

# **LED Drivers for LCD Backlights**

# White LED Driver for large LCD Panels (DCDC Converter type)

# **BD93941EFV BD93941FP**

### **General Description**

BD93941EFV and BD93941FP is a high efficiency driver for white LEDs and designed for large LCDs. These ICs are built-in a boost DCDC converters that employ an array of LEDs as the light source. BD93941EFV and BD93941FP have some protect function against fault conditions, such as the over-voltage protection (OVP), the over current limit protection of DCDC (OCP), the short circuit protection (SCP), the open detection of LED string. Therefore BD93941EFV and BD93941FP are available for the fail-safe design over a wide range output voltage.

### **Key Specification**

■ Operating power supply voltage range: 9.0V to 35.0V

■ LED minimum current 30mA ■ LED maximum current: 200mA

■ Oscillator frequency: 150kHz (RT=100kΩ)

■ Operating Current: 4.5mA (Typ.)

■ Operating temperature range: -40°C to +85°C

### **Applications**

TV, Computer Display, Notebook, LCD Backlighting

 Package
 W(Typ.) x D(Typ.) x H(Max.)

 HTSSOP-B20
 6.50mm x 6.40mm x 1.00mm

 HSOP20
 14.90mm x 7.80mm x 2.10mm

### **Features**

- 4ch LED constant current driver and DC/DC converter
- Maximum LED Current: 200mA
- LED Feedback Voltage: 0.37V (@ADIM=2.5V) so lower heat. Adjustable Feed Back Voltage by following LED Current setting.
- ±2% LED current accuracy (ADIM=2.5V, when each LED is set to 100mA)
- Analog current (Linear) dimming at ADIM pin
- LED pin rating 60V
- Individual detection and individual LED OFF for both open and short circuits
- Built-in ISET pin short-circuit protection circuit
- Set Soft-Start time by external capacitor.
- FET's Gate (N pin) is driven by 5.8V swing
- Built-in Vout discharge circuit for shutdown
- Built-in Vout overvoltage protection (OVP) / reduced voltage protection (SCP) circuit
- HTSSOP-B20 and HSOP20 package with high heat radiation efficiency



Figure 1(a). HTSSOP-B20 (BD93941EFV)



Figure 1(b). HSOP20 (BD93941FP)

### **Typical Application Circuit**

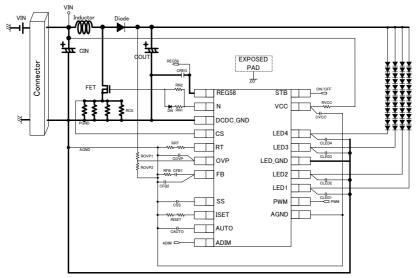


Figure 2. Typical Application Circuit

OProduct structure: Silicon monolithic integrated circuit OThis product has not designed protection against radioactive rays

Absolute maximum ratings (Ta=25°C)

| Parameter   | Symbol                                  | Ratings            | Unit     |
|---|---|--------------------|----------|
| Power supply voltage                                  | VCC                                     | 36                 | V        |
| STB, ADIM, OVP, PWM terminal voltage                  | STB, ADIM, OVP, PWM                     | 36                 | <b>V</b> |
| LED1 to 4 terminal voltage                            | LED1 to 4                               | 60                 | V        |
| AUTO, REG58, CS, N, ISET, SS, FB, RT terminal voltage | AUTO, REG58, CS, N,<br>ISET, SS, FB, RT | 7                  | <b>\</b> |
| Power dissipation (HTSSOP-B20)<br>(HSOP20)            | Pd1<br>Pd2                              | 3.20 *1<br>2.18 *2 | W        |
| Operating temperature range                           | Topr                                    | -40 to +85         | °C       |
| Storage temperature range                             | Tstg                                    | -55 to +150        | °C       |
| Junction temperature                                  | Tjmax                                   | 150                | °C       |

<sup>\*1</sup> Ta = 25°C or more, diminished at -25.6mW/°C in the case of HTSSOP-B20 (when 4-layer / 70.0 mm x 70.0 mm x 1.6 mm board is mounted)

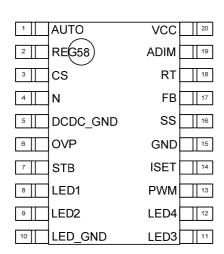
**Operating Ratings** 

| orating realings                              |          |  |            |
|---|----------|--|------------|
| Parameter                                     | Symbol   | Limits   | Unit       |
| VCC supply voltage                            | VCC      | 9.0 to 35.0                                      | V          |
| Min. output current of LED1 to 4              | ILED_MIN | 30   | mA *1      |
| Max. output current of LED1 to 4              | ILED_MAX | 200  | mA<br>*1,2 |
| ADIM input voltage1 (use ADIM function)       | VADIM1   | 0.2 to 2.7 (normal op.)<br>1.0 to 2.7 (start up) | V *3       |
| ADIM input voltage1 (don't use ADIM function) | VADIM2   | REG58 to 35.0                                    | V          |
| DC/DC oscillation frequency                   | Fsw      | 100 to 800                                       | kHz        |
| Min. on-duty time for PWM light modulation    | PWM_MIN  | 30   | μs         |
|   |          |  |            |

<sup>\*1</sup> The amount of current per channel.

# **Pin Configuration**

HTSSOP-B20



HSOP20

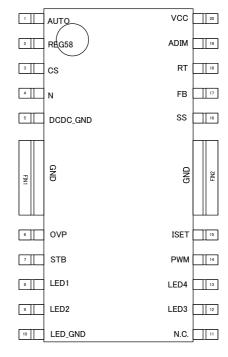


Figure 3. Pin Configuration

<sup>\*2</sup> Ta = 25°C or more, diminished at -17.4mW/°C in the case of HSOP20 (when 4-layer / 70.0 mm x 70.0 mm x 1.6 mm board is mounted)

<sup>\*2</sup> If LED makes significant variations in its reference voltage, the driver will increase power dissipation, resulting in a rise in package temperature.

To avoid this problem, design the board with thorough consideration given to heat radiation measures.

<sup>\*3</sup> To avoid unused LED pin's misdetection, set ADIM within 1.0V to 2.7V at start up stage. After unused LED pin's detection, set ADIM within 0.2V to 2.7V in normal operation.

# Marking diagram and physical dimension

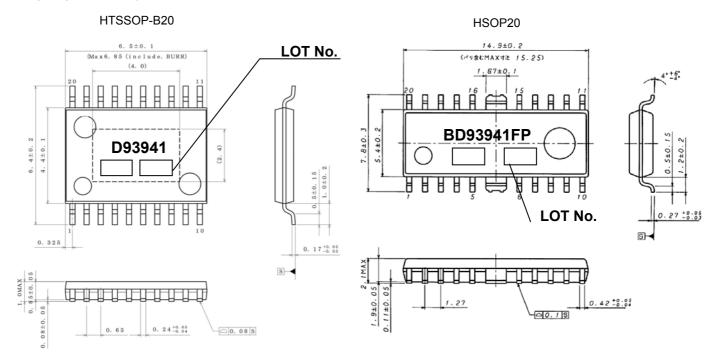


Figure 4. Physical Dimension

Electrical Characteristics (Unless otherwise noted, Ta = 25°C, VCC=24V)

| Parameter                           | Symbol         |       | Limit |           | Unit | Condition                  |
|-------------------------------------|----------------|-------|-------|-----------|------|----------------------------|
| Parameter                           | Min. Typ. Max. |       | Unit  | Condition |      |                            |
| [Whole Device]                      | ,              |       |       |           |      |                            |
| Circuit current while in operation  | ICC            | -     | 4.5   | 9         | mA   | STB=3V,PWM=3V,<br>RT=100kΩ |
| Circuit current while in standby    | ISTB           | -     | 40    | 80        | μΑ   | STB=0V                     |
| [REG58 Block]                       |                |       |       |           |      |                            |
| REG58 Output Voltage                | REG58          | 5.742 | 5.8   | 5.858     | V    | IO=0mA                     |
| REG58 available current             | IREG58         | 15    | -     | -         | mA   |                            |
| [UVLO Block]                        |                |       |       |           |      |                            |
| UVLO release voltage                | VUVLO_VCC      | 6.5   | 7.5   | 8.5       | V    | VCC=SWEEP UP               |
| UVLO hysteresis voltage             | VUHYS_VCC      | 150   | 300   | 600       | mV   | VCC=SWEEP DOWN             |
| [DC/DC Block]                       |                |       |       |           |      |                            |
| Error amp. Reference voltage        | VLED           | 0.35  | 0.37  | 0.39      | V    | ISET=75kΩ, ADIM=2.5V       |
| Oscillation frequency               | fsw            | 142.5 | 150.0 | 157.5     | kHz  | RT=100kΩ                   |
| Max. duty cycle per output of N pin | DMAX           | 83    | 90    | 97        | %    | RT=100kΩ                   |
| On resistance on N pin source side  | RONH           | -     | 4     | 8         | Ω    | ION=-10mA                  |
| On resistance on N pin sink side    | RONL           | -     | 3     | 6         | Ω    | ION=10mA                   |
| SS pin source current               | ISSSO          | -4    | -2    | -1        | uA   | SS=2V                      |
| Soft start completion voltage       | VSS_END        | 3.3   | 3.7   | 4.1       | V    | SS=SWEEP UP                |
| FB sink current                     | IFBSINK        | 50    | 100   | 150       | μA   | LED=2.0V, FB=1.0V          |
| FB source current                   | IFBSOURCE      | -150  | -100  | -50       | μA   | LED=0V, FB=1.0V            |
| Over current detection voltage      | VCS            | 0.40  | 0.45  | 0.50      | V    | CS=SWEEP UP                |

**Electrical Characteristics** (Unless otherwise noted, Ta = 25°C, VCC=24V)

| Parameter.  | O        |      | Limit |       | 11:4 | O a sa distina sa                    |
|---|----------|------|-------|-------|------|--------------------------------------|
| Parameter   | Symbol   | Min. | Тур.  | Max.  | Unit | Condition                            |
| [DC/DC Protection Block]                            |          |      |       | II.   | I    |                                      |
| Overvoltage protection detection voltage            | VOVP     | 2.7  | 3.00  | 3.3   | V    | OVP=SWEEP UP                         |
| Overvoltage protection detection hysteresis voltage | VOVP_HYS | 50   | 100   | 200   | mV   | OVP=SWEEP DOWN                       |
| Short circuit protection detection voltage          | VSCP     | 0.04 | 0.10  | 0.25  | V    | OVP=SWEEP DOWN                       |
| [LED Driver Block]                                  |          |      |       |       |      |                                      |
| LED pin current accuracy 1                          | dILED1   | -2   | -     | 2     | %    | ILED=100mA,<br>(ADIM=2.5V,ISET=75kΩ) |
| LED pin current accuracy 2                          | dILED2   | -3.2 | -     | 3.2   | %    | ILED=70mA,<br>(ADIM=1.75V,ISET=75kΩ) |
| LED pin current accuracy 3                          | dILED3   | -4.6 | -     | 4.6   | %    | ILED=50mA,<br>(ADIM=1.25V,ISET=75kΩ) |
| LED pin current accuracy 4                          | dILED4   | -8   | -     | 8     | %    | ILED=30mA,<br>(ADIM=0.75V,ISET=75kΩ) |
| LED pin current accuracy 5                          | dILED5   | -3   | -     | 3     | %    | ILED=100mA,<br>(ADIM=7V,ISET=75kΩ)   |
| LED pin Leakage Current                             | ILLED    | -2.5 | -     | 2.5   | μA   | LED=60V                              |
| LED open detection voltage                          | VOPEN    | 0.05 | 0.2   | 0.285 | V    | LED=SWEEP DOWN                       |
| LED short detection voltage                         | VSHORT   | 4    | 5     | 6     | V    | LED=SWEEP UP                         |
| ADIM pin Input Current                              | ILADIM   | -2.5 | -     | 2.5   | μA   | ADIM=3V                              |
| [STB Block]   |          |      |       |       |      |                                      |
| STB pin high-level voltage                          | STBH     | 2    | -     | 35    | V    | STB=SWEEP UP                         |
| STB pin low-level voltage                           | STBL     | -0.3 | -     | 0.8   | V    | STB=SWEEP DOWN                       |
| STB pin pull-down resistance                        | RSTB     | 500  | 1000  | 1500  | kΩ   | STB=3.0V                             |
| [PWM Block]   |          |      | II.   | I     |      |                                      |
| PWM pin high-level voltage                          | PWMH     | 2    | -     | 35    | V    | PWM=SWEEP UP                         |
| PWM pin low-level voltage                           | PWML     | -0.3 | -     | 0.8   | V    | PWM= SWEEP DOWN                      |
| PWM pin pull-down resistance                        | RPWM     | 180  | 300   | 420   | kΩ   | PWM=3.0V                             |
| [Failure Indication Block (Open Dr                  | rain)]   |      |       |       |      |                                      |
| AUTO pin source current                             | IAUTO    | -2   | -1    | -0.5  | μΑ   | AUTO=2V                              |
| AUTO pin Detection Voltage                          | VAUTO    | 3.6  | 4.0   | 4.4   | V    | AUTO=SWEEP UP                        |
| Abnormal Detection Timer                            | tCP      |      | 20    |       | ms   | RT=75kΩ                              |

Pin Descriptions (BD93941EFV)

| Pin No | Pin Name | In/Out | Function  | Rating [V] |
|--------|----------|--------|---|------------|
| 1      | AUTO     | Out    | Auto-restart time setting pin                             | -0.3 to 7  |
| 2      | REG58    | Out    | Power supply for N pin                                    | -0.3 to 7  |
| 3      | CS       | In     | DC/DC output current detection and OCP detection pin      | -0.3 to 7  |
| 4      | N        | Out    | DC/DC switching output pin                                | -0.3 to 7  |
| 5      | DCDC_GND | -      | Power GND pin   | -          |
| 6      | OVP      | In     | Overvoltage protection detection pin                      | -0.3 to 36 |
| 7      | STB      | In     | Enable pin  | -0.3 to 36 |
| 8      | LED1     | Out    | Output pin 1 for LED                                      | -0.3 to 60 |
| 9      | LED2     | Out    | Output pin 2 for LED                                      | -0.3 to 60 |
| 10     | LED_GND  | -      | Ground pin for LED  | -          |
| 11     | LED3     | Out    | Output pin 3 for LED                                      | -0.3 to 60 |
| 12     | LED4     | Out    | Output pin 4 for LED                                      | -0.3 to 60 |
| 13     | PWM      | In     | External PWM light modulation signal input pin for LED1-4 | -0.3 to 36 |
| 14     | ISET     | Out    | LED current setting resistor connection pin               | -0.3 to 7  |
| 15     | GND      | -      | Analog GND pin  | -          |
| 16     | SS       | Out    | Soft start pin / LED protection masking time setting pin. | -0.3 to 7  |
| 17     | FB       | In/Out | Error amp output pin                                      | -0.3 to 7  |
| 18     | RT       | Out    | DC/DC drive frequency setting resistor connection pin.    | -0.3 to 7  |
| 19     | ADIM     | In     | Analog dimming DC voltage input pin                       | -0.3 to 36 |
| 20     | VCC      | In     | Power supply pin  | -0.3 to 36 |

Pin Descriptions (BD93941FP)

| Pin No | Pin Name | In/Out | Function  | Rating [V] |
|--------|----------|--------|---|------------|
| 1      | AUTO     | Out    | Auto-restart time setting pin                             | -0.3 ~ 7   |
| 2      | REG58    | Out    | Power supply for N pin                                    | -0.3 ~ 7   |
| 3      | CS       | In     | DC/DC output current detection and OCP detection pin      | -0.3 ~ 7   |
| 4      | N        | In     | DC/DC switching output pin                                | -0.3 ~ 7   |
| 5      | DCDC_GND | -      | Power GND pin   | -          |
| FIN1   | GND      | -      |   | -          |
| 6      | OVP      | In     | Overvoltage protection detection pin                      | -0.3 ~ 36  |
| 7      | STB      | In     | Enable pin  | -0.3 ~ 36  |
| 8      | LED1     | Out    | Output pin 1 for LED                                      | -0.3 ~ 60  |
| 9      | LED2     | Out    | Output pin 2 for LED                                      | -0.3 ~ 60  |
| 10     | LED_GND  | -      | Ground pin for LED  | -          |
| 11     | N.C.     | -      | Non connection pin  | -          |
| 12     | LED3     | Out    | Output pin 3 for LED                                      | -0.3 ~ 60  |
| 13     | LED4     | Out    | Output pin 4 for LED                                      | -0.3 ~ 60  |
| 14     | PWM      | In     | External PWM light modulation signal input pin for LED1-4 | -0.3 ~ 36  |
| 15     | ISET     | Out    | LED current setting resistor connection pin               | -0.3 ~ 7   |
| FIN2   | GND      | -      | Analog GND pin  | -          |
| 16     | SS       | Out    | Soft start pin / LED protection masking time setting pin. | -0.3 ~ 7   |
| 17     | FB       | In/Out | Error amp output pin                                      | -0.3 ~ 7   |
| 18     | RT       | Out    | DC/DC drive frequency setting resistor connection pin.    | -0.3 ~ 7   |
| 19     | ADIM     | In     | Analog dimming DC voltage input pin                       | -0.3 ~ 36  |
| 20     | VCC      | In     | Power supply pin  | -0.3 ~ 36  |

# Pin ESD Type

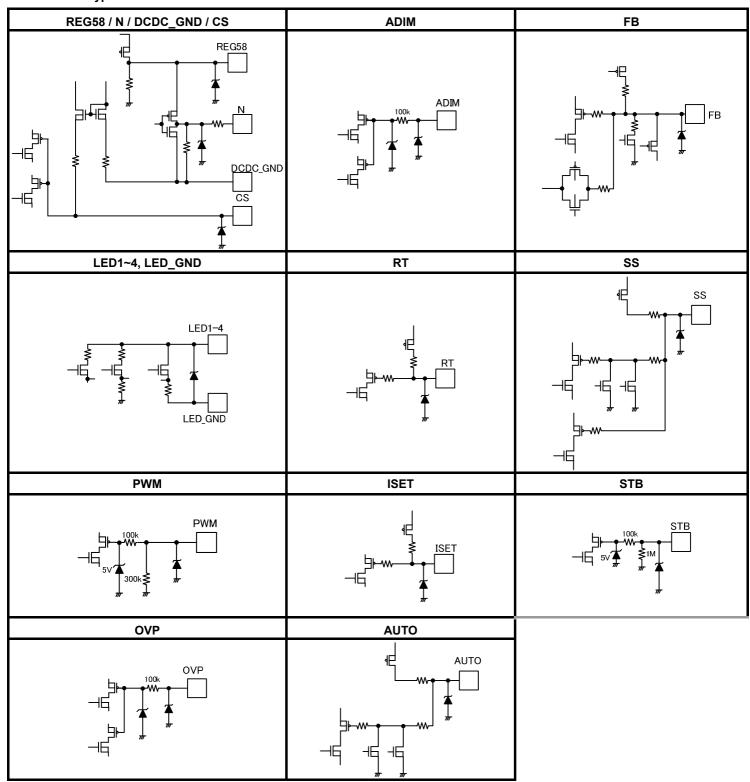


Figure 5. Pin ESD Type

# **Block Diagram**

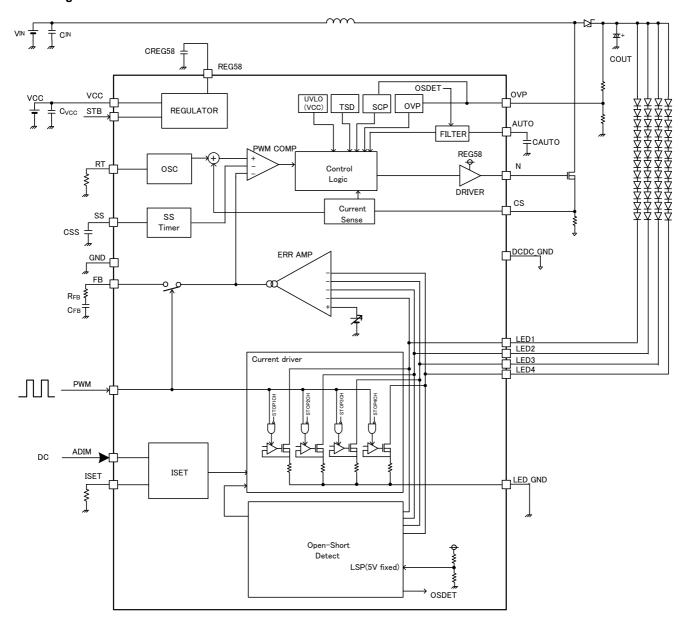


Figure 6. Block Diagram

# **Typical Performance Curve**

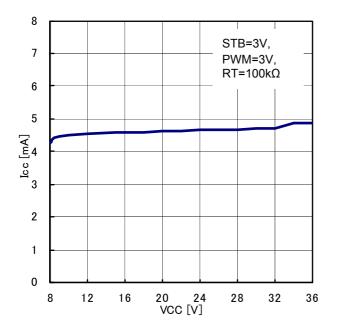


Figure 7. Operating Current (ICC) [mA] vs. VCC[V]

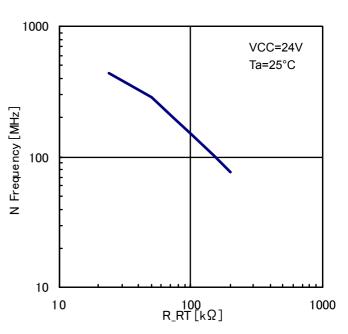


Figure 8. N Frequency [MHz] vs. R\_RT [ $k\Omega$ ]

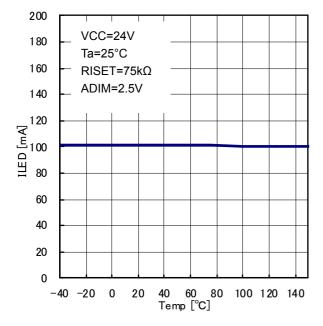


Figure 9. LED Current (ILED) [mA] vs. Temp [°C]

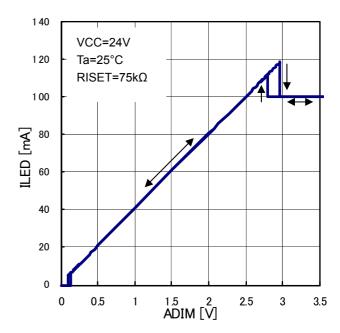


Figure 10. LED Current (ILED) [mA] vs. ADIM[V],

### **Pin Function**

# ∘ AUTO (1pin)

This sets up time till auto-restart time from the point of abnormal detection. Having 1µA constant current charge at external capacitor connected to AUTO pin, it will start again when it becomes over 4.0V (The auto pin is shorted to GND, this IC's protection function operates latched off mode).

# oAuto-restart period vs. AUTO capacitance (Ideal)

$$T_{AUTO} = \frac{4.0[V] \times C_{AUTO}}{1.0 \times 10^{-6} [A]} = 4.0 \times 10^{-6} \times C_{AUTO} [\text{sec}]$$

### o REG58 (2pin)

The REG58 pin is used in the DC/DC converter driver block to output 5.8V power. The maximum operating current is 15mA. Using the REG58 pin at a current higher than 15mA can affect the N pin output pulse, causing the IC to malfunction and leading to heat generation of the IC itself. To avoid this problem, it is recommended to make load setting to the minimum level.

Please place the ceramic capacitor connected to REG58 pin (2.2µF~10µF) closest to REG58-GND pin.

### oCS (3pin)

The CS pin has the following two functions:

### 1. DC/DC current mode current feedback function

Current flowing through the inductor is converted into voltage by the current sensing resistor RCS connected to the CS pin and this voltage is compared with voltage set with the error amplifier to control the DC/DC output voltage.

# 2. Inductor current limit function

The CS pin also incorporates the over current protection (OCP) function. If the CS pin voltage reaches 0.45V (Typ.) or more, switching operation will be forcedly stopped.

# ∘N (4pin)

The N pin is used to output power to the external NMOS gate driver for the DC/DC converter in the amplitude range of approx. 0 to REG58. ON resistances is  $4.0\Omega(typ.)$  in sorrce (H side),  $3.0\Omega(typ.)$  in sink (L side).

Frequency setting can be made with a resistor connected to the RT pin. For details of frequency setting, refer to the description of the RT pin.

### o DCDC\_GND (5pin)

The PGND pin is a power ground pin for the driver block of the output pin N.

# ○OVP (6pin)

The OVP pin is an input pin for over-voltage protection and short circuit protection of DC/DC output voltage. If over-voltage is detected, the OVP pin will stop the DC/DC converter conducting step-up operation. When the short circuit protection (SCP) function is activated, the DC/DC converter will stop operation, and then the timer will start counting. When the timer completes counting the preset period of time, the LED drivers are stopped.

The OVP pin is of the high impedance type and involves no pull-down resistor, resulting in unstable potential in the open-circuited state. To avoid this problem, be sure to make input voltage setting with the use of a resistive divider or otherwise.

# ∘STB (7pin)

The STB pin is used to make setting of turning ON and OFF the IC and allowed for use to reset the IC from shutdown.

Note: The IC state is switched (i.e., the IC is switched between ON and OFF state) according to voltages input in the STB pin. Avoid using the STB pin between two states (0.8 to 2.0V).

# oLED1 - LED4 (8,9,11,12pin)

The LED1 to 4 pins are used to output constant current to LED drivers. Current value setting can be made by connecting a resistor to the ISET pin. For the current value setting procedure, refer to the description of "ISET pin".

If any of the LED pins is put in an erroneous state (e.g. short circuit mode, open circuit mode, or ground short circuit mode), the relevant protection function will be activated.

### oLED GND (10pin)

The LED GND pin is a power ground pin used for the LED driver block.

# ∘*PWM (13pin)*

The PWM pin is used to turn ON and OFF LED drivers. Light can be modulated by changing the duty cycle through the direct input of a PWM light modulation signal. The high and low voltage levels of PWM pin is as listed in the table below:

| State         | PWM Voltage        |
|---------------|--------------------|
| LED ON state  | PWM= 2.0V to 35V   |
| LED OFF state | PWM= -0.3V to 0.8V |

### o ISET (14pin)

The ISET pin is an output current setting resistor. Output current ILED varies in inverse proportion to resistance.

The relation between output current ILED and the resistance of ISET pin connection resistor RISET is given by the following

equation:

$$I_{LED}[mA] = \frac{3000}{R_{ISET}[k\Omega]} \times V_{ADIM}[V]$$
 (ADIM=0.2V to2.7V) 
$$I_{LED}[mA] = \frac{7500}{R_{ISET}[k\Omega]}$$

Output current setting should be made in the range of 30 to 200mA.

It prepares automatically to suitable LED feedback voltage that can output LED current set by ISET pin.

In short LED feedback voltage is dropped when the LED current is small and the IC heating is held automatically.

In case of a large current is needed, raise the LED pin feedback voltage. And it adjusts automatically to LED pin voltage that can be flow large LED current.

The calculation is as below.

$$VLED = 3.7 \times I_{LED} [A] [V]$$

The LED feedback voltage (VLED) is clamped to 0.3V (typ.) when the LED current (ILED) is less than 81.1mA.

ADIM input range is from 0V to 35V. And the range which the LED currents change with linearity is from 0.2V to 2.7V. When it reaches under VISET×0.90V(typ.), the LED current is off to prevent from passing a large current to the LED pin when the RISET is shorted and the ISET pin is shorted to the GND. And as the ISET pin returns to a normal state, the LED current returns.

### ○ GND (15pin)

The GND pin is an internal analog circuit ground of the IC.

# ∘SS (16pin)

The SS pin is used to make setting of soft start time and duty for soft start. It performs constant current charge of 2.0 uA to the external capacitor connected with SS terminal, which enables soft-start of DC/DC converter.

Since the LED protection function (OPEN/SHORT detection) works when the SS terminal voltage reaches 3.7 V (typ.) or higher, it must be set to bring stability to conditions such as DC/DC output voltage and LED constant current drive operation, etc. before the voltage of 3.7 V is detected.

# ∘FB (17pin)

The FB pin is an output pin used for DC/DC current mode control error amplifier. In other words, the FB pin detects the voltages of LED pins (1 to 4) and controls inductor current so that the pin voltage of the LED located in the row with the highest Vf will come to 0.37V (ADIM=2.5V, ILED=100mA). As a result, the pin voltages of other LEDs become higher by Vf variation. After completion of soft start, the FB pin is put into the high-impedance state with the PWM signal being in the low state, thus maintaining the FB voltage.

# ∘RT (18pin)

The RT pin is used to connect a DC/DC frequency setting resistor. DC/DC drive frequency is determined by connecting the RT resistor.

### ODrive frequency vs. RT resistance (Ideal)

$$R_{RT} = \frac{15000}{f_{SW}[kHz]} [k\Omega]$$

When RT is 100kΩ, Fsw is 150kHz(typ.). However, drive frequency setting should be made in the range of 100 kHz to 800 kHz.

# ∘ *ADIM* (19pin)

ADIM pin is for analog dimming. Output current is proportionality with input voltage. Basically, ADIM pin assumes the voltage inputted externally using high accuracy of resistive divider and etc., IC internally is in OPEN (High impedance) condition. Cannot use in an OPEN condition.

If you don't use analog dimming, please connect pull-up resistor to over 5V (for example REG58).

### o VCC (20pin)

The VCC pin is used to supply power for the IC in the range of 9 to 35V.

If the VCC pin voltage reaches 7.5V (Typ.) or more, the IC will initiate operation. If it reaches 7.2V (Typ.) or less, the IC will be shut down.

# Startup operation and soft start (SS) capacitance setting

The following section describes the sequence for the startup of this IC.

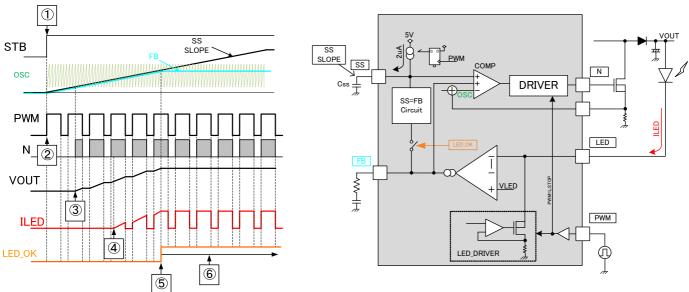


Figure 11. Startup waveform

Figure 12. Circuit behavior at startup

### ODescription of startup sequence

- (1) Set the STB and PWM pin to "ON".
- (2) Set sll systems to "ON", SS charge will be initiated. At this time, a circuit in which SS pin voltage for soft start becomes equal to FB pin voltage operates to equalize the FB pin and SS pin voltages regardless of whether the PWM pin is set to Low or High level.
- (3) Since the FB pin and SS pin reach the lower limit of the internal sawtooth wave of the IC, the DC/DC converter operates to start VOUT voltage rising.
- (4) The Vout voltage continues rising to reach a voltage at which LED current starts flowing.
- (5) When the LED current reaches the set amount of current, isolate the FB circuit from the SS circuit. With this, the startup operation is completed.
- (6) After that, conduct normal operation following the feedback operation sequence with the LED pins.
  If the SS pin voltage reaches 3.7V or more, the LED protection function will be activated to forcedly end the SS and FB equalizing circuit.

### **OSS** capacitance setting procedure

As aforementioned, this IC stops DC/DC converter when the PWM pin is set to Low level and conducts step-up operation only in the section in which the PWM pin is maintained at High level. Consequently, setting the PWM duty cycle to the minimum will extend the startup time. The startup time also varies with application settings of output capacitance, LED current, output voltage, and others.

Startup time at minimum duty cycle can be approximated according to the following method:

Make maeasurement of VOUT startup time with a 100% duty cycle, first. Take this value as "Trise100".

The startup time "Trise min" for the relevant application with the minimum duty cycle is given by the following equation.

$$T_{rise\_min} = \frac{T_{rise\_100}[Sec]}{Min\ Duty[ratio]} [Sec]$$

However, since this calculation method is just for approximation, use it only as a guide.

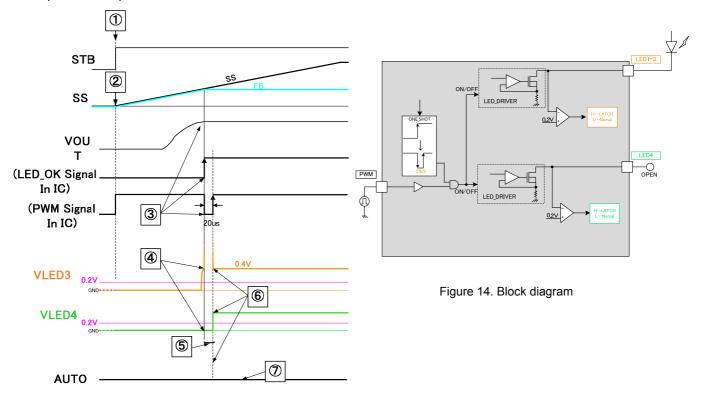
Make setting of time during which the SS pin voltage reaches the FB pin voltage longer than this startup time. Assuming that the FB pin voltage is VFB, the time is given by the following equation:

$$T_{ss} = \frac{C_{ss}[F] \times VFB[V]}{2[\mu A]} \quad [Sec]$$

As a result, it is recommended to make SS capacitance setting so that "Tss" will be greater than "Trise min"

# O About unused LED terminal automatic detecting function

This IC is detected automatically that it is an unused channel by asssuming the LED terminal to be OPEN at starting. It explains the sequence.



# Sequence;

- ① STB=ON
- ② All systems are ON at initial timing of PWM=H. SS starts charging.
- ③ When the output voltage is boosted enough, and enough current flows through the LED, LED\_OK signal is switched in the IC. PWM=L from the Rise timing of this signal for about 20us
- ④ During this PWM=L period, LED pins with LED connections' output voltage becomes 0.2V and above, where as unused LED pins are below 0.2V.
- ⑤ During this time, determination on whether the LED pins are 0.2V above/below is done.
- ⑥ After the determination, unused LED pins are pulled up to 5V.
- 7)The AUTO signal remains "L" level.

In addition, automatic determination of the OPEN decision will only be in SS range, therefore, please set the application so that the step-up/boost be completed before SS> 3.7V.

### LED current setting

Setting of LED output current "ILED" can be made by connecting a resistor RISET to the ISET pin.

# ORISET vs. ILED current relation equation

$$R_{ISET} = \frac{7500}{I_{IED}[mA]} \quad [k\Omega] \tag{ADIM=2.5V}$$

However, LED current setting should be made in the range of 30mA to 200mA.

# [Setting example]

To set ILED current to 100mA, RISET resistance is given by the following equation:

$$R_{ISET} = \frac{7500}{I_{LED}[mA]} = \frac{7500}{100[mA]} = 75 \quad [k\Omega]$$

### DC/DC converter drive frequency setting

DC/DC converter drive frequency is determined by making RT resistance setting.

# ODrive frequency vs. RT resistance (ideal) relation equation

$$R_{RT} = \frac{15000}{f_{SW}[kHz]} \quad [k\Omega]$$

where fsw = DC/DC converter oscillation frequency [kHz]

This equation has become an ideal equation without any correction item included. For accurate frequency settings, thorough verification should be performed on practical sets.

### [Setting example]

To set DC/DC drive frequency "fsw" to 200 kHz, RRT is given by the following equation:

$$R_{RT} = \frac{15000}{f_{sw}[kHz]} = \frac{15000}{200[kHz]} = 75 \quad [k\Omega]$$

### **OVP/SCP Settings**

OVP pin is DC/DC output voltage's over voltage protection and short circuit protection input pin. OVP pin is a high impedance pin with no pull down resistor. Thus, at OPEN state please set the voltage input settings using voltage dividing resistor and such.

Respective OVP pin protection conditions are as below

| Protection<br>Name | Detection<br>Pin | Detection<br>Condition | Cancellation<br>Condition | Timer<br>Operations | Protection Type             |
|--------------------|------------------|------------------------|---------------------------|---------------------|-----------------------------|
| OVP                | OVP              | OVP>3.0V               | OVP<2.9V                  | No                  | DCDC stops during detection |
| SCP                | OVP              | OVP<0.1V               | OVP>0.1V                  | Yes                 | All latch                   |

# **OVP Detection Setting**

VOUT abnormally increase, voltage detected by OVP, VOVP<sub>DET</sub>, R1,R2 settings are as follows

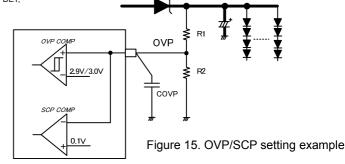
$$R1 = R2[k\Omega] \times \frac{(VOVP_{DET}[V] - 3.0[V])}{3.0[V]} \quad [k\Omega]$$

# **OVP Cancellation Setting**

R1,R2 set from above equation,

OVP cancellation voltage VOVP<sub>CAN</sub> equals to

$$VOVP_{CAN} = 2.9V \times \frac{(R1[k\Omega] + R2[k\Omega])}{R2[k\Omega]}$$
 [V]



VOUT

# **SCP Detection Setting**

When R1,R2 are set using values obtained above, SCP voltage setting is VSCP<sub>DET</sub> is as follows

$$VSCP_{DET} = 0.1V \times \frac{(R1[k\Omega] + R2[k\Omega])}{R2[k\Omega]} \quad [V]$$

### [Setting Example]

VOUT at normal operation 56V, OVP detection voltage VOVP<sub>DET</sub>=68V, R2=10k, R1 is as follows

$$R1 = R2[k\Omega] \times \frac{(VOVP_{DET}[V] - 3.0[V])}{3.0[V]} = 10[k\Omega] \times \frac{(68[V] - 3[V])}{3[V]} = 216.7 \quad [k\Omega]$$

When R1, R2 are set at these values, OVP cancellation voltage, VOVP<sub>CAN</sub>

$$VOVP_{CAN} = 2.9[V] \times \frac{(R1[k\Omega] + R2[k\Omega])}{R2[k\Omega]} = 2.9[V] \times \frac{10[k\Omega] + 216.7[k\Omega]}{10[k\Omega]}[V] = 65.7 \quad [V]$$

In addition, at this R1, R2, SCP detection voltage

$$VSCP_{DET} = 0.1[V] \times \frac{(R1[k\Omega] + R2[k\Omega])}{R2[k\Omega]} = 0.1[V] \times \frac{10[k\Omega] + 216.7[k\Omega]}{10[k\Omega]}[V] = 2.27 \quad [V]$$

To select DC/DC components, give consideration to IC variations as well as individual component variations, and then conduct thorough verification on practical systems.

# **Timer Latch Time Setting**

This IC has a built-in timer latch counter. Timer latch time is set by counting the clock frequency which is set at the RT pin. After

# Timer Latch Time

When various abnormal conditions happen, counting starts from the timing, latch occurs after below time has passed. Furthermore, even if PWM=L, if abnormal condition continues, timer count will not reset.

$$LATCH_{TIME} = 2^{12} \times \frac{R_{RT}}{1.5 \times 10^{10}} = 4096 \times \frac{R_{RT}[k\Omega]}{1.5 \times 10^7}[s]$$

Here, LATCH<sub>TIME</sub> = time until latch condition occurs  $R_{RT}$  = Resistor value connected to RT pin

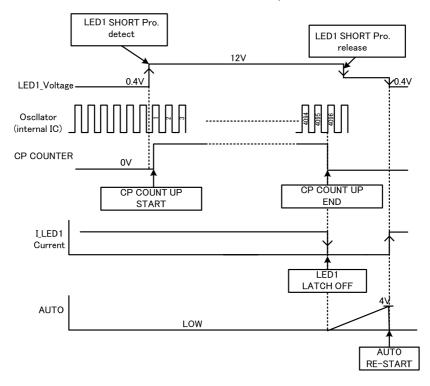


Figure 16. Example of LED Short protection timing chart

# [Setting Example]

Timer latch time when RT=75kohm

$$LATCH_{TIME} = 4096 \times \frac{R_{RT}[k\Omega]}{1.5 \times 10^7} = 4096 \times \frac{75[k\Omega]}{1.5 \times 10^7} = 0.02[s]$$

# **OCP Settings/DCDC Components' Current Capacity Selection Method**

One of the function of CS pin - when its pin voltage>0.45 it stops the DCDC. Thus, RCS resistor value need to be checked after the peak current flow through the inductor is calculated. Furthermore, DCDC external components' current capacity needs to be greater than peak current flowing through this inductor.

# (Inductor peak current Ipeak calculation method)

Firstly, ripple voltage which occurs at the CS pin is decided depending on the DCDC application conditions.

The conditions when made as below;

Output voltage=VOUT[V] LED total current=IOUT[A] DCDC input voltage=VIN[V] DCDC efficiency =n[%]

Total required average input current IIN:

$$I_{IN} = \frac{V_{OUT}[V] \times I_{OUT}[A]}{V_{IN}[V] \times \eta[\%]} \quad [A]$$

Inductor ripple current IL[A] which occurs at inductor L[H] during DCDC drive operation with switching frequency=fsw[Hz] is as follows

$$\Delta IL = \frac{(V_{OUT}[V] - V_{IN}[V]) \times V_{IN}[V]}{L[H] \times V_{OUT}[V] \times f_{SW}[Hz]} \quad [A]$$

Therefore, IL's peak current Ipeak can be calculated using below equation

$$Ipeak = I_{IN}[A] + \frac{\Delta IL[A]}{2} \quad [A] \cdots (1)$$

# (Resistor RCS connected to CS pin selection method)

This Ipeak flows in RCS and generates voltage. (refer to time chart diagram on the right). This voltage value, VCSpeak can be calculated as below

$$VCS_{peak} = Rcs \times Ipeak$$
 [V]

This VCSpeak when reach 0.45V, will stop the DCDC output. Thus when selecting RCS value, below condition needs to be met.

$$Rcs[\Omega] \times Ipeak[A] < 0.45[V]$$

# (DCDC Components' Current Capacity Selection Method)

When OCP reach detection voltage CS=0.45V, locp current

$$I_{OCP} = \frac{0.45[V]}{Rcs[\Omega]} \quad [A] \cdots (2)$$

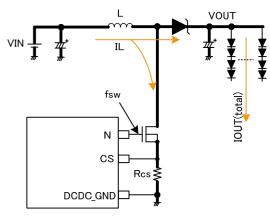


Figure 17. DC/DC convertor application circuit

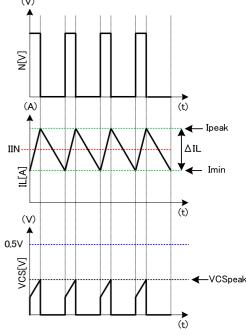


Figure 18. Inductor current waveform

Ipeak current (1), I<sub>OCP</sub> current (2), and components' MAX current capacity needs to satisfy the following

$$I_{\it peak} < I_{\it OCP} < \,$$
 Rated current of components

Above condition needs to be satisfied when selecting DCDC application parts eg. FET, inductor, diode etc. Furthermore, continuous mode is recommended for normal DCDC applications. Inductor's ripple current MIN limit value, Imin becoming

Im 
$$in = I_{IN}[A] - \frac{\Delta IL[A]}{2}[A] > 0$$

Is a condition to be met. If this is not met, it is called discontinuous mode.

# [Setting Example]

Output voltage=VOUT[V]=56V LED total current=IOUT[A]=100mA×4ch=0.40A DCDC input voltage=VIN[V]=14V DCDC efficiency=η[%]=90% Total required average input current IIN:

$$I_{IN}[A] = \frac{V_{OUT}[V] \times I_{OUT}[A]}{V_{IN}[V] \times \eta[\%]} = \frac{56[V] \times 0.40[A]}{14[V] \times 90[\%]} = 1.78 \quad [A]$$

When, DCDC switching frequency =fsw[Hz]=200kHz Inductor L[H]=33uH, Inductor ripple current  $\Delta$ IL[A]:

$$\Delta IL = \frac{(V_{OUT}[V] - V_{IN}[V]) \times V_{IN}[V]}{L[H] \times V_{OUT}[V] \times f_{SW}[Hz]} = \frac{(56[V] - 14[V]) \times 14[V]}{33 \times 10^{-6}[H] \times 56[V] \times 200 \times 10^{3}[Hz]} = 1.59 \quad [A]$$

Thus, IL peak current lpeak becomes

$$Ipeak = I_{IN}[A] + \frac{\Delta IL[A]}{2}[A] = 1.78[A] + \frac{1.59[A]}{2} = 2.58 \quad [A]$$
 ···Peak current calculation result

RCS resistor value when set at 0.10hm

$$VCS_{peak} = Rcs \times Ipeak = 0.10[\Omega] \times 2.58[A] = 0.258$$
 [V] < 0.45V ····RCS resistor consideration

and satisfy the condition.

In addition, OCP detection current  $I_{\text{OCP}}$  at this time is

$$I_{OCP} = \frac{0.45[V]}{0.1[\Omega]} = 4.5 \quad [A]$$

If parts used (FET,INDUCTOR, DIODE etc)'s current capacity<5A,

$$I_{\it peak} < I_{\it OCP} < {
m Rated current of components} = 2.58[A] < 4.5[A] < 5[A]$$

···DCDC current capacity consideration

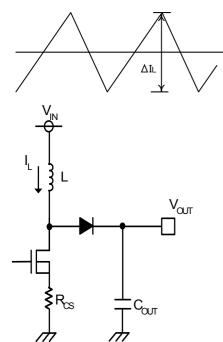
Thus, there is no problem of parts selection as the above condition is satisfied. In addition, IL ripple current minimum limit Imin is

$$\operatorname{Im} in = I_{IN}[A] - \frac{\Delta IL[A]}{2}[A] = 1.78[A] - 0.795[A] = 0.985[A] > 0$$

Thus, will not become discontinuous mode.

To select DC/DC components, give consideration to IC variations as well as individual component variations, and then conduct thorough verification on practical systems.

### Selection of inductor L



The value of inductor has significant influence on the input ripple current. As shown by Equation (1), the larger the inductor and the higher the switching frequency, the inductor ripple current  $\Delta IL$  becomes increasingly lower.

$$\Delta IL = \frac{(V_{OUT} - V_{IN}) \times V_{IN}}{L \times V_{OUT} \times f_{SW}} [A] \qquad (1)$$

Expressing efficiency as shown by Equation (2), peak input current is given as Equation (3).

$$\eta = \frac{V_{OUT} \times I_{OUT}}{V_{IN} \times I_{IN}} \qquad (2)$$

$$IL_{MAX} = I_{IN} + \frac{\Delta IL}{2} = \frac{V_{OUT} \times I_{OUT}}{V_{IN} \times \eta} + \frac{\Delta IL}{2} \qquad (3)$$

where, L: Reactance value [H],

V<sub>OUT</sub>: DC/DC output voltage [V],

V<sub>IN</sub>: Input voltage [V],

I<sub>OUT</sub>: Output load current (total output current) [A],

I<sub>IN</sub>: Input current [A], and F<sub>SW</sub>: Oscillation frequency [Hz]

Figure 19. Inductor current waveform and diagram

Note: If a current in excess of the rated current of the inductor applies to the coil, the inductor will cause magnetic saturation, resulting in efficiency degradation.

Select an inductor with an adequate margin so that peak current will not exceed the rated current of the inductor.

Note: To reduce power dissipation from and increase efficiency of inductor, select an inductor with low resistance component (DCR or ACR).

# Selection of output capacitor Cout

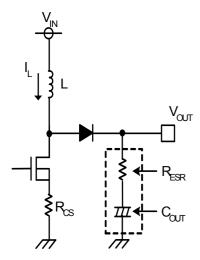


Figure 20. Output capacitor diagram,

Select a capacitor on the output side taking into account the stability region of output voltage and equivalent series resistance necessary to smooth ripple voltage. Note that higher output ripple voltage may result in a drop in LED pin voltage, making it impossible to supply set LED current.

The output ripple voltage  $\Delta V_{OUT}$  is given by Equation (4).

$$\Delta V_{OUT} = ILMAX \times R_{ESR} + \frac{1}{C_{OUT}} \times \frac{I_{OUT}}{\eta} \times \frac{1}{f_{SW}} [V] \qquad \cdot \quad \cdot \quad \cdot \quad (4)$$

where  $R_{ESR}$  = Equivalent series resistance of  $C_{OUT}$ .

Note: Select capacitor ratings with an adequate margin for output voltage.

Note: To use an electrolytic capacitor, an adequate margin should be provided for permissible current. Particularly to apply PWM light modulation to LED, note that a current higher than the set LED current transiently flows.

# **Selection of switching MOSFET transistors**

There will be no problem for switching MOSFET transistors having absolute maximum rating higher than rated current of the inductor L and VF higher than " $C_{OUT}$ " breakdown voltage + Rectifier diode". However, to achieve high-speed switching, select transistors with small gate capacity (injected charge amount).

Note: Rated current larger than over current protection setting current is recommended.

Note: Selecting transistors with low on resistance can obtain high efficiency.

# Selection of rectifier diodes

Select Schottky barrier diodes having current capability higher than the rated current of the inductor L and inverse breakdown voltage higher that C<sub>OUT</sub> breakdown voltage, particularly having low forward voltage VF.

# **Phase Compensation Setting Procedure**

DC/DC converter application for current mode control includes one each of pole fo (phase delay) by CR filer consisting of output capacitor and output resistor (i.e., LED current) and zero (phase lead) fz by the output capacitor and capacitor ESR. Furthermore, the step-up DC/DC converter includes RHP zero "fzRHP" as the second zero. Since the RHP zero has phase delay  $(-90^{\circ})$  characteristics like the pole, the crossover frequency  $f_c$  should be set to not more than RHP zero

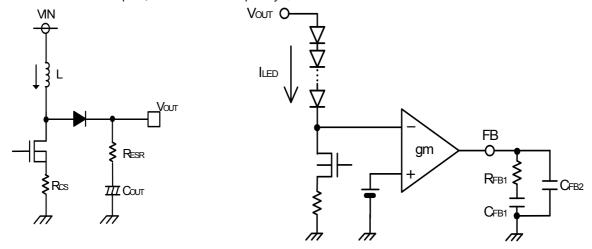


Figure 21. Output stage and error amplifier diagram

Find pole  $f_p$  and RHP zero  $f_{ZRHP}$  of DC/DC converter.

one for and RHP zero fights of DC/DC converter. 
$$f_p = \frac{I_{LED}}{2\pi \times V_{OUT} \times C_{OUT}} [Hz] \qquad f_{ZRHP} = \frac{V_{OUT} \times (1-D)^2}{2\pi \times L \times I_{LED}} [Hz]$$

Where 
$$I_{\!L\!E\!D}$$
 = Total LED current [A], 
$$D = \frac{V_{\scriptscriptstyle OUT} - V_{\scriptscriptstyle I\!N}}{V_{\scriptscriptstyle OUT}}$$

Find phase compensation to be inserted in the error amplifier. (Set  $f_c$  to 1/5 of  $f_{ZRHP}$ .)

$$R_{FB1} = \frac{f_{RHZP} \times R_{CS} \times I_{LED}}{5 \times f_{p} \times gm \times V_{OUT} \times (1 - D)} [\Omega]$$

$$C_{FB1} = \frac{1}{2\pi \times R_{FB1} \times f_C} = \frac{5}{2\pi \times R_{FB1} \times f_{ZRHP}} [F]$$

$$gm = 4.0 \times 10^{-4} [S]$$

iii. Find zero used to compensate ESR (RESR) of COUT (electrolytic capacitor).

$$C_{FB2} = \frac{R_{ESR} \times C_{OUT}}{R_{FB1}} [F]$$

Note: Even if a ceramic capacitor (R<sub>ESR</sub> of the order of milliohms) for C<sub>OUT</sub>, it is recommended to insert C<sub>FB2</sub> for stable operation.

To improve transient response, it is necessary to increase R<sub>FB1</sub> and reduce C<sub>FB1</sub>. However, this improvement reduces a phase margin. To avoid this problem, conduct thorough verification, including variations in external components, on practical systems.

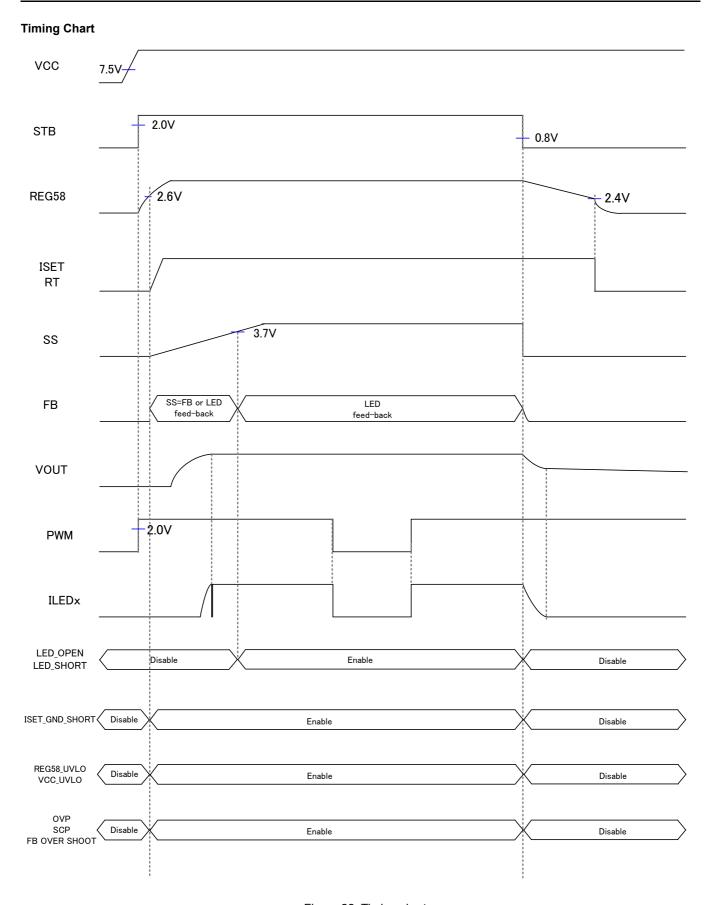


Figure 22. Timing chart

# **List of Protect Function (typ condition)**

| Protection       | Detection | Detection (   | Conditio | ns      | Cancellation | Drotaction Type  |
|------------------|-----------|---------------|----------|---------|--------------|--|
| Name             | Pin       | Detection pin | PWM      | SS      | Conditions   | Protection Type  |
| LED OPEN         | LEDx      | LEDx < 0.2V   | н        | SS>3.7V | LEDx > 0.2V  | Immediately Auto-Restart<br>after detection<br>(Judge periodically<br>whether normal or not) |
| LEDSHORT         | LEDx      | LEDx > 5V     | Н        | SS>3.7V | LEDx < 5V    | Immediately Auto-Restart<br>after detection<br>(Judge periodically<br>whether normal or not) |
| ISET GND         | ISET      | Under         | _        |         | Above        | Auto-Restart   |
| SHORT            | ISET      | ISET×90%      | -        | -       | ISET×90%     | Auto-Nestart   |
| REG58 UVLO       | REG58     | REG58<2.4V    | -        | -       | REG58>2.6V   | Auto-Restart   |
| VCC UVLO         | VCC       | VCC<7.3V      | -        | -       | VCC>7.5V     | Auto-Restart   |
| OVP              | OVP       | OVP>3.0V      | -        | -       | OVP<2.9V     | Auto-Restart   |
| SCP              | OVP       | OVP<0.1V      | -        | -       | OVP>0.1V     | Immediately Auto-Restart<br>after detection<br>(Judge periodically<br>whether normal or not) |
| FB OVER<br>SHOOT | FB        | FB>4V         | -        | -       | FB<3.6V      | Immediately Auto-Restart<br>after detection<br>(Judge periodically<br>whether normal or not) |
| ОСР              | CS        | OCP>0.45V     | -        | -       | -            | Pulse-by-Pulse   |

To clear the latch type, STB should be set to "L" once, and then to "H".

| Protection Function | Operation after the protection function detected |   |                           |  |  |  |
|---------------------|--|---|---------------------------|--|--|--|
| Protection Function | DC/DC  | LED Driver                              | Soft-start                |  |  |  |
| LED OPEN            | Continue to operate                              | Only detects LED, stops after CP1 count | Continue to operate       |  |  |  |
| LEDSHORT            | Continue to operate                              | Only detects LED, stops after CP1 count | Continue to operate       |  |  |  |
| ISET GND SHORT      | Stop immediately                                 | Stop immediately                        | Continue to operate       |  |  |  |
| STB                 | STB Stop immediately                             |   | Discharge immediately     |  |  |  |
| REG58 UVLO          | Stop immediately                                 | Stop immediately                        | Discharge immediately     |  |  |  |
| VCC UVLO            | Stop immediately                                 | Stop immediately                        | Discharge immediately     |  |  |  |
| OVP                 | Stop immediately<br>(N pin only)                 | Continue to operate                     | Continue to operate       |  |  |  |
| SCP                 | SCP Stop immediately (N pin only)                |   | Discharge after CP1 count |  |  |  |
| FB OVER SHOOT       | Stop after CP2 count                             | Stop after CP2 count                    | Continue to operate       |  |  |  |
| OCP                 | N pin limits DUTY                                | Continue to operate                     | Continue to operate       |  |  |  |

<sup>\*</sup> CP1 = 20ms (RT=75K $\Omega$ ) , CP2=1.31s (RT=75K $\Omega$ )

# **Operational Notes**

# 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply terminals.

# 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

# 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

### 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

# 7. Rush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

# 8. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

# 9. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

# **Operational Notes - continued**

### 10. Unused Input Terminals

Input terminals of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input terminals should be connected to the power supply or ground line.

### 11. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

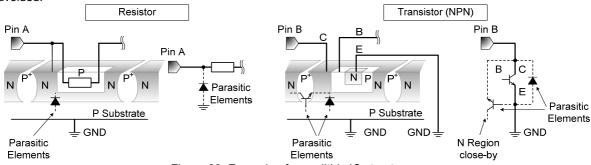


Figure 23. Example of monolithic IC structure

# 12. Ceramic Capacitor

When using a ceramic capacitor, determine the dielectric constant considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

### 13. Area of Safe Operation (ASO)

Operate the IC such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

# 14. Thermal Shutdown Circuit(TSD)

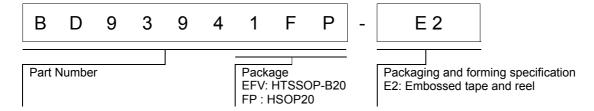
This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF all output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

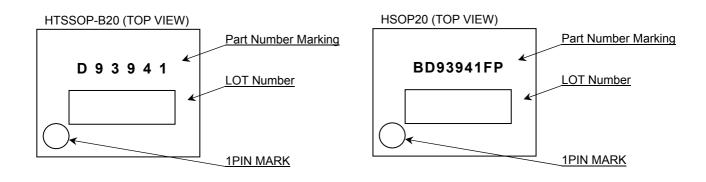
### 15. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

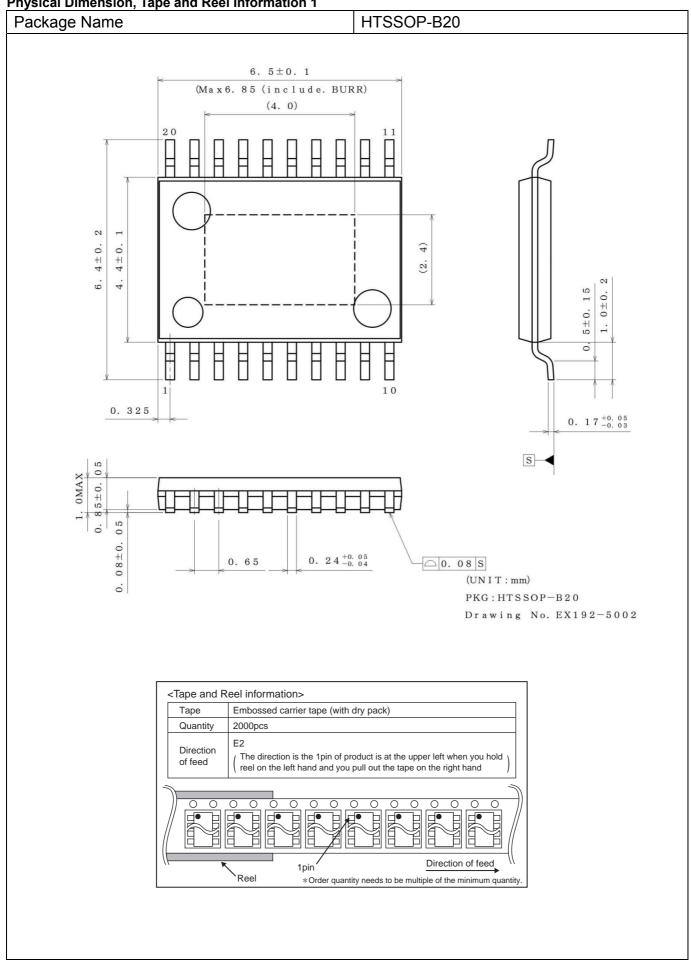
# **Ordering Information**

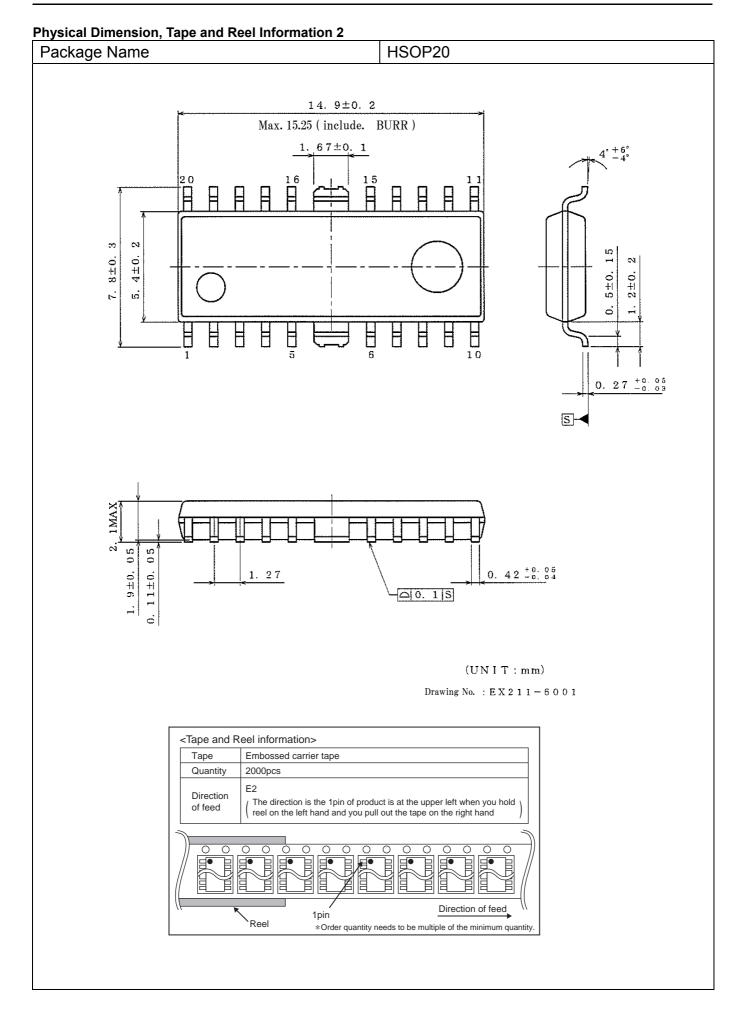


# **Marking Diagrams**



Physical Dimension, Tape and Reel Information 1





# **Revision History**

| Date        | Revision | Changes  |  |  |  |
|-------------|----------|--|--|--|--|
| 10.Oct.2013 | 001      | Draft Version  |  |  |  |
| 24.Dec.2014 | 002      | P.2 ADIM input voltage1 (use ADIM function) Add ADIM range at start up and add note *3 |  |  |  |
| 12.Aug.2015 | 003      | p.3 [REG58 Block] Soft start completion voltage -> REG58 available current             |  |  |  |

# **Notice**

# **Precaution on using ROHM Products**

Our Products are designed and manufactured for application in ordinary electronic equipments (such as AV equipment, OA equipment, telecommunication equipment, home electronic appliances, amusement equipment, etc.). If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment (Note 1), transport equipment, traffic equipment, aircraft/spacecraft, nuclear power controllers, fuel controllers, car equipment including car accessories, safety devices, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM's Products for Specific Applications.

(Note1) Medical Equipment Classification of the Specific Applications

| JÁPAN   | USA        | EU         | CHINA    |  |
|---------|------------|------------|----------|--|
| CLASSⅢ  | CL A CC TT | CLASS II b | CL ACCTI |  |
| CLASSIV | CLASSⅢ     | CLASSⅢ     | CLASSⅢ   |  |

- 2. ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:
  - [a] Installation of protection circuits or other protective devices to improve system safety
  - [b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure
- 3. Our Products are designed and manufactured for use under standard conditions and not under any special or extraordinary environments or conditions, as exemplified below. Accordingly, ROHM shall not be in any way responsible or liable for any damages, expenses or losses arising from the use of any ROHM's Products under any special or extraordinary environments or conditions. If you intend to use our Products under any special or extraordinary environments or conditions (as exemplified below), your independent verification and confirmation of product performance, reliability, etc, prior to use, must be necessary:
  - [a] Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
  - [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
  - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl<sub>2</sub>, H<sub>2</sub>S, NH<sub>3</sub>, SO<sub>2</sub>, and NO<sub>2</sub>
  - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
  - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
  - [f] Sealing or coating our Products with resin or other coating materials
  - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
  - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

# Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

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# **Precautions Regarding Application Examples and External Circuits**

- 1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
- You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

### **Precaution for Electrostatic**

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

# **Precaution for Storage / Transportation**

- 1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
  - [a] the Products are exposed to sea winds or corrosive gases, including Cl2, H2S, NH3, SO2, and NO2
  - [b] the temperature or humidity exceeds those recommended by ROHM
  - [c] the Products are exposed to direct sunshine or condensation
  - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

# **Precaution for Product Label**

QR code printed on ROHM Products label is for ROHM's internal use only.

### **Precaution for Disposition**

When disposing Products please dispose them properly using an authorized industry waste company.

# **Precaution for Foreign Exchange and Foreign Trade act**

Since concerned goods might be fallen under listed items of export control prescribed by Foreign exchange and Foreign trade act, please consult with ROHM in case of export.

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