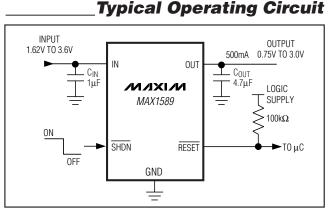
General Description

The MAX1589 low-dropout linear regulator operates from a +1.62V to +3.6V supply and delivers a guaranteed 500mA continuous load current with a low 175mV dropout. The high-accuracy (\pm 0.5%) output voltage is preset to internally trimmed voltages from +0.75V to +3.0V. An active-low, open-drain reset output remains asserted for at least 70ms after the output voltage reaches regulation. This device is offered in 6-pin thin SOT23 and 6-pin 3mm x 3mm thin DFN packages.

An internal PMOS pass transistor maintains low supply current, independent of load and dropout voltage, making the MAX1589 ideal for portable battery-powered equipment such as personal digital assistants (PDAs), digital still cameras, cell phones, cordless phones, and notebook computers. Other features include logic-controlled shutdown, short-circuit protection, and thermaloverload protection.

Applications

Notebook Computers Cellular and PCS Telephones Personal Digital Assistants (PDAs) Hand-Held Computers Digital Still Cameras PCMCIA Cards CD and MP3 Players



M/X/M

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

Features

- Low 1.62V Minimum Input Voltage
- Guaranteed 500mA Output Current
- ♦ ±0.5% Initial Accuracy
- Low 175mV Dropout at 500mA Load
- ♦ 70ms RESET Output Flag
- Supply Current Independent of Load and Dropout Voltage

- Logic-Controlled Shutdown
- Thermal-Overload and Short-Circuit Protection
 Preset Output Voltages (0.75V, 1.0V, 1.2V, 1.3V)
- Preset Output voltages (0.75V, 1.0V, 1.2V, 1.3V, 1.5V, 1.8V, 2.5V, and 3.0V)
- Tiny 6-Pin Thin SOT23 Package (<1.1mm High)</p>
- Thin 6-Pin TDFN Package (<0.8mm High)

Ordering Information

PART*	TEMP RANGE	PIN-PACKAGE
MAX1589EZTT	-40°C to +85°C	6 Thin SOT23-6
MAX1589ETT	-40°C to +85°C	6 TDFN

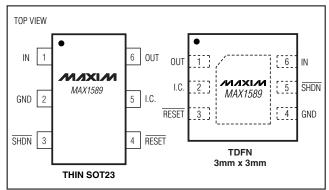
*Insert the desired three-digit suffix (see the Selector Guide) into the blanks to complete the part number. Contact the factory for other output voltages.

Selector Guide

Vout	SUFFIX	ТОР	MARK
(V)	SUFFIX	SOT	TDFN
0.75	075	AAAT	AFJ
1.00	100	AAAU	AFK
1.20	120	—	ATM
1.30	130	AAAV	AFL
1.50	150	AAAW	AFM
1.80	180	AAAX	AFN
2.50	250	AAAY	AFO
3.00	300	AAAZ	AFP

Pin Configurations

Maxim Integrated Products 1



ABSOLUTE MAXIMUM RATINGS

IN, SHDN, RESET to GND.....-0.3V to +4.0V OUT to GND.....-0.3V to (V_{IN} + 0.3V) Output Short-Circuit Duration.....Continuous Continuous Power Dissipation (TA = +70°C) 6-Pin Thin SOT23 (derate 9.1mW/°C above +70°C).....727mW 6-Pin TDFN (derate 24.4mW/°C above +70°C)......1951mW

Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{IN} = (V_{OUT} + 0.5V) \text{ or } V_{IN} = 1.8V$, whichever is greater; $\overline{SHDN} = IN$, $C_{IN} = 1\mu F$, $C_{OUT} = 4.7\mu F$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 1)

PARAMETER	SYMBOL	CON	DITIONS	MIN	ТҮР	MAX	UNITS
Input Voltage	VIN			1.62		3.60	V
Input Undervoltage Lockout	V _{UVLO}	V _{IN} rising (180mV typ	ical hysteresis)	1.30		1.60	V
		$I_{OUT} = 150 \text{mA}, T_{A} = -$	+25°C	-0.5		+0.5	
Output Voltage Accuracy		$I_{OUT} = 1$ mA to 500mA $V_{IN} = (V_{OUT} + 0.5V)$ to		-1.5		+1.5	%
Maximum Output Current	lout	Continuous		500			mA _{RMS}
Current Limit	ILIM	V _{OUT} = 96% of nomin	al value	550	850	1150	mA
		No load			70	140	
Ground Current	IQ	I _{OUT} = 500mA			90		μA
		Dropout (Note 2)			70		
Dropout Voltage	V _{IN} - V _{OUT}	I _{OUT} = 500mA, V _{OUT}	≥ 1.8V (Note 2)		175	350	mV
Load Regulation	ΔV_{LDR}	$I_{OUT} = 1$ mA to 500mA	ł		0.02	0.5	%
Line Regulation	ΔV_{LNR}	$V_{IN} = (V_{OUT} + 0.5V) t$	o +3.6V, I _{OUT} = 100mA	-0.15	+0.01	+0.15	%/V
Output Noise		10Hz to 100kHz, IOUT	- = 10mA		86		μV _{RMS}
PSRR		$f < 1 kHz$, $I_{OUT} = 10 m$	A		70		dB
SHUTDOWN							
Chutdown Supply Current	loss	$\overline{\text{SHDN}} = \text{GND}$	$T_A = +25^{\circ}C$		0.001	1	
Shutdown Supply Current	loff	SHDIN = GIND	$T_A = +85^{\circ}C$		0.01		μA
SHDN Input Logic Levels	VIH	$V_{IN} = 1.62V$ to 3.6V		1.4			V
SHDIN INput Logic Levels	VIL	$V_{IN} = 1.62V$ to 3.6V				0.6	v
SHDN Input Bias Current		$V_{\overline{SHDN}} = 0 \text{ or } 3.6V$	$T_A = +25^{\circ}C$		1	300	nA
SHDN Input Blas Current	I _{SHDN}	V SHDN = 0.013.0V	$T_A = +85^{\circ}C$		5		ПА
Turn-On Delay		From SHDN high to C	UT high, V _{OUT} = 1.5V		90		μs
RESET OUTPUT							
Reset Threshold Accuracy		V _{OUT} falling (1.7% typ	oical hysteresis)	80	82.5	85	%Vout
RESET Output Low Voltage	Mai	$I_{\overline{RESET}} = 100 \mu A$			1.5	100	mV
	V _{OL}	$V_{IN} = +1.0V, I_{\overline{RESET}} =$	= 100µA		3	100	IIIV

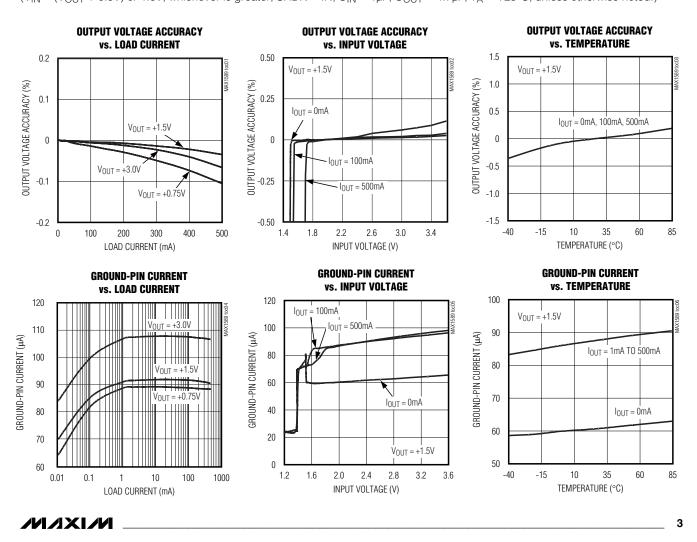
ELECTRICAL CHARACTERISTICS (continued)

 $(V_{IN} = (V_{OUT} + 0.5V) \text{ or } V_{IN} = 1.8V$, whichever is greater; $\overline{SHDN} = IN$, $C_{IN} = 1\mu F$, $C_{OUT} = 4.7\mu F$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 1)

PARAMETER	SYMBOL	CONE	DITIONS	MIN	ТҮР	МАХ	UNITS
RESET Output High	lou	$V_{\overline{RESET}} = 3.6V,$	$T_A = +25^{\circ}C$		0.001	1	μA
Leakage Current	ЮН	RESET not asserted	$T_A = +85^{\circ}C$		0.01		μΑ
Reset Delay	t _{RP}	From V_{OUT} high to \overline{RE}	SET rising	70	100	160	ms
THERMAL PROTECTION							
Thermal-Shutdown Temperature	TSHDN				+165		°C
Thermal-Shutdown Hysteresis	$\Delta T \overline{SHDN}$				15		°C

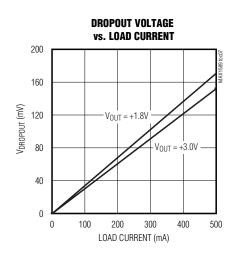
Note 1: Limits are 100% production tested at $T_A = +25^{\circ}C$. Limits over the operating temperature range are guaranteed by design. **Note 2:** The dropout voltage is defined as V_{IN} - V_{OUT} , when V_{OUT} is 4% lower than the value of V_{OUT} when $V_{IN} = V_{OUT} + 0.5V$.

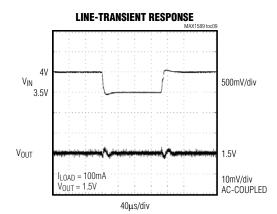
(V_{IN} = (V_{OUT} + 0.5V) or 1.8V, whichever is greater; SHDN = IN, C_{IN} = 1μF, C_{OUT} = 4.7μF, T_A = +25°C, unless otherwise noted.)

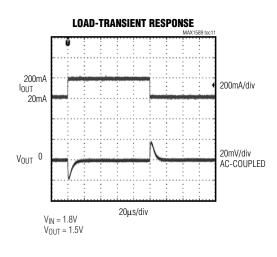


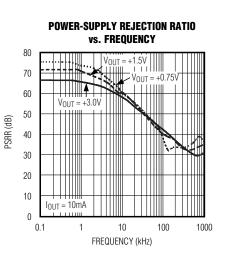
Typical Operating Characteristics (continued)

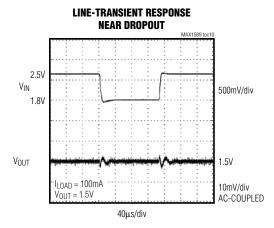
 $(V_{IN} = (V_{OUT} + 0.5V) \text{ or } 1.8V, \text{ whichever is greater; } \overline{SHDN} = IN, C_{IN} = 1\mu F, C_{OUT} = 4.7\mu F, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

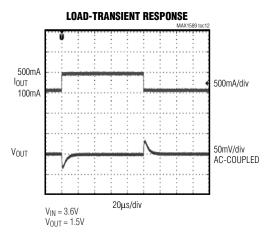










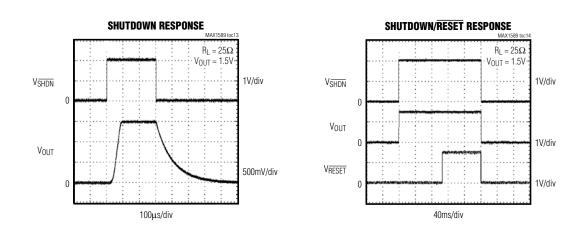


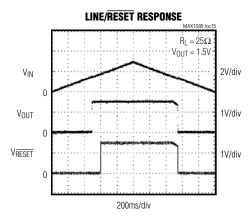
/N/XI/N

MAX1589

Typical Operating Characteristics (continued)

 $(V_{IN} = (V_{OUT} + 0.5V) \text{ or } 1.8V, \text{ whichever is greater; } \overline{SHDN} = IN, C_{IN} = 1\mu F, C_{OUT} = 4.7\mu F, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$





MAX1589

Pin Description

PA	RT	NAME	FUNCTION
SOT23	TDFN	NAME	FUNCTION
1	6	IN	Regulator Input. Supply voltage can range from $+1.62V$ to $+3.6V$. Bypass IN with at least a 1µF ceramic capacitor to GND (see the <i>Capacitor Selection and Regulator Stability</i> section).
2	—	GND	Ground. GND also functions as a heatsink. Solder GND to a large pad or circuit-board ground plane to maximize SOT23 power dissipation.
_	4	GND	Ground
3	5	SHDN	Active-Low Shutdown Input. A logic low reduces supply current to below 1µA. Connect to IN or logic high for normal operation.
4	3	RESET	Active-Low, Open-Drain Reset Output. RESET rises 100ms after the output has achieved regulation. RESET falls immediately if V_{OUT} drops below 82.5% of its nominal voltage, or if the MAX1589 is shut down.
5	2	I.C.	Internally Connected. Leave floating or connect to GND.
6	1	OUT	Regulator Output. Sources up to 500mA. Bypass with a 4.7µF low-ESR ceramic capacitor to GND.
_	Exposed Pad	EP	Ground. EP also functions as a heatsink. Solder EP to a large pad or circuit-board ground plane to maximize TDFN power dissipation.

Detailed Description

The MAX1589 is a low-dropout, low-quiescent-current, high-accuracy linear regulator designed primarily for battery-powered applications. The device supplies loads up to 500mA and is available with preset output voltages from +0.75V to +3.0V. As illustrated in Figure 1, the MAX1589 contains a reference, an error amplifier, a P-channel pass transistor, an internal feedback voltage-divider, and a power-good comparator.

The error amplifier compares the reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the reference voltage, the pass-transistor gate is pulled lower, allowing more current to pass to the output and increasing the output voltage. If the feedback voltage is too high, the passtransistor gate is pulled up, allowing less current to pass to the output.

Internal P-Channel Pass Transistor

The MAX1589 features a 0.33Ω (R_{DS(ON)}) P-channel MOSFET pass transistor. Unlike similar designs using PNP pass transistors, P-channel MOSFETs require no base drive, which reduces quiescent current. PNP-based regulators also waste considerable current in dropout when the pass transistor saturates and use high base-drive currents under large loads. The

MAX1589 does not suffer from these problems and consumes only $90\mu A$ (typ) of quiescent current under heavy loads, as well as in dropout.

Shutdown

Pull $\overline{\text{SHDN}}$ low to enter shutdown. During shutdown, the output is disconnected from the input, an internal 1.5k Ω resistor pulls OUT to GND, RESET is actively pulled low, and supply current drops below 1µA.

RESET Output

The MAX1589's microprocessor (μ P) supervisory circuitry asserts a guaranteed logic-low reset during power-up, power-down, and brownout conditions down to +1V. RESET asserts when V_{OUT} is below the reset threshold and remains asserted for at least 70ms (t_{RP}) after V_{OUT} rises above the reset threshold.

Current Limit

The MAX1589 monitors and controls the pass transistor's gate voltage, limiting the output current to 850mA (typ). If the output current exceeds I_{LIM} , the MAX1589 output voltage drops.

Thermal-Overload Protection

Thermal-overload protection limits total power dissipation in the MAX1589. When the junction temperature exceeds $+165^{\circ}$ C, a thermal sensor turns off the pass

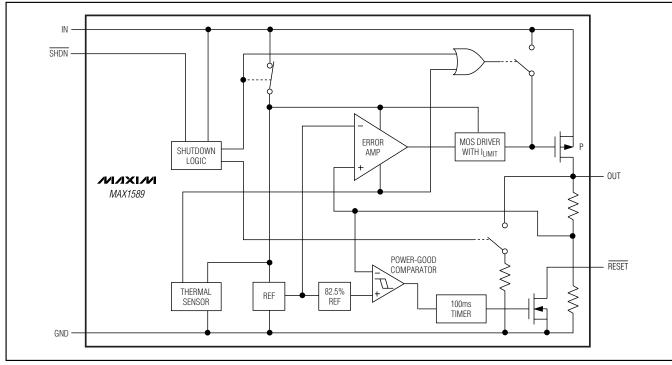


Figure 1. Functional Diagram

transistor, allowing the IC to cool. The thermal sensor turns the pass transistor on again after the junction temperature cools by 15°C, resulting in a pulsed output during continuous thermal-overload conditions. Thermal-overload protection safeguards the MAX1589 in the event of fault conditions. For continuous operation, do not exceed the absolute maximum junctiontemperature rating of +150°C.

Operating Region and Power Dissipation The MAX1589's maximum power dissipation depends on the thermal resistance of the IC package and circuit board, the temperature difference between the die junction and ambient air, and the rate of airflow. The power dissipated in the device is $P = I_{OUT} \times (V_{IN} - V_{OUT})$. The maximum allowed power dissipation is:

$$P_{MAX} = (T_{J(MAX)} - T_{A}) / (\theta_{JC} + \theta_{CA})$$

where $T_{J(MAX)}$ - TA is the temperature difference between the MAX1589 die junction and the surrounding air, θ_{JC} is the thermal resistance of the junction to the case, and θ_{CA} is the thermal resistance from the case through the PC board, copper traces, and other materials to the surrounding air. Typical thermal resistance ($\theta_{JC} + \theta_{JA}$) for a device mounted to a 1in square, 1oz copper pad is

41°C/W for the 3mm x 3mm TDFN package, and 110°C/W for the 6-pin thin SOT23 package. For best heatsinking, expand the copper connected to GND, or the exposed paddle.

The MAX1589 delivers up to 500mA and operates with an input voltage up to +3.6V. However, when using the 6-pin SOT23 version, high output currents can only be sustained when the input-output differential voltage is low, as shown in Figure 2.

The maximum allowed power dissipation for the 6-pin TDFN is 1.951W at $T_A = +70^{\circ}$ C. Figure 3 shows that the maximum input-output differential voltage is not limited by the TDFN package power rating.

Applications Information

Capacitor Selection and Regulator Stability

Capacitors are required at the MAX1589's input and output for stable operation over the full temperature range and with load currents up to 500mA. Connect a 1µF ceramic capacitor between IN and GND and a 4.7µF low-ESR ceramic capacitor between OUT and GND. The input capacitor (CIN) lowers the source impedance of the input supply. Use larger output

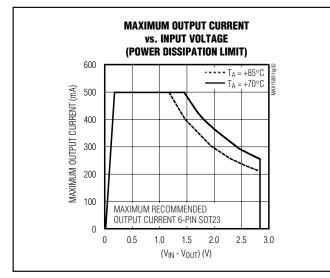


Figure 2. Power Operating Regions for 6-Pin SOT23: Maximum Output Current vs. Input Voltage

capacitors to reduce noise and improve load-transient response, stability, and power-supply rejection.

The output capacitor's equivalent series resistance (ESR) affects stability and output noise. Use output capacitors with an ESR of $30m\Omega$ or less to ensure stability and optimize transient response. Surface-mount ceramic capacitors have very low ESR and are commonly available in values up to 10μ F. Connect C_{IN} and C_{OUT} as close to the MAX1589 as possible to minimize the impact of PC board trace inductance.

Noise, PSRR, and Transient Response The MAX1589 is designed to operate with low dropout voltages and low quiescent currents in battery-powered systems, while still maintaining good noise, transient response, and AC rejection. See the *Typical Operating Characteristics* for a plot of Power-Supply Rejection Ratio (PSRR) vs. Frequency. When operating from noisy sources, improved supply-noise rejection and transient response can be achieved by increasing the values of the input and output bypass capacitors and through passive filtering techniques.

The MAX1589 load-transient response (see the *Typical Operating Characteristics*) shows two components of

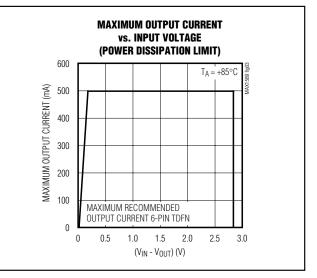


Figure 3. Power Operating Region for 6-Pin TDFN: Maximum Output Current vs. Input Voltage

the output response: a near-zero DC shift from the output impedance due to the load-current change, and the transient response. A typical transient response for a step change in the load current from 100mA to 500mA is 35mV. Increasing the output capacitor's value and decreasing the ESR attenuates the overshoot.

Input-Output (Dropout) Voltage

A regulator's minimum input-output voltage difference (dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this determines the useful end-of-life battery voltage. Because the MAX1589 uses a P-channel MOSFET pass transistor, its dropout voltage is a function of drain-to-source on-resistance (RDS(ON) = 0.33Ω) multiplied by the load current (see the *Typical Operating Characteristics*):

VDROPOUT = VIN - VOUT = $0.33\Omega \times I$ OUT

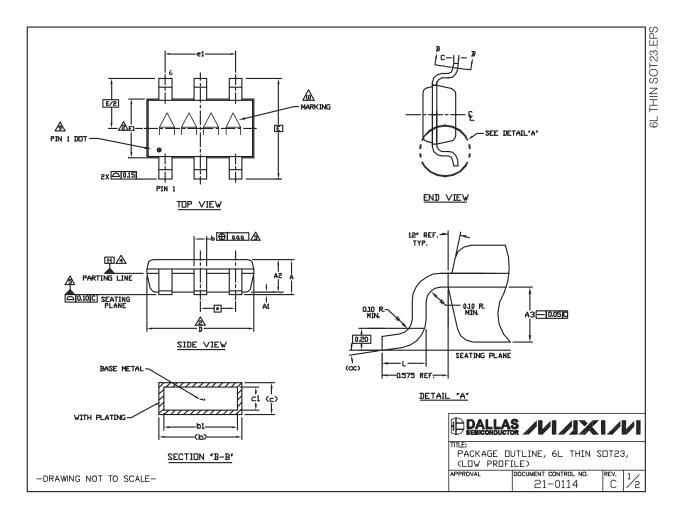
The MAX1589 ground current reduces to 70µA in dropout.

Chip Information

TRANSISTOR COUNT: 2556 PROCESS: BICMOS

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to <u>www.maxim-ic.com/packages</u>.)



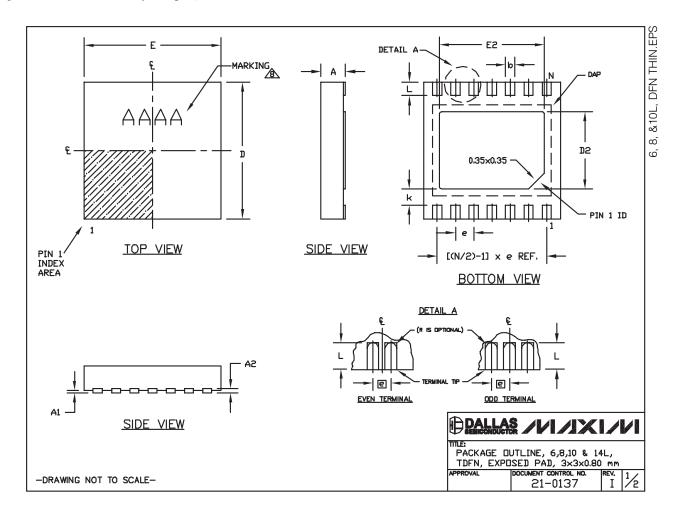
_Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to **www.maxim-ic.com/packages**.)

NDTES				
1. ALL DIMENSIONS ARE IN MILLIMETERS.		SYM	BOLS	
2 'D' AND 'E1' ARE REFERENCE DATUM AND DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS, AND ARE MEASURED AT THE BOTTOM PARTING LINE. MOLD FLASH OR		MIN	NDM	MAX
PROTRUSION SHALL NOT EXCEED 0.15mm ON "D" AND 0.25mm ON "E" PER SIDE.	A	-	-	1.10
3. The lead width dimension does not include dambar protrusion. Allowable	A1	0.00	0.075	0.10
DAMBAR PROTRUSION SHALL BE 0.07mm TOTAL IN EXCESS OF THE LEAD VIDTH DIMENSION AT MAXIMUM NATERIAL CONDITION	A2	0.85	0.88	0.90
4 datum plane "H" located at mold parting line and coincident with lead,	A3		0.50 BSC	
WHERE LEAD EXITS PLASTIC BODY AT THE BOTTOM OF PARTING LINE.	ю	0.30	-	0.45
5. THE LEAD TIPS NUST LINE WITHIN A SPECIFIED TOLERANCE ZONE. THIS	b1	0.25	0.35	0.40
TOLERANCE ZONE IS DEFINED BY TWO PARALLEL LINES, ONE PLANE IS THE SEATING PLANE, DATUM [-C-]; AND THE OTHER PLANE IS AT THE SPECIFIED	с	0.15	-	0.20
DISTANCE FROM (-C-) IN THE DIRECTION INDICATED. FORMED LEADS SHALL BE PLANAR WITH RESPECT TO DNE ANOTHER WITH 0.10mm AT SEATING PLANE.	с1	0.12	0.127	0.15
	D	2.80	2.90	3.00
6. THIS PART IS COMPLIANT WITH JEDEC SPECIFICATION MD-193 EXCEPT FOR THE "e" DIMENSION WHICH IS 0.95mm INSTEAD OF 1.00mm. THIS PART IS IN FULL	Е		2.75 BSC	
COMPLIANCE TO EIAJ SPECIFICATION SC-74,	E1	1.55	1.60	1.65
 COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERNINALS. COPLANARITY SHALL NOT EXCEED 0.08mm. 	L	0.30	0.40	0.50
8. WARPAGE SHALL NOT EXCEED 0.10mm.	e1		1.90 BSC	
<u>^</u>	e		0.95 BSC	
✓9. THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JESD 95-1 PP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE OPTIONAL. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED	00	0*	4*	8*
FEATURE.	aaa		0.20	
10 Marking is for package orientation reference only.	Pkg, c	odesı Z6	-1, Z6-2	
11. ALL DIMENSIONS APPLY TO BOTH LEADED (-> AND LEAD FREE (+> PACKAGE CODES.				
		1140		
			/////	XI/V
	TITLE:			
		AGE DUTLI PROFILE)	NE, 6L TH	IN SOT23,
	APPROVAL		MENT CONTROL N	
-DRAWING NOT TO SCALE-			21-0114	c 7

_Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to <u>www.maxim-ic.com/packages</u>.)



_Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to **www.maxim-ic.com/packages**.)

COMMON	DIMENS	SIONS		PACKAGE VA	RIATI	IONS					
SYMBOL	MIN.	MAX.		PKG. CODE	Ν	D2	E2	е	JEDEC SPEC	b	[(N/2)-1] x e
А	0.70	0.80		T633-2	6	1.50±0.10	2.30±0.10	0.95 BSC	MO229/WEEA	0.40±0.05	1.90 REF
D	2.90	3.10		T833-2	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
E	2.90	3.10		T833-3	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
A1	0.00	0.05		T1033-1	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229/WEED-3	0.25±0.05	2.00 REF
L	0.20	0.40		T1033-2	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229/WEED-3	0.25±0.05	2.00 REF
k	0.25	MIN.		T1433-1	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
A2	0.20	REF.		T1433-2	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
			- 151								
 COPL WARF PACK DRAW N[*] I NUME 	ANARITY AGE SH AGE LEI ING CO S THE BER OF	′ Shall Iall NG NGTH/F NFORMS TOTAL I LEADS	NOT EX T EXCEEL ACKAGE TO JED IUMBER SHOWN	. ANGLES IN CEED 0.08 m O 0.10 mm. WIDTH ARE C EC M0229, E OF LEADS. ARE FOR REF RIENTATION R	m. DNSID XCEP EREN	DERED AS S T DIMENSIO CE ONLY.	NS "D2" AN		C(S). ND T1433-1 & T	1433–2.	
1. ALL [2. COPL 3. WARF 4. PACK 5. DRAW 6. "N" 7. NUME	ANARITY AGE SH AGE LEI ING CO S THE BER OF	′ Shall Iall NG NGTH/F NFORMS TOTAL I LEADS	NOT EX T EXCEEL ACKAGE TO JED IUMBER SHOWN	CEED 0.08 m 0 0.10 mm. WIDTH ARE CO EC MO229, E DF LEADS. ARE FOR REFI	m. DNSID XCEP EREN	DERED AS S T DIMENSIO CE ONLY.	NS "D2" AN			OUTLINE,	6,8,10 & 14L J. 3×3×0.80 m CONTROL NO. [RE

Revision History

Pages changed at Rev 2: 1, 8-12

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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MAX1589

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