f}_{\mathrm{s}}$ |
| 0011 | reserved |  |  |  |  |  |  |  |  |  |  |
| 0100 | reserved |  |  |  |  |  |  |  |  |  |  |
| 0101 | ON | OFF | ON | OFF | 2 | 0 | 9 | 1 | 1 | $20 \times \mathrm{f}_{\text {s }}$ | 0 |
| 0110 | ON | OFF | OFF | ON | 2 | 0 | 9 | 1 | 1 | 0 | $20 \times \mathrm{f}_{\text {s }}$ |
| 0111 | OFF | ON | ON | OFF | 2 | 0 | 9 | 1 | 1 | $20 \times \mathrm{f}_{\text {s }}$ | 0 |
| 1000 | OFF | ON | OFF | ON | 2 | 0 | 9 | 1 | 1 | 0 | $20 \times \mathrm{f}_{\text {s }}$ |
| 1001 | ON | OFF | ON | ON | 1 | 1 | 17 | 1 | 2 | $10 \times \mathrm{f}_{\mathrm{s}}$ | $10 \times \mathrm{f}_{\mathrm{s}}$ |

Table 43. JESD204B configuration table ...continued

| $\begin{aligned} & \text { CFG_SETUP } \\ & \text { [3:0] } \end{aligned}$ | ADC A | ADC B | Lane A | Lane B | $F \underline{[1]}$ | HD[1] | K[1] | M ${ }^{[1]}$ | L[1] | Lane A serial frequency | Lane B serial frequency |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1010 | OFF | ON | ON | ON | 1 | 1 | 17 | 1 | 2 | $10 \times \mathrm{f}_{\mathrm{s}}$ | $10 \times \mathrm{f}_{\mathrm{s}}$ |
| 1011 | reserved |  |  |  |  |  |  |  |  |  |  |
| 1100 | reserved |  |  |  |  |  |  |  |  |  |  |
| 1101 | reserved |  |  |  |  |  |  |  |  |  |  |
| 1110 | reserved |  |  |  |  |  |  |  |  |  |  |
| 1111 | OFF | OFF | OFF | OFF | 2 | 0 | 9 | 2 | 2 | 0 | 0 |

[1] F: Octets per frame clock cycle
HD: High-density mode
K: Frame per multi-frame
M: Converters per device
L: Lane per converter device
For all the configurations, the number of control bit per conversion sample (CS) is 1 , the number of control words per frame clock cycle and link (CF) is 0 , the number of samples transmitter per single converter per frame cycle $(S)$ is 1 and the formula $(F \times K) \geq 17$ is always verified.

Table 44. IP_CTRL1 register (address 0805h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | RESERVED | R/W | 0 | reserved |
| 6 | TRISTATE_CFG_PAD | R/W |  | CFG pad in tri-state mode |
|  |  |  | 0 | CFG Pads in Output mode |
|  |  | R/W |  | CFG Pads in Input mode; operating at power-up |

Table 45. IP_CTRL2 register (address 0806h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 2 | RESERVED | R/W | 001101 | reserved |
| 1 | SWP_LANE_A_B | R/W |  | swaps the lanes |
|  |  |  | 0 | no swap |
|  |  | 1 | lane $A$ and B are inverted |  |
|  |  |  |  |  |

Table 45. IP_CTRL2 register (address 0806h) bit description ...continued Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 0 | SWP_ADC_A_B | R/W |  | swaps the ADC at the input of the frame assembler |
|  |  | 0 | no swap |  |
|  |  | 1 | ADC A and B are inverted |  |

Table 46. IP_PRBS_CTRL register (address 080Bh) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 2 | RESERVED | R/W | 000000 | reserved |
| 1 | PRBS_TYPE | R/W |  | Pseudo-Random Binary Sequence (PRBS) pattern <br> selection |
|  |  |  | $\mathbf{0}$ | PRBS-7; 1+ $\mathbf{x}^{6}+\mathbf{x}^{7}$ |
|  |  |  | 1 | PRBS-23; $1+\mathbf{x}^{18}+\mathbf{x}^{23}$ |

Table 47. JESD204B_CTRL1 register (address 0810h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | LMFC_periodic_rst | R/W |  | LMFC mode definition |
|  |  | 0 | LMFC reset is done once |  |
|  |  | R/W |  | LMFC reset at each SYSREF or SYNC pulse |
| 6 | LMFC_reset_en |  | 0 | LMFC reset selection |
|  |  |  | SYNCBP/N is used (Subclass 2) |  |
|  |  | R/W | 00000 | reserved |
| 5 to 0 | RESERVED |  |  |  |

Table 48. JESD204B_CTRL2 register (address 0811h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | DCS_periodic_rst | R/W |  | DCS mode definition |
|  |  | $\mathbf{0}$ | DCS reset is done once |  |
|  |  | 1 | DCS reset at each SYSREF or SYNC pulse |  |
| 6 | DCS_reset_en |  | $\mathbf{0}$ | DCS reset selection |
|  |  | 1 | DCS reset is disabled |  |
|  |  | R/W | 00000 | DCS reset is enabled |
| 5 to 0 | RESERVED |  |  |  |

Table 49. JESD204B_CTRL3 register (address 0812h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 to 4 | RESERVED | R/W | 0000 | reserved |
| 3 | sync_at_Imfc_en | R/W |  | defines the relation between SYNC and LMFC |
|  |  |  | 0 | SYNC is fetched directly (Subclass 0) |
|  |  |  | 1 | SYNC is taken at next LMFC boundary (Subclass 1 and Subclass 2) |
| 2 | RESERVED | R/W | 0 | reserved |
| 1 | sync_capture_path | R/W |  | selects SYNC mode |
|  |  |  | 0 | Subclass 0 |
|  |  |  | 1 | Subclass 1 and Subclass 2 |
| 0 | RESERVED | R/W | 0 | reserved |

Table 50. IP_DEBUG_OUT1 register (address 0816h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 to 2 | - | - | - | not used |
| 1 to 0 | PATTERN_OUT[9:8] | R/W | 10 | 2 most significant bits of output stage debug word (inserted just before serializer) |

Table 51. IP_DEBUG_OUT2 register (address 0817h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 0 | PATTERN_OUT[7:0] | R/W | 10101010 | 8 least significant bits of output stage debug word <br> (inserted just before serializer) |

Table 52. IP_DEBUG_IN1 register (address 0818h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 0 | PATTERN_IN[15:8] | R/W | 11100110 | 8 most significant bits of input stage debug word <br> (inserted in place of ADC data) |

Table 53. IP_DEBUG_IN2 register (address 0819h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 0 | PATTERN_IN[7:0] | R/W | 11101010 | 8 least significant bits of input stage debug word <br> (inserted in place of ADC data) |

Table 54. IP_TESTMODE register (address 081Bh) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | RESERVED | R/W | 0 | reserved |
| 6 | LOOP_ALIGN | R/W |  | continuous ILA[1] sequence |
|  |  |  | 0 | normal operation |
|  |  | 1 | ILA[1] repeated continuously |  |

Table 54. IP_TESTMODE register (address 081Bh) bit description ...continued Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 5 | DIS_REPL_CHAR | R/W |  | character replacement function selection |
|  |  |  | 0 | normal operation |
|  |  | R/W |  | character replacement disabled |

[1] ILA = Initial Lane Alignment Sequence (see JESD204 JEDEC standard).

Table 55. IP_EXPERT_DOOR register (address 081Ch) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | KEY[7:0] | R/W | $\mathbf{0 0 0 0 0 0 0 0}$ | 8-bit key (0x4a) to enable write access for scrambler <br> (register 0828h) |

Table 56. SYSREF_CFG register (address 081Eh) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 4 | RESERVED | R/W | 0000 | reserved |
| 3 | SYSREF_EN | R/W |  | enables SYSREFP/N path |
|  |  |  | 0 | SYSREFP/N path disabled |
|  |  | R/W |  | SYSREFP/N path enabled |

Table 57. SCR_L register (address 0822h) bit description (IP_EXPERT_DOOR write access needed) Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 | SCR_EN | R/W |  | selects the scrambler function |
|  |  | 0 | scrambler disabled |  |
|  |  | 1 | scrambler enabled |  |
| 6 to 1 | RESERVED | R/W | 000000 | reserved |
| 0 | L |  |  | lanes number minus 1 |
|  |  |  | $\mathbf{1}$ | 1 lane |
|  |  |  | 2 lanes |  |

Table 58. CFG_K register (address 0824h) bit description (IP_EXPERT_DOOR write access needed, address 081Ch)
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 5 | - | - | - | not used |
| 4 to 0 | $\mathrm{~K}[4: 0]$ | R/W | $\mathbf{0 0 0 x} \mathbf{x x x x}$ | Number of frames in a multi-frame. Default value <br> depends on the JESD204B configuration. |

Table 59. IP_OUTBUF00_SWING register (address 086Bh) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 3 | RESERVED[4:0] | R/W | 00000 | reserved |
| 2 to 0 | SWING[2:0] | R/W |  | Configurable lane 0 output current |
|  |  | 000 | $12 \mathrm{~mA} ; \pm 300 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |
|  |  | 001 | $14 \mathrm{~mA} ; \pm 350 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |
|  |  | 010 | $16 \mathrm{~mA} ; \pm 400 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |
|  |  | 011 | $18 \mathrm{~mA} ; \pm 450 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |
|  |  | 100 | $20 \mathrm{~mA} ; \pm 500 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |
|  |  | 101 | $22 \mathrm{~mA} ; \pm 550 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |
|  |  | 110 | $24 \mathrm{~mA} ; \pm 600 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |
|  |  | 111 | $26 \mathrm{~mA} ; \pm 650 \mathrm{mV}(\mathrm{p}-\mathrm{p})$ |  |

Table 60. IP_OUTBUF01_SWING register (address 086Ch) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 to 3 | RESERVED[4:0] | R/W | 00000 | reserved |
| 2 to 0 | SWING[2:0] | R/W |  | Configurable lane 1 output current |
|  |  |  | 000 | 12 mA ; $\pm 300 \mathrm{mV}$ (p-p) |
|  |  |  | 001 | 14 mA ; $\pm 350 \mathrm{mV}$ (p-p) |
|  |  |  | 010 | 16 mA ; $\pm 400 \mathrm{mV}$ (p-p) |
|  |  |  | 011 | $18 \mathrm{~mA} ; \pm 450 \mathrm{mV}$ (p-p) |
|  |  |  | 100 | 20 mA ; $\pm 500 \mathrm{mV}$ (p-p) |
|  |  |  | 101 | 22 mA ; $\pm 550 \mathrm{mV}$ (p-p) |
|  |  |  | 110 | 24 mA ; $\pm 000 \mathrm{mV}$ (p-p) |
|  |  |  | 111 | 26 mA ; $\pm 650 \mathrm{mV}$ (p-p) |

Table 61. IP_LANE_A_0_CTRL register (address 0871h) bit description
Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 5 | RESERVED[2:0] | R/W | 000 | reserved |
| 4 to 3 | LANE_MODE[1:0] | R/W |  | debug option directly before serializer |
|  |  | $\mathbf{0 0}$ | normal mode, ADC path |  |
|  |  | 01 | $0 / 1$ toggle sent over the lanes |  |
|  |  | 10 | IP_DEBUG_OUT value sent over the lanes |  |
|  |  | 11 | 10-bit PRBS pattern is sent over the lane |  |

Table 61. IP_LANE_A_0_CTRL register (address 0871h) bit description ...continued Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 2 | LANE_POL | R/W |  | selects lane polarity |
|  |  | $\mathbf{0}$ | no inversion |  |
|  |  | 1 | lane polarity P/N inverted |  |
| 1 | RESERVED | R/W | 0 | reserved |
| 0 | LANE_PD | R/W |  | Selects lane power mode |
|  |  | $\mathbf{0}$ | lane is powered-up |  |
|  |  | 1 | lane is powered-down |  |

Table 62. IP_LANE_B_0_CTRL register (address 0872h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :---: | :---: | :---: | :---: | :---: |
| 7 to 5 | RESERVED[2:0] | R/W | 000 | reserved |
| 4 to 3 | LANE_MODE[1:0] | R/W |  | debug option directly before serializer |
|  |  |  | 00 | normal mode, ADC path |
|  |  |  | 01 | 0/1 toggle sent over the lanes |
|  |  |  | 10 | IP_DEBUG_OUT value sent over the lanes |
|  |  |  | 11 | 10-bit PRBS pattern is sent over the lane |
| 2 | LANE_POL | R/W |  | selects lane polarity |
|  |  |  | 0 | no inversion |
|  |  |  | 1 | lane polarity P/N inverted |
| 1 | RESERVED | R/W | 0 | reserved |
| 0 | LANE_PD | R/W |  | Selects lane power mode |
|  |  |  | 0 | lane is powered-up |
|  |  |  | 1 | lane is powered-down |

Table 63. IP_ADC_A_0_CTRL register (address 0890h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 6 | RESERVED | R/W | 00 | reserved |
| 5 to 4 | ADC_MODE[1:0] | R/W |  | debug option at ADC output |
|  |  | 00 | normal mode, ADC path |  |
|  |  | 01 | normal mode, ADC path |  |
|  |  | 10 | IP_DEBUG_IN value sent i.s.o. ADC data |  |
|  |  | R/W | 000 | 116-bit PRBS pattern is sent i.s.o. ADC data |
| 3 to 1 | RESERVED | R/W |  | selects ADC power mode |
| 0 | ADC_PD |  | $\mathbf{0}$ | ADC is powered-up |
|  |  |  |  | ADC is powered-down |
|  |  |  |  |  |

Table 64. IP_ADC_B_0_CTRL register (address 0891h) bit description Default settings are shown highlighted.

| Bit | Symbol | Access | Value | Description |
| :--- | :--- | :--- | :--- | :--- |
| 7 to 6 | RESERVED | R/W | 00 | reserved |
| 5 to 4 | ADC_MODE[1:0] | R/W |  | debug option at ADC output |
|  |  | $\mathbf{0 0}$ | normal mode, ADC path |  |
|  |  | 01 | normal mode, ADC path |  |
|  |  | 10 | IP_DEBUG_IN value sent i.s.o. ADC data |  |
|  |  | R/W | 000 | 116-bit PRBS pattern is sent i.s.o. ADC data |
| 3 to 1 | RESERVED | R/W |  | selects ADC power mode |
| 0 | ADC_PD |  | $\mathbf{0}$ | ADC is powered-up |
|  |  |  |  | ADC is powered-down |
|  |  |  |  |  |

## 12. Package outline

VFOFPN56 : plastic thermal enhanced low profile quad flat package; no leads;
56 terminals; resin based; body $8 \times 8 \times 1.35 \mathrm{~mm}$


Fig 43. Package outline PSC-4449 (VFQFPN56)

## 13. Abbreviations

Table 65. Abbreviations

| Acronym | Description |
| :--- | :--- |
| CDC | Analog-to-Digital Converter |
| DAV | Code Division Multiple Access |
| ESD | DAta Valid |
| FFT | ElectroStatic Discharge |
| GSM | Fast Fourier Transform |
| ILA | Global System for Mobile communications |
| IMD3 | Initial Lane Alignment |
| LSB | third order InterMoDulation product |
| LTE | Least Significant Bit |
| LVDS DDR | Long-Term Evolution Voltage Differential Signaling Double Data Rate |
| LVPECL | Low-Voltage Positive Emitter-Coupled Logic |
| MIMO | Multiple Input Multiple Output |
| MSB | Most Significant Bit |
| OTR | OuT-of-Range |
| SFDR | Spurious-Free Dynamic Range |
| SPI | Serial Peripheral Interface |
| SNR | Signal-to-Noise Ratio |
| TD-SCDMA | Time Division-Synchronous Code Division Multiple Access |
| WCDMA | Wideband Code Division Multiple Access |
| WiMAX | Worldwide interoperability for Microwave Access |
| TcIk | Period of the Sampling Clock |

## 14. Revision history

Table 66. Revision history

| Document ID | Release date | Data sheet status | Change notice | Supersedes |
| :--- | :--- | :--- | :--- | :--- |
| ADC1443D_SER v.4.4 | 20140606 | Product data sheet | - Parameter K added <br> - Offset error updated | ADC1443D_SER v.4.3 |
| ADC1443D_SER v.4.3 | 20131113 | Product data sheet | Registers updated | ADC1443D_SER v.4.2 |
| ADC1443D_SER v.4.2 | 20130402 | Product data sheet | new package outline | ADC1443D_SER v.4.1 |
| ADC1443D_SER v.4.1 | 20130227 | Product data sheet | - | ADC1443D_SER v.4.0 |
| ADC1443D_SER v.4.0 | 20130213 | Product data sheet | - | ADC1443D_SER v.3.6 |
| ADC1443D_SER v.3.6 | 20130208 | Data sheet | - | ADC1443D_SER v.3.5 |
| ADC1443D_SER v.3.5 | 20130111 | Data sheet | - | ADC1443D_SER v.3.4 |
| ADC1443D_SER v.3.4 | 20121010 | Data sheet | - | ADC1443D_SER v.3.3 |
| ADC1443D_SER v.3.3 | 20120926 | Objective data sheet | - | ADC1443D_SER v.3.2 |
| ADC1443D_SER v.3.2 | 20120918 | Objective data sheet | - | ADC1443D_SER v.3.1 |
| ADC1443D_SER v.3.1 | 20120911 | Objective data sheet | - | ADC1443D_SER v.3.0 |
| ADC1443D_SER v.3.0 | 20120901 | Objective data sheet | - | ADC1443D_SER v.2.0 |
| Modifications: | $\bullet$ | Text and drawings updated throughout entire data sheet. |  |  |
| ADC1443D_SER v.2.0 | 20120630 | Objective data sheet | - | ADC1443D_SER v.1.1 |
| ADC1443_SER v.1.1 | 20110928 | Objective data sheet | - | ADC1443D_SER v.1 |
| ADC1443D_SER v.1 | 20110901 | Objective data sheet | - | - |

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## Disclaimer

[^1]
[^0]:    1. Typical values measured at $\mathrm{V}_{\mathrm{DDA}}=\mathrm{V}_{\mathrm{DDO}}=1.8 \mathrm{~V} ; \mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}$
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